



US005541607A

United States Patent [19]

[11] Patent Number: **5,541,607**

Reinhardt

[45] Date of Patent: **Jul. 30, 1996**

[54] POLAR DIGITAL BEAMFORMING METHOD AND SYSTEM

[75] Inventor: **Victor S. Reinhardt**, Rancho Palos Verdes, Calif.

[73] Assignee: **Hughes Electronics**, Los Angeles, Calif.

[21] Appl. No.: **349,642**

[22] Filed: **Dec. 5, 1994**

[51] Int. Cl.⁶ **H01Q 3/22**; H01Q 3/24; H01Q 3/26

[52] U.S. Cl. **342/372**; 342/157; 342/81

[58] Field of Search 342/372, 157, 342/81

Primary Examiner—Theodore M. Blum
Attorney, Agent, or Firm—Elizabeth E. Leitereg; Terje Gudmestad; W. K. Denson-Low

[57] ABSTRACT

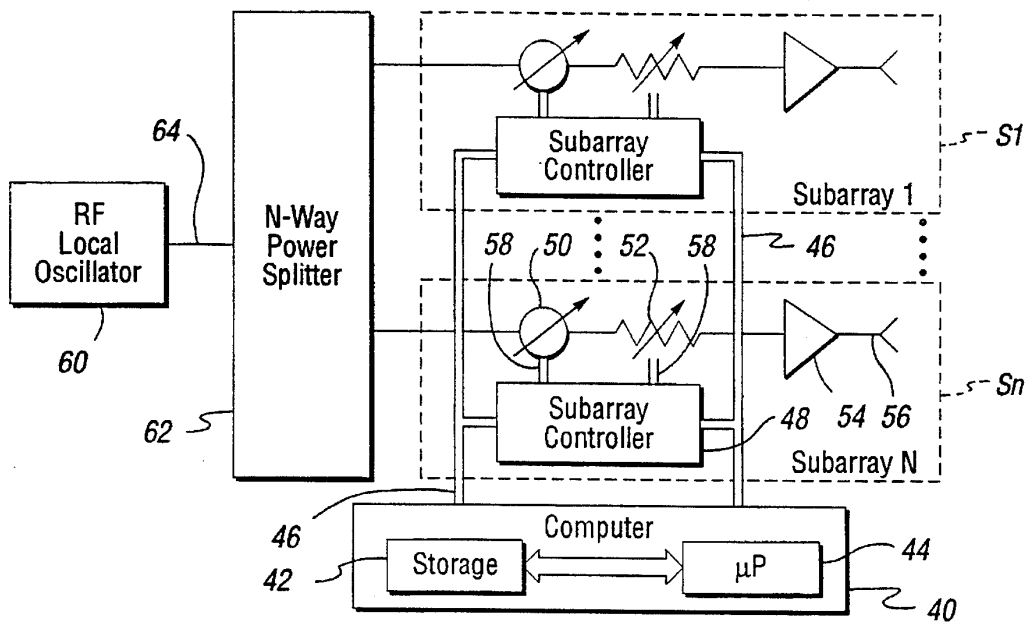
A system and method for polar digital beamforming of at least one independent transmit beam is disclosed. A computer generates a digital signal representing both pointing and modulation information which is communicated to a plurality of subarray controllers which generate the polar weighting signals corresponding to the appropriate antenna element for transmitting. The complex weighting signals may be generated by summing a sequence of complex multiplications or by simply inverting the real and imaginary components of the weighting signal for particular modulation schemes. A phasor may be used in conjunction with an attenuator to modulate a local carrier signal. Alternatively, phasors are utilized without attenuators to increase the efficiency of the power amplifiers. The antenna architecture disclosed permits a single set of phasors and attenuators to be utilized per antenna element regardless of the number of beams to be generated.

[56] References Cited

U.S. PATENT DOCUMENTS

4,965,588	10/1990	Lenormand et al.	342/372
4,983,981	1/1991	Feldman	342/372
5,093,667	3/1992	Andricos	342/372

18 Claims, 3 Drawing Sheets



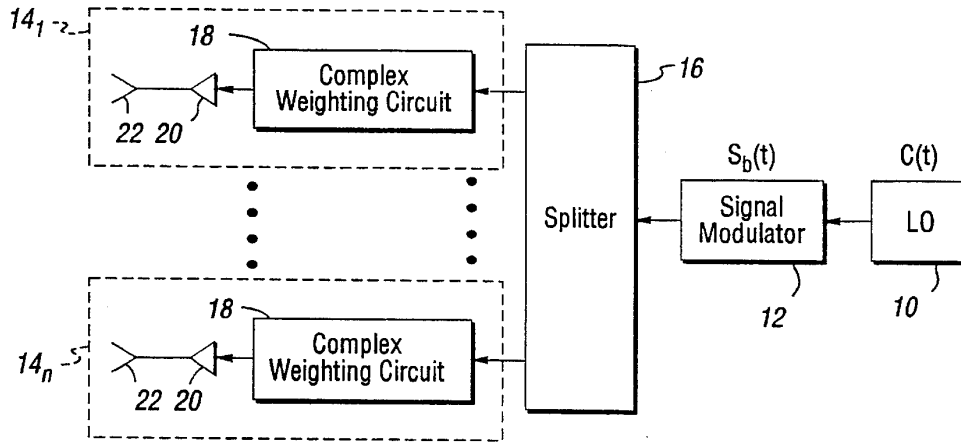


Fig. 1 (PRIOR ART)

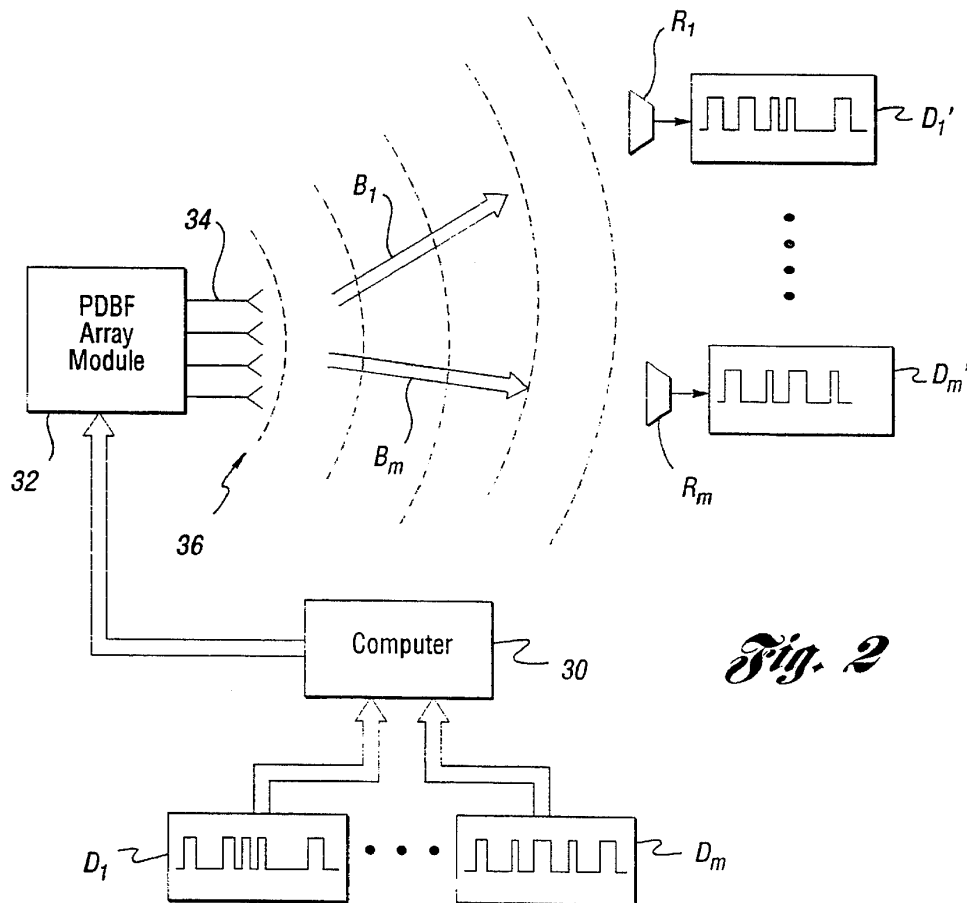


Fig. 2

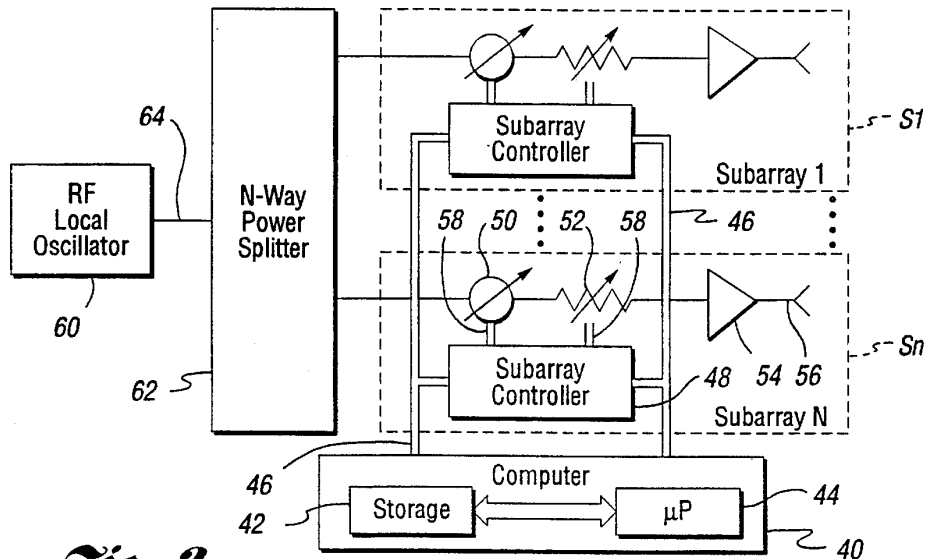


Fig. 3

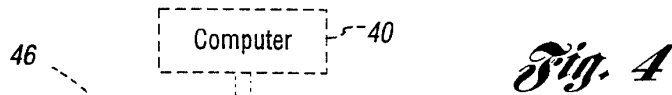
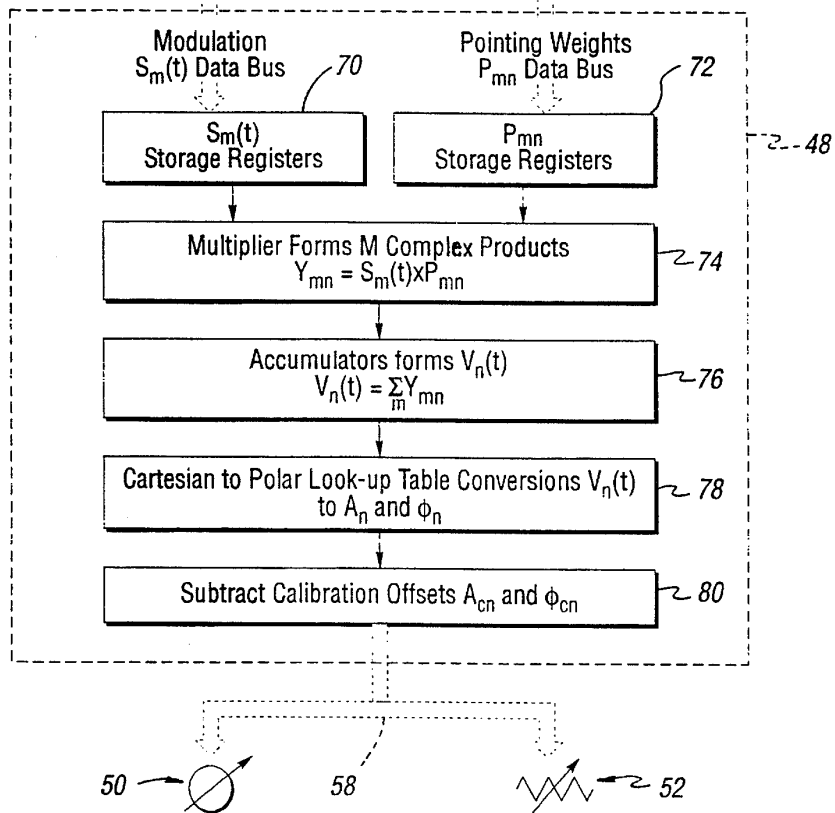


Fig. 4



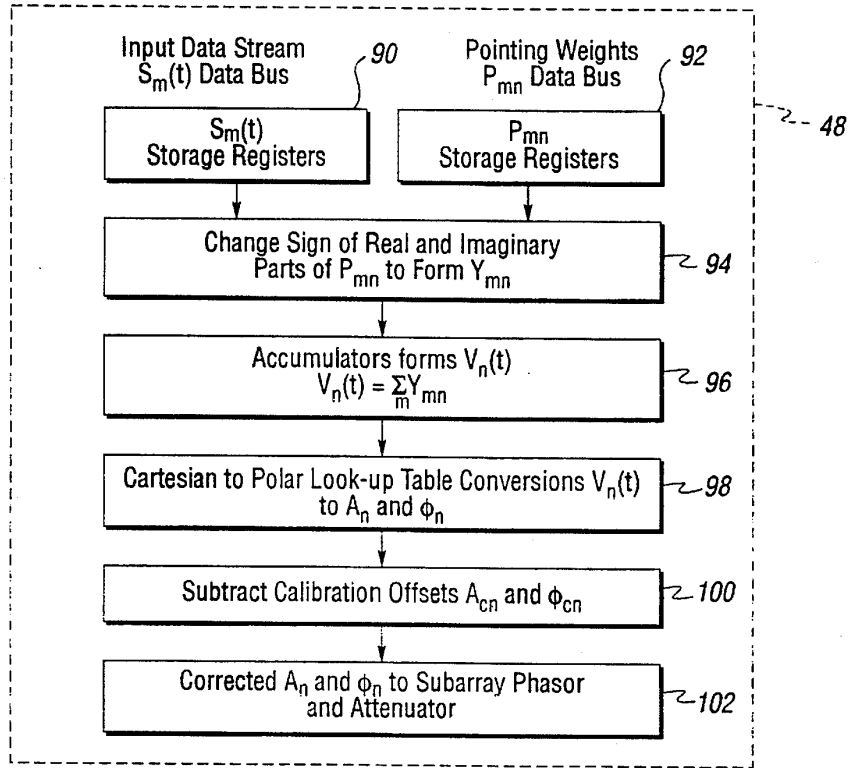


Fig. 5

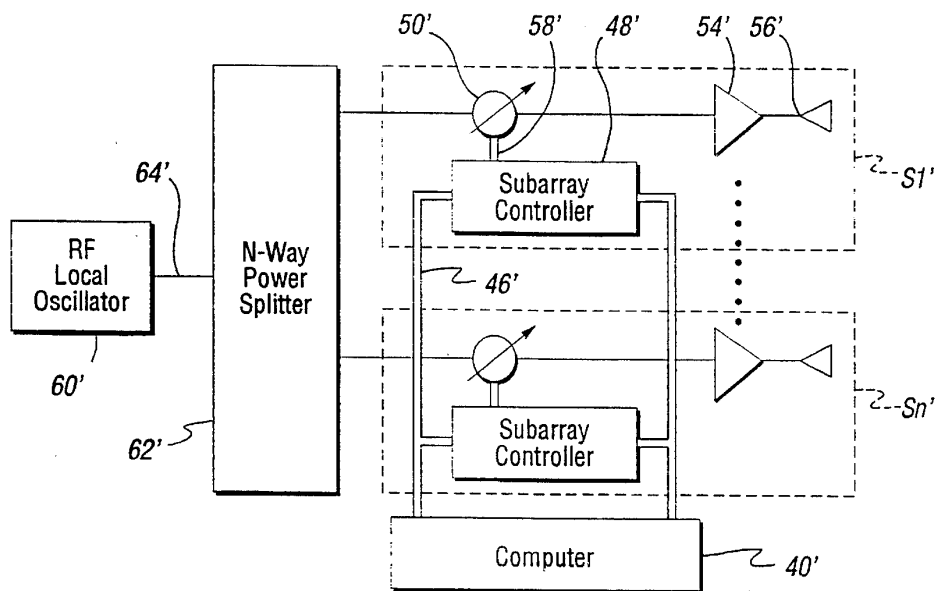


Fig. 6

POLAR DIGITAL BEAMFORMING METHOD AND SYSTEM

TECHNICAL FIELD

This invention relates to transmit phased array antennas and more particularly to a method and system of digital beamforming using a polar element weighting configuration.

BACKGROUND ART

A beamsteered transmit phased array antenna allows electronic steering of the antenna beam direction. This type of antenna system includes a number of individual antenna elements spaced in a regular array. The beam direction of the antenna (i.e., pointing direction) is controlled by the relative phases of the signals radiated by the individual antenna elements. As is known, phased arrays may be used to produce highly directional radiation patterns. Furthermore, performance characteristics normally associated with antennas having large areas can be achieved with a phased array antenna having a comparatively smaller area. Conventional transmit phased array antennas utilize two basic architectures: analog beamforming (ABF) and digital beamforming (DBF).

The basic analog beamforming approach found in the prior art is illustrated in FIG. 1. This system comprises a local radio-frequency (RF) oscillator 10 and an associated signal modulator 12 to produce an RF signal expressed in complex form as:

$$S(t) = S_b(t) \cdot C(t) \quad (1)$$

where $S_b(t)$ is the complex carrier provided by the RF oscillator and given by:

$$C(t) = A_c e^{j\omega_c t} \quad (2)$$

where $S_b(t)$ is the complex baseband waveform generated by the signal modulator. The signal $S(t)$ is then distributed to n subarrays 14₁ to 14_n by a splitter 16. Each subarray consists of a digitally controlled complex weighting circuit 18, a power amplifier 20, and an antenna element 22. Each complex weighting circuit produces a controlled phase and amplitude shift in its corresponding subarray RF signal. The signal is then amplified by power amplifier 18 and radiated by antenna element 22.

If each complex weight is represented by P_n , then the signals at the output of each weighting circuit may be represented by $P_n \cdot S(t)$. The far field radiation pattern will depend upon the number and type of antenna elements, the spacing of the array, and the relative phase and magnitude of the excitation currents applied to the various antenna elements. Generally, the electric field (E-field) generated by the entire phased array is of the form:

$$E(k) = F(k) S(t) \sum_n e^{-jk \cdot r_n} P_n \quad (3)$$

where k is the wave vector, r_n is the position of the n th element, and $F(k)$ is proportional to the E-field generated by a single element. The sum in (3) is maximized in the direction of k when

$$P_n \propto e^{jk \cdot r_n}$$

(assuming approximately equal magnitudes for all the P_n). Thus, the phased array can be electronically steered by manipulating the complex weights P_n .

One of the advantages of a phased array is that a number of beams m can be sent from the same aperture. However, to accomplish this, ABF requires the same number m sets of local oscillators, signal modulators, power splitters, and weighting circuits. At the input of each subarray power amplifier, the m beams are combined to produce a single radiation signal out of each antenna element. The various beam signals then combine in phase in m different directions so as to produce an m -beam output. The resultant E-field of the far field signal is given by:

$$E(k) = F(k) \sum_n e^{-jk \cdot r_n} \sum_m P_{m,n} S_m(t) \quad (5)$$

which represents m independent beams in the far field.

In digital beamforming (DBF), the beam pointing information represented by the complex weights and the modulation information are generated digitally. For one beam, the operation of the complex weighting circuit on the modulated RF signal can be represented as the multiplication of a complex modulation function by a complex weighting number. For multiple beams, these m complex products are summed to produce a single complex number for each subarray. This signal may be represented by:

$$V_n(t) = \sum_m P_{m,n} S_{r,m}(t) \quad (6)$$

where $S_{r,m}(t)$ is either $S_m(t)$ or $S_{b,m}(t)$. One or more digital to analog (D/A) converters are then utilized to produce an analog representation of $V_n(t)$ for each individual antenna element. Thus, only a single set of digitally controlled complex weighting circuits is required thereby eliminating much of the hardware required to generate a similar signal using ABF techniques. The disadvantage of DBF is that a large number of complex multiplications ($m \cdot n$) and complex additions (n) must be performed at a rate equal to the modulation rate. This requires the use of a high speed processor which typically consumes a great deal of power.

Two implementations of DBF have been utilized in the prior art: baseband Cartesian DBF and intermediate frequency (IF) DBF. Cartesian DBF uses a linear in-phase and quadrature (I-Q) modulator and two (2) D/A converters for each complex weighting circuit. The IF DBF technique utilizes D/A converters to directly produce the modulated subarray signals at the intermediate frequency. Upconverters are then required to convert these signals to RF signals. Both Cartesian DBF and IF DBF are characterized by complex implementations which require a significant amount of power. These implementations are not cost effective unless a very large number of beams are required.

SUMMARY OF THE INVENTION

It is, therefore, an object of the present invention to provide a multiple-beam phased array antenna which digitally generates pointing and modulation information and utilizes a simple polar architecture.

A further object of the present invention is to provide a multiple-beam phased array antenna which utilizes a single set of phasors and attenuators per antenna element.

Another object of the present invention is to provide a multiple-beam phased array antenna which utilizes a single set of phasors without attenuators for each antenna element.

Yet another object of the present invention is to provide a multiple-beam phased array antenna which utilizes previously developed phasors, attenuators, and digital Application Specific Integrated Circuits (ASICs) to implement polar digital beamforming.

Explore Litigation Insights

Docket Alarm provides insights to develop a more informed litigation strategy and the peace of mind of knowing you're on top of things.

Real-Time Litigation Alerts



Keep your litigation team up-to-date with **real-time alerts** and advanced team management tools built for the enterprise, all while greatly reducing PACER spend.

Our comprehensive service means we can handle Federal, State, and Administrative courts across the country.

Advanced Docket Research



With over 230 million records, Docket Alarm's cloud-native docket research platform finds what other services can't. Coverage includes Federal, State, plus PTAB, TTAB, ITC and NLRB decisions, all in one place.

Identify arguments that have been successful in the past with full text, pinpoint searching. Link to case law cited within any court document via Fastcase.

Analytics At Your Fingertips



Learn what happened the last time a particular judge, opposing counsel or company faced cases similar to yours.

Advanced out-of-the-box PTAB and TTAB analytics are always at your fingertips.

API

Docket Alarm offers a powerful API (application programming interface) to developers that want to integrate case filings into their apps.

LAW FIRMS

Build custom dashboards for your attorneys and clients with live data direct from the court.

Automate many repetitive legal tasks like conflict checks, document management, and marketing.

FINANCIAL INSTITUTIONS

Litigation and bankruptcy checks for companies and debtors.

E-DISCOVERY AND LEGAL VENDORS

Sync your system to PACER to automate legal marketing.