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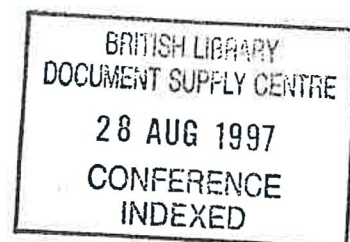
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TABLE of CONTENTS

SESSION A1: Control System Applications I

Session Chair: Hugh VanLandingham

*Feedback Control Solutions to Network Level User-Equilibrium
Real-Time Dynamic Traffic Assignment Problems*

P. Kachroo and K. Ozbay 1

Fuzzy Controller Design Using Genetic Algorithms

W. Hwang and S. Zein-Sabatto 6

SESSION B1: Photonics and Fiber Optics I

Session Chair: Anbo Wang

Design of a Programmable 2-18 GHz Microwave Fiber-Optic Delay Line

L. Wurtz, W. Wheless, and E. Bergsagel 11

Development of Fiber Optic Data Link Model

S. Wunnavva 20

Accuracy of Scalar-Wave Method for Evaluation of Dispersion in Broadband Single-Mode Fibers

C. So and A. Safaai-Jazi 23

Spectral Properties of Multiple Grating Structures

R. Lawson and A. Safaai-Jazi 26

SESSION C1: Digital Design I

Session Chair: James Armstrong

Methodology for VHDL Performance Model Construction and Validation

S. Vuppala, F. Gray, and J. Armstrong 29

Interfacing VHDL Performance Models to Algorithm Partitioning Tools

P. Balasubramanian and F. Gray 36

An Algorithm Based on the Hungarian Method for

Register Reduction during Complex Functional Unit Allocation

T. Lin and W. Cyre 43

SESSION A2: Control System Applications II

Session Chair: Richard Moose

Reducing Response Time in a Distributed Data Warehouse System

B. Panda and W. Perrizo 48

Program Generator for GPIB Instruments

I. Akundi and D. Smith 52

Analysis and Simulation of Nonlinear Systems Using Students' Version of MATLAB

R. Moose, H. VanLandingham, and N. Mclellan 58

SESSION B2: Photonics and Fiber Optics II

Session Chair: Ioannis M. Besieris

Electron-Beam-Addressed Spatial-Light-Modulator-Based Holography

K. Doh, B. Schilling, T. Poon, M. Wu, K. Shinoda, and Y. Suzuki 63

Performance Enhancement in Acoustic-Optic Image Processing

P. Banerjee, D. Cao, and T. Poon 68

Real-Time Image Processing Using Cascaded Acousto-Optic Modulators

D. Dunn, J. Xia, T. Poon, and P. Banerjee 71

SESSION C2: Digital Design II

Session Chair: James Armstrong

<i>A Novel Scheme for Irregular Parallel-Prefix Adders</i> C. Arjhan and R. Deshmukh	74
<i>Computer-Aided-Design Platform For Sequential Systems</i> T. Giurma, D. Welch, and K. MacDonald	79
<i>Generalized Flip-Flop Input Equations Based on a Four-Valued Boolean Algebra</i> J. Tucker and M. Tapia	82
SESSION A3: Speech Processing	
Session Chair: Douglas Linder	
<i>Robust Speech Recognition Techniques using a Radial Basis Function Neural Network for Mobile Applications</i> R. Sankar and N. Sethi	87
<i>Wavelet Based Analysis of Speech Under Stress</i> R. Sarikaya and J. Gowdy	92
<i>Issues in Generating Pronunciation Dictionaries for Voice Interfaces to Spatial Databases</i> J. Ngan and J. Picone	97
<i>Echo Cancellation For Evaluating Speaker Identification Technology</i> A. Ganapathiraju and J. Picone	100
SESSION B3: EM Analysis, Antennas, and Propagation I	
Session Chair: Warren Stutzman	
<i>The Spatial and Temporal Coherence Properties of the Excitation Wavefield of a Dynamic Gaussian Aperture</i> A. Shaarawi	103
<i>Attenuation and Phase Shift of the Electromagnetic Waves due to Moist Snow</i> M. Fares, S. Fares, and C. Ventrice	109
<i>A Comparative Study on the Reconstruction of Localized Pulses</i> M. Abdel-Rahman, I. Besieris, and A. Shaarawi	113
<i>A Comparison of Air Breakdown Criteria for Slightly Divergent Fields</i> J. Donohoe	118
SESSION C3: Computer and Communication Networks I	
Session Chair: Scott Midkiff	
<i>Performance Based Reliability Optimization for Computer Networks</i> S. Ahuja	121
<i>Enforcement Mechanisms for Priority Traffic in ATM Networks</i> Q. Pang and S. Cheng	126
<i>Throughput and Video Performance of Emerging LAN Technologies: Switched Ethernet and LAN emulation over ATM</i> A. Rindos, S. Woollet, D. Cosby, J. Sents, M. Aydemir, M. Vouk, and Z. Ortiz	131
<i>Performance Comparison of FDDI Models</i> M. Sadiku and G. Dempo	135
SESSION A4: Signal Processing and Estimation Techniques	
Session Chair: Joel Donahue	
<i>Investigating the Use of Kalman Filtering Approaches for Dynamic Origin-Destination Trip Table Estimation</i> P. Kachroo, K. Ozbay, and A. Narayanan	138
<i>Estimation of ARMA Parameters Using Third Order Cumulants</i> A. Al-Smadi and D. Wilkes	143
<i>On the Edge Preserving Smoothing Filter</i> C. Hung	146

<i>Acoustic Echo and Noise Canceller Improvements for Hands-Free Telephones</i> S. Sankaran and A. Beex	148
<i>Portable Speech Recognition for the Speech and Hearing Impaired</i> Y. Alsaka, S. Doll, and S. Davis	151
SESSION B4: EM Analysis, Antennas, and Propagation II	
Session Chair: Amr Shaarawi	
<i>All-Fiber Spectral Filter Based on LP01-LP11 Mode Coupling</i> J. Jagannathan and A. Safaai-Jazi	154
<i>Three Coupled Microstrip Lines on Multi-Dielectric Interfaces</i> I. Barseem, E. Hashish, E. Abdallah, and M. El-Said	157
<i>Hybrid model of Electromagnetic Scattering From an Inhomogeneous Dielectric Cylinder of Arbitrary Cross Section</i> M. Kolbehdari and M. Sadiku	162
<i>Finite Element Computation of the Characteristic Impedance of Transmission Lines</i> M. Sadiku and C. Obiozor	166
<i>Effects of Array Element Patterns on Optimum Beamforming in the Presence of Multiple Interferers</i> C. Dietrich and W. Stutzman	169
SESSION C4: Computer and Communication Networks II	
Session Chair: Scott Midkiff	
<i>A Sample Statistical Characterization of the World-Wide Web</i> D. Lee and S. Midkiff	174
<i>An Experimental Investigation of the Internet Integrated Services Model using the Resource Reservation Protocol (RSVP)</i> M. Blackford, J. Austen, and D. Marlow	179
<i>Multi Point Teleconferencing with ISDN</i> S. Wunnavva and K. Mallampati	182
<i>Implementation of CAN/CAN Bridges in Distributed Environments and Performance Analysis of Bridged CAN Systems using SAE Benchmark</i> H. Ekiz, A. Kutlu, and E. Powner	185
<i>Model LAN Scheme and Compressed Data Transfers</i> S. Wunnavva, T. Williams, and M. Choy	188
SESSION A5: Education	
Session Chair: Joseph Tront	
<i>Evolutionary Algorithms for Design</i> H. VanLandingham and S. Sampan	191
<i>Using EPRI Distribution Engineering Workstation as an Educational Tool</i> H. Shaalan and R. Broadwater	196
<i>Power in a Lossy Transmission Line Excited by Sinusoids</i> C. Obiozor and M. Sadiku	199
<i>Distance Learning and use of the Internet and the World Wide Web in Education</i> A. Aguilar, R. Eaglin, I. Batarseh, and N. Bakir	202
<i>A Real-Time Experimental-Analysis System for the Undergraduate Electric Energy Laboratory</i> O. Mohammed and R. Sebastien	207
SESSION A6: Neural and Intelligent Design I	
Session Chair: Hugh VanLandingham	
<i>Intelligent Traffic Monitoring System Using Wireless Cellular Communications</i> R. Sankar and L. Civil	210

<i>Parameter Estimation of Mean Field Annealing Technique</i> K. Sohn and J. Kim	215
<i>On the Behavior of a VQ Algorithm</i> W. Porter and A. Abou-Ali	220
SESSION B6: Electronic Circuits and Materials I	
Session Chair: Edward Wheeler	
<i>Design of Hybrid Converter for Repetitive Charge Mode</i> S. Siddiquie and A. Samra	222
<i>New Low-Voltage Circuits for XOR and XNOR</i> H. Lee and G. Sobelman	225
<i>Deposition and Annealing Studies on Indium Tin Oxide Films</i> K. Sundaram and J. Blanchard	230
<i>CMOS Technology for Ultra-Low Power Circuit Applications</i> C. Salomonson, W. Henley, D. Whittaker, and J. Maimon	233
SESSION C6: Communication Theory and Systems I	
Session Chair: Richard Moose	
<i>A Model of Bandwidth Utilization in Wide-Band Computer Communication Networks</i> C. Katsinis	236
<i>Performance Improvement of CDMA Power Control in Variable Fading Environments</i> S. Won, W. Kim, and I. Jeong	241
<i>Error Floors in Narrowband Digital FM Systems</i> R. Petrovic and S. Filipovic	244
SESSION A7: Neural and Intelligent Design II	
Session Chair: Bogdan Tomescu	
<i>A Heirarchical Approach to Recognition of Finished Wooden Components</i> A. Abbott and Z. Huang	249
<i>Performance Comparisons of Interpolation Algorithms for 2D Images</i> C. Wang, W. Snyder, and G. Bilbro	252
<i>Shape Coded Modulation Using Neural Networks</i> A. Als and J. Sodha	255
<i>A Parallel Approach to Digital Holographic Reconstruction with Twin-Image Removal</i> C. Tu, J. Gerlach, S. Chu, and T. Poon	258
<i>Data Types of Physical Quantities</i> J. Martin	260
SESSION B7: Electronic Circuits and Materials II	
Session Chair: Usha Varshney	
<i>Photoluminescence Spectrum Redshifting of Porous Silicon Layers Prepared from Diamond Scratched Silicon Samples</i> K. Sundaram, J. Blanchard, S. Albin, J. Zheng, and A. Lavarias	264
<i>Doping ZnSe Far From Equilibrium</i> E. Wheeler, N. Brakensiek, J. Boone, and G. Cantwell	267
<i>Electrical and Vibration Analysis of Pre-compressed and Pre-deflected Laminated Piezoelectric Actuators</i> N. Nechitailo, K. Lo, H. Moses, H. Groger, L. Decker, and R. Churchill	271
<i>Characterizing State Devices for Switching Noise</i> R. Hodson, A. Bampton, and P. Knipp	276

SESSION C7: Communication Theory and Systems II

Session Chair: Richard Moose

Modified Chase Algorithms for Reed-Solomon Codes

P. Chiuchiolo and J. Komo 278

Hybrid ARQ with Ratio Threshold Combining for Frequency-Hop Communications Systems

F. Block and C. Baum 281

TelePath: Real-Time Remote Pathology

S. McClellan and T. Winokur 284

Dispersion-Shifted Fibers for Very Long Links

H. Hattori and A. Safaai-Jazi 287

Detection of Ocean Reflected GPS Signals: Theory and Experiment

J. Garrison, S. Katzberg, and C. Howell III 290

SESSION A8: Power and Energy

Session Chair: Subhash Sarkar

Library Based Hierarchical Power Estimation Method Using In/Out Transition Activity

J. Lee and L. Lucke 295

Monitoring and Data Acquisition for a Large Roof-Mounted Photovoltaic Array

M. Begovic, M. Ropp, A. Rohatgi, S. Durand, and A. Rosenthal 298

Identification of Harmonic Sources in Power Distribution Systems

K. Islam and A. Samra 301

Electrical Safety in Medical Institutions, "Neutral Electric Potentiality System in Hospitals"

M. Osman, A. Todorova, and A. Samra 304

Optimum Series Compensated High Voltage Transmission Lines

G. Bellarmine 307

SESSION B8: One and Two Dimensional Signal Processing I

Session Chair: Gail Gray

Surface Blemish Detection from Passive Imagery Using Learned Fuzzy Set Concepts

S. Gurbuz, A. Carver, and R. Schalkoff 310

A Custom Computing Solution to Automated Visual Inspection of Silicon Wafers

P. Athanas, A. Abbott, M. Cherbaka, B. Pudipeddi, and K. Paar 315

Parallel Architecture Selection for Image Tracking Applications

A. Elahi 320

SESSION B9: One and Two Dimensional Signal Processing II

Session Chair: Hugh VanLandingham

Subvector Techniques for DFT Parallel Processing

G. Rodgers 323

Benchmarking of FFT Algorithms

M. Balducci, A. Ganapathiraju, J. Hamaker, J. Picone, A. Choudary, and A. Skjellum 328

Encoding a Hidden Auxiliary Channel onto a Digital Audio Signal Using Psychoacoustic Masking

J. Tilki and A. Beex 331

DSP Based Virtual Power Measurement in Nonsinusoidal Electric Systems

A. Samra and K. Syed 334

Scenic Beauty Estimation of Forestry Images

N. Kalidindi, A. Le, J. Picone, L. Zheng, H. Yaqin, and V. Rudis 337

TUTORIAL ABSTRACTS 340

SYMPOSIUM ABSTRACTS 349

INDEX OF AUTHORS 360

ENCODING A HIDDEN AUXILIARY CHANNEL ONTO A DIGITAL AUDIO SIGNAL USING PSYCHOACOUSTIC MASKING

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Abstract - We report on the development of a method of encoding an auxiliary channel onto a digital audio signal such that it is imperceptible to human observers. The encoding is accomplished by imposing slight and controlled changes on the phase spectrum of short-time signal windows. By employing principles of psychoacoustics the auxiliary channel is masked from human perception. Furthermore, the encoding and decoding are achieved via fast algorithms which allow real time processing. The method is applicable for any digitized audio signal containing voice, music, or other acoustic signals to be heard by humans. Possible signal sources include compact discs, digital television, digital radio, digital telephony, and any other source where an audio signal is in a digital format.

INTRODUCTION

The ability to add extra hidden information to a signal could be useful in many applications. Audio compact discs could be modified to contain artist information, song titles, song lyrics, karaoke information, still images, and perhaps even video clips. In other contexts an extra channel could be used to transmit control information. The want and need for additional bandwidth is ever-increasing, and the capability to add extra information to digital audio can be quite useful even for archival and storage purposes.

The coding method presented in this paper allows the addition of such an auxiliary channel, and it does so in a manner that does not significantly degrade the perceived quality of the primary audio channel. Furthermore, the method does not require changes in the underlying digital format, and so it completely maintains backward compatibility with existing data formats. One can imagine the possibilities for the audio CD example. Existing CD players could still play CDs encoded with the auxiliary channel. With additional signal processing capability however, new CD players could provide the enhanced functionality described above.

Other researchers have aimed at the same objective using methods such as subband coding coupled with adaptive quantization, and subtractively dithered noise-

shaped quantization [1-3]. However, a direct comparison between the schemes in terms of auxiliary channel data rate versus perceived quality of the primary signal could not be performed. Since the other approaches seem to require more computations, our method could be an excellent alternative especially at high sampling rates and where computational capability is a limiting factor.

DESCRIPTION OF CODING METHOD

Encoding

It is well known that humans are relatively insensitive to certain types of phase distortions. We use this fact to encode additional information onto the phase of selected bins of an FFT. The signal bins are selected according to human sensitivity in the corresponding frequency bands, and are spaced to facilitate masking by unmodified neighboring bins in the same spectral region. In particular the lower frequencies (below say 2 kHz), which typically contain the most energy in common audio signals, are left untouched. The stronger low frequency components help to mask the changes made in the higher frequency region. Additionally, the unmodified FFT bins in the higher frequency region help to mask their modified neighbors.

The FFT bins immediately preceding the signal bins are used as references. While keeping the digital audio signal's FFT amplitude the same, the phase of the FFT at the signal bins is discarded, and a new phase is assigned based on the neighboring reference bins, a differential phase change, and the digital information to be encoded. For example, a small clockwise rotation of phase relative to the reference can represent a digital "1" while a counter-clockwise rotation of phase represents a digital "0".

Suppose that the FFT has a complex value of X_r at a reference bin. In polar form this value is represented as

$$X_r = R_r e^{j\theta_r} \quad (1)$$

where R_r is the magnitude and θ_r is the phase. Similarly the value of the FFT at the neighboring signal bin is

$$X_s = R_s e^{j\theta_s} \quad (2)$$

Then if ϕ is the differential phase change, the new FFT value for the signal bin is assigned as

$$\hat{X}_s = R_s e^{j(\theta_s \pm \phi)} \quad (3)$$

Note that it is the phase of the *reference* bin that is used along with the differential phase to replace the phase of the signal bin. Addition of the differential phase represents a digital one and subtraction a digital zero.

Once the phases of the signal bins have been modified, the conjugate phase is assigned to the matching negative frequency bins. An inverse FFT then yields the modified signal block containing both the primary audio and the secondary channel. The process is then repeated for subsequent signal blocks.

Figure 1 shows a small region of an example phase spectrum and its modified version after encoding. The phase at the reference bins is denoted by asterisks, the original phase at the signal bins is denoted by 'o', and the modified phase at the signal bins is denoted by 'x'. This example shows a digital pattern of [0 1 0 0 1].

For a given source sample rate the FFT size is chosen to be short enough in time to maintain the imperceptibility of the introduced phase distortions. However, the FFT size should conversely be chosen as large as possible under the above constraint such that the FFT bin frequencies are as dense as possible. In our limited simulations, FFT sizes corresponding to block lengths of 10 to 50 msec worked well. Smaller blocks often yielded "granular" noise, while larger blocks usually created noticeable distortion of the primary signal.

Decoding

The signal is decoded by comparing the phase at the (known) signal bins to the phase at the neighboring reference bins, and assigning bits based on the direction of relative phase rotation.

Synchronization

The decoding process requires perfect synchronization with the encoded signal blocks. For a storage medium such as a compact disc synchronization is simple and only requires knowledge of the offset of the first signal block. For real-time data streams however, synchronization is more difficult. Several options are available, including decoding candidate blocks until an error detection criterion is satisfied. Error detection can be accomplished with cyclic redundancy checks (CRC) or other efficient error detection schemes [4]. The candidate block alignments are changed one sample at a time until a valid data transmission is detected. The offset at the point of valid data detection provides synchronization for all subsequent signal blocks. Figure 2 shows how zero bit errors result when perfect synchronization is achieved. When synchronization is

incorrect by even a single sample delay or advance several bit errors result (≥ 38 in the case shown).

Another possibility for synchronization is the inclusion of known bit patterns in the data stream. This situation also requires shifts by a single sample at a time until the embedded bit pattern is detected. Such a "control" function for synchronization has been used in a similar context in the analog channel case [5].

SIMULATION RESULTS

The coding method has been tested and works well. The example provided in the paper is based on four seconds of the audio component of a TV commercial, sampled monophonically at a rate of 44.1 kHz, consistent with CD quality audio. The lowpass nature of this audio signal is shown in Figure 3. (Note that the sinusoid at 15.734 kHz is interference from the horizontal synchronization pulse for the video electron beam.) The decoded phase errors would be zero were it not for the quantization back to 16-bits that occurs after the inverse FFT. The phase errors due to quantization are shown in Figure 4. A differential phase of $\pi/8$ was used to modify the phase of every fourth FFT bin above 2 kHz in 2048 point FFT blocks. A secondary channel resulted with a capacity of over 5000 bits per second with little or no degradation in the perceived quality of the primary audio signal.

Higher data rates can be achieved by expanding the frequency region used for coding and using smaller spacing between signal bins in the FFT. However, as would be expected, the perceived quality of the primary signal suffers as the data rate of the auxiliary channel increases.

FUTURE IMPROVEMENTS

Future improvements include using several increments of the differential phase for the forward and reverse rotations of phase, allowing for the coding of multiple bits per FFT bin. Also, pre-quantization of encoded FFT blocks should allow pre-compensation for quantization effects during the encoding process. In addition, the audio signal can be decomposed into subbands, with each subband independently used for coding. This would allow more bits to be encoded in selected subbands based on the relative influence the individual subbands have on overall perception.

Finally, adaptive encoding based on local power levels and frequency content in the primary audio signal can perhaps increase the effective capacity of the auxiliary channel. A related variation would be adaptation of the differential phase change based on the local frequency content and on some measure of phase deviation in the region. This could be used to compensate for the fact that frequency regions with lower energy are more sensitive to

quantization effects than regions with more energy. Note that the phase errors in Figure 4 are largest in the regions of lowest energy, as seen from comparison with Figure 3.

CONCLUSION

In conclusion, we have developed a method of successfully encoding an auxiliary channel onto a digital audio signal. This channel is encoded using psychoacoustic principles such that it is imperceptible to human observers. Both the encoding and decoding are accomplished with the FFT, allowing real time implementations. The coding process is appropriate for any application containing a digitized audio signal.

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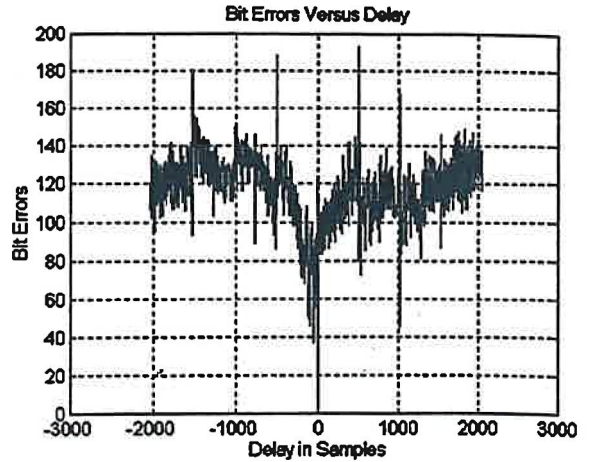


Figure 2. Block Synchronization by Error Detection.

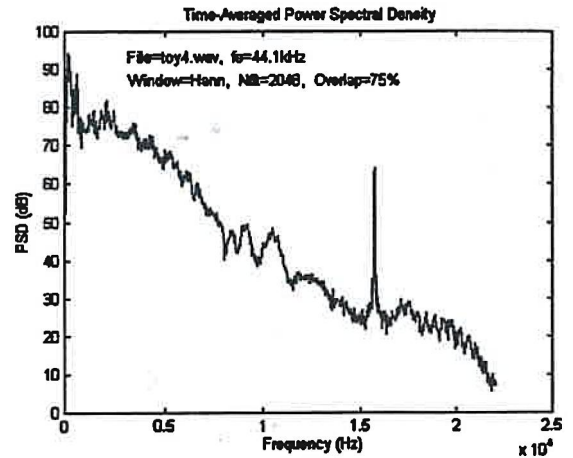


Figure 3. Time-Averaged PSD of Audio Signal.

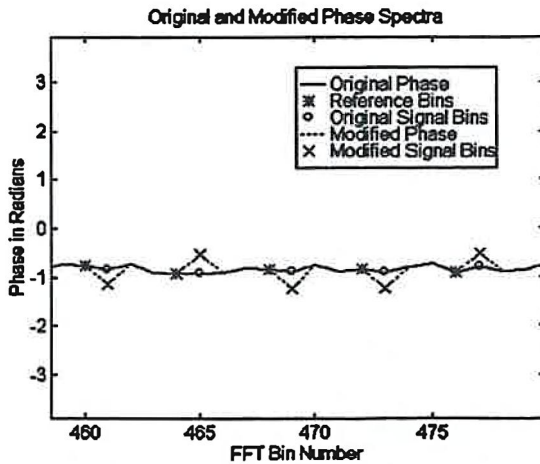


Figure 1. Example Phase Spectra.

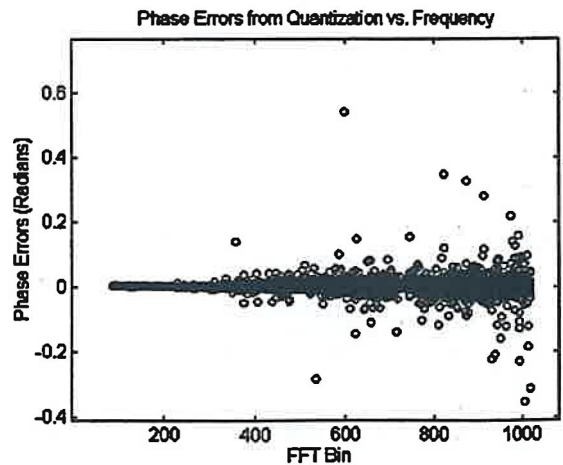


Figure 4. Phase Errors from Quantization vs. Frequency.