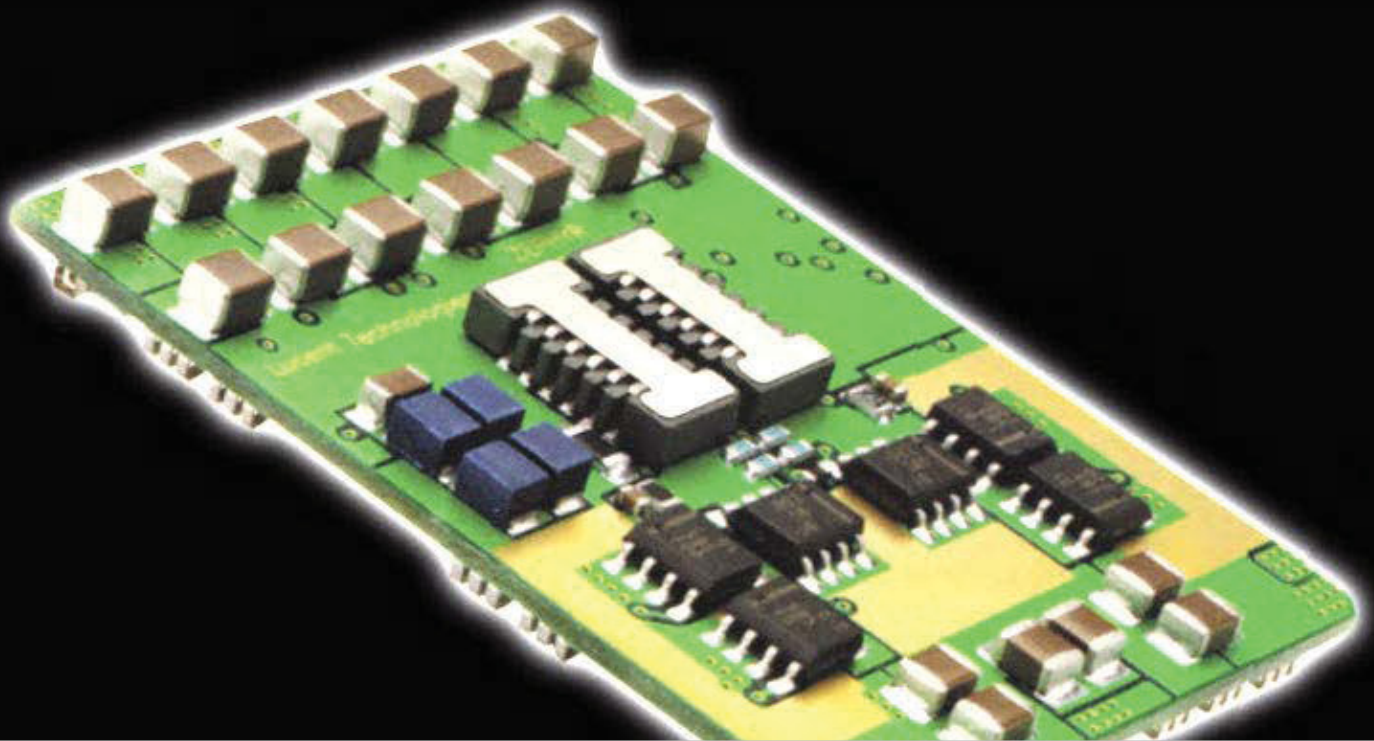


ROBERT W. ERICKSON
Dragan Maksimović

Fundamentals of Power Electronics

Second Edition



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SECOND EDITION

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Contents

Preface	xix
1 Introduction	1
1.1 Introduction to Power Processing	1
1.2 Several Applications of Power Electronics	7
1.3 Elements of Power Electronics	9
References	
I Converters in Equilibrium	11
2 Principles of Steady State Converter Analysis	13
2.1 Introduction	13
2.2 Inductor Volt-Second Balance, Capacitor Charge Balance, and the Small-Ripple Approximation	15
2.3 Boost Converter Example	22
2.4 Ćuk Converter Example	27
2.5 Estimating the Output Voltage Ripple in Converters Containing Two-Pole Low-Pass Filters	31
2.6 Summary of Key Points	34
References	34
Problems	35
3 Steady-State Equivalent Circuit Modeling, Losses, and Efficiency	39
3.1 The DC Transformer Model	39
3.2 Inclusion of Inductor Copper Loss	42
3.3 Construction of Equivalent Circuit Model	45

$$0 = \frac{1}{T_s} \int_0^{T_s} i_C(t) dt = \langle i_C \rangle \quad (2.27)$$

The average value, or dc component, of the capacitor current must be zero in equilibrium.

This should be an intuitive result. If a dc current is applied to a capacitor, then the capacitor will charge continually and its voltage will increase without bound. Likewise, if a dc voltage is applied to an inductor, then the flux will increase continually and the inductor current will increase without bound. Equation (2.27), called the principle of *capacitor amp-second balance* or *capacitor charge balance*, can be used to find the steady-state currents in a switching converter.

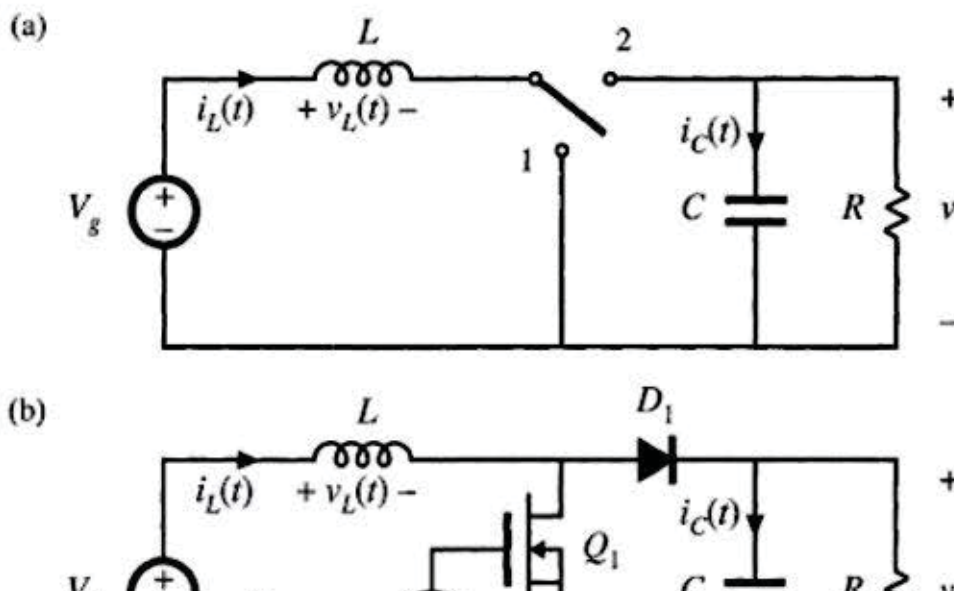
2.3 BOOST CONVERTER EXAMPLE

The boost converter, Fig. 2.13(a), is another well-known switched-mode converter that is capable of producing a dc output voltage greater in magnitude than the dc input voltage. A practical realization of the switch, using a MOSFET and diode, is shown in Fig. 2.13(b). Let us apply the small-ripple approximation and the principles of inductor volt-second balance and capacitor charge balance to find the steady-state output voltage and inductor current for this converter.

With the switch in position 1, the right-hand side of the inductor is connected to ground, resulting in the network of Fig. 2.14(a). The inductor voltage and capacitor current for this subinterval are given by

$$\begin{aligned} v_L &= V_g \\ i_C &= -\frac{v}{R} \end{aligned} \quad (2.28)$$

Use of the linear ripple approximation, $v \approx V$, leads to



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