

## LF-2.7 GHz RF/IF Gain and Phase Detector

AD8302

#### **FEATURES**

Measures Gain/Loss and Phase up to 2.7 GHz Dual Demodulating Log Amps and Phase Detector Input Range –60 dBm to 0 dBm in a 50  $\Omega$  System Accurate Gain Measurement Scaling (30 mV/dB) Typical Nonlinearity < 0.5 dB Accurate Phase Measurement Scaling (10 mV/Degree) Typical Nonlinearity < 1 Degree

Measurement/Controller/Level Comparator Modes
Operates from Supply Voltages of 2.7 V–5.5 V
Stable 1.8 V Reference Voltage Output
Small Signal Envelope Bandwidth from DC to 30 MHz

#### **APPLICATIONS**

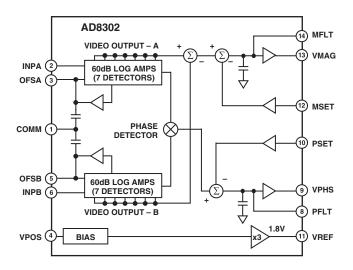
RF/IF PA Linearization
Precise RF Power Control
Remote System Monitoring and Diagnostics
Return Loss/VSWR Measurements
Log Ratio Function for AC Signals

#### PRODUCT DESCRIPTION

The AD8302 is a fully integrated system for measuring gain/loss and phase in numerous receive, transmit, and instrumentation applications. It requires few external components and a single supply of 2.7 V–5.5 V. The ac-coupled input signals can range from –60 dBm to 0 dBm in a 50  $\Omega$  system, from low frequencies up to 2.7 GHz. The outputs provide an accurate measurement of either gain or loss over a  $\pm 30$  dB range scaled to 30 mV/dB, and of phase over a 0°–180° range scaled to 10 mV/degree. Both subsystems have an output bandwidth of 30 MHz, which may optionally be reduced by the addition of external filter capacitors. The AD8302 can be used in controller mode to force the gain and phase of a signal chain toward predetermined setpoints.

The AD8302 comprises a closely matched pair of demodulating logarithmic amplifiers, each having a 60 dB measurement range. By taking the difference of their outputs, a measurement of the magnitude ratio or gain between the two input signals is available. These signals may even be at different frequencies, allowing the measurement of conversion gain or loss. The AD8302 may be used to determine absolute signal level by applying the unknown signal to one input and a calibrated ac reference signal to the other. With the output stage feedback connection disabled, a comparator may be realized, using the setpoint pins MSET and PSET to program the thresholds.

#### FUNCTIONAL BLOCK DIAGRAM



The signal inputs are single-ended, allowing them to be matched and connected directly to a directional coupler. Their input impedance is nominally  $3 \text{ k}\Omega$  at low frequencies.

The AD8302 includes a phase detector of the multiplier type, but with precise phase balance driven by the fully limited signals appearing at the outputs of the two logarithmic amplifiers. Thus, the phase accuracy measurement is independent of signal level over a wide range.

The phase and gain output voltages are simultaneously available at loadable ground referenced outputs over the standard output range of 0 V to 1.8 V. The output drivers can source or sink up to 8 mA. A loadable, stable reference voltage of 1.8 V is available for precise repositioning of the output range by the user.

In controller applications, the connection between the gain output pin VMAG and the setpoint control pin MSET is broken. The desired setpoint is presented to MSET and the VMAG control signal drives an appropriate external variable gain device. Likewise, the feedback path between the phase output pin VPHS and its setpoint control pin PSET may be broken to allow operation as a phase controller.

The AD8302 is fabricated on Analog Devices' proprietary, high performance 25 GHz SOI complementary bipolar IC process. It is available in a 14-lead TSSOP package and operates over a -40°C to +85°C temperature range. An evaluation board is available.

REV. A

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Parameter	Conditions	Min	Typ	Max	Unit
OVERALL FUNCTION Input Frequency Range Gain Measurement Range Phase Measurement Range Reference Voltage Output	$P_{IN}$ at INPA, $P_{IN}$ at INPB = -30 dBm $\phi_{IN}$ at INPA > $\phi_{IN}$ at INPB Pin VREF, -40°C $\leq T_A \leq +85$ °C	>0	±30 ±90 1.8	2700 1.88	MHz dB Degree V
INPUT INTERFACE Input Simplified Equivalent Circuit Input Voltage Range Center of Input Dynamic Range	Pins INPA and INPB To AC Ground, $f \le 500$ MHz AC-Coupled (0 dBV = 1 V rms) re: $50 \Omega$	-73 -60	3  2 -43 -30	-13 0	kΩ  pF dBV dBm dBV dBm
MAGNITUDE OUTPUT Output Voltage Minimum Output Voltage Maximum Center Point of Output (MCP) Output Current Small Signal Envelope Bandwidth Slew Rate Response Time Rise Time Fall Time Settling Time	Pin VMAG $20 \times \text{Log } (V_{\text{INPA}}/V_{\text{INPB}}) = -30 \text{ dB}$ $20 \times \text{Log } (V_{\text{INPA}}/V_{\text{INPB}}) = +30 \text{ dB}$ $V_{\text{INPA}} = V_{\text{INPB}}$ Source/Sink Pin MFLT Open $40 \text{ dB Change, Load } 20 \text{ pF} \  10 \text{ k}\Omega$ Any 20 dB Change, $10\% - 90\%$ Any 20 dB Change, $90\% - 10\%$ Full-Scale 60 dB Change, to 1% Settling		30 1.8 900 8 30 25 50 60 300		mV V mV mA MHz V/µs
PHASE OUTPUT Output Voltage Minimum Output Voltage Maximum Phase Center Point Output Current Drive Slew Rate Small Signal Envelope Bandwidth Response Time	Pin VPHS Phase Difference 180 Degrees Phase Difference 0 Degrees When $\phi_{\text{INPA}} = \phi_{\text{INPB}} \pm 90^{\circ}$ Source/Sink  Any 15 Degree Change, 10%–90% 120 Degree Change $C_{\text{FILT}} = 1 \text{ pF}$ , to 1% Settling		30 1.8 900 8 25 30 40 500		$\begin{array}{c} mV \\ V \\ mV \\ mA \\ V/\mu s \\ MHz \\ ns \\ ns \\ \end{array}$
100 MHz Dynamic Range  Slope Deviation vs. Temperature  Gain Measurement Balance	MAGNITUDE OUTPUT $\pm 1$ dB Linearity $P_{REF} = -30$ dBm ( $V_{REF} = -43$ dBV) $\pm 0.5$ dB Linearity $P_{REF} = -30$ dBm ( $V_{REF} = -43$ dBV) $\pm 0.2$ dB Linearity $P_{REF} = -30$ dBm ( $V_{REF} = -43$ dBV) From Linear Regression Deviation from Output at 25°C $-40$ °C $\leq T_A \leq +85$ °C, $P_{INPA} = P_{INPB} = -30$ dBm Deviation from Best Fit Curve at 25°C $-40$ °C $\leq T_A \leq +85$ °C, $P_{INPA} = \pm 25$ dB, $P_{INPB} = -30$ dBm $P_{INPA} = P_{INPB} = -5$ dBm to $-50$ dBm		58 55 42 29 0.25 0.25		dB dB dB mV/dB dB
Dynamic Range Slope (Absolute Value)	PHASE OUTPUT Less than ±1 Degree Deviation from Best Fit Line Less than 10% Deviation in Instantaneous Slope From Linear Regression about -90° or +90°		145 143 10		Degree Degree mV/Degree
Deviation vs. Temperature	Deviation from Output at $25^{\circ}$ C $-40^{\circ}$ C $\leq T_{A} \leq +85^{\circ}$ C, Delta Phase = 90 Degrees Deviation from Best Fit Curve at $25^{\circ}$ C $-40^{\circ}$ C $\leq T_{A} \leq +85^{\circ}$ C, Delta Phase = $\pm 30$ Degrees		0.7		Degree Degree

## AD8302

Parameter	Conditions	Min	Typ	Max	Unit
900 MHz	MAGNITUDE OUTPUT				
Dynamic Range	$\pm 1$ dB Linearity $P_{REF} = -30$ dBm ( $V_{REF} = -43$ dBV)		58		dB
	$\pm 0.5$ dB Linearity $P_{REF} = -30$ dBm ( $V_{REF} = -43$ dBV)		54		dB
	$\pm 0.2 \text{ dB Linearity } P_{REF} = -30 \text{ dBm } (V_{REF} = -43 \text{ dBV})$		42		dB
Slope	From Linear Regression		28.7		mV/dB
Deviation vs. Temperature	Deviation from Output at 25°C				
	$-40^{\circ}\text{C} \le \text{T}_{\text{A}} \le +85^{\circ}\text{C}, \text{P}_{\text{INPA}} = \text{P}_{\text{INPB}} = -30 \text{ dBm}$		0.25		dB
	Deviation from Best Fit Curve at 25°C				
	$-40^{\circ}\text{C} \le \text{T}_{\text{A}} \le +85^{\circ}\text{C}, \text{P}_{\text{INPA}} = \pm25 \text{ dB}, \text{P}_{\text{INPB}} = -30 \text{ dBm}$		0.25		dB
Gain Measurement Balance	$P_{INPA} = P_{INPB} = -5 \text{ dBm to } -50 \text{ dBm}$		0.2		dB
	PHASE OUTPUT				
Dynamia Panga	Less than ± 1 Degree Deviation from Best Fit Line		1/12		Doggeog
Dynamic Range			143 143		Degree
Clama (Alacalesta Walson)	Less than 10% Deviation in Instantaneous Slope				Degree
Slope (Absolute Value)	From Linear Regression about -90° or +90° Linear Deviation from Best Fit Curve at 25°C		10.1		mV/Degree
Deviation			0.75		Doggeog
	$-40^{\circ}\text{C} \le \text{T}_{\text{A}} \le +85^{\circ}\text{C}$ , Delta Phase = 90 Degrees		0.75		Degree
Diagram Management Dalama	$-40^{\circ}\text{C} \le \text{T}_{\text{A}} \le +85^{\circ}\text{C}$ , Delta Phase = $\pm 30$ Degrees		0.75		Degree
Phase Measurement Balance	Phase @ INPA = Phase @ INPB, $P_{IN} = -5 \text{ dBm to } -50 \text{ dBm}$		0.8		Degree
1900 MHz	MAGNITUDE OUTPUT				
Dynamic Range	$\pm 1$ dB Linearity $P_{REF} = -30$ dBm ( $V_{REF} = -43$ dBV)		57		dB
, c	$\pm 0.5$ dB Linearity $P_{REF} = -30$ dBm ( $V_{REF} = -43$ dBV)		54		dB
	$\pm 0.2 \text{ dB Linearity } P_{REF} = -30 \text{ dBm } (V_{REF} = -43 \text{ dBV})$		42		dB
Slope	From Linear Regression		27.5		mV/dB
Deviation vs. Temperature	Deviation from Output at 25°C				
20 miles (or 10 mp or utal)	$-40^{\circ}\text{C} \le \text{T}_{\text{A}} \le +85^{\circ}\text{C}, \text{ P}_{\text{INPA}} = \text{P}_{\text{INPB}} = -30 \text{ dBm}$		0.27		dB
	Deviation from Best Fit Curve at 25°C		0.2.		42
	$-40^{\circ}\text{C} \le \text{T}_{\text{A}} \le +85^{\circ}\text{C}, \text{ P}_{\text{INPA}} = \pm25 \text{ dB}, \text{ P}_{\text{INPB}} = -30 \text{ dBm}$		0.33		dB
Gain Measurement Balance	$P_{INPA} = P_{INPB} = -5 \text{ dBm to } -50 \text{ dBm}$		0.2		dB
Gain Weasarement Balance			0.2		ub
	PHASE OUTPUT				
Dynamic Range	Less than ±1 Degree Deviation from Best Fit Line		128		Degree
	Less than 10% Deviation in Instantaneous Slope		120		Degree
Slope (Absolute Value)	From Linear Regression about –90° or +90°		10.2		mV/Degree
Deviation	Linear Deviation from Best Fit Curve at 25°C				
	$-40^{\circ}\text{C} \le \text{T}_{\text{A}} \le +85^{\circ}\text{C}$ , Delta Phase = 90 Degrees		0.8		Degree
	$-40^{\circ}\text{C} \le \text{T}_{\text{A}} \le +85^{\circ}\text{C}$ , Delta Phase = $\pm 30$ Degrees		0.8		Degree
Phase Measurement Balance	Phase @ INPA = Phase @ INPB, $P_{IN} = -5 \text{ dBm to } -50 \text{ dBm}$		1		Degree
2200 MHz	MAGNITUDE OUTPUT				
Dynamic Range	$\pm 1$ dB Linearity $P_{REF} = -30$ dBm ( $V_{REF} = -43$ dBV)		53		dB
Dynamic Range	$\pm 0.5$ dB Linearity $P_{REF} = -30$ dBm ( $V_{REF} = -43$ dBV)		51		dB
	$\pm 0.2 \text{ dB Linearity } P_{REF} = -30 \text{ dBm } (V_{REF} = -43 \text{ dBV})$ $\pm 0.2 \text{ dB Linearity } P_{REF} = -30 \text{ dBm } (V_{REF} = -43 \text{ dBV})$		38		dB
Slope	From Linear Regression		27.5		mV/dB
Deviation vs. Temperature	Deviation from Output at 25°C		21.5		III V/GD
Deviation vs. Temperature	$-40^{\circ}\text{C} \le \text{T}_{\text{A}} \le +85^{\circ}\text{C}$ , $P_{\text{INPA}} = P_{\text{INPB}} = -30 \text{ dBm}$		0.28		dB
	Deviation from Best Fit Curve at 25°C		0.20		ub
	$-40^{\circ}$ C $\leq$ T <sub>A</sub> $\leq$ +85°C, P <sub>INPA</sub> = $\pm$ 25 dB, P <sub>INPB</sub> = $-30$ dBm		0.4		dB
Gain Measurement Balance	$P_{\text{INPA}} = P_{\text{INPB}} = -5 \text{ dBm to } -50 \text{ dBm}$		0.4		dB dB
Gain Measurement Balance	r <sub>INPA</sub> - r <sub>INPB</sub> 3 dbiii to -30 dbiii		0.2		ив
	PHASE OUTPUT				
Dynamic Range	Less than ±1 Degree Deviation from Best Fit Line		115		Degree
-	Less than 10% Deviation in Instantaneous Slope		110		Degree
Slope (Absolute Value)	From Linear Regression about -90° or +90°		10		mV/Degree
Deviation	Linear Deviation from Best Fit Curve at 25°C				
	$-40^{\circ}\text{C} \le \text{T}_{\text{A}} \le +85^{\circ}\text{C}$ , Delta Phase = 90 Degrees		0.85		Degree
	$-40^{\circ}$ C $\leq$ T <sub>A</sub> $\leq$ +85°C, Delta Phase = ±30 Degrees		0.9		Degree
DEFENDENCE VOLTA OF					
REFERENCE VOLTAGE	Pin VREF				**
Output Voltage	Load = $2 k\Omega$	1.7	1.8	1.9	V
PSRR	$V_S = 2.7 \text{ V to } 5.5 \text{ V}$		0.25		mV/V
Output Current	Source/Sink (Less than 1% Change)		5		mA
POWER SUPPLY	Pin VPOS				
Supply		2.7	5.0	5.5	v
Operating Current (Quiescent)	$V_S = 5 \text{ V}$		19	25	mA
	,				



### AD8302

#### ABSOLUTE MAXIMUM RATINGS1

Supply Voltage V <sub>S</sub>
PSET, MSET Voltage $V_S + 0.3 \text{ V}$
INPA, INPB Maximum Input3 dBV
Equivalent Power Re. 50 Ω
$\theta_{JA}^2$ 150°C/W
Maximum Junction Temperature
Operating Temperature Range40°C to +85°C
Storage Temperature Range65°C to +150°C
Lead Temperature Range (Soldering 60 sec) 300°C
NOTES

<sup>&</sup>lt;sup>1</sup>Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

#### PIN CONFIGURATION

СОММ 1	•	14	MFLT
INPA 2		13	VMAG
OFSA 3		-	MSET
VPOS 4	TOP VIEW (Not to Scale)	11	VREF
OFSB 5	,	10	PSET
INPB 6		9	VPHS
СОММ 7		8	PFLT

#### PIN FUNCTION DESCRIPTIONS

Pin No.	Mnemonic	Function	Equivalent Circuit
1, 7	COMM	Device Common. Connect to low impedance ground.	
2	INPA	High Input Impedance to Channel A. Must be ac-coupled.	Circuit A
3	OFSA	A capacitor to ground at this pin sets the offset compensation filter corner and provides input decoupling.	Circuit A
4	VPOS	Voltage Supply (V <sub>S</sub> ), 2.7 V to 5.5 V	
5	OFSB	A capacitor to ground at this pin sets the offset compensation filter corner and provides input decoupling.	Circuit A
6	INPB	Input to Channel B. Same structure as INPA.	Circuit A
8	PFLT	Low Pass Filter Terminal for the Phase Output	Circuit E
9	VPHS	Single-Ended Output Proportional to the Phase Difference between INPA and INPB.	Circuit B
10	PSET	Feedback Pin for Scaling of VPHS Output Voltage in Measurement Mode. Apply a setpoint voltage for controller mode.	Circuit D
11	VREF	Internally Generated Reference Voltage (1.8 V Nominal)	Circuit C
12	MSET	Feedback Pin for Scaling of VMAG Output Voltage Measurement Mode. Accepts a set point voltage in controller mode.	Circuit D
13	VMAG	Single-Ended Output. Output voltage proportional to the decibel ratio of signals applied to INPA and INPB.	Circuit B
14	MFLT	Low Pass Filter Terminal for the Magnitude Output	Circuit E

#### **ORDERING GUIDE**

Model	Temperature Range	Package Description	Package Option
AD8302ARU AD8302ARU-REEL AD8302ARU-REEL7 AD8302-EVAL	-40°C to +85°C	Tube, 14-Lead TSSOP 13" Tape and Reel 7" Tape and Reel Evaluation Board	RU-14

#### CAUTION

ESD (electrostatic discharge) sensitive device. Electrostatic charges as high as 4000 V readily accumulate on the human body and test equipment and can discharge without detection. Although the AD8302 features proprietary ESD protection circuitry, permanent damage may occur on devices subjected to high energy electrostatic discharges. Therefore, proper ESD precautions are recommended to avoid performance degradation or loss of functionality.





<sup>&</sup>lt;sup>2</sup>JEDEC 1S Standard (2-layer) board data.

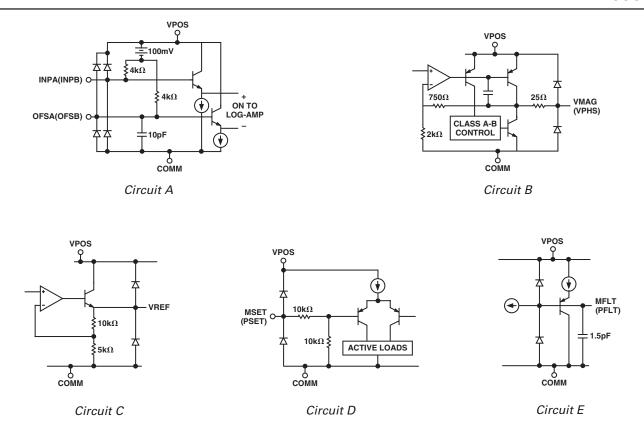


Figure 1. Equivalent Circuits

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