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BAS on

## IP and ATM Convergence\*

Sheraton Hotel, Tuesday Afternoon, Room Gavea A

Proposer: Paulo T. de Sousa, European Commission

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Paulo T. de Sousa, European Commission	

## Symposium on Global Internet: Application and Technology (V03)

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Chair: Ramon Caceres, AT&T Research

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2 A Distributed Algorithm for Multipath Computation .....	1689
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2 Analysis of TCP with Several Bottleneck Nodes .....	1709
Chadi Barakat and Eitan Altman, INRIA, France	
3 The Window Distribution of Multiple TCPs with Random Loss Queues .....	1714
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4 Detecting and Measuring Asymmetric Links in an IP Network .....	1727
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### Buffer Management Mechanisms

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Chair: Jennifer Rexford, AT&T Research

1 Class-Based Buffer Management using Early Fair Drop .....	1732
John Bruno, Banu Ozden, Abraham Silberschatz, and Taroon Mandhana, Lucent Technologies, USA	
2 On the Provision of Integrated QoS Guarantees of Unicast and Multicast Traffic in Input-Queued Switches .....	1742
Ge Nong and Mounir Hamdi, The Hong Kong University of Science and Technology, Hong Kong	

- 3 **Random Early Marking for Internet Congestion Control** ..... 1747  
*David Lapsley, Melbourne Information Technologies; Steven Low, The University of Melbourne, Australia*
- 4 **Flow-valve: Embedding a Safety-valve in RED** ..... 1753  
*Kenjiro Cho, Sony Computer Science Laboratories, Inc., Japan*

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Chair: Dinesh Verma, IBM Watson

- 1 **Integrating WWW Caches and Search Engines** ..... 1763  
*Wagner Meira Jr., Rodrigo Fonseca, Marcio Cesario, and Nivio Ziviani, Federal University of Minas Gerais, Brazil*
- 2 **En Passant: Predicting HTTP/1.1 Traffic** ..... 1768  
*Balachander Krishnamurthy and Jennifer Rexford, AT&T Labs Research, USA*
- 3 **Comparison of Scalable Key Distribution Schemes for Secure Group Communication** ..... 1774  
*Lakshminath Dondeti, University of Nebraska-Lincoln; Sarit Mukherjee, Panasonic Technologies, Inc.; Ashok Samal, University of Nebraska-Lincoln, USA*

## SESSION TM-G106

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11:00-12:30, Sheraton, Room Ipanema

Chair: Edmundo A. de Souza e Silva, Federal University of Rio de Janeiro

- 1 **A Two-Tier Resource Management Model for the Internet** ..... 1779  
*Andreas Terzis, Lan Wang, Jun Ogawa, and Lixia Zhang, University of California at Los Angeles, USA*
- 2 **Bandwidth Assurance Issues for TCP flows in a Differentiated Services Network** ..... 1792  
*Nabil Seddigh, Biswajit Nandy, and Peter Pleda, Computing Technology Labs-Nortel Networks, Canada*
- 3 **Supporting Best-Effort Traffic With Fair Service Curve** ..... 1799  
*T. S. Eugene Ng, Carnegie Mellon University; Donpaul C. Stephens, Carnegie Mellon University/Lucent Technologies; Ion Stoica and Hui Zhang, Carnegie Mellon University, USA*
- 4 **A Quantitative Study of Differentiated Services for the Internet** ..... 1808  
*Sambit Sahu, Don Towsley, and Jim Kurose, University of Massachusetts, USA*

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Chair: Bengt Ahlgren, SICS

- 1 **A TCP-Friendly Congestion Control Scheme for Real-time Packet Video Using Prediction** ..... 1818  
*Yeali Sun and Fu-Ming Tsou, National Taiwan University; Meng Chang Chen, Academia Sinica; Zsehung Tsai, National Taiwan University, Taiwan*
- 2 **A Smoothing Proxy Service for Variable-Bit-Rate Streaming Video** ..... 1823  
*Jennifer Rexford, AT&T Labs Research; Subhabrata Sen, University of Massachusetts; Andrea Basso, AT&T Labs Research, USA*
- 3 **Efficient Mechanisms for Recovering Voice Packets in the Internet** ..... 1830  
*Daniel Ratton Figueiredo and Edmundo de Souza e Silva, Federal University of Rio de Janeiro, Brazil*
- 4 **Measuring Internet Telephony Quality: Where Are We Today?** ..... 1838  
*Olof Hagsand, SICS; Kjell Hanson, Ericsson Business Networks; Ian Marsh, SICS, Sweden*

## SESSION TA-G108

**Internet Traffic Characterization**

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Chair: Sugih Jamin, University of Michigan, USA

- 1 **Investigating the Scaling Behavior, Crossover and Anti-persistence of Internet Packet Delay Dynamics** ..... 1843  
*Qiong Li and David L. Mills, USA*
- 2 **Inference of Internal Loss Rates in the Mbone** ..... 1853  
*Ramón Cáceres and N.G. Duffield, AT&T Labs Research; Sue B. Moon and Don Towsley, University of Massachusetts, USA*
- 3 **Inter-AS Traffic Patterns and Their Implications** ..... 1859  
*Wenja Fang and Larry Peterson, Princeton University, USA*

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4	The Current State and Likely Evolution of the Internet .....	1869
	<i>Andrew Odlyzko, AT&amp;T Labs Research, USA</i>	

## Symposium on Enterprise Application and Services (V03)

### SESSION MM-EN01

#### WEB Based Management

10:45-12:30, Sheraton, Room Vidigal B + C

Organiser: John Buford

Chair: Pradeep Ray

#### 1 Next Generation Computing [Invited Talk]

*Michael Broodie, GTE Labs, USA*

2	Agent and Web-Based Technologies in Network Management .....	1877
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*Matthew J. Wren and Jairo A. Gutierrez, The University of Auckland, New Zealand*

3	A Scalable, Web-based Architecture for Hierarchical Network Management .....	1882
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*Jiahai Yang, Peiyu Wang e Jianping Wu, Tsinghua University, China*

4	Distributed Network Management by HTTP-Based Remote Invocation .....	1889
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*Hwa-Chun Lin and Chien-Hsing Wang, National Tsing Hua University, China*

5	Web-based Customer Service Management System for a Virtual Private Network Service Based on Semi-permanent Connections .....	1894
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*S. De Smit, K. Daenen, and W. Van Leekwijck, Alcatel Corporate Research Center, Belgium*

### SESSION MA-EN02

#### Enterprise Network Design, Implementation and Operation

14:45-18:15, Sheraton, Room Vidigal B + C

Chair: Masayoshi Ejiri

Organiser: Giovanni Pau

1	Topological Design Comparison for Multicast Network .....	1899
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*Takumi Miyoshi and Yoshiaki Tanaka, Waseda University/Telecommunications Advancement Organization of Japan; Kaoru Sezaki, University of Tokio, Japan*

2	Modeling, Simulation, and Verification of an Enterprise Network .....	1905
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*Jonathan Felten, Özgür Gürbüz, and Henry Owen, Georgia Institute of Technology, USA;  
Thomas Grossmann, Gerd Kussman, and Werner Schröck, DeTeSystem Deutsche Telekom Systemlösungen, Germany*

3	Analysis and Implementation of a Transparent Priority Mechanism for LAN Internet Access .....	1910
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*A. Giovanardi and G. Mazzini, University of Ferrara, Italy*

4	Interfaces for Interworking among Intelligent Networks, Computer Telephony, and Voice over IP Systems .....	1916
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*Ryo Takeuchi, Kazuhiro Nagayama, Akira Miura, and Hiroki Tanaka, NTT Information Sharing Platform Laboratories, Japan*

5	Active TCP Control by a Network .....	1921
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*Arata Koike, NTT Information Sharing Platform Laboratories, Japan*

6	A Transactional Approach for Cross-Organizational Cooperation .....	1926
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*M. Munier, K. Benali, and C. Godart, LORIA, France*

7	The Evolution Towards the Information Society, A Market Driven Perspective from the Convair Project in ACTS .....	1932
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*P. Briggs and B. Hopwood, Marconi Communications; E. Garetti, CSELT; P. Polese, ETIC A.I./Alcatel, Belgium*

### SESSION TM-EN03

#### Enterprise Applications and Network Management

9:30-12:30, Sheraton, Room Vidigal B + C

Organiser: Simon Kerridge

Chair: Lundy Lewis

1	On Workflow and Agent Technologies on the Internet/Intranet .....	1937
---	---	------

*Geng-Sheng (G. S.) Kuo and Howard Y. Cheng, National Central University, Taiwan*

2	The COO Operator to Support and Improve the Flexibility of Adaptive Workflows .....	1942
---	---	------

*O. Perrin and C. Godart, LORIA-INRIA, France*

3	Using the Internet to Communicate Software Metrics in a Large Organization .....	1947
	<i>Tony Jokikyyny, Ericsson Research; Casper Lassenius, Helsinki University of Technology, Finland</i>	
4	A Study on Efficient Information Searchers with Agents for Large-Scale Networks .....	1954
	<i>Shinji Sugawara, Katsunori Yamaoka, and Yoshinori Sakai, Tokyo Institute of Technology, Japan</i>	
5	A Web Ontology Description Language .....	1959
	<i>Hicham Ouahid and Ahmed Karmouch, University of Ottawa, Canada</i>	
6	BDI-Oriented Agents for Network Mangement .....	1964
	<i>Morsy M. Cheikhrouhou, Institut Eurécom, France</i>	
7	Security Management Against Cloning Mobile Phones .....	1969
	<i>Mirela Sechi Moretti Annoni Notare, Federal University of Santa Catarina, Brazil; Azzedine Boukerche, University of North Texas, USA; Fernando A. da S. Cruz, Bernardo G. Riso, and Carlos B. Westphall, Federal University of Santa Catarina, Brazil</i>	

SESSION TA-ENO4

Enterprise Service and System Management

14:30-18:00, Sheraton, Room Vidigal B + C

Organiser: G. S. Kuo

Chair: Samir Chatterjee

1	Service Level Management: Definition, Architecture, and Research Challenges .....	1974
	<i>Lundy Lewis, Cabletron Systems, USA; Pradeep Ray, University of New South Wales, Australia</i>	
2	Cooperative Service Management Over Enterprise Networks: A Case Study in Healthcare Environments .....	1979
	<i>Pradeep Ray, Westmead Hospital; Gamin Weerakkody, University of New South Wales, Australia</i>	
3	A Pricing and Accounting Architecture for QoS Guaranteed Services on a Multi-domain Network .....	1984
	<i>Mitsuhiro Nakamura, Fujitsu Ltd., Japan; Masakazu Sato, TAO, Japan; Takeo Hamada, Fujitsu Laboratories of America, USA</i>	
4	Efficient Agent-Based Negotiation for Telecommunications Services .....	1989
	<i>George D. Stamoulis, Dimitrios Kalapsikakis, Anna Kirikoglou, and Costas Courcoubetis, University of Crete, Greece</i>	
5	Managing Users, Applications and Resources with RMON2 .....	1997
	<i>L. P. Gaspar, Federal University of Rio Grande do Sul/Vale do Rio dos Sinos University; L. R. Tarouco, University of Rio Grande do Sul, Brazil</i>	
6	Broadband connectivity Management Service for Multi-domain ATM and SDH Networks .....	2002
	<i>Alex Galis, University College London, United Kingdom</i>	
7	Enterprise Directory Support for Future SNMPV3 Network Management Applications .....	2010
	<i>S. Omani, University of Versailles, France; R. Boutaba, University of Toronto, Canada; O. Cherkaoui, UQAM University, Canada</i>	

Symposium on Multimedia Services and Technology Issues (V04)

SESSION WM-MM01

Multimedia: State of the Art [Invited Session]

9:30-12:30, Intercontinental, Room Agata

1	Content Adaptation Framework: Bringing the Internet to Information Appliances .....	2015
	<i>Rakesh Mahan, John R. Smith and Chung-Sheng Li, IBM T.J. Watson Research Center, USA</i>	
2	Audio and Video Compression on Programmable Chips for Mobile Communication .....	2022
	<i>Takao Nishitani and Ichiro Kuroda, NEC Corporation, Japan</i>	
3	Video Communication Advances and Applications .....	2030
	<i>Bernd Girod, University of Erlangen-Nuremberg, Germany</i>	
4	Advances in Color Image Segmentation .....	2038
	<i>Luca Lucchese and S. K. Mitra, University of California, USA</i>	
5	High Quality Audio for Multimedia: Key Technologies and MPEG Standards .....	2045
	<i>P. Noll, Technical University of Berlin, Germany</i>	
6	The Voice User Interface .....	2051
	<i>Joseph P. Olive, Lucent Technologies, USA</i>	

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## SESSION WA-MM02

### Multimedia Applications and Systems

14:30-16:05, Intercontinental, Room Agata

Organizer: Nelson Fonseca, State University of Campinas, Brazil

Chair: Jeff H. Derby, IBM Corporation

- 1 The Video Conference Service in an Intelligent Broadband Network: Design, Load Balancing and Performance Analysis in the Control Plane ..... 2056  
*F. Cuomo and M. Listanti, University of Rome "La Sapienza", Italy*
- 2 Quality of Perception to Quality of Service Mapping Using a Dynamically Reconfigurable Communication System ..... 2061  
*G. Ghinea, J. P. Thomas, and R. S. Fish, University of Reading, United Kingdom*
- 3 Overview of Risks in Multimedia Broadband Upgrades ..... 2066  
*Kjell Stordahl, Telenor; Nils Kristian Elnegaard, Tele Denmark; Leif Arthun Ims and Borge Tørre Olsen, Telenor, Norway*
- 4 A Stream Relationship Monitor for Adaptive Multimedia Document Retrieval ..... 2071  
*Eduardo Carneiro da Cunha, Luiz Fernando Rust da Costa Carmo, and Luci Pirmez, Federal University of Rio de Janeiro, Brazil*

## SESSION WA-MM03

### Multimedia Services Access and Control

16:25-18:00, Intercontinental, Room Agata

Organizer: Nelson Fonseca, State University of Campinas, Brazil

Chair: Otto Carlos M. B. Duarte, Federal University of Rio de Janeiro, Brazil

- 1 Threshold-Based Admission Policies for Video Services ..... 2076  
*S.-H. Gary Chan, Hong Kong University of Science and Technology, Hong Kong; Fouad A. Tobagi, Stanford University, USA*
- 2 Performance Evaluation of an Adaptive Access Control Scheme in an Integrated Voice/Video/Data CDMA System ..... 2081  
*Seung Sik Choi and Dong-Ho Cho, Korea Advanced Institute of Science and Technology, Korea*
- 3 Resource Reservation Admission Control Algorithm with User Interactions ..... 2086  
*F. Sallabi and A. Karmouch, University of Ottawa, Canada*
- 4 FEC-PSD: A FEC-aware Video Packet Drop Scheme ..... 2091  
*Ahmed Mehaoua, University of Versailles, France; Sijing Zhang, University of Cambridge, United Kingdom; Raouf Boutaba, University of Toronto, Canada*

## Symposium on Advanced Signal Processing for Communications (V04)

### SESSION MA-SP01

#### Antenna and Array Processing

14:45-18:15, Intercontinental, Room Quartzo B

Organisers: Tomohiko Taniguchi, Fujitsu, USA; Naohisa Ohta, Sony, Japan

Chair: Horst Bessai, University of Siegen, Germany

- 1 Downlink Beamforming for Frequency Division Duplex Systems ..... 2097  
*Klaus Hugl, Juha Laurila, and Ernst Bonek, Technische Universität Wien, Austria*
- 2 Field Measurements of High Speed QAM Wireless Transmission Using Equalization and Real-Time Beamforming ..... 2102  
*Jean-François Frigon and Babak Daneshrad, University of California at Los Angeles, USA*
- 3 Distribution Analysis of SIR in Adaptive Antenna Arrays under Rayleigh Correlated Channels ..... 2107  
*Alime (Yanartas) Özyildirim and Yalçın Tanik, Middle East Technical University, Turkey*
- 4 Reducing Capture Effects in Despread-Respread Multi-Target Adaptive Arrays in CDMA Systems ..... 2112  
*Francisco R. P. Cavalcanti, Federal University of Ceará; João M. T. Romano, State University of Campinas-UNICAMP, Brazil*
- 5 Space-Time Iterative MMSE Multiuser Detection in Frequency Selective Coded CDMA Wireless Mobile Channels ..... 2117  
*Joseph Thomas and Evaggelos Geraniotis, University of Maryland, USA*
- 6 Bandwidth-Efficient Pilot Symbol Aided Technique with Diversity Reception in Land Mobile Satellite Fading Channels ..... 2122  
*M. H. Ng and S. W. Cheung, The University of Hong Kong, Hong Kong*
- 7 Implementation of Mobile User Tracking System ..... 2127  
*Ramin Baghaie and Petri Karttunen, Helsinki University of Technology, Finland*

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Equalisation and Channel Estimation

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Organiser: Bin Qiu, Monash University, Australia

Chair: João Marcos Romano, State University of Campinas, Brazil

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*João B. Destro Filho, State University of Campinas-UNICAMP, Brazil; Gérard Favier, CNRS/UNSA, France; João M. Travassos Romano, State University of Campinas-UNICAMP, Brazil*

2 Information Rate Maximizing FIR Transceivers: Filterbank Precoders and Decision-Feedback Equalizers for Block Transmissions over Dispersive Channels ..... 2142  
*Anastasios Stamoulis, Wei Tang, and Georgios B. Giannakis, University of Minnesota, USA*

3 New Results on Second-Order Statistics-Based Blind Identification and Equalization of S.I.M.O. Channels ..... 2147  
*Jitendra K. Tugnait, Auburn University, USA*

4 Comparison of Equalization Techniques in a Wavelet Packet Based Multicarrier Modulation DS-CDMA System ..... 2152  
*Yifeng Zhang, Philips Semiconductors; Jeffrey Dill, Ohio University, USA*

5 Non Linear Channel Estimation in Fading OFDM Systems ..... 2157  
*Fernando D. Nunes and José M. N. Leitão, Instituto Superior Técnico, Portugal*

6 Channel Estimation with Superimposed Pilot Sequence ..... 2162  
*Peter Hoeher, University of Kiel, Germany; Fredrik Tufvesson, Lund University, Sweden*

7 A Reduced-State Viterbi Algorithm for Blind Sequence Estimation of DPSK Sources ..... 2167  
*Tongtong Li, Auburn University; Zhi Ding, University of Iowa, USA*

SESSION TA-SPO3

Communication Signal Processors

14:30-18:00, Intercontinental, Room Quartzo B

Organiser: Jaafar Elmighani, University of Northumbria, Newcastle, United Kingdom

Chair: Naohisa Ohta, Sony Japan

1 A High-Speed Processor for Rectangular-to-Polar Conversion with Applications in Digital Communications ..... 2172  
*Dengwei Fu and Alan N. Willson, Jr., University of California at Los Angeles, USA*

2 Data-Aided ML Parameter Estimators of PSK Burst Modems and Their Systolic VLSI Implementations ..... 2177  
*Yimin Jiang, Farhad B. Verahrami, and Wen-Chun Ting, Hughes Network Systems, Inc.; Robert L. Richmond and John S. Baras, University of Maryland, USA*

3 Radial Basis Function Decision Feedback Equaliser Assisted Burst-by-burst Adaptive Modulation ..... 2183  
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4 Multiple Access Over Digital Subscriber Loops Using Optimal FIR Transmit Filter Bank ..... 2188  
*Ahmed F. Shalash, New Jersey Design Center; Mohammed Nafie, University of Minnesota, USA*

5 Analysis of a Partial Decorrelator in a Multi-cell DS/CDMA System ..... 2193  
*Mohammad Saquib, LSU; Roy Yates, Rutgers University, USA*

6 Optimal Perceptual Binary Allocation for AAC Audio Coder ..... 2198  
*Marcos Perreau Guimaraes, Madeleine Bonnet, and Nicolas Moreau, Université René Descartes, France*

7 Non-Intrusive Whitening of Speech Using Least Mean Square and Divergence Detection Technique ..... 2203  
*Wai Pang Ng, Jaafar M.H. Elmighani, R.A. Cryan, and Simon Broom, University of Northumbria, United Kingdom*

8 Hierarchical Image Transmission System for Telemedicine Using Segmented Wavelet Transform and Golomb-Rice Codes ..... 2208  
*Masayuki Hashimoto, Atsushi Koike, and Syuichi Matsumoto, KDD R&D Laboratories, Japan*

SESSION WM-SPO4

Interference Cancellation

9:30-12:30, Intercontinental, Room Quartzo B

Organisers: Trevor Clarkson, Kings College London; Jaafar Elmighani, University of Northumbria, Newcastle, United Kingdom

Chair: Yeheskel Bar-Ness, New Jersey Institute of Technology, USA

1 Interference Cancellation in W-CDMA Cellular Structures Using Statistical Processing ..... 2213  
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2	Fast Adaptive Interference Cancellation In Chirp Spread Spectrum Systems .....	2218
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3	Multistage Symbol-by-Symbol Bayesian Interference Cancellation for UMTS-CDMA Links Affected by Severe Multipath .....	2223
	<i>R. Cusani and M. Di Felice, University of Rome "La Sapienza", Italy; J. Mattila, Helsinki University of Technology, Finland</i>	
4	Receive-Filtering for GSM Receivers in Interference Limited Environments .....	2228
	<i>Gunnar Fock, Jens Baltersee, and Heinrich Meyr, Aachen University of Technology, Germany; Lihbor Yiin, Lucent Technologies, USA</i>	
5	Interference Cancellation for Downlink TDMA Systems Using Smart Antenna .....	2233
	<i>Christian Ibars and Yeheskel Bar-Ness, New Jersey Institute of Technology, USA</i>	
6	Noise Reduction Based on Local Linear Representation Using Artificial Neural Networks .....	2238
	<i>A. Müller and J. M. H. Elmirghani, University of Northumbria, United Kingdom</i>	
7	Direct Carrier Frequency Compensation in Burst Mode Satellite Communication .....	2244
	<i>S. Thao and M. Bellanger, CNAM; A. Marguinaud and T. de Couasnon, Alcatel, France</i>	

## SESSION WA-SPO5

### Detection

14:30-18:00, Intercontinental, Room Quartzo B

Organisers: Trevor Clarkson, Kings College London, United Kingdom; Bin Qiu, Monash University, Australia

Chair: Jaafar Elmirghani, University of Northumbria, United Kingdom

1	Progressively Powerful Multiuser Detectors for Rapidly Time-varying Multipath Environments .....	2249
	<i>Tamer A. Kadous and Akbar M. Sayeed, University of Wisconsin-Madison, USA</i>	
2	Adaptive Blind MIMO Channel Estimation and Multiuser Detection in DS-CDMA Systems .....	2254
	<i>I-Tai Lu and Jaeyoung Kwak, Polytechnic University, USA</i>	
3	Prediction-Based Decision Feedback Differential Detection for MDPSK .....	2259
	<i>Robert Schober and Wolfgang H. Gerstacker, University of Erlangen-Nuremberg, Germany</i>	
4	Differential Detection of Multilevel Continuous Phase Modulation for Radio Applications .....	2264
	<i>Bernhard Spinnler and Berthold Lankl, Siemens AG; Johannes Huber, University of Erlangen, Germany</i>	
5	Polyhedral Methods for Blind Deterministic Separation of Binary Co-Channel Signals .....	2269
	<i>João Xavier and Victor Barroso, Instituto Superior Técnico, Portugal</i>	
6	Fast, Accurate and Simple Carrier Acquisition Algorithm for OFDM Systems .....	2274
	<i>Zulfiquar Sayeed, Lucent Technologies, USA</i>	
7	Analysis Of OFDM/OQAM Systems Based On The Filterbank Theory .....	2279
	<i>Pierre Siohan and Nicolas Lacaille, CNET/DMR, France</i>	
8	Asynchronous Communication and Symbol Synchronization in Multipoint-to-Point Multicarrier Systems .....	2285
	<i>Ravi Chandran, Mark J. Patton, Peter J. W. Melsa, and Daniel J. Marchok, Tellabs Research Center, USA</i>	

## Symposium on Communication Theory (V05)

### SESSION MM-CT01

#### Equalization

10:45-12:30, Intercontinental, Room Agua Marinha

Chair: Hikmet Sari, Alcatel, France

1	Fractionally-Spaced Prefiltering for Reduced State Equalization .....	2291
	<i>Michael Schmidt and Gerhard P. Fettweis, Dresden University of Technology, Germany</i>	
2	A Novel Reduced-Complexity MAP Equalizer Using Soft-Statistics For Decision-feedback ISI Cancellation .....	2296
	<i>E. Baccarelli and A. Fasano, University of Rome "La Sapienza", Italy; S. Galli, Telcordia Technologies, USA; A. Zucchi, University of Rome "La Sapienza", Italy</i>	
3	Blind Spatio-Temporal Decision Feedback Equalization: A Self-Adaptive Approach .....	2301
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4	Delay Processing vs. Per Survivor Techniques for Equalization with Fading Channels .....	2306
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5 Per Tone Equalization for DMT Receivers ..... 2311  
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SESSION MM-CT02

Multiple-Antenna Systems

10:45-12:30, Intercontinental, Room Onix

Chair: Len Cimini, AT&T, USA

1 Multiple Antennas in Cellular CDMA Systems: Transmission, Detection and Spectral Efficiency ..... 2316  
*Howard Huang, Harish Viswanathan, and G. J. Foschini, Lucent Technologies, USA*

2 New Spread-Spectrum Techniques for Multiple Antenna Transmit Diversity ..... 2321  
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3 Power Control Schemes for TDD Systems with Multiple Transmit and Receive Antennas ..... 2326  
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4 Exact Error Probability Expressions for MRC in Correlated Nakagami Channels with Unequal Fading Parameters and Branch Powers ..... 2331  
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5 Turbo Coded Modulation for Systems with Transmit and Receive Antenna Diversity ..... 2336  
*Andrej Stefanov and Tolga M. Duman, Arizona State University, USA*

SESSION MA-CT03

Turbo Codes & Iterative Decoding I

14:45-18:15, Intercontinental, Room Agua Marinha

Chair: Max Costa, State University of Campinas, Brazil

1 A Capacity-Approaching Hybrid ARQ Scheme Using Turbo Codes ..... 2341  
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2 Performance of Trellis Coded CPM with Iterative Demodulation and Decoding ..... 2346  
*Krishna r. Narayanan, Texas A&M University; Gordon L. Stüber, Georgia Institute of Technology, USA*

3 Semi-Random Interleaver Design Criteria ..... 2352  
*Christine Fragouli and Richard D. Wesel, University of California at Los Angeles, USA*

4 Adapted Iterative Decoding of Product Codes ..... 2357  
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Organizer: Chih-Lin I, AT&T

Moderator: Jesse Russell, AT&T

Jesse Russell, Chairman, Wireless Communications Division, TIA VP, Advanced Communications, AT&T

William Lee, VP & Chief Scientist, Global Technology, Vodafone AirTouch

Bosco Fernandes, Chairman, Information Communications Technologies, UMTS VP & Global Key Account, Siemens

Akio Sasaki, Director, ARIB (Association of Radio Industries and Businesses) VP, NTT-DoCoMo

Shihei Li, Chairman, China Academy of Telecomm Tech Chairman, Beijing Xinwei Telecom Tech Co.

Mark Epstein, Senior VP, Development, QUALCOMM Chairman, International Standards Coordinating Subcommittee, TIA

George Zysman, CTO and VP, Wireless Comm, Lucent

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(\* ) The BAS papers did not follow the review process

## Achieving High Data Rates in CDMA Systems Using BLAST Techniques

Howard Huang, Harish Viswanathan, and G. J. Foschini

Bell Labs, Lucent Technologies; 791 Holmdel-Keyport Rd.; Holmdel, NJ 07030

{hchuang, harishv gif} @ lucent.com

**Abstract-** We evaluate the capacity of a downlink cellular CDMA system where the transmitters use multiple antennas and the receivers use space-time multiuser detection. We discuss a family of transmission techniques which combine transmit diversity and multicode transmission for achieving high data rates. Higher system spectral efficiencies (greater than one) can potentially be achieved if the same spreading code is used on different antennas to transmit independent substreams. In this case multiple antennas at the receiver distinguish the signals based only on their spatial characteristics. In general, the receivers employ an extension of the BLAST (Bell Labs Layered Space-Time) receiver architecture for wideband CDMA signals. Accounting for shadowing and frequency-selective Rayleigh fading, we develop a novel technique for evaluating the capacity of systems with CDMA BLAST detectors and show that the resulting capacities offer significant improvement over conventional single antenna systems.

### I. INTRODUCTION

High-speed data services will be a major application of future wireless networks. Provisions such as multicode transmission and transmit antenna diversity have been recommended for third generation CDMA systems to provide such services. However, using conventional single-antenna matched filter receivers, the overall system capacities are limited.

In reference [1], high spectral efficiencies were demonstrated for a narrowband point-to-point system with multiple antennas at the transmitter and receiver. In this work, we extend these ideas to study the downlink of a cellular CDMA system. High spectral efficiencies are achieved using multiple antennas and multiuser detection at the receivers. In Section II, we investigate a family of transmission techniques using multiple antennas and orthogonal codes. The received signal at the high-speed mobiles is given in Section III. As described in Section IV, these receivers use multiple antennas and a generalization of the V-BLAST algorithm [2] to account for multiaccess interference.

We then determine the resulting spectral efficiency (the number of information bits per chip per sector) from the number of high-speed data users which can be supported in a sector at a given error rate and outage rate. In Section V, we develop a technique based on [3] for determining the system capacity for a multiple antenna system with multiuser detectors, and in Section VI we apply the technique to demonstrate the potential spectral efficiency gains achievable by these enhancements.

### II. TRANSMISSION TECHNIQUES

The base station transmitter has three resources: spreading codes, antennas, and power. Our goal is to allocate the resources efficiently among  $K$  high-speed data users in a manner that minimizes interference and maximizes the system capacity. We

assume that there are  $M$  transmit antennas at the base station indexed by  $m = 1 \dots M$ . High-speed data transmissions are achieved by demultiplexing a single data stream into  $G$  substreams, each with rate  $R$ . In general, a substream's power can be arbitrarily distributed and transmitted over space (transmit antennas) and time. However in order to be compatible with current CDMA proposals, we assume that a substream is spread by a length  $N$  code and transmitted over one or more antennas (more antennas to achieve transmit diversity). If the same substream is simultaneously transmitted from more than one antenna, different codes are used to spread the data onto each antenna since, otherwise, there would be no gain from transmit diversity. On the other hand, different substreams could be transmitted from up to  $M$  antennas using either the same code or different codes. If the same code is used, the receiver would distinguish the substreams based only on their spatial characteristics. If different codes are used, the receiver would use both spatial and code information.

To illustrate the coupling between data substreams, spreading codes and transmit antennas, Figs. 1A - 1D show four assignment options for a given data user's  $G$  substreams. Each row represents a transmit antenna ( $m = 1, \dots, M; M = 4$ ), and each column represents a unique orthogonal code.

- Fig. 1A shows the assignment for same code, no transmit diversity ( $M_t = 1$ ) transmission for  $G = 8$  substreams. Substreams 1 - 4 are transmitted, respectively, from antennas 1 - 4 using the same code. Substreams 5 - 8 are transmitted, respectively, from antennas 1 - 4 using a different code, so a total of  $G/M = 2$  codes is used.
- Fig. 1B shows the assignment for different code, no transmit diversity transmission. The antenna assignments for the substreams is the same as before, but now each substream is assigned a unique code so that a total of  $G = 8$  codes is used.
- Fig. 1C shows the assignment for same code transmission with transmit diversity order  $M_t = 2$ . Each substream is now transmitted twice using different antennas and codes. For the first transmission, the code and antenna assignments are the same as in Fig. 1A. For the second transmission, the antenna assignments are cyclically shifted so that the  $m_t$ th transmission ( $m_t = 1 \dots M_t$ ) of the  $g$ th ( $g = 1 \dots G$ ) substream is through antenna

$$m = (m_t + g - 2) \bmod M + 1 \quad (1)$$

A total of  $GM_t / M = 4$  codes is used.

- Fig. 1D shows the assignment for different code transmission with transmit diversity order  $M_t = 2$ . The antenna assignment is the same as in the corresponding same-code case (using (1)), but now each substream is assigned a unique code so a total of  $GM_t = 16$  codes is used.

$m$	1	1	5
	2	2	6
	3	3	7
	4	4	8

Code A B

Fig. 1A. Same-code transmission,  $M_t = 1$

$m$	1	1		5			
	2		2		6		
	3			3		7	
	4				4		8

Code A B C D E F G H

Fig. 1B. Different-code transmission;  $M_t = 1$

$m$	1	1	5	4	8
	2	2	6	1	5
	3	3	7	2	6
	4	4	8	3	7

Code A B C D

Fig. 1C. Same-code transmission,  $M_t = 2$

$m$	1	1		5				4			8			
	2		2		6				1			5		
	3			3		7				2			6	
	4				4		8				3			7

Code A B C D E F G H I J K L M N O P

Fig. 1D. Different-code transmission;  $M_t = 2$

### III. RECEIVED SIGNAL MODEL AND ASSUMPTIONS

We consider the complex baseband received signal at the  $k$ th high-speed user's receiver. Each high-speed data user has a  $P$  antenna receiver for demodulating its  $G$  substreams. The  $KG$  data substreams are transmitted out of  $M$  antennas, where the antenna assignment for a user's substream is given by (1). The substreams are each spread by a length  $N$  code. The multipath channel induces  $L$  resolvable multipath components, and the complex fading coefficient among all antennas and multipath components are uncorrelated. The delay spread is small compared to the symbol period so that intersymbol interference can be ignored. Following a chip matched filter, the discrete-time complex baseband received signal for a mobile user at its  $p$ th ( $p = 1 \dots P$ ) antenna during a given symbol period can be written as a complex  $N$ -vector:

$$\mathbf{r}_p = \mathbf{S}\tilde{\mathbf{C}}_p\mathbf{A}\mathbf{b} + \mathbf{n}_p \quad (2)$$

where  $\mathbf{S}$  is the real  $N$ -by- $KGM_tL$  code matrix defined by

$$\mathbf{S} \triangleq [s_{1,1,1,1} \dots s_{1,1,1,L} \dots s_{1,1,M,1} \dots s_{1,1,M,L} \dots s_{1,G,1,1} \dots s_{1,G,1,L} \dots s_{1,G,M,1} \dots s_{1,G,M,L} \dots s_{K,1,1,1} \dots s_{K,1,1,L} \dots s_{K,1,M,1} \dots s_{K,1,M,L} \dots s_{K,G,1,1} \dots s_{K,G,1,L} \dots s_{K,G,M,1} \dots s_{K,G,M,L}]$$

and where  $s_{k,g,m,l}$  corresponds to the code sequence of the  $k$ th user's  $g$ th substream's transmission over the antenna for the  $m$ th transmission and  $l$ th multipath component. For  $l = 1$ , unique codes (depending on the transmission technique) are independent and binary-valued. The chips values are  $\pm 1/\sqrt{N}$ , each with probability 1/2. For a given code, its resolvable multipath components  $l = 1 \dots L$  can likewise be modeled as random, binary-valued, and mutually independent codes. Matrix  $\tilde{\mathbf{C}}_p$  is a block diagonal  $KGM_tL$ -by- $KGM_t$  matrix  $\text{diag}(\mathbf{C}_p, \dots, \mathbf{C}_p)$  where  $\mathbf{C}_p$  is the complex  $GM_tL$ -by- $GM_t$  channel matrix defined by

$$\mathbf{C}_p \triangleq \text{diag} \left( \begin{bmatrix} \mathbf{c}_{m(1,1),p} \\ \vdots \\ \mathbf{c}_{m(1,M_t),p} \end{bmatrix}, \dots, \begin{bmatrix} \mathbf{c}_{m(G,1),p} \\ \vdots \\ \mathbf{c}_{m(G,M_t),p} \end{bmatrix} \right)$$

where in turn  $\mathbf{c}_{m(g,m),p}$  is the complex  $L$ -vector corresponding to the multipath channels between transmit antenna  $m$  and the receive antenna  $p$ :  $\mathbf{c}_{m(g,m),p} \triangleq [c_{m,p,1} \dots c_{m,p,L}]^T$ . The transmit antenna assignment  $m$  is given by (1). The channel amplitudes are independent, zero-mean proper complex Gaussian random variables with normalized variance:

$$E(\hat{c}_{m_1,l,p_1}^* \hat{c}_{m_2,l,p_2}) = \begin{cases} 0 & \text{if } m_1 \neq m_2 \text{ or } p_1 \neq p_2 \\ 1 & \text{if } m_1 = m_2 \text{ and } p_1 = p_2 \end{cases} \quad (3)$$

where  $*$  denotes the complex conjugate. Matrix  $\mathbf{A}$  is a  $KG$ -by- $KG$  diagonal matrix of amplitudes defined by  $\mathbf{A} \triangleq \text{diag}(A_{1,1} \dots A_{1,G} \dots A_{K,1} \dots A_{K,G})$ . Vector  $\mathbf{b}$  is the real  $KG$  binary data vector defined by  $\mathbf{b} \triangleq [b_{1,1} \dots b_{1,G} \dots b_{K,1} \dots b_{K,G}]^T$ . Vector  $\mathbf{n}_p$  is the zero-mean complex (circularly symmetric) Gaussian noise  $N$ -vector with i.i.d. components whose real and imaginary components each have variance  $\sigma^2$ .

Note that all the substreams corresponding to a given data user have the same transmit power. We determine the system capacities assuming that the total base station power is fixed. The channel is assumed to be fixed over the duration of the symbol and assumed to be known at the receiver. In other words, we assume that the detectors will have perfect estimates of  $\mathbf{c}_{m,l,p}$ . Channel (and timing) estimates can be obtained from auxiliary pilot channels (one for each transmit antenna), but these are not explicitly accounted for in (2). However, one could interpret  $\mathbf{r}_p$  in (2) as an enhanced received signal where the pilot signals have been removed after channel estimation.

IV. DETECTION TECHNIQUE

The jointly optimum detector is the maximum likelihood space-time multiuser detector which jointly detects all  $KG$  data substreams [4]. Because the complexity of this detector is exponential with respect to the number of substreams, we develop a suboptimal detector based on an extension of the V-BLAST receiver [2] to CDMA spread signals. Shown in Figure 2, the receiver acquires a sufficient statistic vector from the received signal (2) by performing a space-time matched filtering with respect to the codes and channel coefficients [5]. Since the data is binary valued, we then take the real component of each element to obtain the sufficient statistic vector. Letting  $\mathbf{y}$  be the  $KG$ -dimensional sufficient statistic vector, we can write

$$\mathbf{y} = \text{Re} \left[ \sum_{p=1}^P \tilde{\mathbf{C}}_p^H \mathbf{S}^T \mathbf{r}_p \right] = \mathbf{R} \mathbf{A} \mathbf{b} + \mathbf{n} \quad (4)$$

where  $\mathbf{R} \triangleq \sum_{p=1}^P \text{Re} \left[ \tilde{\mathbf{C}}_p^H \mathbf{S}^T \mathbf{S} \tilde{\mathbf{C}}_p \right]$  is the  $KG$ -by- $KG$  space-time code correlation matrix, where the vector  $\mathbf{n}$  is a real  $KG$ -dimensional Gaussian noise vector with covariance  $\sigma^2 \mathbf{R}$ , and where the real operator on vectors and matrices is defined by  $\text{Re}(\mathbf{x}) \triangleq (\mathbf{x} + \mathbf{x}^*) / 2$ . Assuming non-orthogonal space-time codes, each component of  $\mathbf{y}$  is corrupted by multiaccess interference from the  $G - 1$  other substreams destined for that user and the  $(K-1)G$  substreams for the other users. The interuser interference is eliminated using a group decorrelating detector [6], and the remaining intrauser interference is eliminated using the V-BLAST algorithm. For each remaining substream, this algorithm detects the strongest substream using a decorrelator, removes the reconstructed estimate of the strongest (post-decorrelator) substream's signal, and then repeats the process until all  $G$  substreams are detected. We initialize  $\mathbf{X}_p(1) = \tilde{\mathbf{S}} \tilde{\mathbf{C}}_p$  for the  $p$ th antenna on the first iteration. On successive iterations,  $\mathbf{X}_p(j)$  is derived from  $\mathbf{X}_p(j-1)$  by striking out the column corresponding to the strongest remaining substream, designated by the index  $g(j)$ . Defining  $\mathbf{R}(j) \triangleq \sum_{p=1}^P \text{Re} \left[ \mathbf{X}_p^H(j) \mathbf{X}_p(j) \right]$  and

defining  $\tilde{\mathbf{R}}(j) \triangleq \left[ \mathbf{R}^{-1}(j) \right]_{[1:M-j+1, 1:M-j+1]}$  to be upper-left  $(M-j)$ -by- $(M-j)$  submatrix of  $\mathbf{R}^{-1}(j)$ , the conditional bit error rate for the  $g$ th substream of user 1, assuming the desired substream's data bit is one, is

$$P_{1,g(j)}(\sigma) = Q \left( \frac{\left[ \tilde{\mathbf{R}}^{-1}(j) \sum_{p=1}^P \mathbf{X}_p^H(j) \Gamma(j) \right]_{g(j)}}{\sigma \sqrt{\left[ \tilde{\mathbf{R}}^{-1}(j) \right]_{(g(j),g(j))}}} \right) \quad (5)$$

where  $\Gamma(j) \triangleq \tilde{\mathbf{S}} \tilde{\mathbf{C}}_p \mathbf{A} \mathbf{b} - \sum_{i=1}^{j-1} \mathbf{S}_{1,g(i)} \begin{bmatrix} \mathbf{c}_{m(g(i),1),p} \\ \vdots \\ \mathbf{c}_{m(g(i),M),p} \end{bmatrix} A_i \hat{\mathbf{b}}_{1,g(i)}$ , where

$\hat{\mathbf{b}}_{1,g(j)}$  is the bit estimate for the user's  $g(j)$ th bit, and where  $[\mathbf{x}]_k$  denotes the  $k$ th component of the vector  $\mathbf{x}$ . Since the linear transformation  $\mathbf{R}^{-1}$  requires the knowledge of all  $KG$  substream

codes, this information could be transmitted to the receiver on an auxiliary control channel.

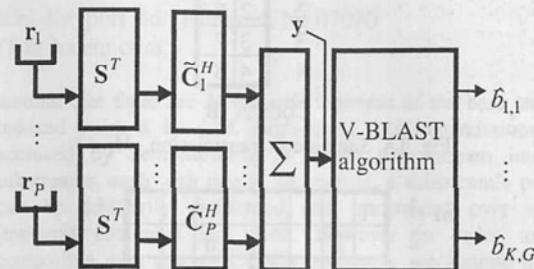


Fig. 2. Block diagram of the space-time decorrelating detector

V. SPECTRAL EFFICIENCY ANALYSIS

In this section we describe the technique for determining capacity in terms of the number of users per sector the system can support. The spectral efficiency is given by the total data throughput per sector divided by the bandwidth. The capacity analysis is based on the approach in [3] for voice, but with suitable modifications to account for multiple transmit and receive antennas and the fact that a decorrelating detector at the front-end of the receiver is used. The capacity analysis consists of two parts: the outage curves and the BER curves.

The outage curves are obtained by randomly distributing a given number of users in each sector of a multiple cell system (we use a 19 cell system with 2 tiers of cells) and then determining the probability that the received  $E_b/N_0$  does not meet the  $E_b/N_0$  requirement for all users. We assume that the cells are divided into three 120 degree sectors, that there is perfect sectorization at the base station (no sidelobe energy), and that there is no soft handoff. The received  $E_b/N_0$  (following the code matched filters and channel combiner but prior to antenna combining) is given by

$$\left( \frac{E_b}{N_0} \right)_{rx} = \frac{\beta S \phi_k \gamma_{k,1} N}{\sum_{b=2}^{19} S \gamma_{k,b} + N_0 W} \quad (6)$$

where  $S$  is the maximum transmit power available at each base station,  $\gamma_{k,b}$  is the shadow fading and path loss from base station  $b$  to user  $k$  which is communicating with cell 1 (the center cell),  $\phi_k$  is the fraction of the total power available in base station 1 to each substream of user  $k$ ,  $N_0$  is the AWGN spectral density,  $W$  is the signal bandwidth,  $N$  is the spread factor,  $\beta$  is the fraction of the base station power available for information, and  $1 - \beta$  is the fraction devoted to the pilot. The denominator term includes the interference and noise terms. The interference term consists only of the out-of-cell interference. Unlike the voice capacity analysis in [3], since a decorrelator detector is used, the in-cell interference is ignored in the outage analysis since it is considered in the BER analysis to derive the target  $E_b/N_0$ . Note that the above equation is independent of the number of transmit and receive antennas since the received  $E_b/N_0$  is measured prior to combining among the  $P$  receive antennas. Since all the substreams of a given user are demodulated at the same location, each one requires the



same transmit power level. Hence the total transmit power to user  $k$  for  $G$  substreams is  $G\phi_k$ .

For successful decoding of the received signal the received  $E_b/N_0$  has to be larger than the  $E_b/N_0$  required to meet the target BER. The base station can choose the power fractions  $G\phi_k$  to each user just sufficiently large to meet the  $E_b/N_0$  requirement so as to support as many users as possible subject to the total power constraint  $\sum_{k=1}^K G\phi_k \leq 1$ . Note that this assumption amounts to having perfect power control in our capacity calculations. An outage event is said to occur if for a given number of users  $K$ , it is not possible for all the users to meet their  $E_b/N_0$  requirement, i.e., to satisfy the condition  $(E_b/N_0)_{rx} = (E_b/N_0)_{req}$ . From (6) this yields

$$\phi_k = \frac{(E_b/N_0)_{req}}{\beta N} \left( \sum_{b=2}^{19} \frac{\gamma_{k,b}}{\gamma_{k,1}} + \frac{N_0 W}{S \gamma_{k,1}} \right), \quad (7)$$

and the outage probability is thus  $\Pr \left\{ \sum_{k=1}^K G\phi_k > 1 \right\}$ .

This is a probability averaged over the shadow fading and the random user locations. For a fixed  $E_b/N_0$ , we plot the outage probability versus the number of users  $K$ . Then for a fixed outage probability, we can plot the measured  $E_b/N_0$  versus  $K$  as shown in Fig. 3, using  $N=128$ ,  $\beta=0.8$ ,  $G=1$ , and activity factor of 1.

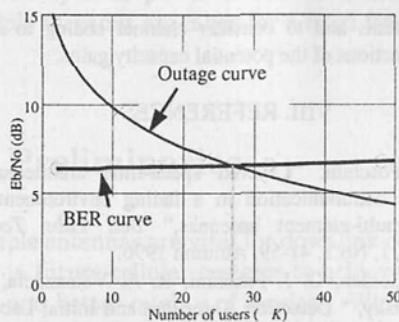


Fig. 3. Determining the number of users from the outage and BER curves

The BER curves determine the required  $E_b/N_0$  to achieve the target BER. As a function of  $E_b/N_0$ , they are generated for a given  $K$ ,  $M$ ,  $M_t$ ,  $L$ , and  $P$  using the received signal model (2) and BER expression (5). In this context the  $E_b/N_0$  corresponding to (6) is given as  $A^2/2\sigma^2$ . This is the signal-to-noise ratio after despreading, multipath combining and transmit diversity combining but prior to receive diversity combining. Hence a required  $E_b/N_0$  value obtained from the BER curves can be directly applied in equation (7) to obtain the appropriate outage curve. The additive noise  $n_p$  in BER calculation plays the role of out-of-cell interference and thermal noise. Note that in the case of a decorrelation type detector the BER does not depend on the received power levels of the interfering users. Hence the calculation of the BER can be independent of the power control scheme used. After plotting the BER versus

$E_b/N_0$  for various values of  $K$ , then for a fixed BER, we can plot the required  $E_b/N_0$  versus  $K$  as shown in Figure 3 ( $L=2$ ,  $M=1$ ,  $P=4$ ,  $N=128$ , using same code transmission).

The resulting system capacity corresponds to the intersection of the outage and BER curves, and given this capacity  $C$  in terms of the number of users, the spectral efficiency is the total sector throughput per bandwidth:  $SE = CGR/W = CG/N$ . The units are given in bits per second per Hertz per sector.

### VI. NUMERICAL RESULTS

Fig. 4 shows the system spectral efficiency versus the number of receive antennas. The top three graphs are for same-code transmission, and the bottom three are for different-code transmission. The columns of graphs correspond to different numbers of transmit antennas ( $M=2, 4, 8$ ), and the curves in each graph are parameterized by the transmit diversity order ( $M_t=1, 2, 4$ ). We fix the number of substreams per data user to be  $G=8$  and consider the spectral efficiency as  $M$ ,  $P$ , and  $M_t$  are varied. Assuming the data rate of a single substream is 9.6 Kbps, each data user receives a rate of 76.8Kbps. We assume a spreading factor of  $N=128$  and  $L=2$  resolvable multipath components.

We observe that the spectral efficiency increases linearly with the number of receive antennas  $P$ . This is because, roughly speaking, the required pre-antenna-combining  $E_b/N_0$  decreases linearly with  $P$ ; hence for a fixed base station transmit power, increasing the number of antennas results in a corresponding increase in spectral efficiency. Note however that the capacity eventually becomes saturated when the set of orthogonal codes is exhausted. For same-code transmission, the number of codes used is  $GM_t/M$ . Hence if the spreading factor is  $N$ , the maximum number of users is  $(NM)/(GM_t)$ , and the maximum spectral efficiency is  $M/M_t$ . For different-code transmission, the number of codes used is  $GM_t$ , and the maximum spectral efficiency is  $1/M_t$ . Therefore while both different-code transmission and transmit diversity improve the link performance, the tradeoff is a lower achievable spectral efficiency. It follows that if  $P$  is large, we would opt to use same-code transmission and no transmit diversity to achieve the highest spectral efficiency. From Fig. 4 we see that with  $P=12$ , the code limited spectral efficiency of 4 bps/Hz per sector is achieved with  $M=4$  and  $M_t=1$ . The capacity improvement with  $M=8$  is negligible considering the additional transmitter and receiver complexity.

If  $P$  is small, we would use different-code transmission and as much transmit diversity as possible (without running out of codes) to improve the link performance. With  $P=2$ , the spectral efficiency is maximized (0.32 bps/Hz per sector) using  $M=2$  antennas,  $M_t=2$ , and different code transmission.

For comparison, we use Monte Carlo simulations to evaluate a voice-only system which uses single antennas at the transmitter and receivers. With a fixed required  $E_b/N_0$  of 7dB and voice activity factor 3/8, this system can support 20 users, resulting in a spectral efficiency of  $20/128 = 0.16$  bps/Hz per sector.

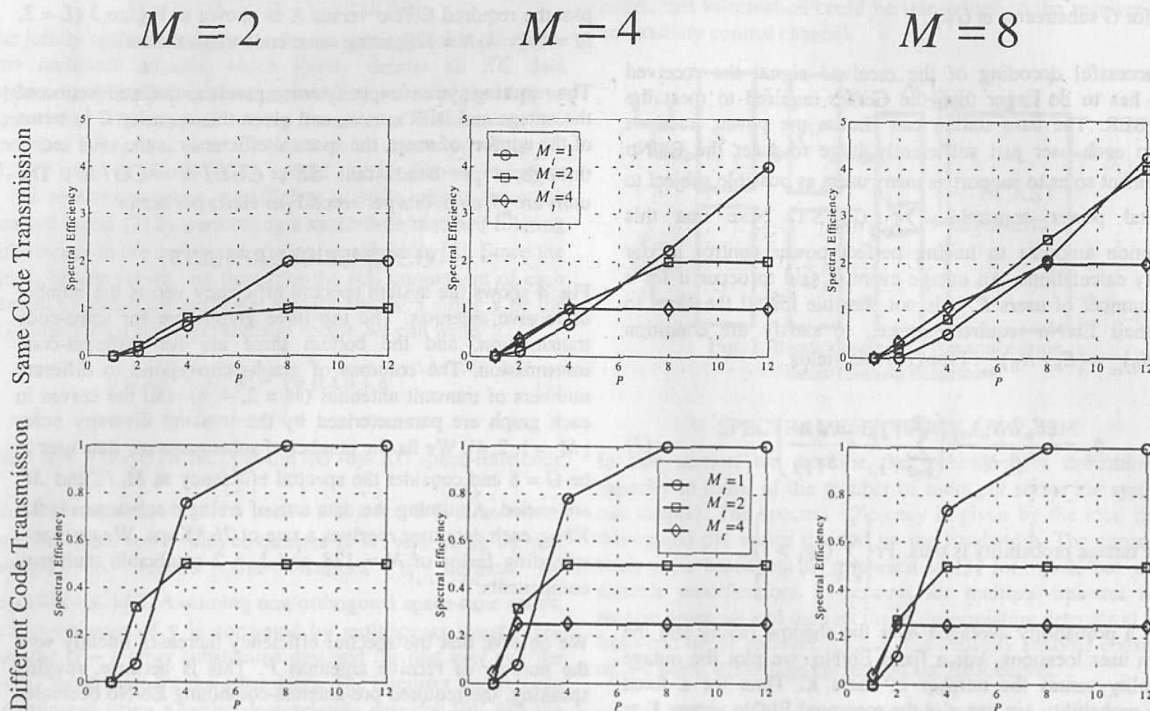


Figure 4. System spectral efficiency

VII. CONCLUSIONS

We studied a high-speed downlink CDMA system which uses multiple antenna transmit diversity, multicode transmission, and space-time detectors. Using a novel technique for evaluating the system capacity, we showed that significant capacity improvements could be achieved. The transmission method which yields the highest system spectral efficiency depends on the number of antennas  $P$  at the receivers. If  $P$  is small so that the maximum potential spectral efficiency is less than one, we would use the codes liberally to provide signal separation via different code transmission and to improve the link performance via transmit diversity. On the other hand, if  $P$  is larger so that the spectral efficiency is potentially greater than one, we would achieve this capacity by conserving codes with same-code transmission and less transmit diversity.

For the range of parameters we considered, the maximum spectral efficiency (4.0 bps/Hz per sector) was achieved using same-code transmission with  $M = 4$  transmit,  $P = 12$  receive antennas, and no transmit diversity.

While we have demonstrated that there is potential for significant capacity gains from using multiple transmit and receive antennas in CDMA systems, the results were based on assumptions such as perfect power control, perfect channel estimation and complex processing at the receiver. It remains for future work to study the effect of non-idealities that occur in

practical systems and to consider channel coding to achieve significant fractions of the potential capacity gains.

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