

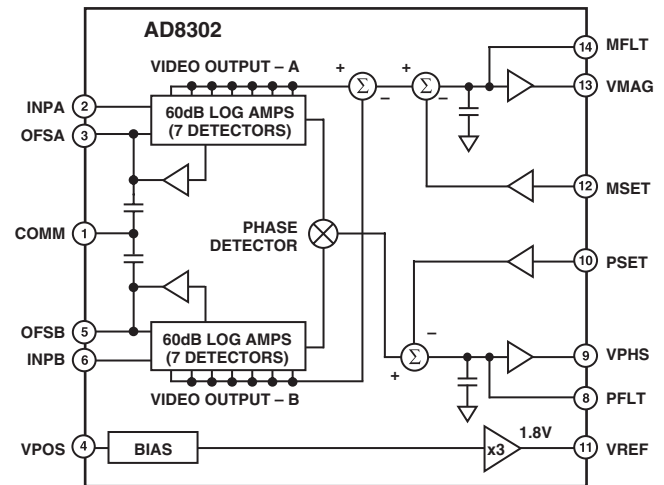
FEATURES

Measures Gain/Loss and Phase up to 2.7 GHz
Dual Demodulating Log Amps and Phase Detector
Input Range –60 dBm to 0 dBm in a 50 Ω System
Accurate Gain Measurement Scaling (30 mV/dB)
Typical Nonlinearity < 0.5 dB
Accurate Phase Measurement Scaling (10 mV/Degree)
Typical Nonlinearity < 1 Degree
Measurement/Controller/Level Comparator Modes
Operates from Supply Voltages of 2.7 V–5.5 V
Stable 1.8 V Reference Voltage Output
Small Signal Envelope Bandwidth from DC to 30 MHz

APPLICATIONS

RF/IF PA Linearization
Precise RF Power Control
Remote System Monitoring and Diagnostics
Return Loss/VSWR Measurements
Log Ratio Function for AC Signals

FUNCTIONAL BLOCK DIAGRAM



PRODUCT DESCRIPTION

The AD8302 is a fully integrated system for measuring gain/loss and phase in numerous receive, transmit, and instrumentation applications. It requires few external components and a single supply of 2.7 V–5.5 V. The ac-coupled input signals can range from –60 dBm to 0 dBm in a 50 Ω system, from low frequencies up to 2.7 GHz. The outputs provide an accurate measurement of either gain or loss over a ± 30 dB range scaled to 30 mV/dB, and of phase over a 0°–180° range scaled to 10 mV/degree. Both subsystems have an output bandwidth of 30 MHz, which may optionally be reduced by the addition of external filter capacitors. The AD8302 can be used in controller mode to force the gain and phase of a signal chain toward predetermined setpoints.

The AD8302 comprises a closely matched pair of demodulating logarithmic amplifiers, each having a 60 dB measurement range. By taking the difference of their outputs, a measurement of the magnitude ratio or gain between the two input signals is available. These signals may even be at different frequencies, allowing the measurement of conversion gain or loss. The AD8302 may be used to determine absolute signal level by applying the unknown signal to one input and a calibrated ac reference signal to the other. With the output stage feedback connection disabled, a comparator may be realized, using the setpoint pins MSET and PSET to program the thresholds.

The signal inputs are single-ended, allowing them to be matched and connected directly to a directional coupler. Their input impedance is nominally 3 k Ω at low frequencies.

The AD8302 includes a phase detector of the multiplier type, but with precise phase balance driven by the fully limited signals appearing at the outputs of the two logarithmic amplifiers. Thus, the phase accuracy measurement is independent of signal level over a wide range.

The phase and gain output voltages are simultaneously available at loadable ground referenced outputs over the standard output range of 0 V to 1.8 V. The output drivers can source or sink up to 8 mA. A loadable, stable reference voltage of 1.8 V is available for precise repositioning of the output range by the user.

In controller applications, the connection between the gain output pin VMAG and the setpoint control pin MSET is broken. The desired setpoint is presented to MSET and the VMAG control signal drives an appropriate external variable gain device. Likewise, the feedback path between the phase output pin VPHS and its setpoint control pin PSET may be broken to allow operation as a phase controller.

The AD8302 is fabricated on Analog Devices' proprietary, high performance 25 GHz SOI complementary bipolar IC process. It is available in a 14-lead TSSOP package and operates over a –40°C to +85°C temperature range. An evaluation board is available.

REV. A

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AD8302—SPECIFICATIONS ($T_A = 25^\circ\text{C}$, $V_S = 5\text{ V}$, VMAG shorted to MSET, VPHS shorted to PSET, $52.3\ \Omega$ shunt resistors connected to INPA and INPB, for Phase measurement $P_{\text{INPA}} = P_{\text{INPB}}$, unless otherwise noted.)

Parameter	Conditions	Min	Typ	Max	Unit	
OVERALL FUNCTION						
Input Frequency Range	P_{IN} at INPA, P_{IN} at INPB = -30 dBm ϕ_{IN} at INPA $> \phi_{\text{IN}}$ at INPB Pin VREF, $-40^{\circ}\text{C} \leq T_{\text{A}} \leq +85^{\circ}\text{C}$	>0		2700	MHz	
Gain Measurement Range			± 30		dB	
Phase Measurement Range			± 90		Degree	
Reference Voltage Output		1.72	1.8	1.88	V	
INPUT INTERFACE						
Input Simplified Equivalent Circuit	Pins INPA and INPB To AC Ground, $f \leq 500\text{ MHz}$ AC-Coupled (0 dBV = 1 V rms) re: $50\ \Omega$	-73	$3\parallel 2$	-13	$\text{k}\Omega\parallel\text{pF}$	
Input Voltage Range		-60		0	dBV	
Center of Input Dynamic Range			-43 -30		dBV dBm	
MAGNITUDE OUTPUT						
Pin VMAG	Pin VMAG $20 \times \text{Log} (V_{\text{INPA}}/V_{\text{INPB}}) = -30\text{ dB}$ $20 \times \text{Log} (V_{\text{INPA}}/V_{\text{INPB}}) = +30\text{ dB}$ $V_{\text{INPA}} = V_{\text{INPB}}$ Source/Sink Pin MFLT Open 40 dB Change, Load $20\text{ pF}\parallel 10\text{ k}\Omega$ Any 20 dB Change, 10%–90% Any 20 dB Change, 90%–10% Full-Scale 60 dB Change, to 1% Settling		30		mV	
Output Voltage Minimum			1.8		V	
Output Voltage Maximum			900		mV	
Center Point of Output (MCP)			8		mA	
Output Current			30		MHz	
Small Signal Envelope Bandwidth			25		V/ μs	
Slew Rate						
Response Time			50		ns	
Rise Time			60		ns	
Fall Time			300		ns	
Settling Time						
PHASE OUTPUT						
Pin VPHS	Pin VPHS Phase Difference 180 Degrees Phase Difference 0 Degrees When $\phi_{\text{INPA}} = \phi_{\text{INPB}} \pm 90^{\circ}$ Source/Sink Any 15 Degree Change, 10%–90% 120 Degree Change $C_{\text{FILT}} = 1\text{ pF}$, to 1% Settling		30		mV	
Output Voltage Minimum			1.8		V	
Output Voltage Maximum			900		mV	
Phase Center Point			8		mA	
Output Current Drive			25		V/ μs	
Slew Rate			30		MHz	
Small Signal Envelope Bandwidth			40		ns	
Response Time			500		ns	
100 MHz						
MAGNITUDE OUTPUT						
Dynamic Range		$\pm 1\text{ dB}$ Linearity $P_{\text{REF}} = -30\text{ dBm}$ ($V_{\text{REF}} = -43\text{ dBV}$) $\pm 0.5\text{ dB}$ Linearity $P_{\text{REF}} = -30\text{ dBm}$ ($V_{\text{REF}} = -43\text{ dBV}$) $\pm 0.2\text{ dB}$ Linearity $P_{\text{REF}} = -30\text{ dBm}$ ($V_{\text{REF}} = -43\text{ dBV}$) From Linear Regression		58		dB
				55		dB
			42		dB	
Slope	Deviation from Linear Regression		29		mV/dB	
Deviation vs. Temperature	Deviation from Output at 25°C $-40^{\circ}\text{C} \leq T_{\text{A}} \leq +85^{\circ}\text{C}$, $P_{\text{INPA}} = P_{\text{INPB}} = -30\text{ dBm}$		0.25		dB	
	Deviation from Best Fit Curve at 25°C $-40^{\circ}\text{C} \leq T_{\text{A}} \leq +85^{\circ}\text{C}$, $P_{\text{INPA}} = \pm 25\text{ dB}$, $P_{\text{INPB}} = -30\text{ dBm}$		0.25		dB	
	Gain Measurement Balance	$P_{\text{INPA}} = P_{\text{INPB}} = -5\text{ dBm}$ to -50 dBm		0.2		dB
PHASE OUTPUT						
Dynamic Range	Less than ± 1 Degree Deviation from Best Fit Line Less than 10% Deviation in Instantaneous Slope From Linear Regression about -90° or $+90^{\circ}$		145		Degree	
			143		Degree	
			10		mV/Degree	
Slope (Absolute Value)	Deviation from Output at 25°C $-40^{\circ}\text{C} \leq T_{\text{A}} \leq +85^{\circ}\text{C}$, Delta Phase = 90 Degrees		0.7		Degree	
Deviation vs. Temperature	Deviation from Best Fit Curve at 25°C $-40^{\circ}\text{C} \leq T_{\text{A}} \leq +85^{\circ}\text{C}$, Delta Phase = ± 30 Degrees		0.7		Degree	

Parameter	Conditions	Min	Typ	Max	Unit
900 MHz	MAGNITUDE OUTPUT				
Dynamic Range	± 1 dB Linearity $P_{REF} = -30$ dBm ($V_{REF} = -43$ dBV)		58		dB
	± 0.5 dB Linearity $P_{REF} = -30$ dBm ($V_{REF} = -43$ dBV)		54		dB
	± 0.2 dB Linearity $P_{REF} = -30$ dBm ($V_{REF} = -43$ dBV)		42		dB
Slope	From Linear Regression		28.7		mV/dB
Deviation vs. Temperature	Deviation from Output at 25°C				
	$-40^{\circ}\text{C} \leq T_A \leq +85^{\circ}\text{C}$, $P_{INPA} = P_{INPB} = -30$ dBm		0.25		dB
	Deviation from Best Fit Curve at 25°C				
	$-40^{\circ}\text{C} \leq T_A \leq +85^{\circ}\text{C}$, $P_{INPA} = \pm 25$ dB, $P_{INPB} = -30$ dBm		0.25		dB
Gain Measurement Balance	$P_{INPA} = P_{INPB} = -5$ dBm to -50 dBm		0.2		dB
	PHASE OUTPUT				
Dynamic Range	Less than ± 1 Degree Deviation from Best Fit Line		143		Degree
	Less than 10% Deviation in Instantaneous Slope		143		Degree
Slope (Absolute Value)	From Linear Regression about -90° or $+90^{\circ}$		10.1		mV/Degree
Deviation	Linear Deviation from Best Fit Curve at 25°C				
	$-40^{\circ}\text{C} \leq T_A \leq +85^{\circ}\text{C}$, Delta Phase = 90 Degrees		0.75		Degree
	$-40^{\circ}\text{C} \leq T_A \leq +85^{\circ}\text{C}$, Delta Phase = ± 30 Degrees		0.75		Degree
Phase Measurement Balance	Phase @ INPA = Phase @ INPB, $P_{IN} = -5$ dBm to -50 dBm		0.8		Degree
1900 MHz	MAGNITUDE OUTPUT				
Dynamic Range	± 1 dB Linearity $P_{REF} = -30$ dBm ($V_{REF} = -43$ dBV)		57		dB
	± 0.5 dB Linearity $P_{REF} = -30$ dBm ($V_{REF} = -43$ dBV)		54		dB
	± 0.2 dB Linearity $P_{REF} = -30$ dBm ($V_{REF} = -43$ dBV)		42		dB
Slope	From Linear Regression		27.5		mV/dB
Deviation vs. Temperature	Deviation from Output at 25°C				
	$-40^{\circ}\text{C} \leq T_A \leq +85^{\circ}\text{C}$, $P_{INPA} = P_{INPB} = -30$ dBm		0.27		dB
	Deviation from Best Fit Curve at 25°C				
	$-40^{\circ}\text{C} \leq T_A \leq +85^{\circ}\text{C}$, $P_{INPA} = \pm 25$ dB, $P_{INPB} = -30$ dBm		0.33		dB
Gain Measurement Balance	$P_{INPA} = P_{INPB} = -5$ dBm to -50 dBm		0.2		dB
	PHASE OUTPUT				
Dynamic Range	Less than ± 1 Degree Deviation from Best Fit Line		128		Degree
	Less than 10% Deviation in Instantaneous Slope		120		Degree
Slope (Absolute Value)	From Linear Regression about -90° or $+90^{\circ}$		10.2		mV/Degree
Deviation	Linear Deviation from Best Fit Curve at 25°C				
	$-40^{\circ}\text{C} \leq T_A \leq +85^{\circ}\text{C}$, Delta Phase = 90 Degrees		0.8		Degree
	$-40^{\circ}\text{C} \leq T_A \leq +85^{\circ}\text{C}$, Delta Phase = ± 30 Degrees		0.8		Degree
Phase Measurement Balance	Phase @ INPA = Phase @ INPB, $P_{IN} = -5$ dBm to -50 dBm		1		Degree
2200 MHz	MAGNITUDE OUTPUT				
Dynamic Range	± 1 dB Linearity $P_{REF} = -30$ dBm ($V_{REF} = -43$ dBV)		53		dB
	± 0.5 dB Linearity $P_{REF} = -30$ dBm ($V_{REF} = -43$ dBV)		51		dB
	± 0.2 dB Linearity $P_{REF} = -30$ dBm ($V_{REF} = -43$ dBV)		38		dB
Slope	From Linear Regression		27.5		mV/dB
Deviation vs. Temperature	Deviation from Output at 25°C				
	$-40^{\circ}\text{C} \leq T_A \leq +85^{\circ}\text{C}$, $P_{INPA} = P_{INPB} = -30$ dBm		0.28		dB
	Deviation from Best Fit Curve at 25°C				
	$-40^{\circ}\text{C} \leq T_A \leq +85^{\circ}\text{C}$, $P_{INPA} = \pm 25$ dB, $P_{INPB} = -30$ dBm		0.4		dB
Gain Measurement Balance	$P_{INPA} = P_{INPB} = -5$ dBm to -50 dBm		0.2		dB
	PHASE OUTPUT				
Dynamic Range	Less than ± 1 Degree Deviation from Best Fit Line		115		Degree
	Less than 10% Deviation in Instantaneous Slope		110		Degree
Slope (Absolute Value)	From Linear Regression about -90° or $+90^{\circ}$		10		mV/Degree
Deviation	Linear Deviation from Best Fit Curve at 25°C				
	$-40^{\circ}\text{C} \leq T_A \leq +85^{\circ}\text{C}$, Delta Phase = 90 Degrees		0.85		Degree
	$-40^{\circ}\text{C} \leq T_A \leq +85^{\circ}\text{C}$, Delta Phase = ± 30 Degrees		0.9		Degree
REFERENCE VOLTAGE	Pin VREF				
Output Voltage	Load = 2 k Ω	1.7	1.8	1.9	V
PSRR	$V_S = 2.7$ V to 5.5 V		0.25		mV/V
Output Current	Source/Sink (Less than 1% Change)		5		mA
POWER SUPPLY	Pin VPOS				
Supply	$V_S = 5$ V	2.7	5.0	5.5	V
Operating Current (Quiescent)	$-40^{\circ}\text{C} \leq T_A \leq +85^{\circ}\text{C}$		19	25	mA

AD8302

ABSOLUTE MAXIMUM RATINGS¹

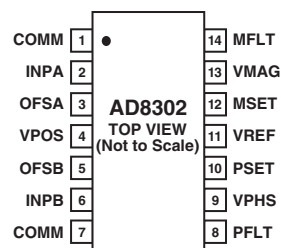
Supply Voltage V_S	5.5 V
PSET, MSET Voltage	$V_S + 0.3$ V
INPA, INPB Maximum Input	−3 dBV
Equivalent Power Re. 50 Ω	10 dBm
θ_{JA}^2	150°C/W
Maximum Junction Temperature	125°C
Operating Temperature Range	−40°C to +85°C
Storage Temperature Range	−65°C to +150°C
Lead Temperature Range (Soldering 60 sec)	300°C

NOTES

¹Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

²JEDEC 1S Standard (2-layer) board data.

PIN CONFIGURATION



PIN FUNCTION DESCRIPTIONS

Pin No.	Mnemonic	Function	Equivalent Circuit
1, 7	COMM	Device Common. Connect to low impedance ground.	
2	INPA	High Input Impedance to Channel A. Must be ac-coupled.	Circuit A
3	OFSA	A capacitor to ground at this pin sets the offset compensation filter corner and provides input decoupling.	Circuit A
4	VPOS	Voltage Supply (V_S), 2.7 V to 5.5 V	
5	OFSB	A capacitor to ground at this pin sets the offset compensation filter corner and provides input decoupling.	Circuit A
6	INPB	Input to Channel B. Same structure as INPA.	Circuit A
8	PFLT	Low Pass Filter Terminal for the Phase Output	Circuit E
9	VPHS	Single-Ended Output Proportional to the Phase Difference between INPA and INPB.	Circuit B
10	PSET	Feedback Pin for Scaling of VPHS Output Voltage in Measurement Mode. Apply a setpoint voltage for controller mode.	Circuit D
11	VREF	Internally Generated Reference Voltage (1.8 V Nominal)	Circuit C
12	MSET	Feedback Pin for Scaling of VMAG Output Voltage Measurement Mode. Accepts a set point voltage in controller mode.	Circuit D
13	VMAG	Single-Ended Output. Output voltage proportional to the decibel ratio of signals applied to INPA and INPB.	Circuit B
14	MFLT	Low Pass Filter Terminal for the Magnitude Output	Circuit E

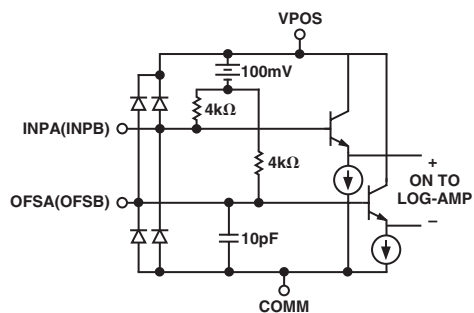
ORDERING GUIDE

Model	Temperature Range	Package Description	Package Option
AD8302ARU AD8302ARU-REEL AD8302ARU-REEL7 AD8302-EVAL	−40°C to +85°C	Tube, 14-Lead TSSOP 13" Tape and Reel 7" Tape and Reel Evaluation Board	RU-14

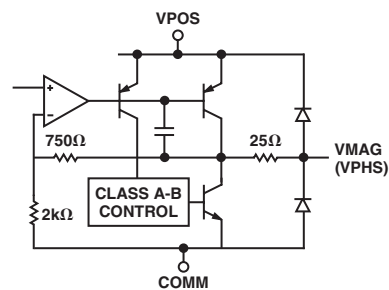
CAUTION

ESD (electrostatic discharge) sensitive device. Electrostatic charges as high as 4000 V readily accumulate on the human body and test equipment and can discharge without detection. Although the AD8302 features proprietary ESD protection circuitry, permanent damage may occur on devices subjected to high energy electrostatic discharges. Therefore, proper ESD precautions are recommended to avoid performance degradation or loss of functionality.

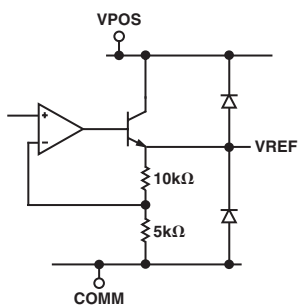




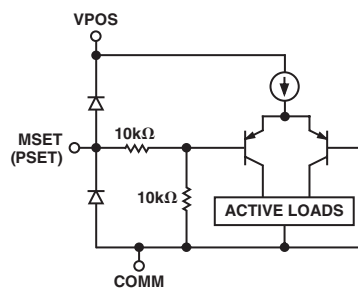
Circuit A



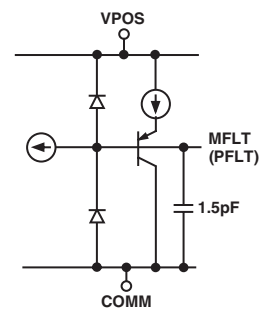
Circuit B



Circuit C



Circuit D



Circuit E

Figure 1. Equivalent Circuits



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