Case 1:15-cv-00616-RGA Document 8 Filed 09/09/15 Page 1 of 2 PageID #: 279

Director of the U.S. P.G Alexandr	1ail Stop 8 Patent and Trademark Of D. Box 1450 ria, VA 22313-1450	TRADEMARK
In Compliance v	with 35 U.S.C. § 290 and/or 15	5 U.S.C. § 1116 you are hereby advised that a court action has been District of Delaware on the following
☐ Trademarks or ☐ P	Patents. (the patent actio	on involves 35 U.S.C. § 292.):
	DATE FILED	U.S. DISTRICT COURT District of Delaware
CKET NO. 15-cv-616-RGA	7/17/2015	IDEFENDANT
NINTIFF ⊋ Delta, LLC		Verizon Communications Inc., Verizon Services Corp., Verizon Online LLC, Verizon Business Network Services Inc., Verizon Delaware LLC, and Verizon Information Technologies LLC
PATENT OR	DATE OF PATENT OR TRADEMARK	HOLDER OF PATENT OR TRADEMARK
TRADEMARK NO. See Attached	OR TALLES	
See Attached		
		1/2) have been included:
		he following patent(s)/ trademark(s) have been included:
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9/9/2015	[
PATENT OR TRADEMARK NO.	DATE OF PATENT OR TRADEMARK	HOLDER OF PATENT OR TRADEMARK
PATENT OR TRADEMARK NO.	DATE OF PATENT	
PATENT OR TRADEMARK NO. 1 US 9,094,268 B2	DATE OF PATENT OR TRADEMARK	HOLDER OF PATENT OR TRADEMARK
PATENT OR TRADEMARK NO. 1 US 9,094,268 B2	DATE OF PATENT OR TRADEMARK	HOLDER OF PATENT OR TRADEMARK
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PATENT OR TRADEMARK NO. 1 US 9,094,268 B2 2	DATE OF PATENT OR TRADEMARK	HOLDER OF PATENT OR TRADEMARK
PATENT OR TRADEMARK NO. 1 US 9,094,268 B2 2 3 4	DATE OF PATENT OR TRADEMARK 7/28/2015	HOLDER OF PATENT OR TRADEMARK TQ Delta, LLC
PATENT OR TRADEMARK NO. 1 US 9,094,268 B2 2 3 4 5	DATE OF PATENT OR TRADEMARK 7/28/2015	HOLDER OF PATENT OR TRADEMARK
PATENT OR TRADEMARK NO. 1 US 9,094,268 B2 2 3 4	DATE OF PATENT OR TRADEMARK 7/28/2015	HOLDER OF PATENT OR TRADEMARK TQ Delta, LLC
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PATENT OR TRADEMARK NO. 1 US 9,094,268 B2 2 3 4 5	DATE OF PATENT OR TRADEMARK 7/28/2015	HOLDER OF PATENT OR TRADEMARK TQ Delta, LLC
PATENT OR TRADEMARK NO. 1 US 9,094,268 B2 2 3 4 5	DATE OF PATENT OR TRADEMARK 7/28/2015	HOLDER OF PATENT OR TRADEMARK TQ Delta, LLC

Case 1:15-cv-00616-RGA Document 8 Filed 09/09/15 Page 2 of 2 PageID #: 280

PATENT OR	DATE OF PATENT	HOLDER OF PATENT OR TRADEMARK
TRADEMARK NO.	OR TRADEMARK	TQ Delta, LLC
1 US 6,961,369 B1	11/1/2005	TO Delta, LLC
2 US 8,718,158 B2	5/6/2014	TQ Delta, LLC
3 US 9,014,243 B2	4/21/2015	TQ Delta, LLC
4 US 7,835,430 B2	11/16/2010	TO Delta, LLC
5 US 8,238,412 B2	8/7/2012 4/30/2013	TQ Delta, LLC
6 US 8,432,956 B2	12/17/2013	TQ Delta, LLC
7 US 8,611,404 B2	12/11/2013	

AO 120 (Rev. 08/10)

TO:

Mail Stop 8 Director of the U.S. Patent and Trademark Office P.O. Box 1450 Alexandria. VA 22313-1450

REPORT ON THE FILING OR DETERMINATION OF AN ACTION REGARDING A PATENT OR TRADEMARK

Alevandr	O. Box 1450 ria, VA 22313-1450			TRADEMA	.KK
In Compliance v	with 35 U.S.C. § 290 and/or 1:	5 U.S.C. § Dis	1116 you are hereby trict of Delaware	advised that a court a	on the following
filed in the U.S. District	Patents. (the patent action				
	DATE FILED		CTRICT COURT	D: L: L of Delevio	ro
15-cv-615-RGA	7/17/2015		DEFENDANT	District of Delawa	<u> </u>
AINTIFF Q Delta, LLC				able Inc. and Tim C	e Warner Cable
PATENT OR TRADEMARK NO.	DATE OF PATENT OR TRADEMARK		HOLDER	R OF PATENT OR T	RADEMARK
See Attached					
	In the above—entitled case, the	he followi	ng patent(s)/ trademarl	k(s) have been includ	led:
		he followi	☐ Answer	Cross Bill	☐ Other Pleading
ATE INCLUDED 9/9/2015 PATENT OR	INCLUDED BY		☐ Answer		☐ Other Pleading
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Case 1:15-cv-00615-RGA Document 8 Filed 09/09/15 Page 2 of 2 PageID #: 263

PATENT OR	DATE OF PATENT	HOLDER OF PATENT OR TRADEMARK
TRADEMARK NO.	OR TRADEMARK	TQ Delta, LLC
1 US 6,961,369 B1	11/1/2005	TQ Delta, LLC
2 US 8,718,158 B2	5/6/2014	TQ Delta, LLC
3 US 9,014,243 B2	4/21/2015	TQ Delta, LLC
4 US 7,835,430 B2	11/16/2010	TQ Delta, LLC
5 US 8,238,412 B2	8/7/2012	TQ Delta, LLC
6 US 8,432,956 B2	4/30/2013	TQ Delta, LLC
7 US 8,611,404 B2	12/17/2013	

AO 120 (Rev. 08/10)

Mail Stop 8 Director of the U.S. Patent and Trademark Office P.O. Box 1450

REPORT ON THE FILING OR DETERMINATION OF AN ACTION REGARDING A PATENT OR

P Alexano	S. Patent and Trademark Offi P.O. Box 1450 dria, VA 22313-1450	TRADEMARK	ATENT OR
Glad in the H.S. Distr	e with 35 U.S.C. § 290 and/or 15 U.S.C. § Middle D. Patents. (the patent action	istrict of Florida, Orlando Strieta	has been on the following
OCKET NO. 6:14-cv-1027	DATE FILED 6/26/2014	U.S. DISTRICT COURT Middle District of Florida, Orlando	Division
6:14-69-1027 AINTIFF Orlando Communication		DEFENDANT LG Electronics, Inc., et al	
PATENT OR TRADEMARK NO.	DATE OF PATENT OR TRADEMARK	HOLDER OF PATENT OR TRADE	EMARK
5,687,1296	11/11/1997	James Arthur Proctor, Jr., James Carl Otto	
6,0009,553	12/28/0199	Dennis Martinez, Thomas Hengeveld, MIch	ael Axford
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	In the above—entitled case, the	following patent(s)/ trademark(s) have been included:	
DATE INCLUDED	INCLUDED BY	ndment	Other Pleading
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DECISION/JUDGEMENT Order of Dismissal			
	Lan	() DEPUTY CLERK	DATE
CLERK Sheryl Loesch	L ·	R. Olsen	9/9/2015

AO 120 (Rev. 08/10)

Mail Stop 8 Director of the U.S. Patent and Trademark Office P.O. Box 1450

REPORT ON THE FILING OR DETERMINATION OF AN

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OCKET NO. 6:14-cv-1026	DATE FILED 6/26/2014	U.S. D	ISTRICT COURT Middle District of Florida, Orlando	Division
_AINTIFF Orlando Communication	s LLC		LG Electronics, Inc., et al	
PATENT OR TRADEMARK NO.	DATE OF PATENT OR TRADEMARK		HOLDER OF PATENT OR TRAD	EMARK
5,687,1296	11/11/1997	1	nes Arthur Proctor, Jr., James Carl Otto	
2 6,0009,553	12/28/0199	De	nnis Martinez, Thomas Hengeveld, MIch	nael Axford
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DATE INCLUDED PATENT OR TRADEMARK NO.	INCLUDED BY	e follows nendment	ng patent(s)/ trademark(s) have been included: Answer	Other Pleading
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In the ab DECISION/JUDGEMENT Order of Dismissal	ove—entitled case, the followin	ng decisio	n has been rendered or judgement issued:	
CLERY](1	BY) DEP	UTY CLERK	DATE
CLERK Sheryl Loesch	1	R. Ols		9/9/2015

Mail Stop 8 Director of the U.S. Patent and Trademark Office

REPORT ON THE FILING OR DETERMINATION OF AN

P	. Patent and Trademark Of .O. Box 1450 Iria, VA 22313-1450		ACTION REGARDING A I TRADEMARK	PATENT OR
filed in the U.S. Distri	with 35 U.S.C. § 290 and/or 15 ict Court Middle Patents. (the patent actio	District c	116 you are hereby advised that a court action f Florida, Orlando Division 35 U.S.C. § 292.):	n has been on the following
_			PRIOT COURT	
OCKET NO. 6:14-cv-1028	DATE FILED 6/26/2014		Middle District of Florida, Online	o Division
AINTIFF		[]	DEFENDANT	
Orlando Communications	s LLC		HTC Corporation, et al	
PATENT OR	DATE OF PATENT	$\frac{1}{1}$	HOLDER OF PATENT OR TRAI	DEMARK
TRADEMARK NO.	OR TRADEMARK	 	es Arthur Proctor, Jr., James Carl Ott	0
5,687,1296	11/11/1997			
6,0009,553	12/28/0199	Denr	is Martinez, Thomas Hengeveld, MIc	Hael Axiolu
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	ove—entitled case, the followin	g uccision	,	
DECISION/JUDGEMENT Order of Dismissal				
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CLERK Chard Loosch	T	R. Olse		9/9/2015
Sheryl Loesch				<u> </u>

AO 120 (Rev. 08/10) Mail Stop 8 TO: Director of the U.S. Patent and Trademark Office

REPORT ON THE FILING OR DETERMINATION OF AN

ſ	P.O. Box 1450 ndria, VA 22313-1450		ACTION REGARDING TRADEMA	
filed in the U.S. Dist	*	Dis	§ 1116 you are hereby advised that a court trict of Delaware	action has been on the following
DOCKET NO.	DATE FILED 7/17/2015	U.S. DI	STRICT COURT District of Delawa	re
PLAINTIFF	771772013	<u> </u>	DEFENDANT	ie .
TQ Delta, LLC			Verizon Communications Inc., Verizon Online LLC, Verizon Business Netwo Delaware LLC, and Verizon Informa	ork Services Inc., Verizon
PATENT OR TRADEMARK NO.	DATE OF PATENT OR TRADEMARK	1	HOLDER OF PATENT OR TI	RADEMARK
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DECISION/JUDGEMENT				1 1 2
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	PATENT OR	DATE OF PATENT	HOLDER OF PATENT OR TRADEMARK
	TRADEMARK NO.	OR TRADEMARK	HOLDER OF FATENT OR TRADEMARK
1	US 6,961,369 B1	11/1/2005	TQ Delta, LLC
2	US 8,718,158 B2	5/6/2014	TQ Delta, LLC
3	US 9,014,243 B2	4/21/2015	TQ Delta, LLC
4	US 7,835,430 B2	11/16/2010	TQ Delta, LLC
5	US 8,238,412 B2	8/7/2012	TQ Delta, LLC
6	US 8,432,956 B2	4/30/2013	TQ Delta, LLC
7	US 8,611,404 B2	12/17/2013	TQ Delta, LLC

AO 120 (Rev. 08/10)			
	Mail Stop 8 J.S. Patent and Trademark (P.O. Box 1450 andria, VA 22313-1450		REPORT ON THE OR DETERMINATION OF AN REGARDING A PATENT OR TRADEMARK
filed in the U.S. Dis	·	Delaware	lvised that a court action has been on the following
DOCKET NO.	DATE FILED 9/4/2013	J.S. DISTRICT COURT	Delaware
PLAINTIFF	1	DEFENDANT	Dolawaro
ROCHE PALO ALTO L GENENTECH, INC.	LC and	WATSON LABOR	RATORIES, INC. – FLORIDA
PATENT OR TRADEMARK NO.	DATE OF PATENT OR TRADEMARK	HOLDER O	F PATENT OR TRADEMARK
1 6,083,953	7/4/2000	Roche Palo Alto LLC	
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CLERK John A Cerino, United States D 844 N. King Str	Pistrict Court	PUTY CLERK	DATE 7/17/15
Wilmington, DE Copy 1—Upon initiation of a Copy 2—Upon filing docume	19801 ection, mail this copy to Directo ent adding patent(s), mail this c	Copy 3—Upon termination of to Director Copy 4—Case	of action, mail this copy to Director file copy

	Mail Stop 8 U.S. Patent and Trademark C P.O. Box 1450 andria, VA 22313-1450	REPORT ON THE FILING OR DETERMINATION OF AN ACTION REGARDING A PATENT OR TRADEMARK
·	nce with 35 U.S.C. § 290 and/or 1: Massacl	5 U.S.C. § 1116 you are hereby advised that a court action has been husetts on the following G Patents or G Trademarks:
DOCKET NO.	DATE FILED 7/17/2015	U.S. DISTRICT COURT Massachusetts
PLAINTIFF		DEFENDANT
BOSTON PROPERTIE	ES LIMITED PARTNERSHIF	P CLAUDETTE MOUSSA, d/b/a Boston Properties Advisors
PATENT OR TRADEMARK NO.	DATE OF PATENT OR TRADEMARK	HOLDER OF PATENT OR TRADEMARK
1 2,527,181	1/8/2002	BOSTON PROPERTIES LIMITED PARTNERSHIP
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UNITED STATES PATENT AND TRADEMARK OFFICE

CERTIFICATE OF CORRECTION

PATENT NO. : 8,718,158 B2 Page 1 of 1

APPLICATION NO. : 13/303417

DATED : May 6, 2014

INVENTOR(S) : Marcos C. Tzannes

It is certified that error appears in the above-identified patent and that said Letters Patent is hereby corrected as shown below:

In the Claims

At Column 11, Claim 15, line 49, delete "the input bit stream," and insert -- at least one bit of the plurality of data bits, --

Signed and Sealed this Twenty-eighth Day of April, 2015

Michelle K. Lee

Michelle K. Lee

Director of the United States Patent and Trademark Office

ATE : 02/04	1/2015	
<u> </u>	JNIT <u>2634</u>	
		on for Appl. No.: <u>13/303.417</u> Patent No.: <u>8,</u> 718,158 B2
OBJECT . Reques	a for Certificate of Correction	
N		CofC mailroom date: 12/04/2014
•	his request for a certif	ficate of correction within 7 days.
FOR IFW FILES:		
	image. No new matt	rrections as shown in the COCIN document(s) in er should be introduced, nor should the scope or
Please complete the using document cod		w) and forward the completed response to scanning
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OR PAPER FILES	=	
Please review the recorrection. Please of	equested changes/co complete this form (se	rrections as shown in the attached certificate of ee below) and forward it with the file to:
Please review the recorrection. Please of Certificates of Randolph School Palm Location	equested changes/corcomplete this form (see of Correction Branchuare – 9D10-A	ee below) and forward it with the file to: h (CofC)
Please review the recorrection. Please correction Please correction Please corrections of the Plandolph Science (Plandolph Science)	equested changes/corcomplete this form (see of Correction Branchuare – 9D10-A	ee below) and forward it with the file to: h (CofC)
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Please review the recorrection. Please correction. Please correction. Please correction. Please correction. Please correction. Please correction. Plant Location in the Classical Correction.	equested changes/corcomplete this form (see of Correction Branch quare – 9D10-A on 7580 aims	ee below) and forward it with the file to: h (CofC) - - - - - - - - - -
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Certificates of Randolph Son Palm Location in the Classicates. Thank You For You have your decision on the appropriate of the page of the	equested changes/corcomplete this form (see of Correction Branch quare – 9D10-A on 7580 aims	ee below) and forward it with the file to: h (CofC) Overginia Tolbert Certificates of Correction Branch (571) 272-0460 Intifled correction(s) is hereby:
Certificates of Randolph Son Palm Location in the Classicates. Thank You For You have your decision on the appropriate of the page of the	equested changes/corcomplete this form (see of Correction Brancinguare – 9D10-A on 7580 aims ur Assistance suing the above-ider propriete box.	ee below) and forward it with the file to: h (CofC) Oirginia Tolbert Certificates of Correction Branch (571) 272-0460 ntified correction(s) is hereby: All changes apply.
Certificates of Randolph Son Palm Location. In the Classicates. In the Classicates of Palm Location Location. Thank You For You have your decision on the appropriate of the Pappropriate Control Lands of the Pappropriate Control Lands of the Pappropriate Lands of the Lands of	equested changes/corcomplete this form (see of Correction Brancinguare – 9D10-A on 7580 aims ur Assistance suing the above-ider propriete box.	ee below) and forward it with the file to: h (CofC) Oirginia Tolbert Certificates of Correction Branch (571) 272-0460 ntified correction(s) is hereby: All changes apply. Specify below which changes do not apply.

PTOL-306 (REV. 7/03)

DISH Exhibit 1002 Page 13

U.S. DEPARTMENT OF COMMERCE Patent and Trademark Office

OATE	: 02/04/2015				
O SPE OF	· VEIGHIAU IV	•			
O SPE OI	: ART UNIT 2634				
SUBJECT	: Request for Certificate of Correction for Appl. No.: 13/303.417 Patent No.: 8,718,158 B2				
		CofC mailroom date: 12/04/2014			
Please resp	ond to this request for a cer	tificate of correction within 7 days.			
FOR IFW F	ILES:				
he IFW app	ew the requested changes/c plication image. No new ma the claims be changed.	orrections as shown in the COCIN document(s) in tter should be introduced, nor should the scope or			
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Please revie correction.	ew the requested changes/c Please complete this form (orrections as shown in the attached certificate of see below) and forward it with the file to:			
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Rand	ificates of Correction Bran dolph Square – 9D10-A ı Location 7580	Cn (COTC)			
Rand Palm	dolph Square – 9D10-A	ch (Co1C)			
Rand Palm	dolph Square – 9D10-A n Location 7580				
Rand Palm	dolph Square – 9D10-A n Location 7580	<u> Virginia Tolbert</u>			
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DISH Exhibit 1002 Page 14

IN THE UNITED STATES PATENT AND TRADEMARK OFFICE

In Re the Application of: Marcos C. Tzannes

Application No.: 13/303,417

Patent No.: 8,718,158

Filed: November 23, 2011

Issued: May 6, 2014

Examiner: PHU, Phuong M.

Confirmation No.: 3035

Art Unit: 2634

Atty. File No.: 6936-47-CON-5

For: SYSTEM AND METHOD FOR SCRAMBLING THE PHASE OF THE CARRIERS IN A MULTICARRIER COMMUNICATIONS SYSTEM

REQUEST FOR CERTIFICATE OF CORRECTION OF PATENT FOR APPLICANT'S MISTAKE (37 C.F.R §1.323)

Attn: Certificate of Corrections Branch Commissioner for Patents P.O. Box 1450 Alexandria, VA 22313-1450

Sir:

This is a request for a Certificate of Correction under 37 C.F.R. 1.323 for the above-identified patent. Attached is form PTO/SB/44. Applicants have determined that there was an error in claim 15 which was the mistake of the Applicant. The correction does not introduce new matter, do not require re-examination, and do not materially affect the scope or meaning of the patent.

• At Column 11, Claim 15, line 49, delete "the input bit stream," and insert -- at least one bit of the plurality of data bits, --

Atty Ref. No.: 6936-47-CON-5

Payment in the amount of \$100.00, as required by 37 C.F.R. § 1.20(a) is being submitted herewith. Although no additional fees are believed due in connection with the filing of this paper, please charge any additional fees deemed necessary or credit any overpayment to Deposit Account No. 19-1970.

Respectfully submitted,

SHERIDAN ROSS P.C.

Date: Upac 14

By:

Jason H. Vick Reg. No. 45,285

1560 Broadway, Suite 1200 Denver, Colorado 80202 Telephone: 303-863-9700 Under the Paperwork Reduction Act of 1995, no persons are required to respond to a collection of information unless it displays a valid OMB control number.

(Also Form PTO-1050)

UNITED STATES PATENT AND TRADEMARK OFFICE

CERTIFICATE OF CORRECTION
Page <u>1</u> of <u>1</u>
PATENT NO. : 8,718,158
APPLICATION NO.: 13/303,417
ISSUE DATE : May 6, 2014
INVENTOR(S) : Marcos C. Tzannes
It is certified that an error appears or errors appear in the above-identified patent and that said Letters Patent is hereby corrected as shown below:
At Column 11, Claim 15, line 49, delete "the input bit stream," and insert at least one bit of the plurality of data bits,

MAILING ADDRESS OF SENDER (Please do not use customer number below):

Jason H. Vick c/o Sheridan Ross, PC 1560 Broadway, Suite 1200 Denver, CO 80202

This collection of information is required by 37 CFR 1.322, 1.323, and 1.324. The information is required to obtain or retain a benefit by the public which is to file (and by the USPTO to process) an application. Confidentiality is governed by 35 U.S.C. 122 and 37 CFR 1.14. This collection is estimated to take 1.0 hour to complete, including gathering, preparing, and submitting the completed application form to the USPTO. Time will vary depending upon the individual case. Any comments on the amount of time you require to complete this form and/or suggestions for reducing this burden, should be sent to the Chief Information Officer, U.S. Patent and Trademark Office, U.S. Department of Commerce, P.O. Box 1450, Alexandria, VA 22313-1450. DO NOT SEND FEES OR COMPLETED FORMS TO THIS ADDRESS. SEND TO: Attention Certificate of Corrections Branch, Commissioner for Patents, P.O. Box 1450, Alexandria, VA 2311-1450. VA 22313-1450.

If you need assistance in completing the form, call 1-800-PTO-9199 and select option 2.

Electronic Patent Application Fee Transmittal						
Application Number:	13	13303417				
Filing Date:	23	23-Nov-2011				
Title of Invention:	SYSTEM AND METHOD FOR SCRAMBLING THE PHASE OF THE CARRIERS IN A MULTICARRIER COMMUNICATIONS SYSTEM					
First Named Inventor/Applicant Name:	Ma	ircos C. Tzannes				
Filer:	Jas	on Vick/Joanne Vos	5			
Attorney Docket Number:	69	36-47-CON-5				
Filed as Large Entity						
Utility under 35 USC 111(a) Filing Fees						
Description		Fee Code	Quantity	Amount	Sub-Total in USD(\$)	
Basic Filing:						
Pages:						
Claims:						
Miscellaneous-Filing:						
Petition:						
Patent-Appeals-and-Interference:						
Post-Allowance-and-Post-Issuance:						
Certificate of Correction 1811 1 100 100						
Extension-of-Time:						

Description	Fee Code	Quantity	Amount	Sub-Total in USD(\$)
Miscellaneous:				
	Tot	al in USD	(\$)	100

Electronic Acknowledgement Receipt					
EFS ID:	20872204				
Application Number:	13303417				
International Application Number:					
Confirmation Number:	3035				
Title of Invention:	SYSTEM AND METHOD FOR SCRAMBLING THE PHASE OF THE CARRIERS IN A MULTICARRIER COMMUNICATIONS SYSTEM				
First Named Inventor/Applicant Name:	Marcos C. Tzannes				
Customer Number:	62574				
Filer:	Jason Vick/Joanne Vos				
Filer Authorized By:	Jason Vick				
Attorney Docket Number:	6936-47-CON-5				
Receipt Date:	04-DEC-2014				
Filing Date:	23-NOV-2011				
Time Stamp:	18:54:53				
Application Type:	Utility under 35 USC 111(a)				

Payment information:

Submitted with Payment	yes
Payment Type	Deposit Account
Payment was successfully received in RAM	\$100
RAM confirmation Number	5356
Deposit Account	191970
Authorized User	

The Director of the USPTO is hereby authorized to charge indicated fees and credit any overpayment as follows:

Charge any Additional Fees required under 37 C.F.R. Section 1.16 (National application filing, search, and examination fees)

Charge any Additional Fees required under 37 C.F.R. Section 1.17 (Patent application and reexamination processing fees)

Charge any Additional Fees required under 37 C.F.R. Section 1.19 (Document supply fees)

Charge any Additional Fees required under 37 C.F.R. Section 1.20 (Post Issuance fees)

Charge any Additional Fees required under 37 C.F.R. Section 1.21 (Miscellaneous fees and charges)

File Listing:

Document Number	Document Description	File Name	File Size(Bytes)/ Message Digest	Multi Part /.zip	Pages (if appl.)			
1	Request for Certificate of Correction	Degreet for Cartificate of Correction Degreet for 2nd COC ndf	Request for 2nd COC.pdf	290374	no	3		
·	nequestroi certificate of correction	' = = - '	06e831637566de56e5429e6eb5be636fea0 54692		J			
Warnings:								
Information:								
2	Fee Worksheet (SB06)	fee-info.pdf	30271	no	2			
2	ree Worksheet (3500)	cebcf357/cbe3a0d3b3f827a83207e5f9b5a0		110	2			
Warnings:								
Information:	Information:							
Total Files Size (in bytes): 320645			20645					

This Acknowledgement Receipt evidences receipt on the noted date by the USPTO of the indicated documents, characterized by the applicant, and including page counts, where applicable. It serves as evidence of receipt similar to a Post Card, as described in MPEP 503.

New Applications Under 35 U.S.C. 111

If a new application is being filed and the application includes the necessary components for a filing date (see 37 CFR 1.53(b)-(d) and MPEP 506), a Filing Receipt (37 CFR 1.54) will be issued in due course and the date shown on this Acknowledgement Receipt will establish the filing date of the application.

National Stage of an International Application under 35 U.S.C. 371

If a timely submission to enter the national stage of an international application is compliant with the conditions of 35 U.S.C. 371 and other applicable requirements a Form PCT/DO/EO/903 indicating acceptance of the application as a national stage submission under 35 U.S.C. 371 will be issued in addition to the Filing Receipt, in due course.

New International Application Filed with the USPTO as a Receiving Office

If a new international application is being filed and the international application includes the necessary components for an international filing date (see PCT Article 11 and MPEP 1810), a Notification of the International Application Number and of the International Filing Date (Form PCT/RO/105) will be issued in due course, subject to prescriptions concerning national security, and the date shown on this Acknowledgement Receipt will establish the international filing date of the application.

UNITED STATES PATENT AND TRADEMARK OFFICE

CERTIFICATE OF CORRECTION

PATENT NO. : 8,718,158 B2 Page 1 of 1

APPLICATION NO. : 13/303417

DATED : May 6, 2014

INVENTOR(S) : Marcos C. Tzannes

It is certified that error appears in the above-identified patent and that said Letters Patent is hereby corrected as shown below:

In the Claims

Column 11, Claim 1, line 9, following "signal;" insert -- and --

Column 11, Claim 5, line 19, delete "Interne" and insert -- Internet --

Signed and Sealed this Sixteenth Day of September, 2014

Michelle K. Lee

Michelle K. Lee

Deputy Director of the United States Patent and Trademark Office

IN THE UNITED STATES PATENT AND TRADEMARK OFFICE

In Re the Application of: Marcos C. Tzannes) Application No.: 13/303,417
Patent No.: 8,718,158	Filed: November 23, 2011
Issued: May 6, 2014	Examiner: PHU, Phuong M.
Confirmation No.: 3035	Art Unit: 2634

Atty. File No.: 6936-47-CON-5

For: SYSTEM AND METHOD FOR SCRAMBLING THE PHASE OF THE CARRIERS IN A MULTICARRIER COMMUNICATIONS SYSTEM

REQUEST FOR CERTIFICATE OF CORRECTION OF PATENT FOR OFFICE'S MISTAKE (37 C.F.R. §1.322)

Attn: Certificate of Corrections Branch Commissioner for Patents P.O. Box 1450 Alexandria, VA 22313-1450

Dear Madam:

This is a request for a Certificate of Correction under 37 C.F.R. 1.322. Attached is form PTO/SB/44. Applicants have discovered there was an error in claims 1 and 5 of the issued patent

- At Column 11, Claim 1, line 9, following "signal;" and insert -- and --
- At Column 11, Claim 5, line 19, delete "interne" and insert -- internet --

The correction to these claims can be found in Applicants' Response to the Non-Final Office Action, filed on December 27, 2013.

Attorney Docket No.: 6936-47-CON-5

Although Applicants do not believe that any additional fees are due in connection with the filing of this paper, please charge any additional fees deemed necessary or credit any overpayment to Deposit Account No. 19-1970:

Respectfully submitted,

SHERIDAN ROSS P.C.

Date: 12 mm /4

Jason H. Vick Reg. No. 45,285

1560 Broadway, Suite 1200

Denver, Colorado 80202 Telephone: 303-863-9700 Approved for use through 08/31/2013, OMB 0651-0033

U.S. Patent and Trademark Office; U.S. DEPARTMENT OF COMMERCE

Under the Paperwork Reduction Act of 1995, no persons are required to respond to a collection of information unless it displays a valid OMB control number. (Also Form PTO-1050)

UNITED STATES PATENT AND TRADEMARK OFFICE

CERTIFICATE OF CORRECTION		
PATENT NO. : 8,718,158	Page <u>1</u>	of <u>1</u>
APPLICATION NO.: 13/303,417		
ISSUE DATE : May 6, 2014		
INVENTOR(S) : Marcos C. Tzannes		
It is certified that an error appears or errors appear in the above-identified patent and the is hereby corrected as shown below:	at said Lette	ers Patent
At Column 11, Claim 1, line 9, following "signal;" and insert and		
At Column 11, Claim 5, line 19, delete "interne" and insert internet		

MAILING ADDRESS OF SENDER (Please do not use customer number below):

Jason H. Vick / Sheridan Ross, PC 1560 Broadway, Suite 1200 Denver, Colorado 80202

This collection of information is required by 37 CFR 1.322, 1.323, and 1.324. The information is required to obtain or retain a benefit by the public which is to file (and by the USPTO to process) an application. Confidentiality is governed by 35 U.S.C. 122 and 37 CFR 1.14. This collection is estimated to take 1.0 hour to complete, including gathering, preparing, and submitting the completed application form to the USPTO. Time will very depending upon the individual case. Any comments on the amount of time you require to complete this form and/or suggestions for reducing this burden, should be sent to the Chief Information Officer, U.S. Patent and Trademark Office, U.S. Department of Commerce, P.O. Box 1450, Alexandria, VA 22313-1450. DO NOT SEND FEES OR COMPLETED FORMS TO THIS ADDRESS. SEND TO: Attention Certificate of Corrections Branch, Commissioner for Patents, P.O. Box 1450, Alexandria, VA 2313-1450. VA 22313-1450.

If you need assistance in completing the form, call 1-800-PTO-9199 and select option 2.

Electronic Acknowledgement Receipt					
EFS ID:	19000114				
Application Number:	13303417				
International Application Number:					
Confirmation Number:	3035				
Title of Invention:	SYSTEM AND METHOD FOR SCRAMBLING THE PHASE OF THE CARRIERS IN A MULTICARRIER COMMUNICATIONS SYSTEM				
First Named Inventor/Applicant Name:	Marcos C. Tzannes				
Customer Number:	62574				
Filer:	Jason Vick/Joanne Vos				
Filer Authorized By:	Jason Vick				
Attorney Docket Number:	6936-47-CON-5				
Receipt Date:	12-MAY-2014				
Filing Date:	23-NOV-2011				
Time Stamp:	12:20:20				
Application Type:	Utility under 35 USC 111(a)				

Payment information:

Submitted with Payment	no

File Listing:

Information:

Document Number	Document Description	File Name	File Size(Bytes)/ Message Digest	Multi Part /.zip	Pages (if appl.)
1	Request for Certificate of Correction	rection.pdf	285839	no	2
'	nequestroi certificate of correction		8fc16c2b44b8455f0a27926eef329b3e9836 49d8		3
Warnings:		'		,	

This Acknowledgement Receipt evidences receipt on the noted date by the USPTO of the indicated documents, characterized by the applicant, and including page counts, where applicable. It serves as evidence of receipt similar to a Post Card, as described in MPEP 503.

New Applications Under 35 U.S.C. 111

If a new application is being filed and the application includes the necessary components for a filing date (see 37 CFR 1.53(b)-(d) and MPEP 506), a Filing Receipt (37 CFR 1.54) will be issued in due course and the date shown on this Acknowledgement Receipt will establish the filing date of the application.

National Stage of an International Application under 35 U.S.C. 371

If a timely submission to enter the national stage of an international application is compliant with the conditions of 35 U.S.C. 371 and other applicable requirements a Form PCT/DO/EO/903 indicating acceptance of the application as a national stage submission under 35 U.S.C. 371 will be issued in addition to the Filing Receipt, in due course.

New International Application Filed with the USPTO as a Receiving Office

If a new international application is being filed and the international application includes the necessary components for an international filing date (see PCT Article 11 and MPEP 1810), a Notification of the International Application Number and of the International Filing Date (Form PCT/RO/105) will be issued in due course, subject to prescriptions concerning national security, and the date shown on this Acknowledgement Receipt will establish the international filing date of the application.



UNITED STATES PATENT AND TRADEMARK OFFICE

UNITED STATES DEPARTMENT OF COMMERCE United States Patent and Trademark Office Address: COMMISSIONER FOR PATENTS

P.O. Box 1450 Alexandria, Virginia 22313-1450 www.uspto.gov

APPLICATION NO. ISSUE DATE PATENT NO. ATTORNEY DOCKET NO. CONFIRMATION NO. 13/303,417 05/06/2014 8718158

62574

04/16/2014

6936-47-CON-5

3035

Jason H. Vick Sheridan Ross, PC Suite # 1200 1560 Broadway Denver, CO 80202

ISSUE NOTIFICATION

The projected patent number and issue date are specified above.

Determination of Patent Term Adjustment under 35 U.S.C. 154 (b)

(application filed on or after May 29, 2000)

The Patent Term Adjustment is 9 day(s). Any patent to issue from the above-identified application will include an indication of the adjustment on the front page.

If a Continued Prosecution Application (CPA) was filed in the above-identified application, the filing date that determines Patent Term Adjustment is the filing date of the most recent CPA.

Applicant will be able to obtain more detailed information by accessing the Patent Application Information Retrieval (PAIR) WEB site (http://pair.uspto.gov).

Any questions regarding the Patent Term Extension or Adjustment determination should be directed to the Office of Patent Legal Administration at (571)-272-7702. Questions relating to issue and publication fee payments should be directed to the Application Assistance Unit (AAU) of the Office of Data Management (ODM) at (571)-272-4200.

APPLICANT(s) (Please see PAIR WEB site http://pair.uspto.gov for additional applicants):

Marcos C. Tzannes, Orinda, CA;

The United States represents the largest, most dynamic marketplace in the world and is an unparalleled location for business investment, innovation, and commercialization of new technologies. The USA offers tremendous resources and advantages for those who invest and manufacture goods here. Through SelectUSA, our nation works to encourage and facilitate business investment. To learn more about why the USA is the best country in the world to develop technology, manufacture products, and grow your business, visit SelectUSA.gov.

IR103 (Rev. 10/09)

UNITED STATES DEPARTMENT OF COMMERCE United States Patent and Trademark Office Address: COMMISSIONER FOR PATENTS P.O. Box 1450 Alexandria, Virginia 22313-1450 www.usplo.gov

APPLICATION NO.	FILING DATE	FILING DATE FIRST NAMED INVENTOR		CONFIRMATION NO.
13/303,417	11/23/2011 Marcos C. Tzannes		6936-47-CON-5	3035
62574 Jason H. Vick	7590 04/07/20:	4	EXAM	IINER
Sheridan Ross, Suite # 1200	PC		PHU, PH	UONG M
1560 Broadway	•		ART UNIT	PAPER NUMBER
Denver, CO 802	202		2634	
			NOTIFICATION DATE	DELIVERY MODE
			04/07/2014	ELECTRONIC

Please find below and/or attached an Office communication concerning this application or proceeding.

The time period for reply, if any, is set in the attached communication.

Notice of the Office communication was sent electronically on above-indicated "Notification Date" to the following e-mail address(es):

jvick@sheridanross.com

Supplemental Notice of Allowability

Application No. 13/303,417	Applicant(s) TZANNES, M	IARCOS C.
Examiner PHUONG PHU	Art Unit 2634	AIA (First Inventor to File) Status No

,	PHUON	G PHU	2634	No No				
The MAILING DATE of this communication appearable communication appe	(OR REM/ or other a GHTS. Th	AINS) CLOSED in this app opropriate communication in application is subject to	lication. If not i will be mailed i	included n due course. THIS				
 This communication is responsive to <u>the IDS and the amend</u> A declaration(s)/affidavit(s) under 37 CFR 1.130(b) was/ 								
	An election was made by the applicant in response to a restriction requirement set forth during the interview on; the restriction requirement and election have been incorporated into this action.							
 The allowed claim(s) is/are <u>21-50</u>. As a result of the allowed Highway program at a participating intellectual property offic http://www.uspto.gov/patents/init_events/pph/index.jsp or set 	e for the c	corresponding application.	For more inform					
4. Acknowledgment is made of a claim for foreign priority under	r 35 U.S.C	C. § 119(a)-(d) or (f).						
Certified copies:								
a) ☐ All b) ☐ Some *c) ☐ None of the:								
1. Certified copies of the priority documents have	been rece	eived.						
2. Certified copies of the priority documents have	been rece	eived in Application No						
3. Copies of the certified copies of the priority doc	cuments h	ave been received in this n	 ational stage a	pplication from the				
International Bureau (PCT Rule 17.2(a)).			_					
* Certified copies not received:								
Applicant has THREE MONTHS FROM THE "MAILING DATE" of noted below. Failure to timely comply will result in ABANDONMI THIS THREE-MONTH PERIOD IS NOT EXTENDABLE.			omplying with t	the requirements				
5. CORRECTED DRAWINGS (as "replacement sheets") must	be submi	tted.						
including changes required by the attached Examiner's Paper No./Mail Date	s Amendm	ent / Comment or in the Of	fice action of					
Identifying indicia such as the application number (see 37 CFR 1, each sheet. Replacement sheet(s) should be labeled as such in the				not the back) of				
 DEPOSIT OF and/or INFORMATION about the deposit of BI attached Examiner's comment regarding REQUIREMENT FO 				ie				
Attachment(s)								
1. Notice of References Cited (PTO-892)		5. Examiner's Amenda	nent/Comment					
2. 🛛 Information Disclosure Statements (PTO/SB/08),		6. Examiner's Stateme	nt of Reasons	for Allowance				
Paper No./Mail Date 3. ☐ Examiner's Comment Regarding Requirement for Deposit		7. 🛛 Other <u><i>PTOL-271</i></u> .						
of Biological Material								
4. ☐ Interview Summary (PTO-413), Paper No./Mail Date								
/PHUONG PHU/								
Primary Examiner, Art Unit 2634								

U.S. Patent and Trademark Office PTOL-37 (Rev. 08-13)

Notice of Allowability

Part of Paper No./Mail Date 20140328

	Application No.	Applicant(s)				
Decrease to Dule 240 Communication	13/303,417	TZANNES, MARCOS C.				
Response to Rule 312 Communication	Examiner	Art Unit				
	PHUONG PHU	2634				
The MAILING DATE of this communication appears on the cover sheet with the correspondence address –						
 1. ☐ The amendment filed on <u>02/25/14</u> under 37 CFR 1.312 h a) ☐ entered. 	as been considered, and has been:					
b) an entered as directed to matters of form not affecting	the scope of the invention.					
c) disapproved because the amendment was filed after the payment of the issue fee. Any amendment filed after the date the issue fee is paid must be accompanied by a petition under 37 CFR 1.313(c)(1) and the required fee to withdraw the application from issue.						
d) disapproved. See explanation below.						
e) entered in part. See explanation below.						
	/PHUONG PHU/ Primary Examiner, Art Uni	t 2634				

U.S. Patent and Trademark Office PTOL-271 (Rev. 04-01)

Reponse to Rule 312 Communication

Part of Paper No. 20140328

Sub	Substitute for form 1449A/PTO INFORMATION DISCLOSURE STATEMENT BY APPLICANT			Complete if Known		
18.1				Application Number	13/303,417	
				Filing Date	November 23, 2011	
S				First Named Inventor	Marcos C. Tzannes	
				Art Unit	2634	
				Examiner Name	PHU, Phuong M.	
Sheet	1	of	1	Attorney Docket Number	6936-47-CON-5	

	U.S. PATENT DOCUMENTS				
Examiner Initials*	Cite No. ¹	Document Number Number-kind Code ^{2 (If known)}	Publication Date MM-DD-YYYY	Name of Patentee of Applicant of Cited Document	Pages, Columns, Lines, Where Relevant Passages or Relevant Figures Appear
	1	6259685	07-10-2001	Rinne et al.	
	2	2003/0128673	07-10-2003	Lee et al.	

UNPUBLISHED U.S. PATENT DOCUMENTS					
Examiner Initials*	Cite No. ¹	Document Number Number-kind Code ^{2 (If known)}	Filing Date MM-DD-YYYY	Name of Applicant of Cited Document	Pages, Columns, Lines, Where Relevant Passages or Relevant Figures Appear

		FO	REIGN PATENT	DOCUMENTS		
Examiner Initials*	Cite No. ¹	Foreign Patent Document Country Code ³ ; Number ⁴ ; Kind Code ⁵ (if known)	MM-DD-YYYY	Name of Patentee or Applicant of Cited Document	Pages, Columns, Lines, Where Relevant Passages or Relevant Figures Appear	T ⁶
	3	EP 1542488	06-15-2005	ERICSSON TELEFON AB L M		

	OTHER ART (Including Author, Title, Date, Pertinent Pages, etc.)				
Examiner Initials*	Cite No. ¹				
		Official Action for U.S. Patent Application No. 13/718,016, mailed February 4, 2014 (Attorney Ref. No.: 6936-47-CON-DIV-CON-3)			

Examiner Signature	/Phuong Phu/	Date Considered	03/28/2014

*EXAMINER: Initial if reference is considered, whether or not citation is in conformance and not considered. Include copy of this form with next communication to applicant.

ALL REFERENCES CONSIDERED EXCEPT WHERE LINED THROUGH. /P.P./

IN THE UNITED STATES PATENT AND TRADEMARK OFFICE

In Re the Application of: Marcos C. Tzannes) Group Art Unit: 2634
Application No.: 13/303,417	Examiner: PHU, Phuong M.
Filed: November 23, 2011	Confirmation No.: 3035
Atty. File No.: 6936-47-CON-5))

For: SYSTEM AND METHOD FOR SCRAMBLING THE PHASE OF THE CARRIERS IN A MULTICARRIER COMMUNICATIONS SYSTEM

AMENDMENT AFTER ALLOWANCE UNDER 37 C.F.R. 1.312

Commissioner for Patents P.O. Box 1450 Alexandria, VA 22313-1450

Madam:

Applicants submit this Amendment After Allowance pursuant to 37 C.F.R. 1.312 in response to the Notice of Allowance having a mailing date of February 7, 2014. While Applicants believe that no fees are due with the filing of this response, the undersigned hereby authorizes the charge of any fees deemed necessary to Deposit Account No. 19-1970.

An amendment may be entered after the mailing of a Notice of Allowance but prior to payment of the issue fee upon recommendation of the primary Examiner. Therefore, it is respectfully requested that the above-referenced application be amended as follows:

1

Amendments to the Specification begin on page 2 of this paper.

Remarks begin on page 3 of this paper.

O.K. TO ENTER. /P.P./ 03/28/14

Attorney Docket No. 6936-47-CON-5

PART B - FEE(S) TRANSMITTAL

Complete and send this form, together with applicable fee(s), to: Mail Mail Stop ISSUE FEE

o: Mail Mail Stop ISSUE FEE
Commissioner for Patents
P.O. Box 1450
Alexandria, Virginia 22313-1450
or Fax (571)-273-2885

INSTRUCTIONS: This form should be used for transmitting the ISSUE FEE and PUBLICATION FEE (if required). Blocks 1 through 5 should be completed where appropriate. All further correspondence including the Patent, advance orders and notification of maintenance fees will be mailed to the current correspondence address as indicated unless corrected below or directed otherwise in Block 1, by (a) specifying a new correspondence address; and/or (b) indicating a separate "FEE ADDRESS" for maintenance fee notifications.

maintenance fee notifications. Note: A certificate of mailing can only be used for domestic mailings of the Fee(s) Transmittal. This certificate cannot be used for any other accompanying papers. Each additional paper, such as an assignment or formal drawing, must have its own certificate of mailing or transmission. CURRENT CORRESPONDENCE ADDRESS (Note: Use Block 1 for any change of address) Certificate of Mailing or Transmission

I hereby certify that this Fee(s) Transmittal is being deposited with the United States Postal Service with sufficient postage for first class mail in an envelope addressed to the Mail Stop ISSUE FEE address above, or being facsimile transmitted to the USPTO (571) 273-2885, on the date indicated below. 62574 02/07/2014 Jason H. Vick Sheridan Ross, PC Suite # 1200 1560 Broadway Denver, CO 80202 (Signatur (Date APPLICATION NO. FILING DATE FIRST NAMED INVENTOR ATTORNEY DOCKET NO. CONFIRMATION NO. 11/23/2011 Marcos C. Tzannes TITLE OF INVENTION: SYSTEM AND METHOD FOR SCRAMBLING THE PHASE OF THE CARRIERS IN A MULTICARRIER COMMUNICATIONS SYSTEM APPLN. TYPE ENTITY STATUS ISSUE FEE DUE PUBLICATION FEE DUE PREV. PAID ISSUE FEE TOTAL FEE(S) DUE DATE DUE UNDISCOUNTED nonprovisional \$960 \$0 \$0 \$960 05/07/2014 EXAMINER ART UNIT CLASS-SUBCLASS PHU, PHUONG M 2634 375-260000 Change of correspondence address or indication of "Fee Address" (37 CFR 1.363). 2. For printing on the patent front page. list (1) The names of up to 3 registered patent attorneys or agents OR, alternatively, Jason H. Vick ☐ Change of correspondence address (or Change of Correspondence Address form PTO/SB/122) attached. 2 Sheridan Ross, PC (2) The name of a single firm (having as a member a "Fee Address" indication (or "Fee Address" Indication form PTO/SB/47; Rev 03-02 or more recent) attached. Use of a Customer Number is required. registered attorney or agent) and the names of up to 2 registered patent attorneys or agents. If no name is listed, no name will be printed. 3. ASSIGNEE NAME AND RESIDENCE DATA TO BE PRINTED ON THE PATENT (print or type) PLEASE NOTE: Unless an assignee is identified below, no assignee data will appear on the patent. If an assignee is identified below, the document has been filed for recordation as set forth in 37 CFR 3.11. Completion of this form is NOT a substitute for filing an assignment. (B) RESIDENCE: (CITY and STATE OR COUNTRY) (A) NAME OF ASSIGNEE TQ DELTA, LLC AUSTIN, TEXAS Please check the appropriate assignee category or categories (will not be printed on the patent): 🔲 Individual 🗹 Corporation or other private group entity 🖵 Government 4a. The following fee(s) are submitted: 4b. Payment of Fee(s): (Please first reapply any previously paid issue fee shown above) Issue Fee A check is enclosed. Publication Fee (No small entity discount permitted) Payment by credit card. Form PTO-2038 is attached. Advance Order - # of Copies The Director is hereby authorized to charge the required fee(s), any deficiency, or credits any overpayment, to Deposit Account Number 19-1970 (enclose an extra copy of this form). 5. Change in Entity Status (from status indicated above) Applicant certifying micro entity status. See 37 CFR 1.29 NOTE: Absent a valid certification of Micro Entity Status (see forms PTO/SB/15A and 15B), issue fee payment in the micro entity amount will not be accepted at the risk of application abandonment. NOTE: If the application was previously under micro entity status, checking this box will be taken to be a notification of loss of entitlement to micro entity status. ☐ Applicant asserting small entity status. See 37 CFR 1.27 Applicant changing to regular undiscounted fee status. NOTE: Checking this box will be taken to be a notification of loss of entitlement to small or micro entity status, as applicable, NOTE: This form must be signed in accordance with 37 CFR 1.31 and 1.33. See 37 CFR 1.4 for signature requirements and certifications 25 Fel 14 Authorized Signature Typed or printed name Jason H. Vick Registration No. 45,285

Page 2 of 3

PTOL-85 Part B (10-13) Approved for use through 10/31/2013,

OMB 0651-0033

U.S. Patent and Trademark Office; U.S. DEPARTMENT OF COMMERCE

IN THE UNITED STATES PATENT AND TRADEMARK OFFICE

In Re the Application of: Marcos C. Tzannes	Group Art Unit: 2634
Application No.: 13/303,417	Examiner: PHU, Phuong M.
Filed: November 23, 2011	Confirmation No.: 3035
Atty. File No.: 6936-47-CON-5))

For: SYSTEM AND METHOD FOR SCRAMBLING THE PHASE OF THE CARRIERS IN A MULTICARRIER COMMUNICATIONS SYSTEM

COMMENTS ON STATEMENT OF REASONS FOR ALLOWANCE

Commissioner for Patents P.O. Box 1450 Alexandria, VA 22313

Madam:

Applicant submits this Comments on Statement of Reasons for Allowance to address further the Notice of Allowability ("Notice") having a mailing date of February 7, 2014.

In the Notice, the Examiner's stated reasons for allowance were that:

Claims 21-50 are allowed.

The following is an examiner's statement of reasons for allowance:

-Regarding independent claim 21, none of prior art of record teaches or suggests, in a multicarrier modulation system including a first transceiver in communication with a second transceiver using a transmission signal having a plurality of carrier signals for modulating a plurality of data bits, each carrier signal having a phase characteristic associated with at least one bit of the plurality of data bits, a method for scrambling the phase characteristics of the carrier signals comprising: associating a carrier signal with a value determined independently of any bit of the plurality of data bits carried by the carrier signal, the value associated with the carrier signal determined by a pseudo-random number generator; determining a phase shift for the carrier signal at least based on the value associated with the carrier signal; modulating at least one bit of the plurality of data bits on the carrier signal; modulating the at least one bit on a second carrier signal of the plurality of carrier signals.

1

Attorney Docket No.: 6936-47-CON-5

-Regarding independent claim 35, none of prior art of record teaches or suggests a multicarrier modulation system including a first transceiver in communication with a second transceiver using a transmission signal having a plurality of carrier signals for modulating a plurality of data bits, each carrier signal having a phase characteristic associated with the input bit stream, the first transceiver operable to: associate a carrier signal with a value determined independently of any bit of the plurality of data bits carried by the carrier signal, the value associated with the carrier signal determined by a pseudo-random number generator; determine a phase shift for the carrier signal at least based on the value associated with the carrier signal; modulate at least one bit of the plurality of data bits on the carrier signal; and modulate the at least one bit on a second carrier signal of the plurality of carrier signals.

Based on the Notice, the patentability of all other independent and dependent claims is assumed to be based upon the elements as set forth in such claims and that such claims meet all criteria for patentability under §101, §102, §103 and §112.

As is clear from MPEP 1302.14,

"The statement [of reasons for allowance] is not intended to necessarily state all the reasons for allowance or all the details why claims are allowed and should not be written to specifically or impliedly state that all the reasons for allowance are set forth."

While the stated Reasons for Allowance may be a stated reason for allowing some independent claims, Applicant submits that some independent claims have a different reason for allowance and that some independent claims have other reasons for allowance.

Specifically, the prior art fails to teach the specific combination of features as recited in the independent claims 21 and 35.

2

Although the Applicant believes that no fees are due for filing this Comments on Statement of Reasons for Allowance, please charge any fees deemed necessary to Deposit Account No. 19-1970.

Respectfully submitted,

SHERIDAN ROSS P.C.

Date: 25 Rf 19

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IN THE UNITED STATES PATENT AND TRADEMARK OFFICE

In Re the Application of: Marcos C. Tzannes) Group Art Unit: 2634
Application No.: 13/303,417) Examiner: PHU, Phuong M.
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For: SYSTEM AND METHOD FOR SCRAMBLING THE PHASE OF THE CARRIERS IN A MULTICARRIER COMMUNICATIONS SYSTEM

AMENDMENT AFTER ALLOWANCE UNDER 37 C.F.R. 1.312

Commissioner for Patents P.O. Box 1450 Alexandria, VA 22313-1450

Madam:

Applicants submit this Amendment After Allowance pursuant to 37 C.F.R. 1.312 in response to the Notice of Allowance having a mailing date of February 7, 2014. While Applicants believe that no fees are due with the filing of this response, the undersigned hereby authorizes the charge of any fees deemed necessary to Deposit Account No. 19-1970.

An amendment may be entered after the mailing of a Notice of Allowance but prior to payment of the issue fee upon recommendation of the primary Examiner. Therefore, it is respectfully requested that the above-referenced application be amended as follows:

Amendments to the Specification begin on page 2 of this paper.

Remarks begin on page 3 of this paper.

AMENDMENTS TO THE SPECIFICATION

Please amend the first paragraph of the application beneath the heading "RELATED APPLICATION":

This application is a Continuation of U.S. Patent Application No. 12/783,725, filed May 20, 2010, now U.S. Patent No. 8,090,008, which is a Continuation of U.S. Patent Application No. 12/255,713, filed October 22, 2008, now U.S. Patent No. 7,769,104, which is a Continuation of U.S. Patent Application No. 11/863,581, filed September 28, 2007, now U.S. Patent No. 7,471,721, which is a Continuation of U.S. Application No. 11/211,535, filed August 26, 2005, now U.S. Patent No. 7,292,627, which is a Continuation of U.S. Patent Application No. 09/710,310, filed November 9, 2000, now U.S. Patent No. 6,961,369, which claims the benefit of the filing date of copending U.S. Provisional Application, Serial No. 60/164,134, filed November 9, 1999, entitled "A Method For Randomizing The Phase Of The Carriers In A Multicarrier Communications System To Reduce The Peak To Average Power Ratio Of The Transmitted Signal," each of which are incorporated herein by reference in their entirety.

REMARKS

By this amendment, the Related Application has been updated. No new matter is believed to be introduced by this amendment.

The Commissioner is hereby authorized to charge to deposit account number 19-1970 any fees under 37 CFR § 1.16 and 1.17 that may be required by this paper and to credit any overpayment to that Account. If any extension of time is required in connection with the filing of this paper and has not been separately requested, such extension is hereby petitioned.

Respectfully submitted,

SHERIDAN ROSS P.C.

Date: 25 Feb 14

Ву: ___

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IN THE UNITED STATES PATENT AND TRADEMARK OFFICE

In Re the Application of:) Group Art Unit: 2634
Marcos C. Tzannes	Confirmation No.: 3035
Serial No.: 13/303,417	Examiner: PHU, Phuong M.
Filed: November 23, 2011	
Atty. File No.: 6936-47-CON-5	SUPPLEMENTAL INFORMATION DISCLOSURE
Entitled: "SYSTEM AND METHOD FOR SCRAMBLING THE PHASE OF THE CARRIERS IN A MULTICARRIER COMMUNICATIONS SYSTEM") STATEMENT) Electronically Submitted)
Commissioner for Patents	
P.O. Box 1450 Alexandria, VA 22313-1450	
Dear Sir: The references cited on attached Form PTC	0-1449 are being called to the attention
of the Examiner. Copies of the cited non-patent and/or foreign Copies of the cited U.S. patents and/or pater Copies of the cited U.S. patents/unpublished publications are not enclosed in accordance with 37 Copies of the cited references are not enclosed in accordance with 37 rademark Office in prior application Serial No. which is relied upon for an earlier filing date under To the best of applicants' belief, the pertinate are believed to be summarized in the attached Enfigures, although applicants do not necessarily vouc Examiner's attention is drawn to the following Serial No filed filed	n references are enclosed herewith. In applications application application are considered applications. In applications are enclosed herewith. In application application application are application are application are application are are application
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pertinence of any reference without the benefit of hindsight, nor should an inference be drawn as to the pertinence of the references based on the order in which they are presented. Submission of this statement should not be taken as an indication that a search has been conducted, or that no better art exists.

It is respectfully requested that the cited information be expressly considered during the prosecution of this application and the references made of record therein.

FEES 37 CFR 1.97(b): No fee is believed due in connection with this submission, because the information disclosure statement submitted herewith is satisfied by one of the following conditions ("X" indicates satisfaction): Within three months of the filing date of a national application other than a continued prosecution application under 37 CFR 1.53(d), or Within three months of the date of entry into the national stage of an international application as set forth in 37 CFR 1.491 or Before the mailing date of a first Office Action on the merits, or Before the mailing of a first Office action after the filing of a request for continued examination under 37 CFR 1.114. Although no fee is believed due, if any fee is deemed due in connection with this submission, please charge such fee to Deposit Account 19-1970. 37 CFR 1.97(c): The information disclosure statement transmitted herewith is being filed after all the above conditions (37 CFR 1.97(b)), but before the mailing date of one of the following conditions: (1) a final action under 37 C.F.R. 1.113 or (2) a notice of allowance under 37 C.F.R. 1.311, or (3) an action that otherwise closes prosecution in the application. This Information Disclosure Statement is accompanied by: A Certification (below) as specified by 37 C.F.R. 1.97(e). Although no fee is believed due, if any fee is deemed due in connection with this submission, please charge such fee to Deposit Account 19-1970. Please charge Deposit Account 19-1970 in the amount of \$180.00 for the fee set forth in 37 C.F.R. 1.17(p) for submission of an information disclosure statement. Please credit any overpayment or charge any underpayment to Deposit Account 19-1970. \boxtimes 37 CFR 1.97(d): This Information Disclosure Statement is being submitted after the period specified in 37 CFR 1.97(c). M This information Disclosure Statement includes a Certification (below) as specified by 37 C.F.R. 1.97(e) AND Applicants hereby requests consideration of the reference(s) disclosed herein. Please charge Deposit Account 19-1970 in the amount of \$180.00 under 37 C.F.R. 1.17(p). Please credit any overpayment or charge any underpayment to Deposit Account 19-1970. Election to pay the fee should not be taken as an indication that applicant(s) cannot execute a

	Certification (37 C.F.R. 1.97(e)) (Applicable only if checked)
	The undersigned certifies that: Each item of information contained in this information disclosure statement was first cited in any communication from a foreign patent office in a counterpart foreign application not more than three months prior to the filing of this statement. 37 C.F.R. 1.97(e)(1). A copy of the communication from the foreign patent office is enclosed.
	OR
	No item of information contained in this information disclosure statement was cited in a communication from a foreign patent office in a counterpart foreign application, and, to the knowledge of the undersigned after making reasonable inquiry, no item of information contained in this Information Disclosure Statement was known to any individual designated in 37 C.F.R. 1.56(c) more than three months prior to the filing of this statement. 37 C.F.R. 1.97(e)(2).
	Respectfully submitted,
	SHERIDAN ROSS P.C.
Date:	By: Jason H. Vick Registration No. 45,285 1560 Broadway, Suite 1200 Denver, Colorado 80202-5141 (303) 863-9700

Sub	stitute for form	1449A/PTO		Comp	olete if Known
		FIGNI BIGG	LOGUE	Application Number	13/303,417
		TION DISC		Filing Date	November 23, 2011
S	ATEME	NT BY API	PLICANT	First Named Inventor	Marcos C. Tzannes
				Art Unit	2634
				Examiner Name	PHU, Phuong M.
Sheet	1	of	1	Attorney Docket Number	6936-47-CON-5

U.S. PATENT DOCUMENTS						
Examiner Initials*	Cite No.1	Document Number Number-kind Code ^{2 (If known)}	Publication Date MM-DD-YYYY	Name of Patentee of Applicant of Cited Document	Pages, Columns, Lines, Where Relevant Passages or Relevant Figures Appear	
	1	6259685	07-10-2001	Rinne et al.		
	2	2003/0128673	07-10-2003	Lee et al.		

UNPUBLISHED U.S. PATENT DOCUMENTS						
Examiner Initials*	Cite No. ¹	Document Number Number-kind Code ^{2 (If known)}	Filing Date MM-DD-YYYY	Name of Applicant of Cited Document	Pages, Columns, Lines, Where Relevant Passages or Relevant Figures Appear	

FOREIGN PATENT DOCUMENTS								
Examiner Initials*	Cite No. ¹	Foreign Patent Document Country Code ³ ; Number ⁴ ; Kind Code ⁵ (if known)	MM-DD-YYYY	Name of Patentee or Applicant of Cited Document	Pages, Columns, Lines, Where Relevant Passages or Relevant Figures Appear	T⁵		
	3	EP 1542488	06-15-2005	ERICSSON TELEFON AB L M				

OTHER ART (Including Author, Title, Date, Pertinent Pages, etc.)							
Examiner Initials*	Cite No. ¹						
		Official Action for U.S. Patent Application No. 13/718,016, mailed February 4, 2014 (Attorney Ref. No.: 6936-47-CON-DIV-CON-3)					

Examiner	Date	
Signature	Considered	

*EXAMINER: Initial if reference is considered, whether or not citation is in conformance and not considered. Include copy of this form with next communication to applicant.





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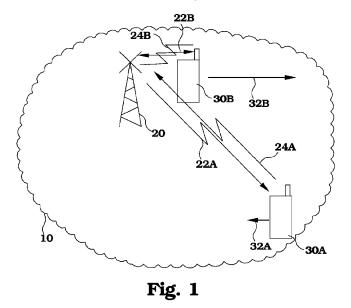
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(54) Method and apparatus for allocating a pilot signal adapted to the channel characteristics

(57) A set of different pilot structures are designed for use in different environments and/or different user behaviours that are expected to occur in a cell. The radio conditions for a user are estimated. Each user is then assigned an area (108A-E) in resource space for its communication, which has a suitable pilot configuration. In one embodiment, the entire resource space is provid-

ed with different pilot structures in different parts (110A-D) in advance and allocation of resources to the users are then performed in order to match estimated radio conditions to the provided pilot structure. In another embodiment, allocation is performed first, and then the actual pilot structure is adapted within the allocated resource space area to suit the environmental conditions.



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Description

TECHNICAL FIELD

[0001] The present invention relates generally to wireless multi-carrier communications systems and in particular to resource allocation and pilot signals of such systems.

BACKGROUND

[0002] In a cellular multi-user, multi-carrier wireless communications system, the base station must accommodate many users that each experiences different channel characteristics due to fading in time and frequency. Furthermore, different users travel at different speeds and thus experience different Doppler shifts. In most wireless systems, e.g. GSM (Global System for Mobile communications), WCDMA (Wideband Code Division Multiple Access), WLAN (Wireless Local Area Network), special well known training sequences or pilot signals are transmitted so that the receiver can estimate the channel parameters sufficiently well for detection of the data signal. Several methods exist to do this, some use user specific pilots and some use common pilots or combinations. Some pilots are code spread and overlaid with user data, others have dedicated time-frequency slots when pilots are transmitted. In any case, some part of the available radio resources must be allocated for pilots resulting in overhead that cannot be used for data. [0003] Today, there are a few multi-carrier systems in use. However, they are not particularly designed for the difficult, ever changing, hard-to-predict multi-user environments that are envisioned for future wireless systems.

[0004] For example, the systems for DVB/DAB (Digital Video Broadcasting/Digital Audio Broadcasting) are broadcast systems that cannot take into account the need for individual users. Such systems must design their pilot structure according to the worst-case scenario so that detection becomes possible even under the worst possible conditions. Such a pilot structure gives rise to a substantial pilot overhead, and is indeed necessary in these worst-case scenarios. However, whenever the situation is better than the worst case, which typically is the case most of the time, the pilot structure is unnecessarily extensive, giving an unnecessary pilot overhead for most users. The pilot overhead can indeed be substantial. This reduces data capacity in the own cell and furthermore increases the interference to the neighbouring cells (so called "pilot pollution").

[0005] Another example of a multi-carrier system is WLAN (i.e. IEEE 802.11a, IEEE 802.11g). Such a system is designed for a limited geographical area in which the users are stationary or slowly moving. The design is not intended for conditions in which the user is moving quickly or for handling mobility in a multi-cellular environment.

SUMMARY

[0006] The main problems with existing solutions are that pilot structures are either not at all suitable for considerably changing radio conditions or that they are designed for worst cases which in turn results in vast pilot overhead and "pilot pollution".

[0007] An objective of the present invention is to provide methods and devices for multi-user multi-carrier wireless communications system, which are capable to provide all users with sufficient pilots without causing unnecessary pilot overhead and pilot pollution. A further objective of the present invention is to provide such methods and devices, which are easy to implement within present and planned wireless systems.

[0008] The above objectives are achieved by methods and devices according to the enclosed patent claims. In general words, a set of different pilot structures are designed for use in different environments and/or different general radio characteristics that are expected to occur in the cell. The radio conditions for a user are estimated, either from direct measurements or from knowledge about the cell characteristics, possibly combined with position information. Each user is then assigned an area in resource space for its communication, which has a suitable pilot configuration. In one embodiment, the entire resource space is provided with different pilot structures in different parts in advance and allocation of resources to the users are then performed in order to match estimated radio conditions to the provided pilot structure. In another embodiment, allocation is performed first, and then the actual pilot structure is adapted within the allocated resource space area to suit the environmental conditions. For best performance, depending on such things as frequency selectivity, time selectivity (e.g. time dispersion and Doppler shift), and path loss the amount of pilot energy should be adapted and the 'distance' between pilots in the time-frequency domain needs to be changed.

[0009] The radio resource space can have different dimensions. In multi-carrier systems, frequency is one dimension. Other dimensions that could be utilised within the present invention are time, code, antenna and/or spatial dimensions. One or several of these dimensions span the radio resource space, in which the present invention is applied.

[0010] By adapting the pilot structure to the environment or set of environments likely to occur in the cell and allocating these pilots to the users most likely to benefit from them, an overall efficiency is achieved. The amount of pilot overhead is then connected to the actual environments being accommodated. Difficult environments require more overhead than simpler ones and hence pilot pollution is reduced on the average.

BRIEF DESCRIPTION OF THE DRAWINGS

[0011] The invention, together with further objects

4

and advantages thereof, may best be understood by making reference to the following description taken together with the accompanying drawings, in which:

FIG. 1 is a schematic illustration of a multi-user wireless communication system;

FIGS. 2A and 2B are illustrations of pilot structures in time-frequency space, and the allocation of different users to subspaces;

FIG. 3A illustrates a radio resource space having a code dimension:

FIG. 3B is an illustration of a pilot structure in the frequency-code sub-space:

FIG. 4 is a flow diagram illustrating an embodiment of a method according to the present invention;

FIGS. 5A, 5B and 6 are diagrams illustrating pilot structures in time-frequency space, and the allocation of different users to subspaces according to embodiments of the present invention;

FIGS. 7A and 7B are flow diagrams illustrating other embodiments of a method according to the present invention:

FIG. 8 is a flow diagram illustrating a part of a further embodiment of a method according to the present invention:

FIGS. 9A to 9C are block diagrams of downlink radio management devices of network nodes according to embodiments of the present invention;

FIG. 10 is a block diagram of uplink radio management devices of network nodes according to embodiments of the present invention;

FIG. 11 is a diagram illustrating pilot structures in time-frequency space having different intensities, and the allocation of different users to subspaces according to an embodiment of the present invention; and

FIGS. 12 and 13 are diagrams illustrating limited data descriptions of regular pilot structure.

DETAILED DESCRIPTION

[0012] In the following description, OFDM (Orthogonal Frequency Division Multiplexing) systems are used for exemplifying the present invention. However, the present invention can also be applied to other multi-carrier wireless communications systems.

[0013] Fig. 1 illustrates a multi-user multi-carrier wireless communications system 10, in this particular embodiment intended to be an OFDM system. Nonexclusive examples of other communications systems, in which the present invention is advantageously applicable, are EFDMA (Interleaved Frequency Division Multiple Access) systems, non-orthogonal or bi-orthogonal multi-carrier systems. A base station or access point 20 communicates with two mobile stations or user equipments 30A, 30B. There is a downlink connection 22A between the access point 20 and the user equipment 30A and an uplink connection 24A between the same

nodes. Likewise, there is a downlink connection 22B between the access point 20 and the user equipment 30B and an uplink connection 24B between the same nodes. User equipment 30A is located at a relatively large distance from the access point 20, but the speed 32A (illustrated as an arrow) of the user equipment 30A is small. User equipment 30b is located closer to the access point 20, but has a high speed 32B (also illustrated as an arrow). The user equipment 30A may have a relatively high need for repetitive pilots in the frequency dimension, since the propagation conditions for the different carriers may differ considerably over the bandwidth in case of multi-path propagation with large delay spread. However, the radio conditions are probably quite slowly varying with time due to the small speed of user equipment 30A. The user equipment 30B is close to the access point, and a pilot on one frequency can probably be used for channel estimations for many neighbouring carriers. However, the radio conditions are probably changing rapidly in time, whereby frequent pilots in time dimension are required.

[0014] Fig. 2A is a diagram of a time-frequency space. This can represent a limited portion of the entire available radio resource space 100 in these two dimensions. Data is transmitted in quantities limited in time and frequency. These data quantities correspond to the small squares 104 in the diagram. Selected ones 102 of these data quantities contain pilot data and are illustrated in the diagram with hatching. The pilot structure is in this embodiment dispersed over the time-frequency space relatively uniformly. With this distribution, one data quantity out of 11 is occupied by pilot data.

[0015] The useful data transmission rate is thereby reduced by 1/11. The users of the user equipments 30A and 30B (Fig. 1) have allocated radio resources within the available radio resource space 100. User equipment 30A is allocated the resource sub-space indicated by 108A, while user equipment 30B is allocated the resource sub-space indicated by 108B. Both users are experiencing the same pilot density and the uniform distribution between the frequency and time dimensions.

[0016] User 30B moves fast. The time between two consecutive pilot messages in time dimension is 11 time slots, and even if information from neighbouring frequencies are used for channel estimation in the meantime, at least 4 time slots will pass between two consecutive updates. The speed of user 30B is so high that this pilot structure is not sufficient for an acceptable quality of service.

[0017] However, arranging the pilot structure as in Fig. 2B will change the situation. Here, there is a new update in time dimension every second time slot, which well supports the fast moving user equipment. Despite this increased density in time direction, the total amount of pilot data quantities is reduced somewhat. Now only one data quantity out of 12 comprises a pilot. The overhead has decreased from 1/11 to 1/12 (about 9%).

[0018] However, user equipment 30A now achieves

problems. This user equipment 30A moves slowly and is of limited use of the frequent updating in time. However, it has need for more closely located pilots in frequency dimension instead. The pilot structure of Fig. 2B becomes very unsuitable for user equipment 30A.

[0019] So far, only two dimensions, time and frequency, have been discussed. Fig. 3A illustrates a radio resource space in three dimensions, time, frequency and code. In such a system, each data quantity will instead correspond to a small cube 104. Generalisation can be performed to higher order spaces, comprising e.g. antenna or space dimensions. In general, any radio resource space in at least two dimensions, of which one is frequency, can be used with the present invention.

[0020] Fig. 3B illustrates a pilot pattern in a frequency-code space for a specified time. In this example 16 different codes are available and also 16 different frequencies. The illustrated pilot pattern leads to that the pilots are transmitted on all frequencies during the specified time duration, however, spread out in the code dimension. One code in each frequency is occupied by a pilot, whereas the remaining 15 codes are used for data transmission.

[0021] As mentioned briefly above, more generally the antenna or spatial dimensions could also be part of the resource space. One example is that different frequency bands are allocated to different beams of a multi-sector or fixed beam site. In this case, the spatial dimension is part of the description since different pilot patterns may be deployed for the different beams that overlap in the spatial domain. With the grouping of resources in terms of antenna sectors or beams the pilots allocated to different users can change dynamically when the user for example moves between sectors and the sectors have different frequency bands allocated to them. In such cases, antenna or spatial dimension can also be used as additional dimensions in a total resource space.

[0022] The flow diagram of Fig. 4 illustrates the main steps of an embodiment of a method according to the present invention. The procedure starts in step 200. In step 202, a number of pilot configurations are provided, which are believed to suit different radio conditions appearing in the cell in question. At least two such pilot configurations are available, i.e. they can be handled by both sides of the transmission connection. The transmitter manages the sending of pilots according to this configurations and the receiver is capable of performing channel estimation based on the at least two pilot configurations. In step 204, an estimation of the radio conditions at the receiver is obtained. This estimation can be provided in many different ways. The actual radio conditions can be measured and evaluated. Another possibility is to assume an estimate from knowledge about the characteristics in the cell and possibly based on e.g. location and/or speed of the receiver relative the transmitter.

[0023] In step 206, a user is allocated resources in

resource space, which have a pilot configuration that is matched to the estimated radio conditions. This matching can be performed in different manners, described more in detail further below. The procedure stops in step 299. Anyone skilled in the art realises that step 202 preferably is performed once, and the provided pilot structures can then be used for any future allocation of users, or re-allocation of existing users.

[0024] A few examples, using OFDM as an example system, will be used to visualise the effect of the present invention. The basic setup in Fig. 5A is assumed as follows. During a certain time period and seen over all frequency resources, the available radio resources constitute a grid of basic resources that can be used for data, control signalling or pilot signals or other signals as discussed earlier. The resolution in frequency dimension is one OFDM carrier and in time it is one OFDM symbol. Pilot symbols are as above depicted with hatched boxes

[0025] The transmitter side, in this example assumed to be the base station, determines a number of different pilot patterns and assigns these pilot patterns to different parts of the entire radio resource space. The pilot patterns may for example be periodically recurring with some period or pseudo-randomly designed. This means that different parts of the radio resource space have a denser or at least differing pilot pattern than other parts. Each pilot pattern is intended to accommodate users experiencing different channel characteristics.

[0026] This is illustrated in Fig. 5A. The entire radio resource space illustrated is divided into four rectangular parts, 110A-D. The resource space part 110A has a pilot pattern, having a dense occurrence in time dimension (every second OFDM symbol at certain carriers), but a more dispersed behaviour in the frequency dimension (only every sixth OFDM carrier). The resource space part 110B has a very diluted pilot pattern, having only one pilot in 36 resource units, evenly spread in time and frequency dimensions. The resource space part 110C is the opposite of part 110A, with a dense pilot pattern in frequency dimension, but sparse in time dimension. Finally, resource space part 110D has a very dense pilot structure in both dimensions, comprising a pilot symbol in every fourth resource unit.

[0027] According to one embodiment of the invention, the users are now allocated to the different parts of the radio resource space dependent on their estimated radio conditions. In other words, whenever a certain user has certain demands, the user is assigned resources in the resource space where pilots with the appropriate density can be utilised for channel estimation. In the situation in Fig. 5A, there are pilot structures suitable for typically four combinations of Doppler and delay spread. In part 110A, the pilot structure is intended for a large Doppler and low delay spread. In part 110B, the pilot structure is intended for a low Doppler and low delay spread. In part 110C, the pilot structure is intended for a low Doppler and high delay spread. In part 110D, the

pilot structure is intended for a high Doppler and high delay spread.

[0028] A first user, having radio conditions demanding a high density of pilots in both dimensions is allocated to the resource sub-space 108A within the part 110D. A second user, only having need for dense pilot in the time dimension is allocated resources in a resource subspace 108B within the part 110A. A third user with very favourable radio conditions is allocated to a resource sub-space 108C in part 110B. Finally, two more users, having high demands on pilot density are given resources in two sub-spaces 108D and 108E, respectively in part 110D. One realises that each user has achieved a pilot pattern that is suited to its individual needs. It is beneficial, e.g. to assign resources for mobiles with certain fast varying channel or Doppler conditions in the dense parts of the pilot pattern and users with more slowly varying conditions in the less dense parts.

[0029] Note that the base station does not need to transmit all pilots at all times. Only pilots that in fact can be utilised by any user needs to be transmitted. If a pilot resource at time of transmission cannot be utilised by any data symbol that some user need to detect with the help of said pilot, then the pilot need not be transmitted. In such a way, the overall pilot pollution is reduced, and so is the average transmission power.

[0030] In Fig. 5B, a further embodiment of the present invention is illustrated. Assume the same situation as was present in Fig. 5A. Three users are occupying all resources in the densest part 110D. If yet another user with need for a very dense pilot configuration appears, the pre-defined pilot configuration plan of Fig. 5A becomes insufficient. However, the new user can be allocated to a free resource sub-space 108F, preferably in connection with the part 110D. This sub-space 108F had originally a pilot pattern according to part 110C, but when allocating the user, the pilot pattern is adjusted to match the demands put by the new user. In such a way, the original pre-determined division into different parts in the resource space can be adapted to the actual need. However, if a good initial configuration is used, most cases are covered and the frequency of adjustments is

[0031] Now, return to the situation of Fig. 5A. If the user having the allocation of sub-space 108E slows down, the estimated radio conditions change, and the need for pilots is reduced. The user can then be reallocated to another sub-space of the resource space, having a more suitable pilot configuration for the new estimated radio conditions, e.g. to part 110C. An alternative is to keep the allocated sub-space but instead change the pilot pattern to a more suitable one for the new conditions.

[0032] The ideas of adjusting or adapting the pilot configuration when needed can also be brought to the extreme end, where no pilot pattern at all is preconfigured for the different parts of the resource space. Instead, there is always an adjustment of pilot pattern for

all users. This is schematically illustrated in Fig. 6. Here, a first user was assigned a sub-space 108A, without associated pre-defined pilot pattern. The pilot pattern was then adjusted according to the actual needs as concluded from the estimated radio conditions. In this case a dense pattern was selected. A second user was allocated to sub-space 108B and subsequently, a suitable pilot pattern was selected for this sub-space. In such a way. all the sub-spaces 108A-F were associated with pilot configurations suitable for each individual need. Subspaces not allocated to any user do not comprise any pilots in such an approach. A user with certain estimated properties is thus allocated to use certain resources and the pilot pattern is designed accordingly. The result is the same as the previous embodiments, pilot patterns and user characteristics are matched.

[0033] The above embodiments can also be expressed in flow diagrams. In Fig. 7A, a flow diagram corresponding to the situation in Fig. 5A is illustrated. The resource space is in step 203 provided with at least two different pre-determined pilot configurations at different parts of the resource space. Step 204 is unchanged compared to Fig. 4. In step 207, the matching of the radio conditions and pilot structures is performed by selecting a suitable resource space.

[0034] The situation in Fig. 5B is illustrated by the flow diagram of Fig. 7B. Also here, pre-defined pilot configurations are associated with different parts of the resource space in step 203. In step 205, it is determined whether there is any available resources in parts that are suitable for the particular estimated radio conditions for the user to be allocated. If there are resources with suitable pilot structures available, the procedure continues to step 207, as in Fig. 7A. If no resource space with appropriate pilot structure is available, any free resource space is allocated in step 209, however, preferably in the vicinity of the part having a suitable pilot pattern, In step 210, the pilot configuration is adapted within the selected resource sub-space to match the estimated radio conditions

[0035] The embodiment illustrated in Fig. 6 can similarly be illustrated by the part flow diagram of Fig. 8. Here, the step 206 in Fig. 4 is described in more detail. In step 208, an area is selected as a resource sub-space for the user. In step 210, the pilot configuration in the selected area is adapted to the need connected to the estimated radio conditions of the user. Note the similarities between Fig. 7B and Fig. 8.

[0036] The present invention can be implemented for wireless communication between any nodes in a communications system. Such nodes can be e.g. user equipment, mobile station, base station, access point or relay. In the examples below, the most straightforward situation with communication between a base station and a user equipment will be discussed as an example. The scope of the claims should, however, not be affected by this example.

[0037] Multi-carrier communication is typically most

applied in downlink connections. In Fig. 9A, a wireless communications system according to an embodiment of the present invention is illustrated. A base station 20 communicates with a mobile terminal 30 via an uplink 24 and a downlink 22 connection. In the downlink communication, the ideas of the present invention are implemented. The base station 20 comprises a downlink control unit 25, which is enlarged in the lower part of Fig. 9A. The downlink control unit 25 is responsible for allocating resources for communication on the downlink 22 between the base station 20 and the mobile terminal 30 and comprises in turn a pilot manager 26 and a radio condition processor 28. Similarly, the mobile terminal or user equipment 30 also comprises a downlink control unit 35, also enlarged in the lower part of Fig. 9A. The downlink control unit 35 comprises a channel estimator 36 and a measurement unit 38 for radio conditions

[0038] The radio conditions measurement unit 38 measures the actual radio conditions at the user equipment 30. Such measurements can comprise e.g. Doppler shift and signal strength as well as power delay profile, channel impulse response, time and frequency selectivity measurements and interference levels. The results of the measurements are transferred to the radio conditions processor 28 of the base station 20, preferably by the uplink communication link 24. The radio conditions processor 28 evaluates the measured conditions and translates it to estimated radio conditions for the user equipment 30. In other words, the radio conditions processor 28 obtains data associated with estimated radio conditions for the user equipment 30. In a basic version, the estimated radio conditions could e.g. comprise two flags, one indicating low or high Doppler shift and one indicating small or large delay spread. When having a radio resource space in frequency and time dimensions, quantities associated with coherence bandwidth and coherence time, respectively, are of interest. The estimated radio conditions are forwarded to the pilot manager 26, which performs the actual selection and/ or adjustment of resource sub-spaces. The pilot manager 26 thus provides access to the use of the different pilot configurations. When pre-defined pilot patterns are used, the pilot manager selects in which part of the multicarrier space the allocated resource sub-space will be placed. Without pre-defined patterns in different parts of the multi-carrier space, the pilot manager 26 comprises functionalities for selecting a multi-carrier sub-space for allocation and functionalities to adapt the pilot pattern of that selected sub-space according to the estimated radio conditions. When the pilot manager has decided what pilot pattern to apply, the user equipment 30 has to be informed about the selection, in order to be able to perform the right channel estimation upon reception of the data. The pilot manager 26 thus comprises means for transferring suitable data to the channel estimator 36

[0039] In Fig. 9B, another embodiment is illustrated, where the base station 20 has the entire responsibility

for the selection of pilot structure. The downlink control unit 25 here also comprises a position estimator 29. The position estimator 29 provides an estimation of the position of the user equipment 30 and preferably also the velocity. This can be performed in any manner, e.g. according to prior art methods, and is not further discussed here. The position is forwarded to the radio condition processor 28. The radio condition processor 28 has access to knowledge about the different environments within the cell. A cell could e.g. cover a first area having generally slowly moving user equipments, and a second area, were the average speed is considerably higher. The position estimation could reveal the location of the user equipment, i.e. if it is situated in the high- or lowspeed area. From such information, the radio condition processor 28 can conclude what radio conditions that should be assumed for the user equipment. Such estimation then forms the base on which the pilot pattern is selected

[0040] In Fig. 9C, yet another embodiment is illustrated. In this embodiment, the user equipment 30 makes more efforts in the procedure to find suitable pilot structures. The downlink control unit 35 here additionally comprises a radio conditions processor 39. This means that both the measurements and the evaluation of the measurements are performed in the user equipment 30. The estimated radio conditions are reported to the base station 20, e.g. in the form of data representing coherence bandwidth and coherence time, respectively. Alternatively, the radio conditions processor 39 can also select an appropriate pilot pattern and transmit a request to use such a pattern to the base station 20. The base station 20 can in such a case either follow the recommendation or overrule it and make an own decision. [0041] Fig. 10 illustrates one possible configuration for uplink communication. The base station 20 comprises an uplink control unit 45, in turn comprising a radio conditions measurement unit 21, a radio conditions processor 28 and a pilot manager 26. The operations of the units are similar to the ones in the downlink case, but adapted for uplink communication instead, i.e. it is the radio conditions of the received signals from the user equipment 30 that are of importance. The pilot manager 26 decides which pilot pattern that is appropriate to apply, and transmits a request to an uplink control unit 55 in the user equipment 30. In a basic version, the uplink control unit 55 simply applies the proposed pilot pattern on its uplink traffic. The uplink control unit 35 of the base station 20 also comprises a channel estimator 27 in order to be able to detect the data sent on the uplink. This channel estimator 27 is also informed about the pilot structure to use.

[0042] Fig. 11 illustrates yet another embodiment of the present invention, in which one makes use of the possibilities to vary the intensity to reduce pilot pollution. In parts 110A and 110D, all or some of the pilot data is marked to be transmitted with a lower (or zero) intensity. If a user equipment using the pilot signals is close to the

base station, the transmission power does not have to be equally high to obtain a reasonable channel estimation compared with user equipments situated further away from the base station, In such a way, it is also possible to vary the pilot intensity throughout the resource space. Such intensity configurations can as above be performed either in advance or as adjustment procedures.

[0043] The pilot symbols can also be transmitted with different power for different classes of users and depending on path loss. The power levels can either be dynamically varying between zero and a given number P_{max} or be defined in advance. Note that a power level equal to zero is equivalent to no pilots for this slot, enabling the use of this slot for other purposes, such as data. If the power is dynamically varying, the power levels have to be signalled to the receiver for appropriate treatment.

[0044] When there are several possible pilot patterns to use in a system, the receiver has to be informed about which one is actually used. If a numbered set of predetermined pilot patterns are used, the identification number of the pilot pattern is sufficient. However, more elaborate systems can use different pilot patterns for different cells and the numbering of patterns can be difficult to manage. In such a case, a solution is to transfer a complete description of the pilot pattern to be used. For regular pilot patterns, the amount of data that is needed to uniquely define the patterns is quite limited.

[0045] In Fig. 12, a pilot pattern is illustrated within a 30 resource sub-space in frequency and time dimensions. The resource sub-space is reported anyway, and is typically defined by frequency and time "coordinates" and the number of frequency DF and time DT slots that are comprised in the sub-space. The pilot pattern is then easily characterised by only three vectors in the (twodimensional) resource space. A first vector V0 defines the "distance" in frequency and time, respectively, between a well-defined position in the sub-space, e.g. the lower left corner as illustrated in the picture, and any pilot data within the pattern. A second vector V1 defines a "relative distance" between the two closest pilots in the pattern. A third vector V2 defines a "relative distance" between the second closest pilots, that is not aligned with the first vector V1. By knowing only these vectors, the entire pilot pattern can easily be calculated. [0046] Also somewhat more complicated patterns can be fit into a similar model. In Fig. 13, a pattern having two neighbour pilots distributed in pairs over the resource space. In order to describe this pattern, only one extra vector is needed, the relative vector between the two pilots in each pair. Anyone skilled in the art realises that with a very limited number of data, rather complex pilot patterns can easily be defined.

[0047] It will be understood by those skilled in the art that various modifications and changes may be made to the present invention without departure from the scope thereof, which is defined by the appended claims.

Claims

- Method for wireless communication in a multi-user, multi-carrier communications system (10), comprising the steps of:
 - allocating a first resource sub-space (108A-F) of an entire multi-carrier resource space (100) for communication between a first node (20, 30; 30A-B) and a second node (20, 30; 30A-B),

characterised by the further step of:

- obtaining data associated with estimated radio conditions for the second node (20, 30; 30A-B).
- providing access to the use of at least two pilot resource configurations, intended for different estimated node radio conditions,

whereby the first resource sub-space is associated a first pilot resource configuration, being in agreement with pilot need for the estimated radio conditions for the second node (20, 30: 30A-B).

25 2. Method according to claim 1, characterised in that the entire multi-carrier resource space (100) being divided into parts (110A-D) having different pilot resource configurations:

whereby the step of allocating comprises the step of selecting the first resource sub-space (108A-F) in a part having a pilot resource configuration suitable for the estimated radio conditions for the second node (20, 30; 30A-B).

- 35 3. Method according to claim 2, **characterised by** the further steps of:
 - selecting, if no resource space part (110A-D)
 having a pilot resource configuration suitable
 for the estimated radio conditions for the second node (20, 30; 30A-B) is available, an arbitrary first multi-carrier resource sub-space; and
 - adapting the pilot resource configuration within the first multi-carrier resource sub-space to suit the estimated radio conditions for the second node (20, 30; 30A-B).
 - 4. Method according to claim 1, characterised by the step of:
 - selecting the first multi-carrier resource subspace; and
 - adapting the pilot resource configuration within the first multi-carrier resource sub-space to suit the estimated radio conditions for the second node (20, 30; 30A-B) after the step of selecting.
 - 5. Method according to any of the claims 1 to 4, char-

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acterised in that the multi-carrier resource space (100) has a time dimension.

- Method according to any of the claims 1 to 5, characterised in that the multi-carrier resource space (100) has a code dimension.
- Method according to any of the claims 1 to 6, characterised in that the multi-carrier resource space (100) has a spatial dimension.
- Method according to any of the claims 1 to 7, characterised in that the step of obtaining in turn comprises the step of estimating a set of estimated radio conditions.
- Method according to claim 8, characterised in that the set of estimated radio conditions comprises at least Doppler conditions.
- 10. Method according to claim 8 or 9, characterised in that the set of estimated radio conditions comprises at least delay spread conditions.
- 11. Method according to any of the claims 8 to 10, characterised in that the step of estimating is based on position and/or velocity information concerning the second node (20, 30; 30A-B).
- 12. Method according to any of the claims 1 to 11, characterised in that the step of obtaining comprises the step of receiving instructions and/or suggestions about preferred pilot resource configuration.
- 13. Method according to any of the claims 1 to 12, characterised in that the first node is selected from the group of:

user equipment; mobile station; base station; access point; and relay.

14. Method according to any of the claims 1 to 13, characterised in that the second node is selected from the group of:

user equipment; mobile station; base station; access point; and relay.

15. Method according to any of the claims 1 to 14, characterised in that resources of the first resource sub-space are allocated for downlink communication (22; 22A-B).

- 16. Method according to claim 15, characterised in that the step of obtaining data associated with estimated radio conditions for the second node (20, 30; 30A-B) is performed in a base station or access point (20).
- 17. Method according to claim 16, characterised by the further step of transferring data characterising the first pilot resource configuration from the base station or access point (20) to the second node (20, 30; 30A-B).
- **18.** Method according to any of the claims 1 to 12, **characterised in that** resources of the first resource sub-space are allocated for uplink communication (24; 24A-B).
- 19. Method according to claim 18, characterised in that the step of obtaining data associated with estimated radio conditions for the second node (30; 30A-B) is performed in a base station or access point (20), followed by the step of transferring the data associated with estimated radio conditions for the second node (30; 30A-B) to the second node (30; 30A-B).
- 20. Method according to claim 18, characterised in that the step of obtaining data associated with estimated radio conditions for the second node (30; 30A-B) is performed in the second node (30; 30A-B).
- 21. Method according to claim 20, characterised by the further step of transferring data characterising the first pilot resource configuration from the second node (30; 30A-B) to the first node (20, 30; 30A-B).
- 22. Method according to any of the claims 1 to 21, characterised in that retraining from transmitting pilots in areas of the entire multi-carrier resource space (100) not being allocated.
- Method according to any of the claims 1 to 22, characterised in that the wireless communication utilises OFDM.
- 24. Method according to any of the claims 1 to 23, **characterised in that** the available at least two pilot resource configurations comprises different distribution patterns of pilot symbols in the multi-carrier resource space (100).
- **25.** Method according to claim 24, **characterised in that** the available at least two pilot resource configurations further comprises transmission of pilot symbols with differing intensity.
- 26. Arrangement being or comprising a first node (20,

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30; 30A-B) of a multi-user, multi-carrier wireless communications system (10), the first node (20, 30; 30A-B) comprising:

means (25) for allocating a first resource subspace of an entire multi-carrier resource space (100) for communication between the first node (20, 30; 30A-B) and a second node (20, 30; 30A-B).

characterised by

- means (28, 29, 38, 39) for obtaining data associated with estimated radio conditions for the second node (20, 30; 30A-B), and
- means (26) for providing access to the use of at least two pilot resource configurations, intended for different estimated node radio conditions

whereby the first resource sub-space comprises a first pilot resource configuration, being in agreement with pilot need for the estimated radio conditions for the second node (20, 30; 30A-B).

27. Arrangement according to claim 26, characterised in that the entire multi-carrier resource space (100) being divided into parts (110A-D) having different pilot resource configurations;

whereby the means (25) for allocating being arranged for selecting the first resource sub-space in a part having a pilot resource configuration suitable for the estimated radio conditions for the second node (20, 30; 30A-B).

28. Arrangement according to claim 26, characterised in that the first node (20, 30; 30A-B) further comprises:

means for selecting the first multi-carrier resource sub-space; and means for adapting the pilot resource configu-

ration within the first multi-carrier resource subration within the first multi-carrier resource subspace to suit the estimated radio conditions for the second node (20, 30; 30A-B), the means for adapting being connected to an output of the means for selecting.

29. Arrangement according to any of the claims 26 to 28,

characterised in that the first node (20, 30; 30A-B) further comprises:

means for transferring data characterising the first pilot resource configuration from the first node (20, 30; 30A-B) to the second node (20, 30; 30A-B).

30. Arrangement according to any of the claims 26 to 29.

characterised in that the means (28, 29, 38, 39) for obtaining data associated with estimated radio conditions for the second node (20, 30; 30A-B) in turn comprises a receiver for receiving instructions and/or suggestions about preferred pilot resource configuration from the second node (20, 30; 30A-B).

10 31. Arrangement according to any of the claims 26 to

characterised in that the arrangement is a wireless communications system (10).

15 **32.** Arrangement according to any of the claims 26 to 31.

characterised in that the arrangement utilises OFDM.

20 **33.** Arrangement according to any of the claims 26 to

characterised in that the first node is selected from the group of:

user equipment; mobile station; base station; access point; and relay.

34. Arrangement according to any of the claims 26 to

characterised in that the second node is selected from the group of:

user equipment; mobile station; base station; access point; and relay.

- 35. User equipment capable of being connected to a multi-user, multi-carrier wireless communications system, comprising:
 - means (35) for communication between the user equipment (30) and a node (20, 30; 30A-B) utilising a first resource sub-space (108A-F) of an entire multi-carrier resource space (100),

characterised in that

the first resource sub-space (108A-F) comprises a first pilot resource configuration, out of a set of at least two different pilot resource configurations, and

whereby the first pilot resource configuration

being in agreement with pilot need for estimated radio conditions for the user equipment (30).

36. User equipment according to claim 35, **character**ised by

- receiver for receiving data characterising the first pilot resource configuration from the node (20, 30; 30A-B);
- means (36) for channel estimation, connected 10 to the receiver,

whereby the means for channel estimation (36) is arranged to perform channel estimation based on the received data characterising the first pilot resource configuration.

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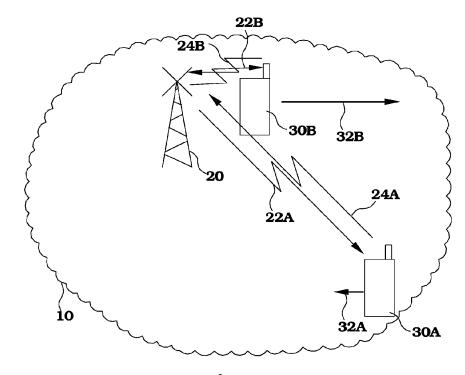
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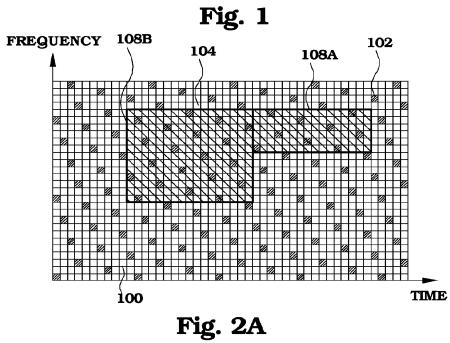
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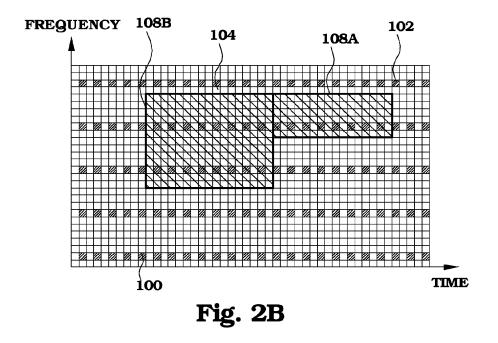
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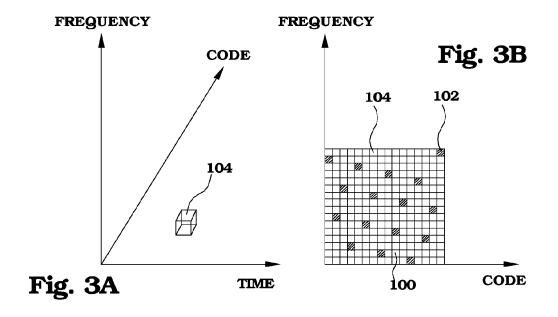
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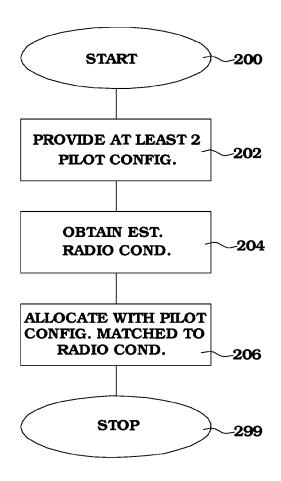
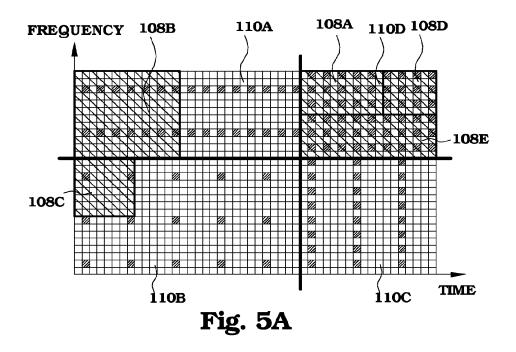
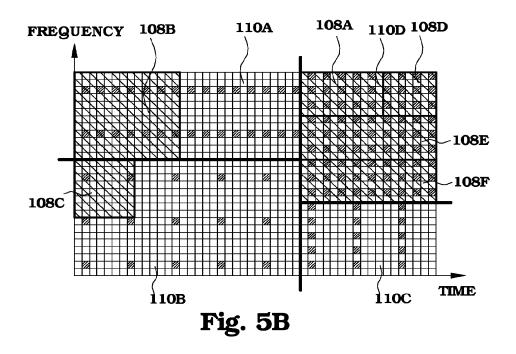
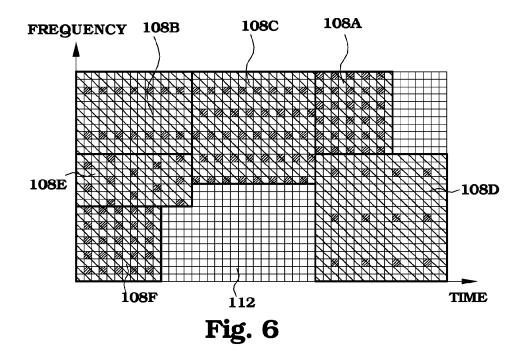


Fig. 4







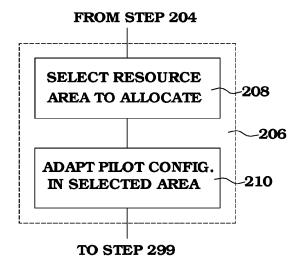


Fig. 8

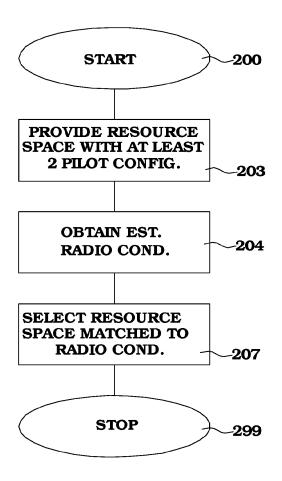


Fig. 7A

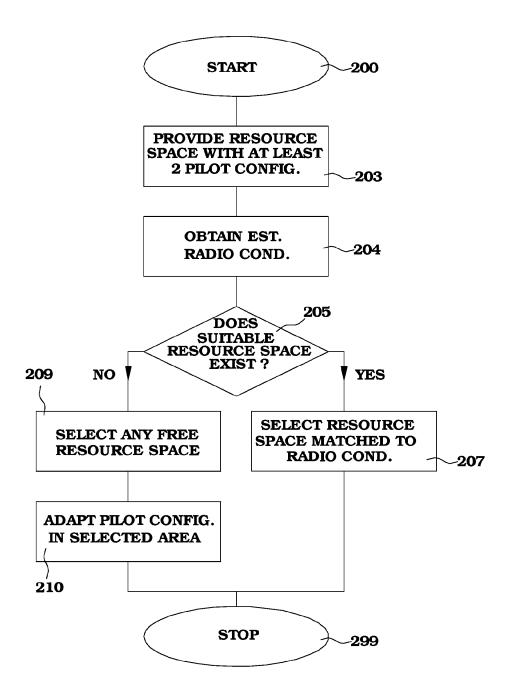
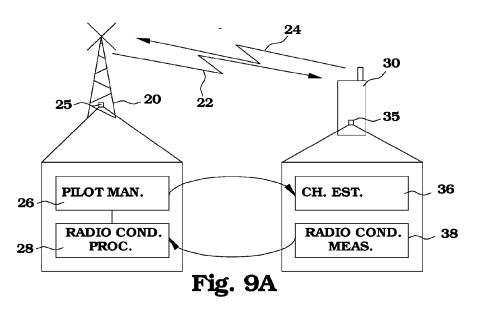
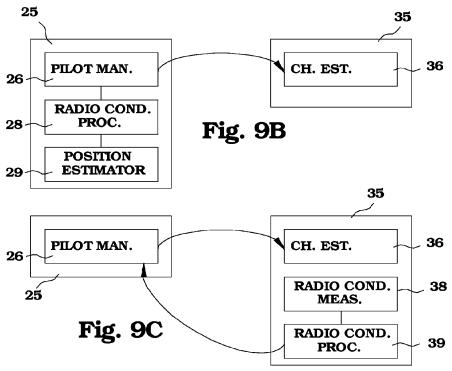


Fig. 7B





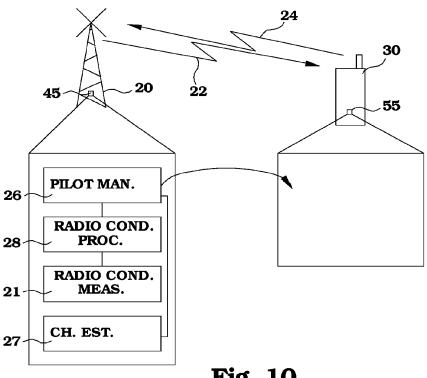
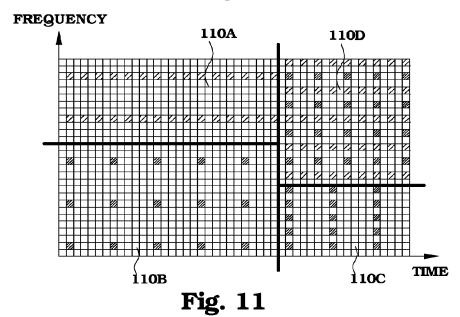
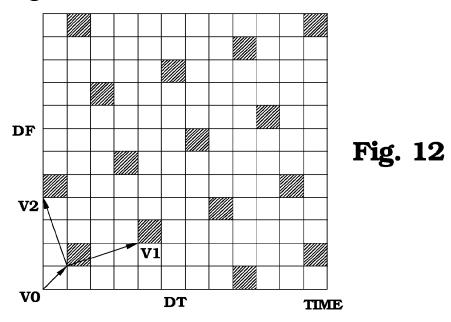


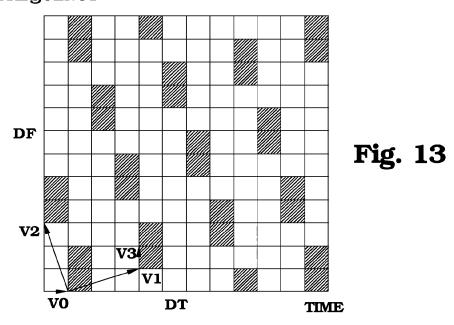
Fig. 10



FREQUENCY



FREQUENCY





EUROPEAN SEARCH REPORT

Application Number EP 03 10 4661

	DOCUMENTS CONSID	ERED TO BE RELEVA	NT	
Category	Citation of document with ir of relevant passa	dication, where appropriate, ges	Relevant to claim	CLASSIFICATION OF THE APPLICATION (Int.CI.7)
X Y	US 6 452 936 B1 (SH 17 September 2002 (* column 1, line 1- * column 2, line 38 * column 3, line 4- * column 4, line 8- * column 5 * * column 6 * * column 7, line 11 * figures 1,3 *	2002-09-17) 26 * - column 3, line 3 61 * 67 *	35,36 1-34	H04Q7/38
Y	US 2003/215021 A1 (MARTIN) 20 November * paragraphs [0021] * paragraphs [0031]	2003 (2003-11-20) -[0024] *	IR 1-34	
A	EP 0 862 343 A (LUC 2 September 1998 (1 * abstract *	ENT TECHNOLOGIES IN 998-09-02)	1-36	
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	The present search report has be	Date of completion o' the s	earch	Examiner
	MUNICH	2 June 2004	Me	le, M
X : parti Y : parti docu A : tech O : non	NTEGORY OF CITED DOCUMENTS ioularly relevant if taken alone cularly relevant if combined with anoth ment of the same category nological background written disolosure mediate document	E : earlier pa after the t er D : documer L : documer	nt cited in the application nt cited for other reasons of the same patent famil	lished on, or

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ANNEX TO THE EUROPEAN SEARCH REPORT ON EUROPEAN PATENT APPLICATION NO.

EP 03 10 4661

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02-06-2004

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Electronic Patent Application Fee Transmittal									
Application Number:	13303417								
Filing Date:	23	-Nov-2011							
Title of Invention:	SYSTEM AND METHOD FOR SCRAMBLING THE PHASE OF THE CARRIERS I MULTICARRIER COMMUNICATIONS SYSTEM								
First Named Inventor/Applicant Name:	Marcos C. Tzannes								
Filer:	Jason Vick/Joanne Vos								
Attorney Docket Number: 6936-47-CON-5									
Filed as Large Entity									
Utility under 35 USC 111(a) Filing Fees									
Description		Fee Code	Quantity	Amount	Sub-Total in USD(\$)				
Basic Filing:									
Pages:									
Claims:									
Miscellaneous-Filing:									
Petition:									
Patent-Appeals-and-Interference:									
Post-Allowance-and-Post-Issuance:									
Utility Appl Issue Fee		1501	1	960	960				
Extension-of-Time:									

Description	Fee Code	Quantity	Amount	Sub-Total in USD(\$)	
Miscellaneous:					
Submission- Information Disclosure Stmt	1806	1	180	180	
	Total in USD (\$)			1140	

Electronic Acknowledgement Receipt				
EFS ID:	18294218			
Application Number:	13303417			
International Application Number:				
Confirmation Number:	3035			
Title of Invention:	SYSTEM AND METHOD FOR SCRAMBLING THE PHASE OF THE CARRIERS IN A MULTICARRIER COMMUNICATIONS SYSTEM			
First Named Inventor/Applicant Name:	Marcos C. Tzannes			
Customer Number:	62574			
Filer:	Jason Vick/Joanne Vos			
Filer Authorized By:	Jason Vick			
Attorney Docket Number:	6936-47-CON-5			
Receipt Date:	25-FEB-2014			
Filing Date:	23-NOV-2011			
Time Stamp:	14:01:55			
Application Type:	Utility under 35 USC 111(a)			

Payment information:

Submitted with Payment	yes
Payment Type	Deposit Account
Payment was successfully received in RAM	\$1140
RAM confirmation Number	249
Deposit Account	191970
Authorized User	

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File Listing:

Document Number	Document Description	File Name	File Size(Bytes)/ Message Digest	Multi Pages Part /.zip (if appl	
1	Issue Fee Payment (PTO-85B)	Issue_Fee_Transmittal.pdf	227155 b6a74aaa391bd196786dd0dc6423483667	no	1
Warnings:			d33909		
Information:					
2	Post Allowance Communication -	Comments_on_Reason_for_All		no	3
	Incoming	owance.pdf	94ea5a0b5ada8cbf09da30777d789395510 b664c		
Warnings:					
Information:			<u> </u>		
3		AMEND_312.pdf	297156	yes	3
			b75ab89ce0af4bac88062ad9db082dc0fa7 21f7d		
	Multip	part Description/PDF files in .	zip description		
	Document De	Document Description			nd
	Amendment after Notice of	1	1		
	Specification		2	2	
	Applicant Arguments/Remarks Made in an Amendment		3	3	
Warnings:					
Information:					
4	IDS_07.pdf		555765	yes	4
		155_07.pai	d96986a1b9b4f7909d66726a5e017a7e456 9cfad	yes	
	Multip	zip description			
	Document Description		Start	End	
	Transmittal Letter		1	3	
	Information Disclosure Statement (IDS) Form (SB08)		4	4	
Warnings:			1		
Information:					
5	Non Patent Literature	6936-47-CON-DIV- CON-3_OA_02-04-2014.pdf	338482	no	10
3			67262593446be6c0506363b2e58f469f458 efce1	no	

Warnings:					
Information:					
6	Foreign Reference	EP1542488.pdf	1049528	no	22
	, oreign neterence	E. 15 12 1661pg.	d0d92b19247aafa4b5c91118d9aaf656e40 27db5		
Warnings:					
Information:					
7	Fee Worksheet (SB06)	fee-info.pdf	32315	no	2
,	r ee worksneet (3000)	ree-imo.pui	87859fcf61d3dd6dc238b6cd800acdf70495 afe9	110	
Warnings:					
Information:					
	Total Files Size (in bytes			57046	

This Acknowledgement Receipt evidences receipt on the noted date by the USPTO of the indicated documents, characterized by the applicant, and including page counts, where applicable. It serves as evidence of receipt similar to a Post Card, as described in MPEP 503.

New Applications Under 35 U.S.C. 111

If a new application is being filed and the application includes the necessary components for a filing date (see 37 CFR 1.53(b)-(d) and MPEP 506), a Filing Receipt (37 CFR 1.54) will be issued in due course and the date shown on this Acknowledgement Receipt will establish the filing date of the application.

National Stage of an International Application under 35 U.S.C. 371

If a timely submission to enter the national stage of an international application is compliant with the conditions of 35 U.S.C. 371 and other applicable requirements a Form PCT/DO/EO/903 indicating acceptance of the application as a national stage submission under 35 U.S.C. 371 will be issued in addition to the Filing Receipt, in due course.

New International Application Filed with the USPTO as a Receiving Office

If a new international application is being filed and the international application includes the necessary components for an international filing date (see PCT Article 11 and MPEP 1810), a Notification of the International Application Number and of the International Filing Date (Form PCT/RO/105) will be issued in due course, subject to prescriptions concerning national security, and the date shown on this Acknowledgement Receipt will establish the international filing date of the application.

UNITED STATES DEPARTMENT OF COMMERCE United States Patent and Trademark Office Address: COMMISSIONER FOR PATENTS P.O. Box 1450 Alexandria, Virginia 22313-1450 www.uspto.gov

NOTICE OF ALLOWANCE AND FEE(S) DUE

62574 7590 Jason H. Vick Sheridan Ross, PC Suite # 1200 1560 Broadway Denver, CO 80202

02/07/2014

EXAMINER

PHU, PHUONG M

ART UNIT PAPER NUMBER

2634

DATE MAILED: 02/07/2014

APPLICATION NO.	FILING DATE	FIRST NAMED INVENTOR	ATTORNEY DOCKET NO.	CONFIRMATION NO.
-----------------	-------------	----------------------	---------------------	------------------

13/303,417 11/23/2011 Marcos C. Tzannes 6936-47-CON-5 3035

TITLE OF INVENTION: SYSTEM AND METHOD FOR SCRAMBLING THE PHASE OF THE CARRIERS IN A MULTICARRIER COMMUNICATIONS SYSTEM

APPLN. TYPE	ENTITY STATUS	ISSUE FEE DUE	PUBLICATION FEE DUE	PREV. PAID ISSUE FEE	TOTAL FEE(S) DUE	DATE DUE
nonprovisional	UNDISCOUNTED	\$960	\$0	\$0	\$960	05/07/2014

THE APPLICATION IDENTIFIED ABOVE HAS BEEN EXAMINED AND IS ALLOWED FOR ISSUANCE AS A PATENT. PROSECUTION ON THE MERITS IS CLOSED. THIS NOTICE OF ALLOWANCE IS NOT A GRANT OF PATENT RIGHTS. THIS APPLICATION IS SUBJECT TO WITHDRAWAL FROM ISSUE AT THE INITIATIVE OF THE OFFICE OR UPON PETITION BY THE APPLICANT. SEE 37 CFR 1.313 AND MPEP 1308.

THE ISSUE FEE AND PUBLICATION FEE (IF REQUIRED) MUST BE PAID WITHIN THREE MONTHS FROM THE MAILING DATE OF THIS NOTICE OR THIS APPLICATION SHALL BE REGARDED AS ABANDONED. THIS STATUTORY PERIOD CANNOT BE EXTENDED. SEE 35 U.S.C. 151. THE ISSUE FEE DUE INDICATED ABOVE DOES NOT REFLECT A CREDIT FOR ANY PREVIOUSLY PAID ISSUE FEE IN THIS APPLICATION. IF AN ISSUE FEE HAS PREVIOUSLY BEEN PAID IN THIS APPLICATION (AS SHOWN ABOVE), THE RETURN OF PART B OF THIS FORM WILL BE CONSIDERED A REQUEST TO REAPPLY THE PREVIOUSLY PAID ISSUE FEE TOWARD THE ISSUE FEE NOW DUE.

HOW TO REPLY TO THIS NOTICE:

I. Review the ENTITY STATUS shown above. If the ENTITY STATUS is shown as SMALL or MICRO, verify whether entitlement to that entity status still applies.

If the ENTITY STATUS is the same as shown above, pay the TOTAL FEE(S) DUE shown above.

If the ENTITY STATUS is changed from that shown above, on PART B - FEE(S) TRANSMITTAL, complete section number 5 titled "Change in Entity Status (from status indicated above)".

For purposes of this notice, small entity fees are 1/2 the amount of undiscounted fees, and micro entity fees are 1/2 the amount of small entity fees.

II. PART B - FEE(S) TRANSMITTAL, or its equivalent, must be completed and returned to the United States Patent and Trademark Office (USPTO) with your ISSUE FEE and PUBLICATION FEE (if required). If you are charging the fee(s) to your deposit account, section "4b" of Part B - Fee(s) Transmittal should be completed and an extra copy of the form should be submitted. If an equivalent of Part B is filed, a request to reapply a previously paid issue fee must be clearly made, and delays in processing may occur due to the difficulty in recognizing the paper as an equivalent of Part B.

III. All communications regarding this application must give the application number. Please direct all communications prior to issuance to Mail Stop ISSUE FEE unless advised to the contrary.

IMPORTANT REMINDER: Utility patents issuing on applications filed on or after Dec. 12, 1980 may require payment of maintenance fees. It is patentee's responsibility to ensure timely payment of maintenance fees when due.

Page 1 of 3

PART B - FEE(S) TRANSMITTAL

Complete and send this form, together with applicable fee(s), to: Mail

Mail Stop ISSUE FEE
Commissioner for Patents
P.O. Box 1450 Alexandria, Virginia 22313-1450 or <u>Fax</u> (571)-273-2885

INSTRUCTIONS: This form should be used for transmitting the ISSUE FEE and PUBLICATION FEE (if required). Blocks 1 through 5 should be completed where appropriate. All further correspondence including the Patent, advance orders and notification of maintenance fees will be mailed to the current correspondence address as indicated unless corrected below or directed otherwise in Block 1, by (a) specifying a new correspondence address; and/or (b) indicating a separate "FEE ADDRESS" for maintenance fee notifications.

CURRENT CORRESPOND	DENCE ADDRESS (Note: Use B	ock 1 for any change of address)	Fee(s) Transmittal. Thi rs. Each additiona	s certificate cannot be use	for domestic mailings of the d for any other accompanying ment or formal drawing, must h.
Jason H. Vick Sheridan Ross, I Suite # 1200		7/2014	I he Stat addi tran	reby certify that the	tificate of Mailing or Tra is Fee(s) Transmittal is be ith sufficient postage for Stop ISSUE FEE addre TO (571) 273-2885, on the	nsmission ing deposited with the United first class mail in an envelope ss above, or being facsimile date indicated below.
1560 Broadway						(Depositor's name)
Denver, CO 802	202					(Signature)
						(Date)
APPLICATION NO.	FILING DATE		FIRST NAMED INVENTOR		ATTORNEY DOCKET NO.	CONFIRMATION NO.
13/303,417	11/23/2011		Marcos C. Tzannes		6936-47-CON-5	3035
TITLE OF INVENTI COMMUNICATIONS S		METHOD FOR SCR	AMBLING THE PHAS	E OF THE CA	ARRIERS IN A MULT	TICARRIER
APPLN. TYPE	ENTITY STATUS	ISSUE FEE DUE	PUBLICATION FEE DUE	PREV. PAID ISSUE	E FEE TOTAL FEE(S) DU	JE DATE DUE
nonprovisional	UNDISCOUNTED	\$960	\$0	\$0	\$960	05/07/2014
EXAM	MINER	ART UNIT	CLASS-SUBCLASS			
PHU, PH	UONG M	2634	375-260000			
"Fee Address" inc PTO/SB/47; Rev 03-0 Number is required. 3. ASSIGNEE NAME A PLEASE NOTE: Un recordation as set fort (A) NAME OF ASSI	ND RESIDENCE DAT. less an assignee is ident th in 37 CFR 3.11. Com GNEE	" Indication form ed. Use of a Customer A TO BE PRINTED ON fified below, no assignee poletion of this form is NO	T a substitute for filing an (B) RESIDENCE: (CITY	rely, e firm (having as a gent) and the nam rneys or agents. If printed. be) atent. If an assign assignment. and STATE OR C	member a 2es of up to no name is 3ee is identified below, the COUNTRY)	document has been filed for
4a. The following fee(s) Issue Fee	are subfillited:	41	 b. Payment of Fee(s): (Please) A check is enclosed. 	se tirst reapply at	iy previousiy paid issue i	ee snown above)
Publication Fee (1	No small entity discount	permitted)	Payment by credit car			
Advance Order - #	of Copies		The Director is hereby overpayment, to Depo	authorized to char sit Account Numbe	ge the required fee(s), any er (enclose	deficiency, or credits any e an extra copy of this form).
_ ~ .	atus (from status indicate		NOTE: Absent a valid ce	rtification of Micro	Entity Status (see forms P	TO/SB/15A and 15B), issue
Applicant assertin	g small entity status. See	37 CFR 1.27	NOTE: If the application	was previously und	der micro entity status, che	of application abandonment. cking this box will be taken
Applicant changir	ng to regular undiscounte	d fee status.	to be a notification of los	of entitlement to r	micro entity status.	ntitlement to small or micro
NOTE: This forms must be		with 27 CED 1 21 and 1 2	entity status, as applicable 3. See 37 CFR 1.4 for signs		and soutifications	
NOTE: This form must t	be signed in accordance	vitii 57 CFR 1.31 and 1.3.	5. See 57 CFR 1.4 for signs	nure requirements	and certifications.	
Authorized Signature				Date		
Typed or printed name	ne			Registration N	Īo	
			Page 2 of 3			
PTOL-85 Part B (10-13)	Approved for use through	sh 10/31/2013	8	S Patent and Trac	dem 1000 Troppe	P. M. W. AL AC AC A M. P. AC



UNITED STATES PATENT AND TRADEMARK OFFICE

UNITED STATES DEPARTMENT OF COMMERCE United States Patent and Trademark Office Address: COMMISSIONER FOR PATENTS P.O. Box 1450 Alexandra, Virgina 22313-1450 www.uspto.gov

APPLICATION NO.	FILING DATE	FIRST NAMED INVENTOR	ATTORNEY DOCKET NO.	CONFIRMATION NO.
13/303,417	11/23/2011	Marcos C. Tzannes	6936-47-CON-5	3035
62574 75	90 02/07/2014		EXAM	INER
Jason H. Vick Sheridan Ross, PC			PHU, PH	UONG M
Suite # 1200			ART UNIT	PAPER NUMBER
1560 Broadway			2634	
Denver, CO 80202			DATE MAIL ED: 02/07/201	4

Determination of Patent Term Adjustment under 35 U.S.C. 154 (b)

(application filed on or after May 29, 2000)

The Patent Term Adjustment to date is 80 day(s). If the issue fee is paid on the date that is three months after the mailing date of this notice and the patent issues on the Tuesday before the date that is 28 weeks (six and a half months) after the mailing date of this notice, the Patent Term Adjustment will be 80 day(s).

If a Continued Prosecution Application (CPA) was filed in the above-identified application, the filing date that determines Patent Term Adjustment is the filing date of the most recent CPA.

Applicant will be able to obtain more detailed information by accessing the Patent Application Information Retrieval (PAIR) WEB site (http://pair.uspto.gov).

Any questions regarding the Patent Term Extension or Adjustment determination should be directed to the Office of Patent Legal Administration at (571)-272-7702. Questions relating to issue and publication fee payments should be directed to the Customer Service Center of the Office of Patent Publication at 1-(888)-786-0101 or (571)-272-4200.

OMB Clearance and PRA Burden Statement for PTOL-85 Part B

The Paperwork Reduction Act (PRA) of 1995 requires Federal agencies to obtain Office of Management and Budget approval before requesting most types of information from the public. When OMB approves an agency request to collect information from the public, OMB (i) provides a valid OMB Control Number and expiration date for the agency to display on the instrument that will be used to collect the information and (ii) requires the agency to inform the public about the OMB Control Number's legal significance in accordance with 5 CFR 1320.5(b).

The information collected by PTOL-85 Part B is required by 37 CFR 1.311. The information is required to obtain or retain a benefit by the public which is to file (and by the USPTO to process) an application. Confidentiality is governed by 35 U.S.C. 122 and 37 CFR 1.14. This collection is estimated to take 12 minutes to complete, including gathering, preparing, and submitting the completed application form to the USPTO. Time will vary depending upon the individual case. Any comments on the amount of time you require to complete this form and/or suggestions for reducing this burden, should be sent to the Chief Information Officer, U.S. Patent and Trademark Office, U.S. Department of Commerce, P.O. Box 1450, Alexandria, Virginia 22313-1450. DO NOT SEND FEES OR COMPLETED FORMS TO THIS ADDRESS. SEND TO: Commissioner for Patents, P.O. Box 1450, Alexandria, Virginia 22313-1450. Under the Paperwork Reduction Act of 1995, no persons are required to respond to a collection of information unless it displays a valid OMB control number.

Privacy Act Statement

The Privacy Act of 1974 (P.L. 93-579) requires that you be given certain information in connection with your submission of the attached form related to a patent application or patent. Accordingly, pursuant to the requirements of the Act, please be advised that: (1) the general authority for the collection of this information is 35 U.S.C. 2(b)(2); (2) furnishing of the information solicited is voluntary; and (3) the principal purpose for which the information is used by the U.S. Patent and Trademark Office is to process and/or examine your submission related to a patent application or patent. If you do not furnish the requested information, the U.S. Patent and Trademark Office may not be able to process and/or examine your submission, which may result in termination of proceedings or abandonment of the application or expiration of the patent.

The information provided by you in this form will be subject to the following routine uses:

- 1. The information on this form will be treated confidentially to the extent allowed under the Freedom of Information Act (5 U.S.C. 552) and the Privacy Act (5 U.S.C 552a). Records from this system of records may be disclosed to the Department of Justice to determine whether disclosure of these records is required by the Freedom of Information Act.
- A record from this system of records may be disclosed, as a routine use, in the course of presenting evidence to a court, magistrate, or administrative tribunal, including disclosures to opposing counsel in the course of settlement negotiations.
- 3. A record in this system of records may be disclosed, as a routine use, to a Member of Congress submitting a request involving an individual, to whom the record pertains, when the individual has requested assistance from the Member with respect to the subject matter of the record.
- 4. A record in this system of records may be disclosed, as a routine use, to a contractor of the Agency having need for the information in order to perform a contract. Recipients of information shall be required to comply with the requirements of the Privacy Act of 1974, as amended, pursuant to 5 U.S.C. 552a(m).
- 5. A record related to an International Application filed under the Patent Cooperation Treaty in this system of records may be disclosed, as a routine use, to the International Bureau of the World Intellectual Property Organization, pursuant to the Patent Cooperation Treaty.
- 6. A record in this system of records may be disclosed, as a routine use, to another federal agency for purposes of National Security review (35 U.S.C. 181) and for review pursuant to the Atomic Energy Act (42 U.S.C. 218(c)).
- 7. A record from this system of records may be disclosed, as a routine use, to the Administrator, General Services, or his/her designee, during an inspection of records conducted by GSA as part of that agency's responsibility to recommend improvements in records management practices and programs, under authority of 44 U.S.C. 2904 and 2906. Such disclosure shall be made in accordance with the GSA regulations governing inspection of records for this purpose, and any other relevant (i.e., GSA or Commerce) directive. Such disclosure shall not be used to make determinations about individuals.
- 8. A record from this system of records may be disclosed, as a routine use, to the public after either publication of the application pursuant to 35 U.S.C. 122(b) or issuance of a patent pursuant to 35 U.S.C. 151. Further, a record may be disclosed, subject to the limitations of 37 CFR 1.14, as a routine use, to the public if the record was filed in an application which became abandoned or in which the proceedings were terminated and which application is referenced by either a published application, an application open to public inspection or an issued patent.
- 9. A record from this system of records may be disclosed, as a routine use, to a Federal, State, or local law enforcement agency, if the USPTO becomes aware of a violation or potential violation of law or regulation.

	Applicant(s	nt(s) ES, MARCOS C.	
Notice of Allowability	13/303,417 Examiner PHUONG PHU	Art Unit 2634	AIA (First Inventor to File) Status
The MAILING DATE of this communication appear All claims being allowable, PROSECUTION ON THE MERITS IS (herewith (or previously mailed), a Notice of Allowance (PTOL-85) of NOTICE OF ALLOWABILITY IS NOT A GRANT OF PATENT RIC of the Office or upon petition by the applicant. See 37 CFR 1.313	OR REMAINS) CLOSED in this apport of the appropriate communication GHTS. This application is subject to	olication. If no will be mailed	t included I in due course. THIS
1. A declaration(s)/affidavit(s) under 37 CFR 1.130(b) was/			
 An election was made by the applicant in response to a restr requirement and election have been incorporated into this ac 	· -	ne interview o	n; the restriction
 The allowed claim(s) is/are <u>21-50</u>. As a result of the allowed Highway program at a participating intellectual property office http://www.uspto.gov/patents/init_events/pph/index.jsp or ser 	e for the corresponding application.	For more info	
a) All b) Some *c) None of the: 1. Certified copies of the priority documents have 2. Certified copies of the priority documents have 3. Copies of the certified copies of the priority documents have International Bureau (PCT Rule 17.2(a)). * Certified copies not received: Applicant has THREE MONTHS FROM THE "MAILING DATE" of noted below. Failure to timely comply will result in ABANDONMETHIS THREE-MONTH PERIOD IS NOT EXTENDABLE.	been received in Application No uments have been received in this r	national stage	
5. \square CORRECTED DRAWINGS (as "replacement sheets") must			
including changes required by the attached Examiner's Paper No./Mail Date			
Identifying indicia such as the application number (see 37 CFR 1.6 each sheet. Replacement sheet(s) should be labeled as such in th			(not the back) of
 DEPOSIT OF and/or INFORMATION about the deposit of BI attached Examiner's comment regarding REQUIREMENT FOR 			the
Attachment(s) 1. ☐ Notice of References Cited (PTO-892) 2. ☑ Information Disclosure Statements (PTO/SB/08), Paper No./Mail Date 08/05/13, 08/05/13 3. ☐ Examiner's Comment Regarding Requirement for Deposit of Biological Material 4. ☐ Interview Summary (PTO-413), Paper No./Mail Date	5. ☐ Examiner's Amendr 6. ☑ Examiner's Stateme 7. ☐ Other		
/PHUONG PHU/ Primary Examiner, Art Unit 2634			

U.S. Patent and Trademark Office PTOL-37 (Rev. 08-13)

Notice of Allowability

Part of Paper No./Mail Date 20140203

Application/Control Number: 13/303,417 Page 2

Art Unit: 2634

DETAILED ACTION

1. The present application is being examined under the pre-AIA first to invent provisions.

2. This Office Action is responsive to the Amendment filed on 12/27/13. Accordingly, claims 21-50 are currently pending; and claims 1-20 and 51 are canceled.

REASONS FOR ALLOWANCE

3. Claims 21-50 are allowed.

4. The following is an examiner's statement of reasons for allowance:

-Regarding independent claim 21, none of prior art of record teaches or suggests, in a multicarrier modulation system including a first transceiver in communication with a second transceiver using a transmission signal having a plurality of carrier signals for modulating a plurality of data bits, each carrier signal having a phase characteristic associated with at least one bit of the plurality of data bits, a method for scrambling the phase characteristics of the carrier signals comprising: associating a carrier signal with a value determined independently of any bit of the plurality of data bits carried by the carrier signal, the value associated with the carrier signal determined by a pseudo-random number generator; determining a phase shift for the carrier signal at least based on the value associated with the carrier signal; modulating at least one bit of the plurality of data bits on the carrier signal; modulating the at least one bit on a second carrier signal of the plurality of carrier signals.

-Regarding independent claim 35, none of prior art of record teaches or suggests a multicarrier modulation system including a first transceiver in communication with a second transceiver using a transmission signal having a plurality of carrier signals for modulating a

Application/Control Number: 13/303,417

Art Unit: 2634

plurality of data bits, each carrier signal having a phase characteristic associated with the input bit stream, the first transceiver operable to: associate a carrier signal with a value determined independently of any bit of the plurality of data bits carried by the carrier signal, the value associated with the carrier signal determined by a pseudo-random number generator; determine a phase shift for the carrier signal at least based on the value associated with the carrier signal; modulate at least one bit of the plurality of data bits on the carrier signal; and modulate the at least one bit on a second carrier signal of the plurality of carrier signals.

Any comments considered necessary by applicant must be submitted no later than the payment of the issue fee and, to avoid processing delays, should preferably accompany the issue fee. Such submissions should be clearly labeled "Comments on Statement of Reasons for Allowance."

Any inquiry concerning this communication or earlier communications from the examiner should be directed to PHUONG PHU whose telephone number is (571)272-3009. The examiner can normally be reached on M-F (8:00 AM - 4:30 PM).

If attempts to reach the examiner by telephone are unsuccessful, the examiner's supervisor, Daniel Washburn can be reached on 571-272-5551. The fax phone number for the organization where this application or proceeding is assigned is 571-273-8300.

Page 3

Application/Control Number: 13/303,417

Art Unit: 2634

Information regarding the status of an application may be obtained from the Patent Application Information Retrieval (PAIR) system. Status information for published applications may be obtained from either Private PAIR or Public PAIR. Status information for unpublished applications is available through Private PAIR only. For more information about the PAIR system, see http://pair-direct.uspto.gov. Should you have questions on access to the Private PAIR system, contact the Electronic Business Center (EBC) at 866-217-9197 (toll-free). If you would like assistance from a USPTO Customer Service Representative or access to the automated information system, call 800-786-9199 (IN USA OR CANADA) or 571-272-1000.

Phuong Phu Primary Examiner Art Unit 2634

/PHUONG PHU/ Primary Examiner, Art Unit 2634 Page 4

Sub	Substitute for form 1449A/PTO			Complete if Known		
		TION DIOC	N COURT	Application Number	13/303,417	
			CLOSURE	Filing Date	November 23, 2011	
SI	STATEMENT BY APPLICANT			First Named Inventor	Marcos C. Tzannes	
				Art Unit	2634	
				Examiner Name	PHU, Phuong M.	
Sheet	1	of	1	Attorney Docket Number	6936-47-CON-5	

			U.S. PATENT DO	CUMENTS	
Examiner Initials*	Cite No.1	Document Number Number-kind Code ^{2 (If known)}	Publication Date MM-DD-YYYY	Name of Patentee of Applicant of Cited Document	Pages, Columns, Lines, Where Relevant Passages or Relevant Figures Appear
	1	2013/0136160	05-30-2013	Tzannes	

	FOREIGN PATENT DOCUMENTS					
Examiner Initials*		Foreign Patent Document Country Code ³ ; Number ⁴ ; Kind Code ⁵ (<i>if known</i>)		Name of Patentee or Applicant of Cited Document	Pages, Columns, Lines, Where Relevant Passages or Relevant Figures Appear	T ⁶

		OTHER ART (Including Author, Title, Date, Pertinent Pages, etc.)	
Examiner Initials*	Cite No. ¹		

Signature Considered UZ/UJ/ZU14	Examiner Signature	/Phuong Phu/	Date Considered	02/03/2014
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*EXAMINER: Initial if reference is considered, whether or not citation is in conformance and not considered. Include copy of this form with next communication to applicant.

ALL REFERENCES CONSIDERED EXCEPT WHERE LINED THROUGH. /P.P./

Sub	Substitute for form 1449A/PTO			Complete if Known		
181		FIGNI DIG		Application Number	13/303,417	
			CLOSURE	Filing Date	November 23, 2011	
Si	AIEME	NT BY AP	PLICANI	First Named Inventor	Marcos C, Tzannes	
				Art Unit	2634	
				Examiner Name	PHU, Phuong M.	
Sheet	1	of	1	Attorney Docket Number	6936-47-CON-5	

	U.S. PATENT DOCUMENTS						
Examiner Initials*	Cite No. ¹	Document Number Number-kind Code ^{2 (If known)}	Publication Date MM-DD-YYYY	Name of Patentee of Applicant of Cited Document	Pages, Columns, Lines, Where Relevant Passages or Relevant Figures Appear		
	1	5903614	05-11-1999	Suzuki et al.			
	2	5914932	06-22-1999	Suzuki et al.			

FOREIGN PATENT DOCUMENTS						
Examiner Initials*		Foreign Patent Document Country Code ³ ; Number ⁴ ; Kind Code ⁵ (if known)			Pages, Columns, Lines, Where Relevant Passages or Relevant Figures Appear	L _e

		OTHER ART (Including Author, Title, Date, Pertinent Pages, etc.)
Examiner Initials*	Cite No. ¹	
		Official Action for U.S. Patent Application No. 13/718,016, mailed May 9, 2013 (Attorney Ref. No.: 6936-47-CON-DIV-CON-3)

Examiner Signature /Phuong Phu/	Date Considered	02/03/2014
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*EXAMINER: Initial if reference is considered, whether or not citation is in conformance and not considered. Include copy of this form with next communication to applicant.

ALL REFERENCES CONSIDERED EXCEPT WHERE LINED THROUGH. /P.P./

	Application/Control No.	Applicant(s)/Patent Under Reexamination
Index of Claims	13303417	TZANNES, MARCOS C.
	Examiner	Art Unit
	PHUONG PHU	2634

✓	Rejected	-	Cancelled	N	Non-Elected	Α	Appeal
=	Allowed	÷	Restricted	ı	Interference	0	Objected

☐ Claims r	enumbered	in the same	order as pre	esented by	applicant		□ СРА	□ т.с	D. 🗆	R.1.47
CLA	MIM					DATE				
Final	Original	06/29/2013	02/03/2014							
1	21	✓	=							
2	22	✓	=							
3	23	✓	=							
4	24	✓	=							
5	25	✓	=							
6	26	✓	=							
7	27	✓	=							
8	28	✓	=							
9	29	✓	=							
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19	39	✓	=							
20	40	✓	=							
21	41	✓	=							
22	42	✓	=							
23	43	✓	=							
24	44	✓	=							
25	45	✓	=							
26	46	✓	=							
27	47	✓	=							
28	48	✓	=							
29	49	✓	=							
30	50	✓	=							
	51	✓	-							

U.S. Patent and Trademark Office Part of Paper No.: 20140203

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CPC Combination Sets											
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NONE	Total Clain	ns Allowed:			
(Assistant Examiner)	(Date)	30			
/PHUONG PHU/ Primary Examiner.Art Unit 2634	02/03/14	O.G. Print Claim(s)	O.G. Print Figure		
(Primary Examiner)	(Date)	1	1		

U.S. Patent and Trademark Office Part of Paper No. 20140203

	Application/Control No.	Applicant(s)/Patent Under Reexamination
Issue Classification	13303417	TZANNES, MARCOS C.
	Examiner	Art Unit
	PHUONG PHU	2634

	US O	RIGINAL C	LASSIFIC	ATION						INTERNATIONAL	CLA	SSI	FIC	ATIO	NC	
	CLASS SUBCLASS								С	LAIMED		NON-CLAIMED				
375	5 260				75 260			260 H 0 4 K 1/10 (2006.01.01)								
	С	ROSS REF	ERENCE	RENCE(S)				H 0 4 L 27/								
CLASS	SU	IBCLASS (ON	E SUBCLAS	SUBCLASS PER BLOCK)				1								
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(Assistant Examiner)	(Date)	30				
/PHUONG PHU/ Primary Examiner.Art Unit 2634	02/03/14	O.G. Print Claim(s)	O.G. Print Figure			
(Primary Examiner)	(Date)	1	1			

U.S. Patent and Trademark Office Part of Paper No. 20140203

	Claims renumbered in the same order as presented by applica								СР	'A [] T.D.		R.1.	47	
Final	Original	Final	Original	Final	Original	Final	Original	Final	Original	Final	Original	Final	Original	Final	Original
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3	23	19	39												
4	24	20	40												
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6	26	22	42												
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11	31	27	47												
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14	34	30	50												
15	35														
16	36														

NONE		Total Claims Allowed:				
(Assistant Examiner)	(Date)	30				
/PHUONG PHU/ Primary Examiner.Art Unit 2634	02/03/14	O.G. Print Claim(s)	O.G. Print Figure			
(Primary Examiner)	(Date)	1	1			

U.S. Patent and Trademark Office Part of Paper No. 20140203

Search Notes

Application/Control No.	Applicant(s)/Patent Under Reexamination		
13303417	TZANNES, MARCOS C.		
Examiner	Art Unit		
PHUONG PHU	2634		

CPC- SEARCHED		
Symbol	Date	Examiner

CPC COMBINATION SETS - SEARC	CHED	
Symbol	Date	Examiner

US CLASSIFICATION SEARCHED			
Class	Subclass	Date	Examiner
375	222, 219, 220, 295, 316, 259, 260	06/26/13	PP
370	203, 527, 208	06/26/13	PP

SEARCH NOTES		
Search Notes	Date	Examiner
EAST	06/26/13	PP

INTERFERENCE SEARCH				
US Class/ CPC Symbol	US Subclass / CPC Group	Date	Examiner	
375	260, 222, 219, 295	02/03/14	PP	

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U.S. Patent and Trademark Office Part of Paper No.: 20140203

IN THE UNITED STATES PATENT AND TRADEMARK OFFICE

In Re the Application of: Marcos C. Tzannes) Group Art Unit: 2634
Application No.: 13/303,417	Examiner: PHU, Phuong M.
Filed: November 23, 2011	Confirmation No.: 3035
Atty. File No.: 6936-47-CON-5)

For: SYSTEM AND METHOD FOR SCRAMBLING THE PHASE OF THE CARRIERS IN A MULTICARRIER COMMUNICATIONS SYSTEM

AMENDMENT AND RESPONSE

Commissioner for Patents P.O. Box 1450 Alexandria, VA 22313-1450

Madam:

Applicants submit this Amendment and Response to address the Office Action having a mailing date of July 5, 2013. Please credit any overpayment or charge any underpayment to Deposit Account No. 19-1970.

Please amend the above-identified patent application as follows:

Amendments to the Claims are shown in the listing of claims which begins on page 2 of this paper.

Remarks begin on page 8 of this paper.

Amendments to the Claims:

This listing of claims will replace all prior versions, and listings, of claims in the application:

Listing of Claims:

1. - 20. (Cancelled)

21. (Currently Amended) In a multicarrier modulation system including a first transceiver in communication with a second transceiver using a transmission signal having a plurality of carrier signals for modulating a plurality of data bitsan input bit stream, each carrier signal having a phase characteristic associated with at least one bit of the plurality of data bitsthe input bit stream, a method for scrambling the phase characteristics of the carrier signals comprising:

transmitting the <u>plurality of data bitsbit stream</u> from the first transceiver to the second transceiver; wherein:

associating a carrier signal each carrier signal is associated with a value determined independently of any bit value of the plurality of data bits bit stream carried by the that respective carrier signal, the value associated with the each carrier signal determined by a pseudo-random number generator;

<u>determining</u> a phase shift for <u>the</u>each carrier signal is at least based on: the value associated with <u>the</u>that carrier signal, and;

a combining of a phase for each carrier signal with the phase characteristic of that respective carrier signal so as to substantially scramble the phase characteristics of the plurality of carrier signals, and

modulating at least one bit of the plurality of data bits on the multiple carrier signal; and signals corresponding to the plurality of phase shifted and scrambled carrier signals are used by the first transceiver to modulate a same bit value of the bit stream; and

modulating the at least one bit on a second carrier signal of the plurality of carrier signals.

wherein:

the first and second transceivers communicate over one or more of a pair of twisted wires of a telephone subscriber line, quad cable and fiber optic, and the first and second transceivers transport video.

- 22. (Previously Presented) The method of claim 21, wherein one or more of the first transceiver and second transceiver are cable transceivers.
- 23. (Previously Presented) The method of claim 21, wherein one or more of the first transceiver and second transceiver are VDSL transceivers.
- 24. (Previously Presented) The method of claim 21, wherein the first and second transceivers are multicarrier DSL transceivers.
- 25. (Previously Presented) The method of claim 21, wherein the first and second transceivers are used for high speed internet access.
- 26. (Currently Amended) The method of claim 21, further comprising, in the first transceiver, independently deriving the values associated with each carrier using a second pseudo-random number generator in the second transceiver.
- 27. (Currently Amended) The method of claim 26, wherein-further comprising using in the first and second transceivers use a same seed for the <u>first and second</u> pseudo-random number generatorgenerators and the value of the seed is transmitted from the first transceiver to the second transceiver.
- 28. (Previously Presented) The method of claim 26, wherein the first and second transceivers are wireless transceivers.
- 29. (Previously Presented) The method of claim 26, wherein the first and second transceivers are cable transceivers.

Attorney Docket No.: 6936-47-CON-5

- 30. (Previously Presented) The method of claim 26, wherein the first and second transceivers are DSL transceivers connected using a pair of twisted wires of a telephone subscriber system.
- 31. (Previously Presented) The method of claim 30, wherein the first and second transceivers are VDSL transceivers.
- 32. (Previously Presented) The method of claim 26, wherein the first and second transceivers are multicarrier DSL transceivers.
- 33. (Previously Presented) The method of claim 26, wherein the first and second transceivers are also used for transport high speed internet access.
- 34. (Previously Presented) The method of claim 21, wherein the first and the second transceivers include digital signal processors.
- 35. (Currently Amended) A multicarrier modulation system including a first transceiver in communication with a second transceiver using a transmission signal having a plurality of carrier signals for modulating a plurality of data bitsan input bit stream, each carrier signal having a phase characteristic associated with the input bit stream, the first transceiver capable of transmitting to the second transceiver the plurality of bits and operable tobit stream, wherein:

associate a carrier signal with a value determined independently of any bit of the plurality of data bits carried by the carrier signal, the value associated with the carrier signal determined by a pseudo-random number generator;

determine a phase shift for the carrier signal at least based on the value associated with the carrier signal;

modulate at least one bit of the plurality of data bits on the carrier signal; and modulate the at least one bit on a second carrier signal of the plurality of carrier signals.

each carrier signal is associated with a value determined independently of any bit

value of the bit stream carried by that respective carrier signal, the value associated with each carrier signal determined by a pseudo random number generator,

a phase shift for each carrier signal is at least based on:

the value associated with that respective carrier signal, and

a combining of a phase shift for each carrier signal with the phase characteristic of that respective carrier signal so as to substantially scramble the phase characteristics of the plurality of carrier signals, and

multiple carrier signals corresponding to the plurality of phase shifted and scrambled carrier signals are used by the first transceiver to modulate a same bit value of the bit stream; and

wherein:

the first and second transceivers communicate over one or more of a pair of twisted wires of a telephone subscriber line, quad cable and fiber optic, and

the first and second transceivers transport video.

- 36. (Previously Presented) The system of claim 35, wherein one or more of the first transceiver and second transceiver are cable transceivers.
- 37. (Previously Presented) The system of claim 35, wherein one or more of the first transceiver and second transceiver are VDSL transceivers.
- 38. (Previously Presented) The system of claim 35, wherein the first and second transceivers are multicarrier DSL transceivers.
- 39. (Previously Presented) The system of claim 35, wherein the first and second transceivers are used for high speed internet access.
- 40. (Previously Presented) The system of claim 35, wherein the first transceiver independently derives the values associated with each carrier using a second pseudo-random number generator in the first transceiver.

- 41. (Currently Amended) The system of claim 40, <u>using inwherein</u> the first and second transceivers use a same seed for <u>the first and second</u> the pseudo-random number generatorgenerators and the value of the seed is transmitted from the first transceiver to the second transceiver.
- 42. (Previously Presented) The system of claim 40, wherein the first and second transceivers are wireless transceivers.
- 43. (Previously Presented) The system of claim 40, wherein the first and second transceivers are cable transceivers.
- 44. (Previously Presented) The system of claim 40, wherein the first and second transceivers are DSL transceivers connected using a pair of twisted wires of a telephone subscriber system.
- 45. (Previously Presented) The system of claim 44, wherein the first and second transceivers are VDSL transceivers.
- 46. (Previously Presented) The system of claim 40, wherein the first and second transceivers are multicarrier DSL transceivers.
- 47. (Previously Presented) The system of claim 40, wherein the first and second transceivers are also used for high speed internet access.
- 48. (Previously Presented) The method of claim 35, wherein the first and second transceivers each include digital signal processors.
- 49. (Previously Presented) The method of claim 21, wherein the video is video-on demand.

- 50. (Previously Presented) The method of claim 35, wherein the video is video-on demand.
 - 51. (Cancelled)

REMARKS

Applicant respectfully requests reconsideration of this application as amended.

Claims 21-50 are pending. Claim 51 is cancelled without prejudice or disclaimer.

By the above amendments, and without concession as to the propriety of the outstanding rejections, the rejected claims have been amended to address the outstanding 35 U.S.C. §112, second paragraph, rejection. Withdrawal of the rejection is respectfully requested.

Similarly, Applicant respectfully submits the 35 U.S.C. §112, second paragraph, rejection of claims 21-34, 49 and 51 has been overcome by virtue of the above amendments. Withdrawal of the rejection is respectfully requested.

Regarding the art-based rejection, amended independent claim 21 recites: ...a multicarrier modulation system including a first transceiver in communication with a second transceiver using a transmission signal having a plurality of carrier signals for modulating a plurality of data bits, each carrier signal having a phase characteristic associated with at least one bit of the plurality of data bits, a method for scrambling the phase characteristics of the carrier signals comprising:

transmitting the plurality of data bits from the first transceiver to the second transceiver; associating a carrier signal with a value determined independently of any bit of the plurality of data bits carried by the carrier signal, the value associated with the carrier signal determined by a pseudo-random number generator;

determining a phase shift for the carrier signal at least based on the value associated with the carrier signal; and

modulating at least one bit of the plurality of data bits on the carrier signal; and modulating the at least one bit on a second carrier signal of the plurality of carrier signals. Wu does not disclose the claimed combination of features.

More specifically, Wu at least fails to teach, suggest or disclose the claimed transmitting, associating, determining and modulating steps. Rather, Wu is generally directed toward modem data communications between telecommunications networks and remote or end user installations over twisted metallic wire pairs.

With comparable arguments being applicable to similar features in the other independent claim, Applicant respectfully submits all independent claims are patentably distinguishable from the reference. The dependent claims are even further distinguishable at least based on the above and the additional feature(s) recited therein.

With all rejections having been overcome, Applicant respectfully submits the application is in condition for allowance.

A prompt notice of allowance is respectfully solicited.

Should the Examiner believe anything further is desirable in order to place the application in even better condition for allowance, the Examiner is encouraged to contact Applicants undersigned representative at the telephone number listed below.

The Commissioner is hereby authorized to charge to deposit account number 19-1970 any fees under 37 CFR § 1.16 and 1.17 that may be required by this paper and to credit any overpayment to that Account. If any extension of time is required in connection with the filing of this paper and has not been separately requested, such extension is hereby petitioned.

Respectfully submitted,

SHERIDAN ROSS P.C.

Date: 27 Dec /3

Jason H. Vick

Reg. No. 45,285

1560 Broadway, Suite 1200

Denver, Colorado 80202 Telephone: 303-863-9700

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PTO/SB/22 (03-13)
Approved for use through 3/31/2013. OMB 0651-0031
U.S. Patent and Trademark Office; U.S. DEPARTMENT OF COMMERCE
Under the Paperwork Reduction Act of 1995, no persons are required to respond to a collection of information unless it displays a valid OMB control number.

Docket Number (Optional)					mber (Optional)		
PETITION FOR EXTENSION	ER 37 CFR 1	.136(a)	6936-47	7-CON-5			
Application Number 13/303,417	Filed Nove	mber 2	23, 201	11			
For SYSTEM AND METHOD FOR SCRAMB	LING THE PHASE O	F THE CARRIERS	IN A MULTIC	CARRIER CO	OMMUNICATIONS SYSTEM		
Art Unit 2634		Examiner PH	U, Phu	iong M			
This is a request under the provisions of 37 CF	R 1.136(a) to extend	the period for filing a	a reply in the	above-ident	ified application.		
The requested extension and fee are as follow	s (check time period c	desired and enter the	appropriate	fee below):			
	Fee Sm	all Entity Fee	Micro Enti	ty Fee			
One month (37 CFR 1.17(a)(1))	\$200	\$100	\$50		\$		
Two months (37 CFR 1.17(a)(2))	\$600	\$300	\$150)	\$		
✓ Three months (37 CFR 1.17(a)(3))	\$1,400	\$700	\$350)	_{\$} 1400		
Four months (37 CFR 1.17(a)(4))	\$2,200	\$1,100	\$550)	\$		
Five months (37 CFR 1.17(a)(5))	\$3,000	\$1,500	\$750)	\$		
Applicant asserts small entity status.	Applicant asserts small entity status. See 37 CFR 1.27.						
Applicant certifies micro entity status.	See 37 CFR 1.29.						
Form PTO/SB/15A or B or equivalent must	either be enclosed or ha	ve been submitted pre-	viously.				
A check in the amount of the fee is enclosed.							
Payment by credit card. Form PTO-2038 is attached.							
The Director has already been author							
The Director is hereby authorized to charge any fees which may be required, or credit any overpayment, to							
Deposit Account Number 19-1970							
Payment made via EFS-Web.							
WARNING: Information on this form may b credit card information and authorization or	ecome public. Credit n PTO-2038.	card information s	should not b	e included	on this form. Provide		
I am the							
applicant/inventor.							
assignee of record of the er			.73(b) statem	nent is enclo	sed (Form PTO/SB/96).		
\checkmark attorney or agent of record. Registration number $45,285$							
attorney or agent acting und	der 37 CFR 1-34. Reg	istration number			· · · · · · · · · · · · · · · · · · ·		
		_ 27	1)00 /-	3			
Signature				Date			
Jason H. Vick		303-863-9					
Typed or printed name Telephone Number NOTE: This form must be signed in accordance with 37 CFR 1.33. See 37 CFR 1.4 for signature requirements and certifications, Submit							
multiple forms if more than one signature is required, see below*. representation of 1 forms are submitted.							

This collection of information is required by 37 CFR 1.136(a). The information is required to obtain or retain a benefit by the public, which is to file (and by the USPTO to process) an application. Confidentiality is governed by 35 U.S.C. 122 and 37 CFR 1.11 and 1.14. This collection is estimated to take 6 minutes to complete, including gathering, preparing, and submitting the completed application form to the USPTO. Time will vary depending upon the individual case. Any comments on the amount of time you require to complete this form and/or suggestions for reducing this burden should be sent to the Chief Information Officer, U.S. Patent and Trademark Office, U.S. Department of Commerce, P.O. Box 1450, Hayandria, VA 22313-1450. DO NOT SEND FEES OR COMPLETED FORMS TO THIS ADDRESS. SEND TO: Commissioner for Patents, P.O. Box 1450, Alexandria, VA 22313-1450.

If you need assistance in completing the form, call 1-800-PTO-9199 and select option 2.

Electronic Patent Application Fee Transmittal					
Application Number:	ication Number: 13303417				
Filing Date:	23-Nov-2011				
Title of Invention: SYSTEM AND METHOD FOR SCRAMBLING THE PHASE OF THE CARRIE MULTICARRIER COMMUNICATIONS SYSTEM		F THE CARRIERS IN A			
First Named Inventor/Applicant Name:	Marcos C. Tzannes				
Filer:	Jason Vick/Joanne Vos				
Attorney Docket Number: 6936-47-CON-5					
Filed as Large Entity					
Utility under 35 USC 111(a) Filing Fees					
Description		Fee Code	Quantity	Amount	Sub-Total in USD(\$)
Basic Filing:					
Pages:					
Claims:					
Miscellaneous-Filing:					
Petition:					
Patent-Appeals-and-Interference:					
Post-Allowance-and-Post-Issuance:					
Extension-of-Time:					
Extension - 3 months with \$0 paid 1253 1 1400 1400					1400

Description	Fee Code	Quantity	Amount	Sub-Total in USD(\$)
Miscellaneous:				
	Tot	al in USD	(\$)	1400

Electronic Acknowledgement Receipt				
EFS ID:	17771466			
Application Number:	13303417			
International Application Number:				
Confirmation Number:	3035			
Title of Invention:	SYSTEM AND METHOD FOR SCRAMBLING THE PHASE OF THE CARRIERS IN A MULTICARRIER COMMUNICATIONS SYSTEM			
First Named Inventor/Applicant Name:	Marcos C. Tzannes			
Customer Number:	62574			
Filer:	Jason Vick/Joanne Vos			
Filer Authorized By:	Jason Vick			
Attorney Docket Number:	6936-47-CON-5			
Receipt Date:	27-DEC-2013			
Filing Date:	23-NOV-2011			
Time Stamp:	18:22:22			
Application Type:	Utility under 35 USC 111(a)			

Payment information:

Submitted with Payment	yes
Payment Type	Deposit Account
Payment was successfully received in RAM	\$1400
RAM confirmation Number	3685
Deposit Account	191970
Authorized User	

The Director of the USPTO is hereby authorized to charge indicated fees and credit any overpayment as follows:

Charge any Additional Fees required under 37 C.F.R. Section 1.16 (National application filing, search, and examination fees)

Charge any Additional Fees required under 37 C.F.R. Section 1.17 (Patent application and reexamination processing fees)

Charge any Additional Fees required under 37 C.F.R. Section 1.19 (Document supply fees)

Charge any Additional Fees required under 37 C.F.R. Section 1.21 (Miscellaneous fees and charges)

File Listing:

,					
Document Number	Document Description	File Name	File Size(Bytes)/ Message Digest	Multi Part /.zip	Pages (if appl.)
1		AMEND_01.pdf	1239160	yes	10
·		, <u>-</u>	92eedf2be5e19cf91f74acc401711f78c148d 999	,	
	Multip	part Description/PDF files in	zip description		
	Document De	scription	Start	E	nd
	Amendment/Req. Reconsiderat	1		1	
	Claims	2		7	
	Applicant Arguments/Remarks	Made in an Amendment	8		9
	Extension o	fTime	10	10 10	
Warnings:					
Information:					
2	Fee Worksheet (SB06) fee-info.pdf -		30707	no	2
			61739572fed80188d7b4072bcf9673dd4b2 0a476		
Warnings:					
Information:					
		Total Files Size (in bytes): 12	69867	

This Acknowledgement Receipt evidences receipt on the noted date by the USPTO of the indicated documents, characterized by the applicant, and including page counts, where applicable. It serves as evidence of receipt similar to a Post Card, as described in MPEP 503.

New Applications Under 35 U.S.C. 111

If a new application is being filed and the application includes the necessary components for a filing date (see 37 CFR 1.53(b)-(d) and MPEP 506), a Filing Receipt (37 CFR 1.54) will be issued in due course and the date shown on this Acknowledgement Receipt will establish the filing date of the application.

National Stage of an International Application under 35 U.S.C. 371

If a timely submission to enter the national stage of an international application is compliant with the conditions of 35 U.S.C. 371 and other applicable requirements a Form PCT/DO/EO/903 indicating acceptance of the application as a national stage submission under 35 U.S.C. 371 will be issued in addition to the Filing Receipt, in due course.

New International Application Filed with the USPTO as a Receiving Office

If a new international application is being filed and the international application includes the necessary components for an international filing date (see PCT Article 11 and MPEP 1810), a Notification of the International Application Number and of the International Filing Date (Form PCT/RO/105) will be issued in due course, subject to prescriptions concerning national security, and the date shown on this Acknowledgement Receipt will establish the international filing date of the application.

Under the Panerwork Reduction Act of 1995, no persons are required to respond to a collection of information unless it displays a valid OMR

P	PATENT APPLICATION FEE DETERMINATION RECORD Substitute for Form PTO-875						or Docket Number /303,417	Filing Date 11/23/2011	To be Mailed
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				APPLICA	ATION AS FIL	ED – PAR	TI		
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	FOR	N	IUMBER FI	_ED	NUMBER EXTRA		RATE (\$)	F	EE (\$)
	BASIC FEE (37 CFR 1.16(a), (b),	or (c))	N/A		N/A		N/A		
	SEARCH FEE (37 CFR 1.16(k), (i), (or (m))	N/A		N/A		N/A		
	EXAMINATION FE (37 CFR 1.16(o), (p),		N/A		N/A		N/A		
	TAL CLAIMS CFR 1.16(i))		mir	nus 20 = *			X \$ =		
	EPENDENT CLAIM CFR 1.16(h))	S	m	inus 3 = *			X \$ =		
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* If t	he difference in colu	ımn 1 is less than	zero, ente	r "0" in column 2.			TOTAL		
		(Column 1)		(Column 2)	ON AS AMEN		ART II		
INT.	12/27/2013	CLAIMS REMAINING AFTER AMENDMENT		HIGHEST NUMBER PREVIOUSLY PAID FOR	PRESENT EX	TRA	RATE (\$)	A DDITK	DNAL FEE (\$)
AMENDMENT	Total (37 CFR 1.16(i))	* 30	Minus	** 31	= 0		x \$80 =		0
EN I	Independent (37 CFR 1.16(h))	* 2	Minus	***3	= 0		× \$420 =		0
AM	Application Si	ze Fee (37 CFR	1.16(s))						
	FIRST PRESEN	ITATION OF MULTI	PLE DEPEN	DENT CLAIM (37 CFF	R 1.16(j))				
							TOTAL ADD'L FE	E	0
		(Column 1)		(Column 2)	(Column 3)			
		CLAIMS REMAINING AFTER AMENDMENT		HIGHEST NUMBER PREVIOUSLY PAID FOR	PRESENT EX	TRA	RATE (\$)	ADDITIO	ONAL FEE (\$)
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This collection of information is required by 37 CFR 1.16. The information is required to obtain or retain a benefit by the public which is to file (and by the USPTO to process) an application. Confidentiality is governed by 35 U.S.C. 122 and 37 CFR 1.14. This collection is estimated to take 12 minutes to complete, including gathering, preparing, and submitting the completed application form to the USPTO. Time will vary depending upon the individual case. Any comments on the amount of time you require to complete this form and/or suggestions for reducing this burden, should be sent to the Chief Information Officer, U.S. Patent and Trademark Office, U.S. Department of Commerce, P.O. Box 1450, Alexandria, VA 22313-1450. DO NOT SEND FEES OR COMPLETED FORMS TO THIS ADDRESS. SEND TO: Commissioner for Patents, P.O. Box 1450, Alexandria, VA 22313-1450.

If you need assistance in completing the form, call 1-800-PTO-9199 and select option 2.

Sub	Substitute for form 1449A/PTO			Сотр	olete if Known
18.1		TION DIO		Application Number	13/303,417
INFORMATION DISCLOSURE				Filing Date	November 23, 2011
S	STATEMENT BY APPLICANT			First Named Inventor	Marcos C. Tzannes
				Art Unit	2634
		Examiner Name	PHU, Phuong M.		
Sheet	1	of	1	Attorney Docket Number	6936-47-CON-5

	U.S. PATENT DOCUMENTS							
Examiner Initials* No.1 Document Number Number-kind Code 2 (If known) Publication Date MM-DD-YYYY Number-kind Code 2 (If known) Publication Date Applicant of Cited Document Relevant Passages or Rele								
	1	5903614	05-11-1999	Suzuki et al.				
	2	5914932	06-22-1999	Suzuki et al.				

FOREIGN PATENT DOCUMENTS						
Examiner Initials*	Cite No. ¹	Foreign Patent Document Country Code ³ ; Number ⁴ ; Kind Code ⁵ (if known)		Name of Patentee or Applicant of Cited Document	Pages, Columns, Lines, Where Relevant Passages or Relevant Figures Appear	T⁵

OTHER ART (Including Author, Title, Date, Pertinent Pages, etc.)							
Examiner Initials*	Cite No. ¹						
		Official Action for U.S. Patent Application No. 13/718,016, mailed May 9, 2013 (Attorney Ref. No.: 6936-47-CON-DIV-CON-3)					

Examiner	Date	
Examinor	Date	
I Signature I	I Consider	ed l
- 3	 Contider	74

*EXAMINER: Initial if reference is considered, whether or not citation is in conformance and not considered. Include copy of this form with next communication to applicant.

IN THE UNITED STATES PATENT AND TRADEMARK OFFICE

In Re the Application of:	Group Art Unit:	2634
TZANNES, Marcos C.	Confirmation No	o.: 3035
Serial No.: 13/303,417	Examiner: PHU	J, Phuong M.
Filed: November 23, 2011		
Atty. File No.: 6936-47-CON-5		EMENTAL NA PIGGL OGLIDE
Entitled: "System and Method for Scrambling the		ON DISCLOSURE TEMENT
Phase of the Carriers in a Multicarrier Communications System"		ally Submitted
	2,000	
Commissioner for Patents		
P.O. Box 1450 Alexandria, VA 22313-1450		
Alexandria, VA 22313-1430		
Dear Sir:		
The references cited on attached Form PTC	1449 are being calle	ed to the attention
of the Examiner.		
Copies of the cited non-patent and/or foreign	eferences are enclos	sed herewith.
Copies of the cited U.S. patents and/or paten	applications are encl	losed herewith.
Copies of the cited U.S. patents/patent appli	ation publications ar	re not enclosed in
accordance with 37 C.F.R. § 1.98(a).		
Copies of the cited references are not end	osed, in accordance	e with 37 C.F.R.
§ 1.98(d), because the references were cited by	r submitted to the	U.S. Patent and
Trademark Office in prior application Serial No	filed	,
which is relied upon for an earlier filing date under	U.S.C. § 120.	
To the best of applicants' belief, the pertine	ce of the foreign-lar	nguage references
are believed to be summarized in the attached E	glish abstracts and/	or in the figures,
although applicants do not necessarily vouch for the	ccuracy of the trans	lation.
Examiner's attention is drawn to the following	grelated applications	3:
• Serial No. <u>13/718,016</u> filed <u>Dec. 18, 20</u>	2 (Attorney Ref. N	o. <u>6936-47-CO</u> N-
DIV-CON-3)		
Other:		
Submission of the above information is not	tended as an admiss	sion that any item
is citable under the statutes or rules to support		
represents analogous art, or that those skilled in the	art would refer to	or recognize the
pertinence of any reference without the benefit of		

drawn as to the pertinence of the references based on the order in which they are presented. Submission of this statement should not be taken as an indication that a search has been conducted, or that no better art exists.

It is respectfully requested that the cited information be expressly considered during the prosecution of this application and the references made of record therein.

FEES

37 CFR 1.97(b): No fee is believed due in connection with this submission, because the information disclosure statement submitted herewith is satisfied by one of the following conditions ("X" indicates satisfaction): Within three months of the filing date of a national application other than a continued prosecution application under 37 CFR 1.53(d), or Within three months of the date of entry of the national stage as set forth in § 1.491 in an international application, or Before the mailing date of a first Office Action on the merits, or Before the mailing of a first Office action after the filing of a request for continued examination under 37 CFR 1.114. Although no fee is believed due, if any fee is deemed due in connection with this submission, please charge such fee to Deposit Account 19-1970.
37 CFR 1.97(c): The information disclosure statement transmitted herewith is being filed after all the above conditions (37 CFR 1.97(b)), but before the mailing date of any one of the following conditions: (1) a final action under 37 C.F.R. 1.113, or (2) a notice of allowance under 37 C.F.R. 1.311, or (3) an action that otherwise closes prosecution in the application. This Information Disclosure Statement is accompanied by: A Certification (below) as specified by 37 C.F.R. 1.97(e). Although no fee is believed due, if any fee is deemed due in connection with this submission, please charge such fee to Deposit Account 19-1970. OR Please charge Deposit Account 19-1970 in the amount of \$180.00 for the fee set forth in 37 C.F.R. 1.17(p) for submission of an information disclosure statement. Please credit any overpayment or charge any underpayment to Deposit Account 19-1970.
37 CFR 1.97(d): This Information Disclosure Statement is being submitted after the period specified in 37 CFR 1.97(e). This information Disclosure Statement includes a Certification (below) as specified by 37 C.F.R. 1.97(e) AND Applicants hereby requests consideration of the reference(s) disclosed herein. Please charge Deposit Account 19-1970 in the amount of \$180.00 under 37 C.F.R. 1.17(p). Please credit any overpayment or charge any underpayment to Deposit Account 19-1970. Election to pay the fee should not be taken as an indication that applicant(s) cannot execute a certification.

	Certification (37 C.F.R. 1.97(e)) (Applicable only if checked)
	The undersigned certifies that: Each item of information contained in this information disclosure statement was first cited in any communication from a foreign patent office in a counterpart foreign application not more than three months prior to the filing of this statement. 37 C.F.R. 1.97(e)(1). A copy of the communication from the foreign patent office is enclosed.
	OR
	No item of information contained in this information disclosure statement was cited in a communication from a foreign patent office in a counterpart foreign application, and, to the knowledge of the undersigned after making reasonable inquiry, no item of information contained in this Information Disclosure Statement was known to any individual designated in 37 C.F.R. 1.56(c) more than three months prior to the filing of this statement. 37 C.F.R. 1.97(e)(2).
	Respectfully submitted,
	SHERIDAN ROSS P.C.
Date:	By: Jason H. Vick Registration No. 45,285 1560 Broadway, Suite 1200 Denver, Colorado 80202-5141 (303) 863-9700

Sub	Substitute for form 1449A/PTO			Complete if Known		
18.1		TION DIOC	N OOUDE	Application Number	13/303,417	
			CLOSURE	Filing Date	November 23, 2011	
SI	STATEMENT BY APPLICANT			First Named Inventor	Marcos C. Tzannes	
				Art Unit	2634	
				Examiner Name	PHU, Phuong M.	
Sheet	1	of	1	Attorney Docket Number	6936-47-CON-5	

U.S. PATENT DOCUMENTS								
Examiner Cite Document Number No.1 Number-kind Code 2 (It known)		Publication Date MM-DD-YYYY	Name of Patentee of Applicant of Cited Document	Pages, Columns, Lines, Where Relevant Passages or Relevant Figures Appear				
	1	2013/0136160	05-30-2013	Tzannes				

FOREIGN PATENT DOCUMENTS								
Examiner Initials*		Foreign Patent Document Country Code ³ ; Number ⁴ ; Kind Code ⁵ (if known)		Name of Patentee or Applicant of Cited Document	Pages, Columns, Lines, Where Relevant Passages or Relevant Figures Appear	T ⁶		

		OTHER ART (Including Author, Title, Date, Pertinent Pages, etc.)	
Examiner Initials*	Cite No. ¹		

		l
Examiner	Date	
Signature	Considered	

*EXAMINER: Initial if reference is considered, whether or not citation is in conformance and not considered. Include copy of this form with next communication to applicant.

Electronic Patent Application Fee Transmittal							
Application Number:	13303417						
Filing Date:	23-Nov-2011						
Title of Invention:		SYSTEM AND METHOD FOR SCRAMBLING THE PHASE OF THE CARRIERS IN A MULTICARRIER COMMUNICATIONS SYSTEM					
First Named Inventor/Applicant Name:	Ma	rcos C. Tzannes					
Filer:	Jas	on Vick/Joanne Vos	5				
Attorney Docket Number:	69:	36-47-CON-5					
Filed as Large Entity							
Utility under 35 USC 111(a) Filing Fees							
Description		Fee Code	Quantity	Amount	Sub-Total in USD(\$)		
Basic Filing:							
Pages:							
Claims:	Claims:						
Miscellaneous-Filing:							
Petition:							
Patent-Appeals-and-Interference:							
Post-Allowance-and-Post-Issuance:							
Extension-of-Time:							

Description	Fee Code	Quantity	Amount	Sub-Total in USD(\$)
Miscellaneous:				
Submission-Information Disclosure Stmt	1806	1	180	180
	Total in USD (\$)			180

Electronic Acknowledgement Receipt			
EFS ID:	16498509		
Application Number:	13303417		
International Application Number:			
Confirmation Number:	3035		
Title of Invention:	SYSTEM AND METHOD FOR SCRAMBLING THE PHASE OF THE CARRIERS IN A MULTICARRIER COMMUNICATIONS SYSTEM		
First Named Inventor/Applicant Name:	Marcos C. Tzannes		
Customer Number:	62574		
Filer:	Jason Vick/Joanne Vos		
Filer Authorized By:	Jason Vick		
Attorney Docket Number:	6936-47-CON-5		
Receipt Date:	05-AUG-2013		
Filing Date:	23-NOV-2011		
Time Stamp:	13:23:15		
Application Type:	Utility under 35 USC 111(a)		

Payment information:

Submitted with Payment	yes
Payment Type	Deposit Account
Payment was successfully received in RAM	\$180
RAM confirmation Number	10832
Deposit Account	191970
Authorized User	

The Director of the USPTO is hereby authorized to charge indicated fees and credit any overpayment as follows:

Charge any Additional Fees required under 37 C.F.R. Section 1.16 (National application filing, search, and examination fees)

Charge any Additional Fees required under 37 C.F.R. Section 1.17 (Patent application and reexamination processing fees)

Charge any Additional Fees required under 37 C.F.R. Section 1.19 (Document supply fees)

Charge any Additional Fees required under 37 C.F.R. Section 1.21 (Miscellaneous fees and charges)

File Listing:

Document Number	Document Description	File Name	File Size(Bytes)/ Message Digest	Multi Part /.zip	Pages (if appl.)	
1		IDS_05_Certified.pdf	343294	yes	Vac	4
'		103_03_certified.pdf	4162d418ffb83af28e71c35686bdd82cb39a e741		4	
	Multip	part Description/PDF files in .	zip description			
	Document De	scription	Start	Ei	nd	
	Transmittal	Letter	1	3		
	Information Disclosure Stater	ment (IDS) Form (SB08)	4		4	
Warnings:						
Information:						
2	Non Patent Literature	6936-47-CON-DIV-	260677	no	7	
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	Transmittal	Letter	1		3	
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Information:						
4	Fee Worksheet (SB06)	fee-info.pdf	30683	no	2	
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Warnings:						
Information:						
		Total Files Size (in bytes)	96	59778		

This Acknowledgement Receipt evidences receipt on the noted date by the USPTO of the indicated documents, characterized by the applicant, and including page counts, where applicable. It serves as evidence of receipt similar to a Post Card, as described in MPEP 503.

New Applications Under 35 U.S.C. 111

If a new application is being filed and the application includes the necessary components for a filing date (see 37 CFR 1.53(b)-(d) and MPEP 506), a Filing Receipt (37 CFR 1.54) will be issued in due course and the date shown on this Acknowledgement Receipt will establish the filing date of the application.

National Stage of an International Application under 35 U.S.C. 371

If a timely submission to enter the national stage of an international application is compliant with the conditions of 35 U.S.C. 371 and other applicable requirements a Form PCT/DO/EO/903 indicating acceptance of the application as a national stage submission under 35 U.S.C. 371 will be issued in addition to the Filing Receipt, in due course.

New International Application Filed with the USPTO as a Receiving Office

If a new international application is being filed and the international application includes the necessary components for an international filing date (see PCT Article 11 and MPEP 1810), a Notification of the International Application Number and of the International Filing Date (Form PCT/RO/105) will be issued in due course, subject to prescriptions concerning national security, and the date shown on this Acknowledgement Receipt will establish the international filing date of the application.

IN THE UNITED STATES PATENT AND TRADEMARK OFFICE

In Re the Application of:) Group Art Unit: 2634
TZANNES, Marcos C.	Confirmation No.: 3035
Serial No.: 13/303,417	Examiner: PHU, Phuong M.
Filed: November 23, 2011	
Atty. File No.: 6936-47-CON-5) SUPPLEMENTAL) INFORMATION DISCLOSURE
Entitled: "System and Method for Scrambling the Phase of the Carriers in a Multicarrier Communications System") INFORMATION DISCLOSURE STATEMENT) Electronically Submitted
Commissioner for Patents P.O. Box 1450 Alexandria, VA 22313-1450 Dear Sir:	
The references cited on attached Form PTC	1440 are being called to the attention
of the Examiner.	5-1449 are being cance to the attention
Copies of the cited non-patent and/or foreign Copies of the cited U.S. patents and/or patent Copies of the cited U.S. patents/patent appl accordance with 37 C.F.R. § 1.98(a). Copies of the cited references are not en	at applications are enclosed herewith.
§ 1.98(d), because the references were cited by	or submitted to the U.S. Patent and
Trademark Office in prior application Serial No	
which is relied upon for an earlier filing date under	
To the best of applicants' belief, the pertine	
are believed to be summarized in the attached E	-
although applicants do not necessarily vouch for the	· ·
Examiner's attention is drawn to the followi	
• Serial No. <u>13/718,016</u> filed <u>Dec. 18, 20</u>	112 (Attorney Ref. No. <u>6936-47-CON-</u>
DIV-CON-3)	
Submission of the above information is not	intended as an admission that any item
is citable under the statutes or rules to support	•
represents analogous art, or that those skilled in t	
pertinence of any reference without the benefit of	-

drawn as to the pertinence of the references based on the order in which they are presented. Submission of this statement should not be taken as an indication that a search has been conducted, or that no better art exists.

It is respectfully requested that the cited information be expressly considered during the prosecution of this application and the references made of record therein.

FEES

37 CFR 1.97(b): No fee is believed due in connection with this submission, because the information disclosure statement submitted herewith is satisfied by one of the following conditions ("X" indicates satisfaction): Within three months of the filing date of a national application other than a continued prosecution application under 37 CFR 1.53(d), or Within three months of the date of entry of the national stage as set forth in § 1.491 in an international application, or Before the mailing date of a first Office Action on the merits, or Before the mailing of a first Office action after the filing of a request for continued examination under 37 CFR 1.114. Although no fee is believed due, if any fee is deemed due in connection with this submission, please charge such fee to Deposit Account 19-1970.
37 CFR 1.97(c): The information disclosure statement transmitted herewith is being filed after all the above conditions (37 CFR 1.97(b)), but before the mailing date of any one of the following conditions: (1) a final action under 37 C.F.R. 1.113, or (2) a notice of allowance under 37 C.F.R. 1.311, or (3) an action that otherwise closes prosecution in the application. This Information Disclosure Statement is accompanied by: A Certification (below) as specified by 37 C.F.R. 1.97(e). Although no fee is believed due, if any fee is deemed due in connection with this submission, please charge such fee to Deposit Account 19-1970. OR Please charge Deposit Account 19-1970 in the amount of \$180.00 for the fee set forth in 37 C.F.R. 1.17(p) for submission of an information disclosure statement. Please credit any overpayment or charge any underpayment to Deposit Account 19-1970.
37 CFR 1.97(d): This Information Disclosure Statement is being submitted after the period specified in 37 CFR 1.97(c). This information Disclosure Statement includes a Certification (below) as specified by 37 C.F.R. 1.97(e) AND Applicants hereby requests consideration of the reference(s) disclosed herein. Please charge Deposit Account 19-1970 in the amount of \$180.00 under 37 C.F.R. 1.17(p). Please credit any overpayment or charge any underpayment to Deposit Account 19-1970. Election to pay the fee should not be taken as an indication that applicant(s) cannot execute a certification.

	Certification (37 C.F.R. 1.97(e)) (Applicable only if checked)
	The undersigned certifies that: Each item of information contained in this information disclosure statement was first cited in any communication from a foreign patent office in a counterpart foreign application not more than three months prior to the filing of this statement. 37 C.F.R. 1.97(e)(1). A copy of the communication from the foreign patent office is enclosed.
	OR
	No item of information contained in this information disclosure statement was cited in a communication from a foreign patent office in a counterpart foreign application, and, to the knowledge of the undersigned after making reasonable inquiry, no item of information contained in this Information Disclosure Statement was known to any individual designated in 37 C.F.R. 1.56(c) more than three months prior to the filing of this statement. 37 C.F.R. 1.97(e)(2).
	Respectfully submitted,
	SHERIDAN ROSS P.C.
Date:	By: Jason H. Vick Registration No. 45,285 1560 Broadway, Suite 1200 Denver, Colorado 80202-5141 (303) 863-9700

UNITED STATES DEPARTMENT OF COMMERCE United States Patent and Trademark Office Address: COMMISSIONER FOR PATENTS P.O. Box 1450 Alexandria, Virginia 22313-1450 www.usplo.gov

APPLICATION NO.	FILING DATE	FIRST NAMED INVENTOR	ATTORNEY DOCKET NO.	CONFIRMATION NO.
13/303,417	11/23/2011	Marcos C. Tzannes	6936-47-CON-5	3035
62574 Jason H. Vick	7590 07/05/20:	13	EXAM	IINER
Sheridan Ross, Suite # 1200	PC		PHU, PH	UONG M
1560 Broadway	•		ART UNIT	PAPER NUMBER
Denver, CO 802	202		2634	_
			NOTIFICATION DATE	DELIVERY MODE
			07/05/2013	ELECTRONIC

Please find below and/or attached an Office communication concerning this application or proceeding.

The time period for reply, if any, is set in the attached communication.

Notice of the Office communication was sent electronically on above-indicated "Notification Date" to the following e-mail address(es):

jvick@sheridanross.com

Application No. 13/303,417 Applicant(s) TZANNES, MARCOS C.					
Office Action Summary	Examiner PHUONG PHU	Art Unit 2634	AIA (First Inventor to File) Status No		
The MAILING DATE of this communication Period for Reply	n appears on the cover sheet with	the corresponder	nce address		
A SHORTENED STATUTORY PERIOD FOR RIWHICHEVER IS LONGER, FROM THE MAILIN. Extensions of time may be available under the provisions of 37 CF after SIX (6) MONTHS from the mailing date of this communicatio. If NO period for reply is specified above, the maximum statutory propulation of the provision of the provi	G DATE OF THIS COMMUNIC, FR 1.136(a). In no event, however, may a rep. n. eriod will apply and will expire SIX (6) MONT! statute, cause the application to become ABA	ATION. bly be timely filed HS from the mailing date NDONED (35 U.S.C. § 1	of this communication.		
Status					
1) Responsive to communication(s) filed on a A declaration(s)/affidavit(s) under 37 CFF		<u></u>			
·—	This action is non-final.				
3) An election was made by the applicant in I			ing the interview on		
the restriction requirement and ele Since this application is in condition for all closed in accordance with the practice und	owance except for formal matte	rs, prosecution as			
Disposition of Claims					
5) Claim(s) 21-51 is/are pending in the applic 5a) Of the above claim(s) is/are with 6) Claim(s) is/are allowed. 7) Claim(s) 21-51 is/are rejected. 8) Claim(s) is/are objected to. 9) Claim(s) are subject to restriction a * If any claims have been determined allowable, you may participating intellectual property office for the corresponding http://www.uspto.gov/patents/init_events/pph/index.jsp or Application Papers	ndrawn from consideration. Ind/or election requirement. be eligible to benefit from the Pater ing application. For more informatio	n, please see	hway program at a		
 10) The specification is objected to by the Examination 11) The drawing(s) filed on <u>23 November 2011</u> Applicant may not request that any objection to 	<u>1</u> is/are: a) \boxtimes accepted or b) \square				
Replacement drawing sheet(s) including the co	= , ,				
Priority under 35 U.S.C. § 119 12) Acknowledgment is made of a claim for for Certified copies: a) All b) Some * c) None of the: 1. Certified copies of the priority docured copies of the priority docured copies of the priority docured copies of the certified copies of the application from the International But * See the attached detailed Office action for a light copies of the action for a light copies of the attached detailed Office action for a light copies of the attache	ments have been received. ments have been received in Ap e priority documents have been ureau (PCT Rule 17.2(a)).	oplication No received in this Na	 ational Stage		
Attachment(s) 1) ☑ Notice of References Cited (PTO-892)	3)	mmary (PTO-413)			
Information Disclosure Statement(s) (PTO/SB/08) Paper No(s)/Mail Date <u>See Continuation Sheet</u> .	· · · · · · · · · · · · · · · · · · ·	Mail Date			

U.S. Patent and Trademark Office PTOL-326 (Rev. 05-13)

Office Action Summary

Part of Paper No./Mail Date 20130628

Application No. 13/303,417

 $\label{localization} Continuation of Attachment (s) 2). Information Disclosure Statement (s) (PTO/SB/08), Paper No(s)/Mail Date :12/08/11,03/12/12,03/22/12,10/30/12.$

Application/Control Number: 13/303,417 Page 2

Art Unit: 2634

DETAILED ACTION

1. This Office Action is responsive to the Preliminary Amendment filed on 06/21/12. Accordingly, claims 21-51 are currently pending; and claims 1-20 are canceled.

Claim Rejections - 35 USC § 112

The following is a quotation of 35 U.S.C. 112(b):
 (b) CONCLUSION.—The specification shall conclude with one or more claims particularly pointing out and distinctly claiming the subject matter which the inventor or a joint inventor regards as the invention.

The following is a quotation of 35 U.S.C. 112 (pre-AIA), second paragraph: The specification shall conclude with one or more claims particularly pointing out and distinctly claiming the subject matter which the applicant regards as his invention.

3. Claims 21-51 are rejected under 35 U.S.C. 112(b) or 35 U.S.C. 112 (pre-AIA), second paragraph, as being indefinite for failing to particularly point out and distinctly claim the subject matter which the inventor or a joint inventor, or for pre-AIA the applicant regards as the invention.

Claim 21, line 6, recites the limitation "transmitting the bit stream". It is unclear whether the "the bit stream" in the limitation refers to the "input bit stream" (see line 3) prior or after being modulated via "modulating". It is herein suggested that the limitation is changed to -- transmitting the modulated bit stream--.

Similarly, in claim 35, line 5, the limitation "the bit stream" is herein suggested to be changed to -- the modulated bit stream--.

Similarly, in claim 51, line 6, the limitation "the bit stream" is herein suggested to be changed to -- the modulated bit stream--.

Claim 21, lines 7, 10, 12, recites the limitation "each carrier signal". It is unclear whether the limitation refers each carrier signal of the "plurality of carrier signals", previously-cited on

Art Unit: 2634

lines 2-3 of the claim. It is herein suggested that the limitation is changed to -- each carrier signal of the plurality of carrier signals--.

Similarly, claim 35, lines 6, 9, 11, recites the limitation "each carrier signal". It is unclear whether the limitation refers each carrier signal of the "plurality of carrier signals", previously-cited on lines 2-3 of the claim. It is herein suggested that the limitation is changed to -- each carrier signal of the plurality of carrier signals--.

Similarly, claim 51, line 7, 10, 12, recites the limitation "each carrier signal". It is unclear whether the limitation refers each carrier signal of the "plurality of carrier signals", previously-cited on lines 2-3 of the claim. It is herein suggested that the limitation is changed to - each carrier signal of the plurality of carrier signals—.

Claim 35 claims a "multicarrier modulation system". The claim further recites an element "a pseudo-random number generator" (see line 8). It is unclear in the claim on whether said element is included in the claimed "multicarrier modulation system". It is herein suggested that the element is changed to -- a pseudo-random number generator included in the multicarrier modulation system--.

Claim 35 further recites a procedure "a phase shift for each carrier signal" (see line 9). It is unclear in the claim on whether said procedure is performed by the claimed "multicarrier modulation system". It is herein suggested that the procedure is changed to -- a phase shift for each carrier signal, performed by the multicarrier modulation system,--.

Claim 21 claims a "method for scrambling". The claim further recites a procedure "a phase shift for each carrier signal" (see line 10). It is unclear in the claim on whether said

Page 3

Art Unit: 2634

procedure is a step performed by the claimed method. It is herein suggested that the procedure is changed to -- a phase shift for each carrier signal, performed by the method,--.

Similarly, claim 51 claims a "method for scrambling". The claim further recites a procedure "a phase shift for each carrier signal" (see line 10). It is unclear in the claim on whether said procedure is a step performed by the claimed method. It is herein suggested that the procedure is changed to -- a phase shift for each carrier signal, performed by the method,--.

Claims (if any), depended on above claims, are therefore also rejected.

4. Claims 21-34, 49 and 51 are rejected under 35 U.S.C. 112(b) or 35 U.S.C. 112 (pre-AIA), second paragraph, as being incomplete for omitting essential steps, such omission amounting to a gap between the steps. See MPEP § 2172.01.

Claim 21 claim a method for "scrambling the phase characteristics of the carrier signals".

Claim 21 omits essential step(s) showing operational/functional interrelationships of procedure

"transmitting the bit stream from the first transceiver to the second transceiver" with procedure

"scrambling the phase characteristics of the carrier signals" that the claim intends to claim. The

omissions render the claim vague on how the procedure "scrambling the phase characteristics of
the carrier signals" is obtained in term of the interrelationship with the procedure "transmitting
the bit stream from the first transceiver to the second transceiver".

Similarly, claim 51 claim a method for "scrambling the phase characteristics of the carrier signals". Claim 51 omits essential step(s) showing operational/functional interrelationships of procedure "transmitting the bit stream from the first transceiver" with procedure "scrambling the phase characteristics of the carrier signals" that the claim intends to claim. The omissions render the claim vague on how the procedure "scrambling the phase

Page 4

Application/Control Number: 13/303,417 Page 5

Art Unit: 2634

characteristics of the carrier signals" is obtained in term of the interrelationship with the procedure "transmitting the bit stream from the first transceiver".

Claims (if any), depended on above claims, are therefore also rejected.

Claim Rejections - 35 USC § 102

5. The following is a quotation of the appropriate paragraphs of pre-AIA 35 U.S.C. 102 that form the basis for the rejections under this section made in this Office action:

A person shall be entitled to a patent unless –

(e) the invention was described in (1) an application for patent, published under section 122(b), by another filed in the United States before the invention by the applicant for patent or (2) a patent granted on an application for patent by another filed in the United States before the invention by the applicant for patent, except that an international application filed under the treaty defined in section 351(a) shall have the effects for purposes of this subsection of an application filed in the United States only if the international application designated the United States and was published under Article 21(2) of such treaty in the English language.

6. Claims 21-25, 29-39, 43-51 are rejected under pre-AIA 35 U.S.C. 102(e) as being anticipated by Wu et al (2002/0172146).

-Regarding claim 21, Wu et al teaches in a multicarrier modulation system including a first transceiver (e.g., shown in figure 6) at one end of a communication with a second transceiver (inherently included at the other end of the communication) using a transmission signal having a plurality of carrier signals "sub-carriers" for modulating an input bit stream (outputted from (PC INTERFACE)), each carrier signal (inherently having its phase/frequency characteristic) associated with the input bit stream in such a way that the input stream is modulated to the plurality of carrier signals for transmissions, a method comprising: transmitting, via (AFE), the bit stream, after being modulated, from the first transceiver to the second transceiver, wherein: the first and second transceivers communicate over a pair of twisted wires

Art Unit: 2634

of a telephone subscriber line, and the first and second transceivers are configurable to transport video, (see [0017, 0043]).

Note that the clause "wherein each carrier signal is associated with a value determined independently of any bit value of the bit stream carried by that respective carrier signal, the value associated with each carrier signal determined by a pseudo-random number generator, a phase shift for each carrier signal is at least based on: the value associated with that carrier signal, and a combining of a phase for each carrier signal with the phase characteristic of that respective carrier signal so as to substantially scramble the phase characteristics of the plurality of carrier signals, and multiple carrier signals corresponding to the plurality of phase shifted and scrambled carrier signals are used by the first transceiver to modulate a same bit value of the bit stream", recited in the claim, is not given here any patentable weight over Wu et al because the clause merely states an application of "each carrier signal" and does not further limit the claimed method that Wu et al anticipated. See *Texas Instruments Inc. v. International Trade*Commission, 26 USPQ2d 1010 (Fed. Cir. 1993); Griffin v. Bertina, 62 USPQ2d 1431 (Fed. Cir. 2002); Amazon.com Inc. v. Barnesandnoble.com Inc., 57 USPQ2d 1747 (Fed. Cir. 2001).

-Regarding claim 22, as applied to claim 21, Wu et al teaches that one or more of the first transceiver and second transceiver are cable transceivers, (see [0041-0043]).

-Regarding claim 23, as applied to claim 21, Wu et al teaches that one or more of the first transceiver and second transceiver are DSL transceivers, (considered here equivalent with the limitation "VDSL transceivers", (see [0041-0043]).

Page 6

Art Unit: 2634

Page 7

-Regarding claim 24, as applied to claim 21, Wu et al teaches that one or more of the first transceiver and second transceiver are multicarrier DSL transceivers, see [0041-0043]).

-Regarding claim 25, Wu et al teaches that the first and second transceivers are usable for

high-speed network access, (considered here equivalent with the limitation "high speed internet

access"), (see [0017]).

-Claim 29 is rejected with similar reasons set forth for claim 22

-Regarding claim 30, Wu et al teaches that the first and the second transceivers are

configurable to be DSL transceivers connected using a pair of twisted wires of a telephone

subscriber system, (see [0041-0043]).

-Claim 31 is rejected with similar reasons set forth for claim 23.

-Claim 32 is rejected with similar reasons set forth for claim 24.

-Claim 33 is rejected with similar reasons set forth for claim 25.

-Regarding claim 34, Wu et al teaches that the first and the second transceivers are

configurable to include digital signal processors, (see [0018]).

-Regarding claim 49, in Wu et al, the video is video on demand, (see [0017]).

-Regarding claim 51, as applied to claim 21, Wu et al teaches a Wu et al teaches in a

multicarrier modulation system including a first transceiver (e.g., shown in figure 6) at one end

of a communication with a second transceiver (inherently included at the other end of the

Art Unit: 2634

communication) using a transmission signal having a plurality of carrier signals "sub-carriers" for modulating an input bit stream (outputted from (PC INTERFACE)), each carrier signal (inherently having its phase/frequency characteristic) associated with the input bit stream in such a way that the input stream is modulated to the plurality of carrier signals for transmissions, a method comprising: transmitting, via (AFE), the bit stream, after being modulated, from the first transceiver to the second transceiver, wherein: the first and second transceivers communicate over a pair of twisted wires of a telephone subscriber line, and the first and second transceivers are configurable to transport video, (see [0017, 0043]).

Note that the clause "wherein each carrier signal is associated with a value determined independently of any bit value of the bit stream carried by that respective carrier signal, the value associated with each carrier signal determined by a pseudo-random number generator, a phase shift for each carrier signal is at least based on: the value associated with that carrier signal, and a combining of a phase for each carrier signal with the phase characteristic of that respective carrier signal so as to substantially scramble the phase characteristics of the plurality of carrier signals, and multiple carrier signals corresponding to the plurality of phase shifted and scrambled carrier signals are used by the first transceiver to modulate a same bit value of the bit stream", recited in the claim, is not given here any patentable weight over Wu et al because the clause merely states an application of "each carrier signal" and does not further limit the claimed method that Wu et al anticipated. See *Texas Instruments Inc. v. International Trade*Commission, 26 USPQ2d 1010 (Fed. Cir. 1993); Griffin v. Bertina, 62 USPQ2d 1431 (Fed. Cir. 2002); Amazon.com Inc. v. Barnesandnoble.com Inc., 57 USPQ2d 1747 (Fed. Cir. 2001).

Page 8

Art Unit: 2634

-Regarding claim 35, as applied to claims 21-25, 28-34, 49 set forth above and herein incorporated, Wu et al teaches a multicarrier modulation system configurable of including a first transceiver (see figure 6) in communication with a second transceiver using a transmission signal having a plurality of carrier signals for modulating an input bit stream, each carrier signal having a phase characteristic associated with the input bit stream, the first transceiver capable of transmitting to the second transceiver the bit stream, wherein: the first and second transceivers communicate over one or more of a pair of twisted wires of a telephone subscriber line, quad cable and fiber optic, and the first and second transceivers transport video.

Note that the clause "wherein each carrier signal is associated with a value determined independently of any bit value of the bit stream carried by that respective carrier signal, the value associated with each carrier signal determined by a pseudo-random number generator, a phase shift for each carrier signal is at least based on: the value associated with that carrier signal, and a combining of a phase for each carrier signal with the phase characteristic of that respective carrier signal so as to substantially scramble the phase characteristics of the plurality of carrier signals, and multiple carrier signals corresponding to the plurality of phase shifted and scrambled carrier signals are used by the first transceiver to modulate a same bit value of the bit stream", recited in the claim, is not given here any patentable weight over Wu et al because the clause merely states an application of "each carrier signal" and does not further limit the claimed system that Wu et al anticipated. See *Texas Instruments Inc. v. International Trade*Commission, 26 USPQ2d 1010 (Fed. Cir. 1993); Griffin v. Bertina, 62 USPQ2d 1431 (Fed. Cir. 2002); Amazon.com Inc. v. Barnesandnoble.com Inc., 57 USPQ2d 1747 (Fed. Cir. 2001).

Page 9

Application/Control Number: 13/303,417 Page 10

Art Unit: 2634

-Claim 36 is rejected with similar reasons set forth for claim 22.

-Claim 37 is rejected with similar reasons set forth for claim 23.

-Claim 38 is rejected with similar reasons set forth for claim 24.

-Claim 39 is rejected with similar reasons set forth for claim 25.

-Claim 43 is rejected with similar reasons set forth for claim 22.

-Claim 44 is rejected with similar reasons set forth for claim 30.

-Claim 45 is rejected with similar reasons set forth for claim 23.

-Claim 46 is rejected with similar reasons set forth for claim 24.

-Claim 47 is rejected with similar reasons set forth for claim 25.

-Claim 48 is rejected with similar reasons set forth for claim 34.

-Claim 50 is rejected with similar reasons set forth for claim 49.

Conclusion

- 7. References 5742679 and 6781951 are additionally cited because they are pertinent to the claimed method/system.
- 8. Any inquiry concerning this communication or earlier communications from the examiner should be directed to PHUONG PHU whose telephone number is (571)272-3009. The examiner can normally be reached on M-F (8:00 AM 4:30 PM).

If attempts to reach the examiner by telephone are unsuccessful, the examiner's supervisor, Daniel Washburn can be reached on 571-272-5551. The fax phone number for the organization where this application or proceeding is assigned is 571-273-8300.

Application/Control Number: 13/303,417 Page 11

Art Unit: 2634

Information regarding the status of an application may be obtained from the Patent Application Information Retrieval (PAIR) system. Status information for published applications may be obtained from either Private PAIR or Public PAIR. Status information for unpublished applications is available through Private PAIR only. For more information about the PAIR system, see http://pair-direct.uspto.gov. Should you have questions on access to the Private PAIR system, contact the Electronic Business Center (EBC) at 866-217-9197 (toll-free). If you would like assistance from a USPTO Customer Service Representative or access to the automated information system, call 800-786-9199 (IN USA OR CANADA) or 571-272-1000.

Phuong Phu Primary Examiner Art Unit 2634

/Phuong Phu/ Primary Examiner, Art Unit 2634

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A copy of this reference is not being furnished with this Office action. (See MPEP § 707.05(a).) Dates in MM-YYYY format are publication dates. Classifications may be US or foreign.

U.S. Patent and Trademark Office PTO-892 (Rev. 01-2001)

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Part of Paper No. 20130628

Application/Control No. Search Notes 13303417 Examiner PHUONG PHU Applicant(s)/Patent Under Reexamination TZANNES, MARCOS C. Art Unit 2634

CPC- SEARCHED			
Symbol	Date	Examiner	

CPC COMBINATION SETS - SEARCHED		
Symbol	Date	Examiner

US CLASSIFICATION SEARCHED			
Class	Subclass	Date	Examiner
375	222, 219, 220, 295, 316, 259, 260	06/26/13	PP
370	203, 527, 208	06/26/13	PP

SEARCH NOTES		
Search Notes	Date	Examiner
EAST	06/26/13	PP

	INTERFERENCE SEARCH		
US Class/ CPC Symbol	US Subclass / CPC Group	Date	Examiner
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		RUL	E							
	APPLICANTS Marcos C. Tzannes, Orinda, CA;									
* CONTINUING DATA ********************************* This application is a CON of 12/783,725 05/20/2010 PAT 8090008 which is a CON of 12/255,713 10/22/2008 PAT 7769104 which is a CON of 11/863,581 09/28/2007 PAT 7471721 which is a CON of 11/211,535 08/26/2005 PAT 7292627 which is a CON of 09/710,310 11/09/2000 PAT 6961369 which claims benefit of 60/164,134 11/09/1999 * FOREIGN APPLICATIONS ************************************										
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Sheridan Suite # 12 1560 Broa Denver, C	Jason H. Vick Sheridan Ross, PC Suite # 1200 1560 Broadway Denver, CO 80202 UNITED STATES									
TITLE										
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Sub	stitute for form 1	449A/PTO		Comp	lete if Known
				Application Number	13/303,417
			CLOSURE	Filing Date	November 23, 2011
ST	TATEMEN	IT BY AP	PLICANT	First Named Inventor	Marcos C. Tzannes
				Art Unit	2611
				Examiner Name	Not Yet Assigned
Sheet	1	of	1	Attorney Docket Number	5550-47-CON-5

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Publication Date MM-DD-YYYY	Name of Patentee of Applicant of Cited Document	Pages, Columns, Lines, Where Relevant Passages or Relevant Figures Appear
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Examiner nitials*	Cite No. ¹	Foreign Patent Document Country Code ³ ; Number ⁴ ; Kind Code ⁵ (<i>if known</i>)		Applicant of Cited Document	Pages, Columns, Lines, Where Relevant Passages or Relevant Figures Appear	T ⁶

		OTHER ART (Including Author, Title, Date, Pertinent Pages, etc.)
Examiner Initials*	Cite No.1	
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	2	Invitation to Restrict or to Pay Additional Fees for International (PCT) Patent Application No. PCT/US00/30958, mailed Nov. 14, 2001 (Attorney Ref. No. 5550-47-PCT)

Examiner Signature	/Phuong Phu/	Date 06/25/2013 Considered	

*EXAMINER: Initial if reference is considered, whether or not citation is in conformance and not considered. Include copy of this form with next communication to applicant.

ALL REFERENCES CONSIDERED EXCEPT WHERE LINED THROUGH. /P.P./

Sub	stitute for form '	1449A/PTO		Comp	lete if Known
				Application Number	13/303,417
			LOSURE	Filing Date	November 23, 2011
SI	TATEME	NT BY AP	PLICANT	First Named Inventor	Marcos C. Tzannes
				Art Unit	2611
				Examiner Name	Not Yet Assigned
Sheet	1	of	1	Attorney Docket Number	5550-47-CON-5

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Examiner Signature	/Phuong Phu/	Date Considered	06/25/2013

*EXAMINER: Initial if reference is considered, whether or not citation is in conformance and not considered. Include copy of this form with next communication to applicant.

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INFORMATION DISCLOSURE				Application Number	13/303,417	
				Filing Date	November 23, 2011	
SI	AIEME	NT BY AP	PLICANT	First Named Inventor	Marcos C. Tzannes	
				Art Unit	2634	
				Examiner Name	Williams, Lawrence B.	
Sheet	1	of	1	Attorney Docket Number	6936-47-CON-5	

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Examiner Initials*	Cite No. ¹	Document Number Number-kind Code ² (If known)	Publication Date MM-DD-YYYY	Name of Patentee of Applicant of Cited Document	Pages, Columns, Lines, Where Relevant Passages or Relevant Figures Appear						
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Examiner Initials*	Cite No. ¹	Foreign Patent Document Country Code³; Number⁴; Kind Code⁵ (if known)		Name of Patentee or Applicant of Cited Document	Pages, Columns, Lines, Where Relevant Passages or Relevant Figures Appear	Т ⁶					

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Examiner Signature /Phuong Phu/	Date Considered 06/25/2013	
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	Application/Control No.	Applicant(s)/Patent Under Reexamination
Index of Claims	13303417	TZANNES, MARCOS C.
	Examiner	Art Unit
	PHUONG PHU	2634

✓	Rejected	-	Cancelled		N	Non-Elected		Α	Appeal
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Final	Original	06/29/2013								
	21	✓								
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	51	√								

U.S. Patent and Trademark Office Part of Paper No.: 20130628

EAST Search History

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L4	1	wo-0135591-\$.did.	US-PGPUB; USPAT; EPO; DERWENT	OR	OFF	2013/06/26 15:55
L5	3098	Marcos.in. or Tzannes.in.	US-PGPUB; USPAT; EPO; DERWENT	OR	OFF	2013/06/26 16:14
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18.1		TION DIO	N COURT	Application Number	13/303,417		
			CLOSURE	Filing Date	November 23, 2011		
51	AIEME	MIRAMA	PLICANT	First Named Inventor	Marcos C. Tzannes		
				Art Unit	2611		
				Examiner Name	Not Yet Assigned		
Sheet	1	of	4	Attorney Docket Number	5550-47-CON-5		

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Examiner Initials*	Cite No. ¹	Document Number Number-kind Code ^{2 (If known)}	Publication Date MM-DD-YYYY	Name of Patentee of Applicant of Cited Document	Pages, Columns, Lines, Where Relevant Passages or Relevant Figures Appear
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Examiner	Date	
Signature	Considered	

Subs	stitute for form	1449A/PTO		Complete if Known		
1811		TION DIGG	N OCUDE	Application Number	13/303,417	
		_	CLOSURE	Filing Date	November 23, 2011	
STATEMENT BY APPLICANT				First Named Inventor	Marcos C. Tzannes	
				Art Unit	2611	
				Examiner Name	Not Yet Assigned	
Sheet	2	of	4	Attorney Docket Number	5550-47-CON-5	

33	2010/0290507	11/18/10	Tzannes	
34	13/284549	(Filed 10-28-2011)	Tzannes	

		FO	REIGN PATEN	T DOCUMENTS		
Examiner Initials*	Cite No. ¹	Foreign Patent Document Country Code ³ ; Number ⁴ ; Kind Code ⁵ (if known)	Publication Date MM-DD-YYYY	Name of Patentee or Applicant of Cited Document	Pages, Columns, Lines, Where Relevant Passages or Relevant Figures Appear	T ⁶
	35	EP 0584534	03/02/94	ALCATEL ITALIA		
	36	EP 0719004	06/26/96	MATSUSHITA ELECTRIC IND CO LTD		
	37	GB 2330491	04/21/99	BRITISH BROADCASTING CORP		
	38	JP H10(1998)-084329		NIPPON HOSO KYOKAI		(Translated Abstract and partial translation)
	39	JP H08(1996)-321820	12/03/96	MATSUSHITA ELECTRIC IND CO LTD		(Translated Abstract)
	40	WO 98/32065	07/23/98	FORTRESS TECHNOLOGIES INC		
	41	WO 99/22463	05/06/99	MOTOROLA INC		
	42	WO 99/29078	06/10/99	TELIA AB		

Examiner	Date	
	Date	
Signature	Considered	
<u> </u>	Considered	

Sub	stitute for form	1449A/PTO		Complete if Known		
INI		TION DIOC	N OOUDE	Application Number	13/303,417	
		TION DISC		Filing Date	November 23, 2011	
51	AIEME	NT BY AP	PLICANI	First Named Inventor	Marcos C. Tzannes	
				Art Unit	2611	
				Examiner Name	Not Yet Assigned	
Sheet	3	of	4	Attorney Docket Number	5550-47-CON-5	

	OTHER ART (Including Author, Title, Date, Pertinent Pages, etc.)
Cite No. ¹	
43	BÄUML et al., "Reducing The Peak-To-Average Power Ratio Of Multicarrier Modulation By Selected Mapping," Electronics Letters, GB, IEE Stevenage, Vol. 32(22), Oct. 24, 1996, pp. 2056-2057, XP000643915 ISSN: 0013-5194
44	HENKEL, "Analog Codes for Peak-to-Average Ratio Reduction," in Proceedings 3rd ITG Conf. Source and Channel Coding, Munich, Germany, Jan. 2000, 5 pages
45	NARAHASHI et al., "New phasing scheme of N multiple carriers for reducing peak-to-average power ratio," Electronics Letters, Aug. 1994, Vol. 30(17), pp. 1382-83
46	TELLADO et al., "Revisiting DMT's Peak-to-Average Ratio," Antwerp, Apr. 20-24, 1998, pp. 1-14
47	TELLAMBURA, "A coding technique for reducing peak-to-average power ratio in OFDM," In the Proceedings of Global Telecommunications Conference, IEEE, Nov. 1998, pp. 2783-2787
48	TELLAMBURA, "Phase optimisation criterion for reducing peak-to-average power ratio in OFDM," Electronics Letters, Jan. 1998, Vol. 34(2), pp. 169-170
49	VAN EETVELT et al., "Peak to average power reduction for OFDM schemes by selective scrambling," Electronics Letters, Oct. 1996, Vol. 32(21), pp. 1963-64
50	Copy of Annex to Form PCT/ISA/206 for PCT/US00/30958, Mar. 23, 2001, 3 pages (5550-47-PCT)
51	International Search Report for International (PCT) Patent Application No. PCT/US00/30958, mailed Jun. 12, 2001 (Attorney Ref. No. 5550-47-PCT)
52	International Preliminary Examination Report for International (PCT) Patent Application No. PCT/US00/30958, completed March 4, 2002 (Attorney Ref. No. 5550-47-PCT)
53	Written Opinion for International (PCT) Patent Application No. PCT/US00/30958, mailed Dec. 18, 2001 (Attorney Ref. No. 5550-47-PCT)
54	Notification of Reasons for Refusal (including translation) for Japanese Patent Application No. 2001-537217, date of dispatch, Mar. 3, 2008 (Attorney Ref. No. 5550-47-PJP)
55	Decision of Refusal (including translation) for Japanese Patent Application No. 2001-537217, date of dispatch, Nov. 4, 2008 (Attorney Ref. No. 5550-47-PJP)
	No.1 43 44 45 46 47 48 49 50 51 52 53 54

Examiner	Date	
Signature	Considered	

Sub	stitute for form	1449A/PTO		Complete if Known		
		rion Dioc	N COURT	Application Number	13/303,417	
			CLOSURE	Filing Date	November 23, 2011	
S	IAIEME	NI BY AP	PLICANT	First Named Inventor	Marcos C. Tzannes	
				Art Unit	2611	
				Examiner Name	Not Yet Assigned	
Sheet	4	of	4	Attorney Docket Number	5550-47-CON-5	

56	Notice of Preliminary Rejection (translation only) for Korean Patent Application No. 7005830/2002 dated Nov. 22, 2006 (Attorney Ref. No. 5550-47-PKR)
57	Official Action for U.S. Patent Application No. 09/710,310, mailed May 4, 2004 (Attorney Ref. No. 5550-47)
58	Notice of Allowance for U.S. Patent Application No. 09/710,310, mailed Jul 5, 2005 (Attorney Ref. No. 5550-47)
59	Notice of Allowance for U.S. Patent Application No. 11/211,535, mailed Sep. 6, 2007 (Attorney Ref. No. 5550-47-CON)
60	Notice of Allowance for U.S. Patent Application No. 11/860,080, mailed Oct. 17, 2011 (Attorney Ref. No. 5550-47-CON-DIV)
61	Official Action for U.S. Patent Application No. 11/863,581, mailed Feb. 6, 2008 (Attorney Ref. No. 5550-47-CON-2)
62	Notice of Allowance for U.S. Patent Application No. 11/863,581, mailed Oct. 8, 2008 (Attorney Ref. No. 5550-47-CON-2)
63	Official Action for U.S. Patent Application No. 12/255,713, mailed Oct. 15, 2009 (Attorney Ref. No. 5550-47-CON-3)
64	Notice of Allowance for U.S. Patent Application No. 12/255,713, mailed May 18, 2010 (Attorney Ref. No. 5550-47-CON-3)
65	Notice of Allowance for U.S. Patent Application No. 12/783,725, mailed Nov. 17, 2011 (Attorney Ref. No. 5550-47-CON-4)

Examiner Signature	/Phuong Phu/	Date Considered	06/25/2013

ALL REFERENCES CONSIDERED EXCEPT WHERE LINED THROUGH. /P.P./

Exhibit 1002 Page 140

POWER OF ATTORNEY TO PROSECUTE APPLICATIONS BEFORE THE USPTO

I hereby revoke all previous powers of attorney given in the application identified in the attached statement under 37 CFR 3.73(c).									
I hereby appoint:									
	Practi								
	OR	tioners associated with Customer Number: 62574							
	Practi	rractitioner(s) named below (if more than ten patent practitioners are to be named, then a customer number must be used):							
		Name Re			atlon Name ber			Registration Number	
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any a	nd all pate	nt applica) to represent the unders itions assigned <u>only</u> to th ecordance with 37 CFR 3	e undersigne	the United S d according	tates Pater to the USP	nt and Trademark Offi TO assignment record	ce (USPTO) ds or assignm	In connection with ents documents
			pondence address for th		identified in	the attache	d statement under 37	GER 3.73(c)	to:
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	The a	iddress as	ssociated with Customer	Number:	62574	ļ		•	
OR	Flrm or								
	Individual	Name						4	
	Address							T	
	City		, , , , , , , , , , , , , , , , , , , ,	State			Zip		
	Country								
	Telephon	е				Email			
Assignee Name and Address: TQ DELTA, LLC 805 Las Cimas Parkway, Suite 240 Austin, Texas 78746									
A copy of this form, together with a statement under 37 CFR 3.73(c) (Form PTO/AIA/96 or equivalent) is required to be Filed in each application in which this form is used. The statement under 37 CFR 3.73(c) may be completed by one of The practitioners appointed in this form, and must identify the application in which this Power of Attorney is to be filed.									
SIGNATURE of Assignee of Record The individual whose signature and title is supplied below is authorized to act on behalf of the assignee									
Signature ? care 4		Ken	Kere		Date 10/4/17				
Name Mark K		K. Roche	Roche			Telephone 512-609-1810			
Title	Title Managing Director								

This collection of information is required by 37 CFR 1.31, 1.32 and 1.33. The information is required to obtain or retain a benefit by the public which is to file (and by the USPTO to process) an application. Confidentiality is governed by 35 U.S.C. 122 and 37 CFR 1.11 and 1.14. This collection is estimated to take 3 minutes to complete, including gathering, preparing, and submitting the completed application form to the USPTO. Time will vary depending upon the individual case. Any comments on the amount of time you require to complete this form and/or suggestions for reducing this burden, should be sent to the Chief Information Officer, U.S. Patent and Trademark Office, U.S. Department of Commerce, P.O. Box 1450, Alexandria, VA 22313-1450. DO NOT SEND FEES OR COMPLETED FORMS TO THIS ADDRESS, SEND TO: Commissioner for Patents, P.O. Box 1450, Alexandria, VA 22313-1450.

If you need assistance in completing the form, call 1-800-PTO-9199 and select option 2.

IN THE UNITED STATES PATENT AND TRADEMARK OFFICE

In Re the Application of: Marcos C. Tzannes	Group Art Unit: 2634
Application No.: 13/303,417	Examiner: WILLIAMS, Lawrence B.
Filed: November 23, 2011	Confirmation No.: 3035
Atty. File No.: 6936-47-CON-5))

For: SYSTEM AND METHOD FOR SCRAMBLING THE PHASE OF THE CARRIERS IN A MULTICARRIER COMMUNICATIONS SYSTEM

ASSERTION OF ENTITLEMENT TO SMALL ENTITY STATUS

Commissioner for Patents P.O. Box 1450 Alexandria, VA 22313-1450

Madam:

In accordance with MPEP §§ 509.02 and 509.03 and 37 CFR 1.27, this document is being filed to inform the U.S. Patent Office of the change of status for the above-identified patent application from large entity status to small entity status. All fees paid to date have been paid as large entity status. No fees have yet been paid as small entity status. Due to the sale of the referenced application, the Applicant is now entitled to small entity status.

We respectfully request that small entity status be granted for the above-referenced patent application.

Please contact the undersigned if there are any questions regarding this notification.

Respectfully submitted,

SHERIDAN ROSS P.C.

Date: 3,001/2 Jason H. Vick, Reg. No. 45,285

1560 Broadway, Suite 1200 Denver, Colorado 80202

Telephone: 303-863-9700

Electronic Acknowledgement Receipt					
EFS ID:	14107684				
Application Number:	13303417				
International Application Number:					
Confirmation Number:	3035				
Title of Invention:	SYSTEM AND METHOD FOR SCRAMBLING THE PHASE OF THE CARRIERS IN A MULTICARRIER COMMUNICATIONS SYSTEM				
First Named Inventor/Applicant Name:	Marcos C. Tzannes				
Customer Number:	62574				
Filer:	Jason Vick/Joanne Vos				
Filer Authorized By:	Jason Vick				
Attorney Docket Number:	6936-47-CON-5				
Receipt Date:	30-OCT-2012				
Filing Date:	23-NOV-2011				
Time Stamp:	18:07:53				
Application Type:	Utility under 35 USC 111(a)				

Payment information:

Submitted with Payment	no
File Listing:	

Document Number	Document Description	File Name	File Size(Bytes)/ Message Digest	Multi Part /.zip	Pages (if appl.)
1		Statement_Under_373c_w_PO A.pdf	346123	yes	3
'			52e9e086d15eeab73c2a30e31851e1ee9a0 6a29b		

	Multipart Description/PDF files in .zip description						
	Document Des	Start	End				
	Assignee showing of owner	1	2				
	Power of Att	3	3				
Warnings:							
Information:	Information:						
2	Miscellaneous Incoming Letter	r Change_of_Entity_Status.pdf	83965	no 2	1		
-	iniscellaneous incoming zerei		1ba885ce96af79fa622cc4023c5185c2fc632 b4d				
Warnings:							
Information:							
	Total Files Size (in bytes): 430088						

This Acknowledgement Receipt evidences receipt on the noted date by the USPTO of the indicated documents, characterized by the applicant, and including page counts, where applicable. It serves as evidence of receipt similar to a Post Card, as described in MPEP 503.

New Applications Under 35 U.S.C. 111

If a new application is being filed and the application includes the necessary components for a filing date (see 37 CFR 1.53(b)-(d) and MPEP 506), a Filing Receipt (37 CFR 1.54) will be issued in due course and the date shown on this Acknowledgement Receipt will establish the filing date of the application.

National Stage of an International Application under 35 U.S.C. 371

If a timely submission to enter the national stage of an international application is compliant with the conditions of 35 U.S.C. 371 and other applicable requirements a Form PCT/DO/EO/903 indicating acceptance of the application as a national stage submission under 35 U.S.C. 371 will be issued in addition to the Filing Receipt, in due course.

New International Application Filed with the USPTO as a Receiving Office

If a new international application is being filed and the international application includes the necessary components for an international filing date (see PCT Article 11 and MPEP 1810), a Notification of the International Application Number and of the International Filing Date (Form PCT/RO/105) will be issued in due course, subject to prescriptions concerning national security, and the date shown on this Acknowledgement Receipt will establish the international filing date of the application.

PTO/AIA/96 (08-12)
Approved for use through 01/31/2013. OMB 0651-0031
U.S. Patent and Trademark Office; U.S. DEPARTMENT OF COMMERCE
Under the Paperwork Reduction Act of 1995, no persons are required to respond to a collection of information unless it displays a valid OMB control number.

STATEMENT UNDER 37 CFR 3.73(c)
Applicant/Patent Owner: TQ DELTA, LLC
Application No./Patent No.: 13/303,417 Filed/Issue Date: November 23, 2011
Titled: SYSTEM AND METHOD FOR SCRAMBLING THE PHASE OF THE CARRIERS IN A MULTICARRIER COMMUNICATIONS SYSTEM
TQ DELTA, LLC , a Corporation
(Name of Assignee) (Type of Assignee, e.g., corporation, partnership, university, government agency, etc.)
states that, for the patent application/patent identified above, it is (choose one of options 1, 2, 3 or 4 below):
1. The assignee of the entire right, title, and interest.
2. An assignee of less than the entire right, title, and interest (check applicable box):
The extent (by percentage) of its ownership interest is
There are unspecified percentages of ownership. The other parties, including inventors, who together own the entir right, title and interest are:
Additional Statement(s) by the owner(s) holding the balance of the interest <u>must be submitted</u> to account for the enti- right, title, and interest.
3. The assignee of an undivided interest in the entirety (a complete assignment from one of the joint inventors was made)
The other parties, including inventors, who together own the entire right, title, and interest are: Additional Statement(s) by the owner(s) holding the balance of the interest must be submitted to account for the entire
right, title, and interest.
4. The recipient, via a court proceeding or the like (e.g., bankruptcy, probate), of an undivided interest in the entirety (a complete transfer of ownership interest was made). The certified document(s) showing the transfer is attached.
The interest identified in option 1, 2 or 3 above (not option 4) is evidenced by either (choose one of options A or B below):
A. An assignment from the inventor(s) of the patent application/patent identified above. The assignment was recorded in the United States Patent and Trademark Office at Reel, Frame, or for which a copy thereof is attached.
B. A chain of title from the inventor(s), of the patent application/patent identified above, to the current assignee as follows:
1. From: Marcos C. Tzannes To: AWARE, INC.
The document was recorded in the United States Patent and Trademark Office at
Reel 010877 , Frame 0307 , or for which a copy thereof is attached.
2. From: AWARE, INC. To: _TQ DELTA, LLC
The document was recorded in the United States Patent and Trademark Office at Reel 029154, Frame 0937, or for which a copy thereof is attached.

[Page 1 of 2]

Page 1 of 2]
This collection of information is required by 37 CFR 3.73(b). The information is required to obtain or retain a benefit by the public which is to file (and by the USPTO to process) an application. Confidentiality is governed by 35 U.S.C. 122 and 37 CFR 1.11 and 1.14. This collection is estimated to take 12 minutes to complete, including gathering, preparing, and submitting the completed application form to the USPTO. Time will vary depending upon the individual case. Any comments on the amount of time you require to complete this form and/or suggestions for reducing this burden, should be sent to the Chief Information Officer, U.S. Patent and Trademark Office, U.S. Department of Commerce, P.O. Box 1450, Alexandria, VA 22313-1450. DO NOT SEND FEES OR COMPLETED FORMS TO THIS ADDRESS. SEND TO: Commissioner for Patents, P.O. Box 1450, Alexandria, VA 22313-1450.

If you need assistance in completing the form, call 1-800-PTO-9199 and select option 2.

PTO/AIA/96 (08-12)

Approved for use through 01/31/2013. OMB 0651-0031

U.S. Patent and Trademark Office; U.S. DEPARTMENT OF COMMERCE

Under the Paperwork Reduction Act of 1995, no persons are required to respond to a collection of information unless it displays a valid OMB control number.

		STATEMENT	UNDER 37 CFR 3.73(c)	
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3. From:			To: ed States Patent and Tradema	
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4. From:			To:	
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5. From:			To:	
	The document was	recorded in the Unite	ed States Patent and Tradema	rk Office at
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Additio	nal documents in the	chain of title are liste	ed on a supplemental sheet(s)	
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✓ As requir	ed by 37 CFR 3.73(c	c)(1)(i), the document	arv evidence of the chain of ti	tle from the original owner to the
			or recordation pursuant to 37	
)) must be submitted to Assignment
DIVISION	in accordance with 3	7 OFFI Part 3, to reco	ro the assignment in the recor	rds of the USPTO. See MPEP 302.08]
The undersigned	l (whose title is suppl	ied below) is authoriz	ed to act on behalf of the assi	gnee.
				30 60 12
Signature				<u> විය රය '/2</u> Date
Jason H. ∖	/ick			45,285
Printed or Typed	Name		0 (0)	Title or Registration Number

[Page 2 of 2]

Substitute for form 1449A/PTO				Сотр	olete if Known
-	IEODIA:	TION DIGG		Application Number	13/303,417
		TION DISC		Filing Date	November 23, 2011
S	STATEMENT BY APPLICANT			First Named Inventor	Marcos C. Tzannes
				Art Unit	2634
				Examiner Name	Williams, Lawrence B.
Sheet	1	of	1	Attorney Docket Number	6936-47-CON-5

	U.S. PATENT DOCUMENTS					
Examiner Initials*	Cite No. ¹	Document Number Number-kind Code ^{2 (if known)}	Publication Date MM-DD-YYYY	Name of Patentee of Applicant of Cited Document	Pages, Columns, Lines, Where Relevant Passages or Relevant Figures Appear	
	1	2012/0195353	08/02/2012	Tzannes		

FOREIGN PATENT DOCUMENTS						
Examiner nitials*	Cite No. ¹	Foreign Patent Document Country Code³; Number⁴; Kind Code⁵ (if known)		Name of Patentee or Applicant of Cited Document	Pages, Columns, Lines, Where Relevant Passages or Relevant Figures Appear	T ⁶

	OTHER ART (Including Author, Title, Date, Pertinent Pages, etc.)					
Examiner Initials*	Cite No. ¹					
		Notice of Allowance for U.S. Patent Application No. 13/439,605, mailed Sept. 14, 2012 (Attorney Ref. No. 6936-47-CON-DIV-CON-2)				

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Examiner	Date	
Signature		
Olgitature	Considered	

*EXAMINER: Initial if reference is considered, whether or not citation is in conformance and not considered. Include copy of this form with next communication to applicant.

Electronic Acknowledgement Receipt				
EFS ID:	14107729			
Application Number:	13303417			
International Application Number:				
Confirmation Number:	3035			
Title of Invention:	SYSTEM AND METHOD FOR SCRAMBLING THE PHASE OF THE CARRIERS IN A MULTICARRIER COMMUNICATIONS SYSTEM			
First Named Inventor/Applicant Name:	Marcos C. Tzannes			
Customer Number:	62574			
Filer:	Jason Vick/Joanne Vos			
Filer Authorized By:	Jason Vick			
Attorney Docket Number:	6936-47-CON-5			
Receipt Date:	30-OCT-2012			
Filing Date:	23-NOV-2011			
Time Stamp:	18:11:29			
Application Type:	Utility under 35 USC 111(a)			

Payment information:

Submitted with Payment	no
File Listing:	

Document Number	Document Description	File Name	File Size(Bytes)/ Message Digest	Multi Part /.zip	Pages (if appl.)
1		IDS 04.pdf	347637	ves	4
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	Multipart Description/PDF files in .zip description					
	Document D	Start	E	nd		
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	Information Disclosure State	4		4		
Warnings:						
Information:						
2	Non Patent Literature	6936-47-CON-DIV-	12281123	no	14	
2	Non ratent Literature	CON-2_NOA_09-14-2012.pdf	9fb2a21023a4f52f8086817c10114c2a00c0 1a6a	110	17	
Warnings:						
Information:						
		Total Files Size (in bytes)	126	528760		

This Acknowledgement Receipt evidences receipt on the noted date by the USPTO of the indicated documents, characterized by the applicant, and including page counts, where applicable. It serves as evidence of receipt similar to a Post Card, as described in MPEP 503.

New Applications Under 35 U.S.C. 111

If a new application is being filed and the application includes the necessary components for a filing date (see 37 CFR 1.53(b)-(d) and MPEP 506), a Filing Receipt (37 CFR 1.54) will be issued in due course and the date shown on this Acknowledgement Receipt will establish the filing date of the application.

National Stage of an International Application under 35 U.S.C. 371

If a timely submission to enter the national stage of an international application is compliant with the conditions of 35 U.S.C. 371 and other applicable requirements a Form PCT/DO/EO/903 indicating acceptance of the application as a national stage submission under 35 U.S.C. 371 will be issued in addition to the Filing Receipt, in due course.

New International Application Filed with the USPTO as a Receiving Office

If a new international application is being filed and the international application includes the necessary components for an international filing date (see PCT Article 11 and MPEP 1810), a Notification of the International Application Number and of the International Filing Date (Form PCT/RO/105) will be issued in due course, subject to prescriptions concerning national security, and the date shown on this Acknowledgement Receipt will establish the international filing date of the application.

IN THE UNITED STATES PATENT AND TRADEMARK OFFICE

In Re the Application of:) Group Art Unit: 2634
TZANNES, Marcos C.	Confirmation No.: 3035
Serial No.: 13/303,417) Examiner: Williams, Lawrence B.
Filed: November 23, 2011)
Atty. File No.: 6936-47-CON-5) SUPPLEMENTAL) INFORMATION DISCLOSURE
Entitled: "System and Method for Scrambling the Phase of the Carriers in a Multicarrier) INFORMATION DISCLOSURE STATEMENT
Communications System"	Electronically Submitted
Commissioner for Patents P.O. Box 1450 Alexandria, VA 22313-1450	
Dear Sir:	
The references cited on attached Form PTC	0-1449 are being called to the attention
of the Examiner.	
Copies of the cited non-patent and/or foreign	n references are enclosed herewith.
Copies of the cited U.S. patents and/or patent	at applications are enclosed herewith.
Copies of the cited U.S. patents/patent appl	ication publications are not enclosed in
accordance with 37 C.F.R. § 1.98(a).	
Copies of the cited references are not en	closed, in accordance with 37 C.F.R.
§ 1.98(d), because the references were cited by	or submitted to the U.S. Patent and
Trademark Office in prior application Serial No	filed,
which is relied upon for an earlier filing date under	35 U.S.C. § 120.
To the best of applicants' belief, the pertine	ence of the foreign-language references
are believed to be summarized in the attached E	inglish abstracts and/or in the figures,
although applicants do not necessarily vouch for the	accuracy of the translation.
Examiner's attention is drawn to the following	ng related applications:
Serial No filed	(Attorney Ref. No)
Serial No filed	(Attorney Ref. No)
Serial No filed Other:	
Submission of the above information is not	intended as an admission that any item
is citable under the statutes or rules to support	a rejection, that any item disclosed
represents analogous art, or that those skilled in t	he art would refer to or recognize the
pertinence of any reference without the benefit of	hindsight, nor should an inference be

drawn as to the pertinence of the references based on the order in which they are presented. Submission of this statement should not be taken as an indication that a search has been conducted, or that no better art exists.

It is respectfully requested that the cited information be expressly considered during the prosecution of this application and the references made of record therein.

FEES

37 CFR 1.97(b): No fee is believed due in connection with this submission, because the information disclosure statement submitted herewith is satisfied by one of the following conditions ("X" indicates satisfaction): Within three months of the filing date of a national application other than a continued prosecution application under 37 CFR 1.53(d), or Within three months of the date of entry of the national stage as set forth in § 1.491 in an international application, or Before the mailing date of a first Office Action on the merits, or Before the mailing of a first Office action after the filing of a request for continued examination under 37 CFR 1.114. Although no fee is believed due, if any fee is deemed due in connection with this submission, please charge such fee to Deposit Account 19-1970.
37 CFR 1.97(e): The information disclosure statement transmitted herewith is being filed after all the above conditions (37 CFR 1.97(b)), but before the mailing date of any one of the following conditions: (1) a final action under 37 C.F.R. 1.113, or (2) a notice of allowance under 37 C.F.R. 1.311, or (3) an action that otherwise closes prosecution in the application. This Information Disclosure Statement is accompanied by: A Certification (below) as specified by 37 C.F.R. 1.97(e). Although no fee is believed due, if any fee is deemed due in connection with this submission, please charge such fee to Deposit Account 19-1970. OR Please charge Deposit Account 19-1970 in the amount of \$180.00 for the fee set forth in 37 C.F.R. 1.17(p) for submission of an information disclosure statement. Please credit any overpayment or charge any underpayment to Deposit Account 19-1970.
37 CFR 1.97(d): This Information Disclosure Statement is being submitted after the period specified in 37 CFR 1.97(c). This information Disclosure Statement includes a Certification (below) as specified by 37 C.F.R. 1.97(e) AND Applicants hereby requests consideration of the reference(s) disclosed herein. Please charge Deposit Account 19-1970 in the amount of \$180.00 under 37 C.F.R. 1.17(p). Please credit any overpayment or charge any underpayment to Deposit Account 19-1970. Election to pay the fee should not be taken as an indication that applicant(s) cannot execute a certification.

Certification (37 C.F.R. 1.97(e)) (Applicable only if checked)
The undersigned certifies that: Each item of information contained in this information disclosure statement was first cited in any communication from a foreign patent office in a counterpart foreign application not more than three months prior to the filing of this statement. 37 C.F.R. 1.97(e)(1). A copy of the communication from the foreign patent office is enclosed.
OR
No item of information contained in this information disclosure statement was cited in a communication from a foreign patent office in a counterpart foreign application, and, to the knowledge of the undersigned after making reasonable inquiry, no item of information contained in this Information Disclosure Statement was known to any individual designated in 37 C.F.R. 1.56(c) more than three months prior to the filing of this statement. 37 C.F.R. 1.97(e)(2).
Respectfully submitted,
SHERIDAN ROSS P.C.
By: Jason H. Vick Registration No. 45,285 1560 Broadway, Suite 1200 Denver, Colorado 80202-5141 (303) 863-9700

IN THE UNITED STATES PATENT AND TRADEMARK OFFICE

In Re the Application of: Marcos C. Tzannes	Group Art Unit: 2611
Application No.: 13/303,417	Examiner: WILLIAMS, Lawrence B.
Filed: November 23, 2011	Confirmation No.: 3035
Atty. File No.: 5550-47-CON-5)

For: SYSTEM AND METHOD FOR SCRAMBLING THE PHASE OF THE CARRIERS IN A MULTICARRIER COMMUNICATIONS SYSTEM

Commissioner for Patents P.O. Box 1450 Alexandria, VA 22313

PRELIMINARY AMENDMENT

Dear Sir:

Prior to the initial review of the above-identified patent application by the Examiner, please enter the following Preliminary Amendment. Although Applicants do not believe that any fees are due based upon the filing of this Preliminary Amendment, please charge any such fees to Deposit Account 19-1970.

Please amend the above-identified patent application as follows:

Amendments to the Claims are shown in the listing of claims which begin on page 2 of this paper.

Remarks begin on page 7 of this paper.

1

Attorney Docket No.: 5550-47-CON-5

AMENDMENTS TO THE CLAIMS

This listing of claims will replace all prior versions, and listings, of claims in the application.

Listing of Claims:

1. - 20. (Cancelled)

21. (New) In a multicarrier modulation system including a first transceiver in communication with a second transceiver using a transmission signal having a plurality of carrier signals for modulating an input bit stream, each carrier signal having a phase characteristic associated with the input bit stream, a method for scrambling the phase characteristics of the carrier signals comprising:

transmitting the bit stream from the first transceiver to the second transceiver, wherein:

each carrier signal is associated with a value determined independently of any bit
value of the bit stream carried by that respective carrier signal, the value associated with each
carrier signal determined by a pseudo-random number generator,

a phase shift for each carrier signal is at least based on:

the value associated with that carrier signal, and

a combining of a phase for each carrier signal with the phase characteristic of that respective carrier signal so as to substantially scramble the phase characteristics of the plurality of carrier signals, and

multiple carrier signals corresponding to the plurality of phase shifted and scrambled carrier signals are used by the first transceiver to modulate a same bit value of the bit stream; and

wherein:

the first and second transceivers communicate over one or more of a pair of twisted wires of a telephone subscriber line, quad cable and fiber optic, and the first and second transceivers transport video.

22. (New) The method of claim 21, wherein one or more of the first transceiver and second transceiver are cable transceivers.

2

Attorney Docket No.: 5550-47-CON-5

- 23. (New) The method of claim 21, wherein one or more of the first transceiver and second transceiver are VDSL transceivers.
- 24. (New) The method of claim 21, wherein the first and second transceivers are multicarrier DSL transceivers.
- 25. (New) The method of claim 21, wherein the first and second transceivers are used for high speed internet access.
- 26. (New) The method of claim 21, further comprising, in the first transceiver, independently deriving the values associated with each carrier using a second pseudo-random number generator in the second transceiver.
- 27. (New) The method of claim 26, wherein the first and second transceivers use a same seed for the pseudo-random number generator.
- 28. (New) The method of claim 26, wherein the first and second transceivers are wireless transceivers.
- 29. (New) The method of claim 26, wherein the first and second transceivers are cable transceivers.
- 30. (New) The method of claim 26, wherein the first and second transceivers are DSL transceivers connected using a pair of twisted wires of a telephone subscriber system.
- 31. (New) The method of claim 30, wherein the first and second transceivers are VDSL transceivers.
- 32. (New) The method of claim 26, wherein the first and second transceivers are multicarrier DSL transceivers.

3

- 33. (New) The method of claim 26, wherein the first and second transceivers are also used for transport high speed internet access.
- 34. (New) The method of claim 21, wherein the first and the second transceivers include digital signal processors.
- 35. (New) A multicarrier modulation system including a first transceiver in communication with a second transceiver using a transmission signal having a plurality of carrier signals for modulating an input bit stream, each carrier signal having a phase characteristic associated with the input bit stream, the first transceiver capable of transmitting to the second transceiver the bit stream, wherein:

each carrier signal is associated with a value determined independently of any bit value of the bit stream carried by that respective carrier signal, the value associated with each carrier signal determined by a pseudo-random number generator,

a phase shift for each carrier signal is at least based on:

the value associated with that respective carrier signal, and a combining of a phase shift for each carrier signal with the phase characteristic of that respective carrier signal so as to substantially scramble the phase characteristics of the plurality of carrier signals, and

multiple carrier signals corresponding to the plurality of phase shifted and scrambled carrier signals are used by the first transceiver to modulate a same bit value of the bit stream; and

wherein:

the first and second transceivers communicate over one or more of a pair of twisted wires of a telephone subscriber line, quad cable and fiber optic, and the first and second transceivers transport video.

36. (New) The system of claim 35, wherein one or more of the first transceiver and second transceiver are cable transceivers.

4

Attorney Docket No.: 5550-47-CON-5

- 37. (New) The system of claim 35, wherein one or more of the first transceiver and second transceiver are VDSL transceivers.
- 38. (New) The system of claim 35, wherein the first and second transceivers are multicarrier DSL transceivers.
- 39. (New) The system of claim 35, wherein the first and second transceivers are used for high speed internet access.
- 40. (New) The system of claim 35, wherein the first transceiver independently derives the values associated with each carrier using a second pseudo-random number generator in the first transceiver.
- 41. (New) The system of claim 40, wherein the first and second transceivers use a same seed for the pseudo-random number generator.
- 42. (New) The system of claim 40, wherein the first and second transceivers are wireless transceivers.
- 43. (New) The system of claim 40, wherein the first and second transceivers are cable transceivers.
- 44. (New) The system of claim 40, wherein the first and second transceivers are DSL transceivers connected using a pair of twisted wires of a telephone subscriber system.
- 45. (New) The system of claim 44, wherein the first and second transceivers are VDSL transceivers.
- 46. (New) The system of claim 40, wherein the first and second transceivers are multicarrier DSL transceivers.

- 47. (New) The system of claim 40, wherein the first and second transceivers are also used for high speed internet access.
- 48. (New) The method of claim 35, wherein the first and second transceivers each include digital signal processors.
 - 49. (New) The method of claim 21, wherein the video is video-on demand.
 - 50. (New) The method of claim 35, wherein the video is video-on demand.
- 51. (New) In a multicarrier modulation system including a first transceiver in communication with a second transceiver using a transmission signal having a plurality of carrier signals for modulating an input bit stream, each carrier signal having a phase characteristic associated with the input bit stream, a method for scrambling the phase characteristics of the carrier signals comprising:

transmitting the bit stream from the first transceiver, wherein:

each carrier signal is associated with a value determined independently of any bit value of the bit stream carried by that respective carrier signal, the value associated with each carrier signal determined by a pseudo-random number generator,

a phase shift for each carrier signal is at least based on:

the value associated with that carrier signal, and

a combining of a phase for each carrier signal with the phase characteristic of that respective carrier signal so as to substantially scramble the phase

characteristics of the plurality of carrier signals, and

multiple carrier signals corresponding to the plurality of phase shifted and scrambled carrier signals are used by the first transceiver to modulate a same bit value of the bit stream; and

wherein:

the first and second transceivers communicate over one or more of a pair of twisted wires of a telephone subscriber line, quad cable and fiber optic, and

the first and second transceivers are used for internet access.

REMARKS/ARGUMENTS

By this amendment, claims 1-20 are canceled without prejudice or disclaimer and new claims 21-51 are added.

Applicant requests examination on the merits.

Applicant believes that the pending claims are in condition for allowance and such disposition is respectfully requested. In the event that a telephone conversation would further prosecution and/or expedite allowance, the Examiner is invited to contact the undersigned.

The Commissioner is hereby authorized to charge to Deposit Account No. 19-1970 any fees under 37 C.F.R. §§ 1.16 and 1.17 that may be required by this paper and to credit any overpayment to that Account. If any extension of time is required in connection with the filing of this paper and has not been separately requested, such extension is hereby Petitioned.

Respectfully submitted,

SHERIDAN ROSS P.C.

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(303) 863-9700

Date: 215m /2

7

Attorney Docket No.: 5550-47-CON-5

Electronic Patent Application Fee Transmittal										
Application Number:	13:	303417								
Filing Date:	23	23-Nov-2011								
Title of Invention:	SYSTEM AND METHOD FOR SCRAMBLING THE PHASE OF THE CARRIERS IN A MULTICARRIER COMMUNICATIONS SYSTEM Marcos C. Tzannes									
First Named Inventor/Applicant Name:	Ma	rcos C. Tzannes								
Filer:	Jas	on Vick/Joanne Vos	5							
Attorney Docket Number:	55:	50-47-CON-5								
Filed as Large Entity										
Utility under 35 USC 111(a) Filing Fees										
Description		Fee Code	Quantity	Amount	Sub-Total in USD(\$)					
Basic Filing:										
Pages:										
Claims:										
Claims in excess of 20		1202	11	60	660					
Miscellaneous-Filing:										
Petition:										
Patent-Appeals-and-Interference:										
Post-Allowance-and-Post-Issuance:										
Extension-of-Time:										

Description	Fee Code	Quantity	Amount	Sub-Total in USD(\$)
Miscellaneous:				
	Tot	al in USD	(\$)	660

Electronic Ack	knowledgement Receipt
EFS ID:	13076649
Application Number:	13303417
International Application Number:	
Confirmation Number:	3035
Title of Invention:	SYSTEM AND METHOD FOR SCRAMBLING THE PHASE OF THE CARRIERS IN A MULTICARRIER COMMUNICATIONS SYSTEM
First Named Inventor/Applicant Name:	Marcos C. Tzannes
Customer Number:	62574
Filer:	Jason Vick/Joanne Vos
Filer Authorized By:	Jason Vick
Attorney Docket Number:	5550-47-CON-5
Receipt Date:	21-JUN-2012
Filing Date:	23-NOV-2011
Time Stamp:	17:56:26
Application Type:	Utility under 35 USC 111(a)

Payment information:

Submitted with Payment	yes
Payment Type	Deposit Account
Payment was successfully received in RAM	\$660
RAM confirmation Number	5333
Deposit Account	191970
Authorized User	

The Director of the USPTO is hereby authorized to charge indicated fees and credit any overpayment as follows:

Charge any Additional Fees required under 37 C.F.R. Section 1.16 (National application filing, search, and examination fees)

Charge any Additional Fees required under 37 C.F.R. Section 1.17 (Patent application and reexamination processing fees)

Charge any Additional Fees required under 37 C.F.R. Section 1.19 (Document supply fees)

Charge any Additional Fees required under 37 C.F.R. Section 1.21 (Miscellaneous fees and charges)

File Listing:

Document Number	Document Description	File Name	File Size(Bytes)/ Message Digest	Multi Part /.zip	Pages (if appl.)
1		AMEND_PRELIM_02.pdf	522326	yes	7
'		//////////////////////////////////////	6701413188046477ecc1af1b594f24c8a93d 9c0b	yes	,
	Multip	part Description/PDF files in	.zip description		
	Document De	scription	Start	E	nd
	Preliminary Am	1		1	
	Claims	S	2	6	
	Applicant Arguments/Remarks	Made in an Amendment	7	7	
Warnings:					
Information:					
2	Fee Worksheet (SB06)	fee-info.pdf	30316	no	2
	,		ab26f4ccfa240f0383a65479fb953278cccd3 017		
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Information:					
		Total Files Size (in bytes	55	52642	

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New Applications Under 35 U.S.C. 111

If a new application is being filed and the application includes the necessary components for a filing date (see 37 CFR 1.53(b)-(d) and MPEP 506), a Filing Receipt (37 CFR 1.54) will be issued in due course and the date shown on this Acknowledgement Receipt will establish the filing date of the application.

National Stage of an International Application under 35 U.S.C. 371

If a timely submission to enter the national stage of an international application is compliant with the conditions of 35 U.S.C. 371 and other applicable requirements a Form PCT/DO/EO/903 indicating acceptance of the application as a national stage submission under 35 U.S.C. 371 will be issued in addition to the Filing Receipt, in due course.

New International Application Filed with the USPTO as a Receiving Office

If a new international application is being filed and the international application includes the necessary components for an international filing date (see PCT Article 11 and MPEP 1810), a Notification of the International Application Number and of the International Filing Date (Form PCT/RO/105) will be issued in due course, subject to prescriptions concerning national security, and the date shown on this Acknowledgement Receipt will establish the international filing date of the application.

Under the Paperwork Reduction Act of 1995, no persons are required to respond to a collection of information unless it displays a valid OMB control number

P	PATENT APPLICATION FEE DETERMINATION RECORD Substitute for Form PTO-875							Docket Number 03,417		ing Date 23/2011	To be Mailed
	Al	PPLICATION	AS FILE		(Column 2)		SMALL	ENTITY 🗌	OR		HER THAN ALL ENTITY
	FOR	N	IUMBER FII	_ED N U	MBER EXTRA	Т	RATE (\$)	FEE (\$)		RATE (\$)	FEE (\$)
	BASIC FEE (37 CFR 1.16(a), (b), or (c))			N/A	1	N/A			N/A		
	SEARCH FEE (37 CFR 1.16(k), (i), (i		N/A		N/A	1	N/A		1	N/A	
	EXAMINATION FE (37 CFR 1.16(o), (p),	E	N/A		N/A	1	N/A		1	N/A	
	ΓAL CLAIMS CFR 1.16(i))		mir	nus 20 = *		1	X \$ =		OR	X \$ =	
IND	EPENDENT CLAIM	S	m	inus 3 = *		1	X \$ =			X \$ =	
(37 CFR 1.16(h)) APPLICATION SIZE FEE (37 CFR 1.16(s)) APPLICATION SIZE FEE (37 CFR 1.16(s)) If the specification and drawings exceed 100 sheets of paper, the application size fee due is \$250 (\$125 for small entity) for each additional 50 sheets or fraction thereof. See 35 U.S.C. 41(a)(1)(G) and 37 CFR 1.16(s).											
	MULTIPLE DEPEN	IDENT CLAIM PF	RESENT (3	7 CFR 1.16(j))							
* If 1	the difference in colu	ımn 1 is less thar	zero, ente	r "0" in column 2.			TOTAL			TOTAL	
	APP	(Column 1)	AMEN	DED — PART II (Column 2)	(Column 3)		SMAL	L ENTITY	OR		ER THAN ALL ENTITY
AMENDMENT	06/21/2012	CLAIMS REMAINING AFTER AMENDMENT		HIGHEST NUMBER PREVIOUSLY PAID FOR	PRESENT EXTRA		RATE (\$)	ADDITIONAL FEE (\$)		RATE (\$)	ADDITIONAL FEE (\$)
ME	Total (37 CFR 1.16(i))	∗ 31	Minus	** 20	= 11		X \$ =		OR	X \$60=	660
N.	Independent (37 CFR 1.16(h))	* 3	Minus	***3	= 0]	X \$ =		OR	X \$250=	0
√ME	Application Si	ze Fee (37 CFR	1.16(s))]					
	FIRST PRESEN	ITATION OF MULTI	PLE DEPEN	DENT CLAIM (37 CF	FR 1.16(j))	1			OR		
							TOTAL ADD'L FEE		OR	TOTAL ADD'L FEE	660
		(Column 1)		(Column 2)	(Column 3)						
T		CLAIMS REMAINING AFTER AMENDMENT		HIGHEST NUMBER PREVIOUSLY PAID FOR	PRESENT EXTRA		RATE (\$)	ADDITIONAL FEE (\$)		RATE (\$)	ADDITIONAL FEE (\$)
ENT	Total (37 CFR 1.16(i))	*	Minus	**	=		X \$ =		OR	X \$ =	
DM	Independent (37 CFR 1.16(h))	*	Minus	***	=]	X \$ =		OR	X \$ =	
EN	Application Si	ze Fee (37 CFR	1.16(s))								
AMENDME	FIRST PRESEN	ITATION OF MULTI	PLE DEPEN	DENT CLAIM (37 CF	FR 1.16(j))	1			OR		
						•	TOTAL ADD'L FEE		OR	TOTAL ADD'L FEE	
** If *** I	the entry in column the "Highest Numbe f the "Highest Numb "Highest Number P	er Previously Paid per Previously Pai	l For" IN TI d For" IN T	HIS SPACE is less HIS SPACE is les	s than 20, enter "20 s than 3, enter "3".		/MONIC	nstrument Ex QUE BRUNSO	ON/	er:	

This collection of information is required by 37 CFR 1.16. The information is required to obtain or retain a benefit by the public which is to file (and by the USPTO to process) an application. Confidentiality is governed by 35 U.S.C. 122 and 37 CFR 1.14. This collection is estimated to take 12 minutes to complete, including gathering, preparing, and submitting the completed application form to the USPTO. Time will vary depending upon the individual case. Any comments on the amount of time you require to complete this form and/or suggestions for reducing this burden, should be sent to the Chief Information Officer, U.S. Patent and Trademark Office, U.S. Department of Commerce, P.O. Box 1450, Alexandria, VA 22313-1450. DO NOT SEND FEES OR COMPLETED FORMS TO THIS

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ATTY. DOCKET NO./TITLE APPLICATION NUMBER FILING OR 371(C) DATE FIRST NAMED APPLICANT

5550-47-CON-5

CONFIRMATION NO. 3035 PUBLICATION NOTICE

Marcos C. Tzannes

62574 Jason H. Vick Sheridan Ross, PC Suite # 1200 1560 Broadway Denver, CO 80202

13/303,417



Title:SYSTEM AND METHOD FOR SCRAMBLING THE PHASE OF THE CARRIERS IN A MULTICARRIER COMMUNICATIONS SYSTEM

Publication No.US-2012-0069878-A1 Publication Date:03/22/2012

NOTICE OF PUBLICATION OF APPLICATION

The above-identified application will be electronically published as a patent application publication pursuant to 37 CFR 1.211, et seq. The patent application publication number and publication date are set forth above.

The publication may be accessed through the USPTO's publically available Searchable Databases via the Internet at www.uspto.gov. The direct link to access the publication is currently http://www.uspto.gov/patft/.

The publication process established by the Office does not provide for mailing a copy of the publication to applicant. A copy of the publication may be obtained from the Office upon payment of the appropriate fee set forth in 37 CFR 1.19(a)(1). Orders for copies of patent application publications are handled by the USPTO's Office of Public Records. The Office of Public Records can be reached by telephone at (703) 308-9726 or (800) 972-6382, by facsimile at (703) 305-8759, by mail addressed to the United States Patent and Trademark Office, Office of Public Records, Alexandria, VA 22313-1450 or via the Internet.

In addition, information on the status of the application, including the mailing date of Office actions and the dates of receipt of correspondence filed in the Office, may also be accessed via the Internet through the Patent Electronic Business Center at www.uspto.gov using the public side of the Patent Application Information and Retrieval (PAIR) system. The direct link to access this status information is currently http://pair.uspto.gov/. Prior to publication, such status information is confidential and may only be obtained by applicant using the private side of PAIR.

Further assistance in electronically accessing the publication, or about PAIR, is available by calling the Patent Electronic Business Center at 1-866-217-9197.

Office of Data Managment, Application Assistance Unit (571) 272-4000, or (571) 272-4200, or 1-888-786-0101

page 1 of 1

Sub	stitute for form	1449A/PTO		Comp	lete if Known
				Application Number	13/303,417
	INFORMATION DISCLOSURE STATEMENT BY APPLICANT			Filing Date	November 23, 2011
S	STATEMENT BY APPLICANT		PLICANT	First Named Inventor	Marcos C. Tzannes
				Art Unit	2611
				Examiner Name	Not Yet Assigned
Sheet	- 1	of	1	Attorney Docket Number	5550-47-CON-5

	U.S. PATENT DOCUMENTS								
Examiner Initials*	Cite No.1	Document Number Number-kind Code ² (If known)	Publication Date MM-DD-YYYY	Name of Patentee of Applicant of Cited Document	Pages, Columns, Lines, Where Relevant Passages or Relevant Figures Appear				
	1	5937010	08-10-1999	Petranovich et al.					

	FOREIGN PATENT DOCUMENTS									
Examiner Initials*	Cite No. ¹	Foreign Patent Document Country Code ³ ; Number ⁴ ; Kind Code ⁵ (<i>if known</i>)		Name of Patentee or Applicant of Cited Document	Pages, Columns, Lines, Where Relevant Passages or Relevant Figures Appear	Т ⁶				

	OTHER ART (Including Author, Title, Date, Pertinent Pages, etc.)						
Examiner Initials*	Cite No. ¹						
		Notice of Allowance for U.S. Patent Application No. 13/284,549, mailed March 20, 2012 (Attorney Ref. No. 5550-47-CON-DIV-CON)					

T	Date	
Examiner	Date	
01	Concidered	
Signature	Considered	

*EXAMINER: Initial if reference is considered, whether or not citation is in conformance and not considered. Include copy of this form with next communication to applicant.

Electronic Ack	knowledgement Receipt		
EFS ID:	12373166		
Application Number:	13303417		
International Application Number:			
Confirmation Number:	3035		
Title of Invention:	SYSTEM AND METHOD FOR SCRAMBLING THE PHASE OF THE CARRIERS IN A MULTICARRIER COMMUNICATIONS SYSTEM		
First Named Inventor/Applicant Name:	Marcos C. Tzannes		
Customer Number:	62574		
Filer:	Jason Vick/Joanne Vos		
Filer Authorized By:	Jason Vick		
Attorney Docket Number:	5550-47-CON-5		
Receipt Date:	22-MAR-2012		
Filing Date:	23-NOV-2011		
Time Stamp:	17:42:04		
Application Type:	Utility under 35 USC 111(a)		

Payment information:

Submitted with Payment	no
File Linking and	

File Listing:

Document Number	Document Description	File Name	File Size(Bytes)/ Message Digest	Multi Part /.zip	Pages (if appl.)
1		IDS 03.pdf	331904	V05	4
'		iD3_03.βui	7f55633356e0882915f78def325d8bcb7f5a 479c	yes	4

	Multip	Multipart Description/PDF files in .zip description					
	Document Description		Start	End			
	Transmittal	1		3			
	Information Disclosure Stater	4	4				
Warnings:			1				
Information:							
2	Non Patent Literature	5550-47-CON-DIV-	1675020	no	33		
	North atent Electrical	CON_NOA_03-20-2012.pdf	0291b0a566ffbabfdcc7ac055302a5dc2468 c96e	110	33		
Warnings:							
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New Applications Under 35 U.S.C. 111

If a new application is being filed and the application includes the necessary components for a filing date (see 37 CFR 1.53(b)-(d) and MPEP 506), a Filing Receipt (37 CFR 1.54) will be issued in due course and the date shown on this Acknowledgement Receipt will establish the filing date of the application.

National Stage of an International Application under 35 U.S.C. 371

If a timely submission to enter the national stage of an international application is compliant with the conditions of 35 U.S.C. 371 and other applicable requirements a Form PCT/DO/EO/903 indicating acceptance of the application as a national stage submission under 35 U.S.C. 371 will be issued in addition to the Filing Receipt, in due course.

New International Application Filed with the USPTO as a Receiving Office

If a new international application is being filed and the international application includes the necessary components for an international filing date (see PCT Article 11 and MPEP 1810), a Notification of the International Application Number and of the International Filing Date (Form PCT/RO/105) will be issued in due course, subject to prescriptions concerning national security, and the date shown on this Acknowledgement Receipt will establish the international filing date of the application.

IN THE UNITED STATES PATENT AND TRADEMARK OFFICE

n Re the Application of:) Group Art Unit: 2611
TZANNES, Marcos C.	Confirmation No.: 3035
Serial No.: 13/303,417) Examiner: Not Yet Assigned
Filed: November 23, 2011 Atty. File No.: 5550-47-CON-5 Entitled: "System and Method for Scrambling the Phase of the Carriers in a Multicarrier) SUPPLEMENTAL) INFORMATION DISCLOSURE) STATEMENT)
Communications System" Commissioner for Patents P.O. Box 1450 Alexandria, VA 22313-1450	Electronically Submitted
Dear Sir:	
The references cited on attached Form PTO of the Examiner. Copies of the cited non-patent and/or foreign Copies of the cited U.S. patents and/or paten Copies of the cited U.S. patents/patent applies	references are enclosed herewith. t applications are enclosed herewith.
accordance with 37 C.F.R. § 1.98(a).	·
Copies of the cited references are not ene § 1.98(d), because the references were cited by	
Trademark Office in prior application Serial No	
which is relied upon for an earlier filing date under	35 U.S.C. § 120.
To the best of applicants' belief, the pertine	ence of the foreign-language references
are believed to be summarized in the attached E	inglish abstracts and/or in the figures,
although applicants do not necessarily vouch for the	e accuracy of the translation.
Examiner's attention is drawn to the following	
Serial No filed	(Attorney Ref. No)
Serial No filed	(Attorney Ref. No)
Other:	
Submission of the above information is not	
is citable under the statutes or rules to support	
represents analogous art, or that those skilled in	the art would refer to or recognize the
pertinence of any reference without the benefit of	f hindsight, nor should an inference be

drawn as to the pertinence of the references based on the order in which they are presented. Submission of this statement should not be taken as an indication that a search has been conducted, or that no better art exists.

It is respectfully requested that the cited information be expressly considered during the prosecution of this application and the references made of record therein.

FEES

\boxtimes	37 CFR 1.97(b): No fee is believed due in connection with this submission, because the information disclosure statement submitted herewith is satisfied by one of the following conditions ("X" indicates satisfaction):
l	Within three months of the filing date of a national application other than a continued prosecution application under 37 CFR 1.53(d), or
	Within three months of the date of entry of the national stage as set forth in § 1.491 in an international application, or
	Before the mailing date of a first Office Action on the merits, or
	Before the mailing of a first Office action after the filing of a request for continued examination under 37 CFR 1.114.
	Although no fee is believed due, if any fee is deemed due in connection with this submission, please charge such fee to Deposit Account 19-1970.
	37 CFR 1.97(c): The information disclosure statement transmitted herewith is being filed after all the above conditions (37 CFR 1.97(b)), but before the mailing date of any one of the following conditions: (1) a final action under 37 C.F.R. 1.113, or (2) a notice of allowance under 37 C.F.R. 1.311, or (3) an action that otherwise closes prosecution in the application. This Information Disclosure Statement is accompanied by: A Certification (below) as specified by 37 C.F.R. 1.97(e). Although no fee is believed due, if any fee is deemed due in connection with this submission, please charge such fee to Deposit Account 19-1970. OR Please charge Deposit Account 19-1970 in the amount of \$180.00 for the fee set forth in 37 C.F.R. 1.17(p) for submission of an information disclosure statement. Please credit any overpayment or charge any underpayment to Deposit Account 19-1970.
	37 CFR 1.97(d): This Information Disclosure Statement is being submitted after the period specified in 37 CFR 1.97(c). This information Disclosure Statement includes a Certification (below) as specified by 37 C.F.R. 1.97(e) AND Applicants hereby requests consideration of the reference(s) disclosed herein. Please charge Deposit Account 19-1970 in the amount of \$180.00 under 37 C.F.R. 1.17(p). Please credit any overpayment or charge any underpayment to
	Deposit Account 19-1970. Election to pay the fee should not be taken as an indication that applicant(s) cannot execute a certification.

Certification (37 C.F.R. 1.97(e)) (Applicable only if checked)
The undersigned certifies that: Each item of information contained in this information disclosure statement was first cited in any communication from a foreign patent office in a counterpart foreign application not more than three months prior to the filing of this statement. 37 C.F.R. 1.97(e)(1). A copy of the communication from the foreign patent office is enclosed.
OR
No item of information contained in this information disclosure statement was cited in a communication from a foreign patent office in a counterpart foreign application, and, to the knowledge of the undersigned after making reasonable inquiry, no item of information contained in this Information Disclosure Statement was known to any individual designated in 37 C.F.R. 1.56(c) more than three months prior to the filing of this statement. 37 C.F.R. 1.97(e)(2).
Respectfully submitted,
SHERIDAN ROSS P.C.
By: Jason H. Vick Registration No. 45,285 1560 Broadway, Suite 1200 Denver, Colorado 80202-5141 (303) 863-9700

Subs	stitute for form 1	449A/PTO		Comp	lete if Known
				Application Number	13/303,417
	Substitute for form 1449A/PTO INFORMATION DISCLOSU STATEMENT BY APPLICA Sheet 1 of		Filing Date	November 23, 2011	
ST	ATEME	NT BY AP	PLICANT	First Named Inventor	Marcos C. Tzannes
				Art Unit	2611
				Examiner Name	Not Yet Assigned
Sheet	1	of	1	Attorney Docket Number	5550-47-CON-5

Dublication Date		
Publication Date MM-DD-YYYY	Name of Patentee of Applicant of Cited Document	Pages, Columns, Lines, Where Relevant Passages or Relevant Figures Appear
=	, 2 (If known) MM-DD-YYYY	MM-DD-YYYY Applicant of Cited Document

	FOREIGN PATENT DOCUMENTS						
Examiner Initials*	Cite No. ¹	Foreign Patent Document Country Code ³ ; Number ⁴ ; Kind Code ⁵ (<i>if known</i>)		Applicant of Cited Document	Pages, Columns, Lines, Where Relevant Passages or Relevant Figures Appear	Т ⁶	

	OTHER ART (Including Author, Title, Date, Pertinent Pages, etc.)							
Examiner Initials*	Cite No. ¹							
	1	Invitation to Pay Additional Fees for International (PCT) Patent Application No. PCT/US00/30958, mailed March 23, 2001 (Attorney Ref. No. 5550-47-PCT)						
	2	Invitation to Restrict or to Pay Additional Fees for International (PCT) Patent Application No. PCT/US00/30958, mailed Nov. 14, 2001 (Attorney Ref. No. 5550-47-PCT)						

Examiner	Date
	Considered
Signature	Considered

^{*}EXAMINER: Initial if reference is considered, whether or not citation is in conformance and not considered. Include copy of this form with next communication to applicant.

PATENT COOPERATION TREA.

From the INTERNATIONAL SEARCHING AUTHORITY	PCT					
To: TESTA, HURWITZ & THIBEAULT, L.L.P. Attn. Rodriguez, M. A. High Street Tower 125 High Street Boston, Massachusetts 02110 UNITED STATES OF AMERICA	INVITATION TO PAY ADDITIONAL FEES (PCT Article 17(3)(a) and Rule 40.1)					
	Date of mailing (day/month/year) 23/03/2001					
Applicant's or agent's file reference AWR-017PC 081513-49	PAYMENT DUE within 45 XXXXXXs/days from the above date of mailing					
International application No. PCT/US 00/ 30958	International filing date (day/month/year) 09/11/2000					
AWARE, INC.						
1. This International Searching Authority (i) considers that there are						
 (ii) X has carried out a partial international search (see Annex) will establish the international search report on those parts of the international application which relate to the invention first mentioned in claims Nos.: 1-13,20-30,37-38 (iii) will establish the international search report on the other parts of the international application only if, and to the extent to which, additional fees are paid 						
2. The applicant is hereby invited, within the time limit indicated above, to pay the amount indicated below:						
3. Claim(s) Nos. Article 17(2)(b) because of defects under Article 17(2)(a)	have been found to be unsearchable under and therefore have not been included with any invention.					
Name and mailing address of the International Searching Authority European Patent Office, P.B. 5818 Patentlaan 2 NL-2280 HV Rijswijk Tel. (+31-70) 340-2040, Tx. 31 651 epo nl, Fax: (+31-70) 340-3016	Authorized officer Hans Pettersson					

Form PCT/ISA/206 (July 1992)

This International Searching Authority found multiple (groups of) inventions in this international application, as follows:

1. Claims: 1-13,20-30, 37-38

carrier phase scrambling

2. Claims: 14-19,31-36

information throughput improvement

3. Claim: 39

error correction

The prior art document D1 has been identified as patent EP719004, "OFDM system with additional protection again multipath effects", published

This document describes an OFDM system which uses a pseudo-random signal generator for rotating the phase of the OFDM carriers .

Here follows an analysis of the claims:

Claims 1-3, 6-12, 20,21,24-30, 37-38:

The subject matter of these claims has been entirely disclosed by the prior art document, then the new features with respect to the prior art have to be searched within the dependent claims 4-5, 13-19, 22-23, 31-36, 39

Claims 4, 5, 13, 22, 23 : the New Features with respect to the prior art are: the transmission of the rotation data to the receiver, the synchronization of the transmitter and the receiver through the transmission of the rotation data, the additional digital bit scrambling of the data source. Considering D1 and Rule 13.2 PCT the previous New Features must be considered Special Technical Features.

The objective problem solved by these Special Technical Features is to avoid the loss of synchronization between the transmitter and the receiver.

Claims 14-19, 31-36:

the New Feature with respect to the prior art is: sending of a predetermined signal when clipping is detected. Considering D1 and Rule 13.2 PCT the previous New Feature must be considered a Special Technical Feature. The objective problem solved by this Special Technical Feature is to assure an information throughput during the clipping event.

Claim 39:

the New Features with respect to the prior art are: comparison of the DMT received signal with a predetermined one, discard/demodulate the received signal in accordance with the previous comparison. Considering D1 and Rule 13.2 PCT the previous New Features must be

considered Special Technical Features. The objective problem solved by these Special Technical Features is to error correct the received signal.

Being the objective problems solved by the different groups of claims different and being the Special Technical Features of the different groups nor the same nor corresponding, the subject application shows a lack of unity and 3 different inventions can be individuated.

1st invention : claims 1 (and dependent 2-13), 20 (and dep. 21-30), 37 (and dep. 38).

2nd invention : claims 1 (and dep. 14-19), 20 (and dep. 31-36)

3rd invention : claim 39

Ar : to Form PCT/ISA/206 COMMUNICA...ON RELATING TO THE RESULTS OF THE PARTIAL INTERNATIONAL SEARCH

International Application No PCT/US 00/30958

- 1.The present communication is an Annex to the invitation to pay additional fees (Form PCT/ISA/206). It shows the results of the international search established on the parts of the international application which relate to the invention first mentioned in claims Nos.:
- 1-13,20-30,37,38 2. This communication is not the international search report which will be established according to Article 18 and Rule 43.
- 3.If the applicant does not pay any additional search fees, the information appearing in this communication will be considered as the result of the international search and will be included as such in the international search report.
- 4.If the applicant pays additional fees, the international search report will contain both the information appearing in this communication and the results of the international search on other parts of the international application for which such fees will have been paid.

C. DOCUM	ENTS CONSIDERED TO BE RELEVANT	
Category °	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
Х	EP 0 719 004 A (MATSUSHITA ELECTRIC IND CO LTD) 26 June 1996 (1996-06-26)	1-3, 6-12,20, 21, 24-30, 37,38
	column 3, line 28 - line 46	, , , , ,
Υ.	column 14, line 39 - line 55 column 16, line 4 - line 6	13
Y	WO 99 29078 A (BAHLENBERG GUNNAR; HAAKANSSON SIWERT (SE); LJUNGGREN LIS MARIE (SE) 10 June 1999 (1999-06-10) page 5, line 18 - line 23	13
X	BAUML R W ET AL: "REDUCING THE PEAK-TO-AVERAGE POWER RATIO OF MULTICARRIER MODULATIONBY SELECTED MAPPING" ELECTRONICS LETTERS,GB,IEE STEVENAGE, vol. 32, no. 22, 24 October 1996 (1996-10-24), pages 2056-2057, XP000643915 ISSN: 0013-5194 page 2056, right-hand column, paragraphs 7,8 -page 2057, left-hand column, paragraph 3	1,2,4,5, 20,22, 23,37,38
А	WO 98 32065 A (FORTRESS TECHNOLOGIES INC) 23 July 1998 (1998-07-23) page 16, line 30 -page 17, line 3 -/	8-12, 26-30
	the standard of the sadding th	lated in cases
X Fu	ther documents are listed in the continuation of box C. X Patent far	mily members are listed in annex.
"A" docun	nent defining the general state of theart which is not cited to under invention cited to under invention	t published after theinternational filing date e and not in conflict with theapplication but rstand the principle or theory underlying the
filing "L" docum which citati "O" docum other "P" docum	document but published on or after the international date "X" document of p cannot be concent which may throw doubts on priority claim(s) or involve an in	articular relevance; the claimed invention nsidered novel or cannot be considered to ventive step when the document is taken alone articular relevance; the claimed invention nsidered to involve an inventive step when the combined with one or more othersuch docucombination being obvious to aperson skilled

Form PCT/ISA/206 (Annex, first sheet) (July 1992)

1

Ar to Form PCT/ISA/206 COMMUNICALION RELATING TO THE RESULTS OF THE PARTIAL INTERNATIONAL SEARCH

International Application No
PCT/US 00/30958

Category °	ation) DOCUMENTS CONSIDERED TO BE RELEVANT Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
alegoly -	onation of document, with indication, where appropriate, of the relevant passages	nelevant to claim No.
4	WO 99 22463 A (MOTOROLA INC) 6 May 1999 (1999-05-06) page 7, line 29 - line 32	8-12, 26-30

Form PCT/ISA/206 (Annex, continuation sheet)(July 1992)

Patent Family Annex

Information on patent family members

International Application No PCT/US 00/30958

Patent document cited in search report		Publication date	Patent family member(s)		Publication date	
EP 0719004	Α	26-06-1996	JP US	8321820 A 5682376 A	03-12-1996 28-10-1997	
WO 9929078	Α	10-06-1999	SE	9704497 A	04-06-1999	
WO 9832065	Α	23-07-1998	AU EP	5811598 A 0951767 A	07-08-1998 27-10-1999	
WO 9922463	Α	06-05-1999	US US	6044276 A 6061574 A	28-03-2000 09-05-2000	

Form PCT/ISA/206 (patent family annex) (July 1992)

	PATENT COOPE	RATION TRE	A./ REC	EIVE
From the INTERNATIONAL PRELIMINARY E. To: VICK, Jason H. Nixon Peabody LLP 8180 Greensboro Drive,Suite 80 McLean,Virginia 22102 ETATS-UNIS D'AMERIQUE	Cartimon	tion "	PCT NVITATION TO RESTRICT TO PAY ADDITIONAL FEE OT Article 34(3) (a) and Rul 14.11.2001	S
Applicant's or agent's file reference 081513-49		REPLY OR PAYMENT DUE	within 1 month(s) from the above date of mailin	g
International application No. PCT/US00/30958	International filing date (da 09/11/2000	1 ay/month/year)	Priority date (day/month/year) 09/11/1999	
International Patent classification (IPC) of H04L27/26	or national Patent classification:		t	
Applicant				
AWARE, INC.				
4 This leads a sign of the continuous	A			

- This International Examining Authority
 - considers that the international application does not comply with the requirements of unity of invention (Rule 13.1, 13.2 and 13.3) for the reasons indicated in the Annex.
 - therefore considers that there are 2 inventions claimed in the international application as indicated in the Annex.
 - (iii) recalls that claims relating to inventions in respect of which no international search report has been established need not be the subject of international preliminary examination (Rule 66.1 (e)).
- 2. Consequently the applicant is hereby invited, within the time limit indicated above, to restrict the claims as suggested under item 3, below, or to pay the amount indicated below:

eur 1533.00 eur 1533.00 number of additional inventions total amount of additional fees Fee per additional invention

The applicant is informed that, according to Rule 68.3 (c), the payment of any additional fee may be made under protest, i.e. a reasonned statement to the effect that the international application complies with the requirement of unity of invention or that the amount of the required additional fee is excessive.

- 3. If the applicant opts to restrict the claims, this Authority suggests the restriction possibilities indicated in the Annex, which in its opinion would be in compliance with the requirement of unity of invention.
- 4. In the absence of any response from the applicant, this Authority will establish the international preliminary examination report on those parts of the international application indicated in the Annex which, in the opinion of this Authority appear to relate to the main invention.

Name and mailing adress of the international preliminary examination authority: European Patent Office

D-80298 Munich Tel. +49 89 2399 - 0 Tx: 523656 epmu d

Fax: +49 89 2399 - 4465

Authorized officer

Pajatakis, E

Telephone No. +49 89 2399-8898



Form PCT/IPEA/405 (July 1992)

- 1. Independent Claims 1, 20 and 37 are directed to scrambling the phase characteristics of carrier signals. This is done by phase shifting a carrier signal by an amount computed from a value which is independent from any input bit value carried by the carrier signal.
 - On the other hand independent Claim 39 is directed to avoiding demodulation errors due to clipping of transmitted signals as result of non linear distortions. This is done by comparing received DMT symbols with a bit-value pattern. DMT symbols matching the bit-value pattern are discarded.
- 2. Thus, the application comprises two groups of independent claims with different features which are based on different concepts.
- 3. If a further fee is not paid, the examination will be carried out on the invention mentioned first in the claims.

Electronic Acknowledgement Receipt						
EFS ID:	12284970					
Application Number:	13303417					
International Application Number:						
Confirmation Number:	3035					
Title of Invention:	SYSTEM AND METHOD FOR SCRAMBLING THE PHASE OF THE CARRIERS IN A MULTICARRIER COMMUNICATIONS SYSTEM					
First Named Inventor/Applicant Name:	Marcos C. Tzannes					
Customer Number:	62574					
Filer:	Jason Vick/Joanne Vos					
Filer Authorized By:	Jason Vick					
Attorney Docket Number:	5550-47-CON-5					
Receipt Date:	12-MAR-2012					
Filing Date:	23-NOV-2011					
Time Stamp:	18:30:07					
Application Type:	Utility under 35 USC 111(a)					

Payment information:

Submitted with	n Payment	no				
File Listing	:					
Document Number	Document Description		File Name File Size(Bytes)/ Multi Message Digest Part /.zip (Pages (if appl.)
1			IDS 02.pdf	339986	yes	4
'			103_02.pd1	63dd0027bec5a6ff0f06576246bd4cff54146 707	yes	7

	Multipart Description/PDF files in .zip description							
	Document	Description	Start	En	d			
	Transmit	1	3					
	Information Disclosure St	4	4					
Warnings:								
Information:								
2	Non Patent Literature	5550-47- PCT_Invite_to_Pay_Fees_03-23	540050	no	6			
		-2001.pdf	227f1e5d63ca3b470ec978bce1dd87ff72f9 0ce6					
Warnings:								
Information:								
3	Non Patent Literature	5550-47- PCT_Invite_to_Pay_Fees_11-14	206677	no	2			
		-2001.pdf	8fb2afa439380e92f947f86b614a23392161 df70					
Warnings:		·		•				
Information:								
		Total Files Size (in bytes)	1086	5713				

This Acknowledgement Receipt evidences receipt on the noted date by the USPTO of the indicated documents, characterized by the applicant, and including page counts, where applicable. It serves as evidence of receipt similar to a Post Card, as described in MPEP 503.

New Applications Under 35 U.S.C. 111

If a new application is being filed and the application includes the necessary components for a filing date (see 37 CFR 1.53(b)-(d) and MPEP 506), a Filing Receipt (37 CFR 1.54) will be issued in due course and the date shown on this Acknowledgement Receipt will establish the filing date of the application.

National Stage of an International Application under 35 U.S.C. 371

If a timely submission to enter the national stage of an international application is compliant with the conditions of 35 U.S.C. 371 and other applicable requirements a Form PCT/DO/EO/903 indicating acceptance of the application as a national stage submission under 35 U.S.C. 371 will be issued in addition to the Filing Receipt, in due course.

New International Application Filed with the USPTO as a Receiving Office

If a new international application is being filed and the international application includes the necessary components for an international filing date (see PCT Article 11 and MPEP 1810), a Notification of the International Application Number and of the International Filing Date (Form PCT/RO/105) will be issued in due course, subject to prescriptions concerning national security, and the date shown on this Acknowledgement Receipt will establish the international filing date of the application.

IN THE UNITED STATES PATENT AND TRADEMARK OFFICE

In Re the Application of:) Group Art Unit: 2611
TZANNES, Marcos C.	Confirmation No.: 3035
Serial No.: 13/303,417) Examiner: Not Yet Assigned
Filed: November 23, 2011)
Atty. File No.: 5550-47-CON-5	SUPPLEMENTAL INFORMATION DISCLOSURE
Entitled: "System and Method for Scrambling the Phase of the Carriers in a Multicarrier Communications System") STATEMENT) Electronically Submitted
Commissioner for Patents P.O. Box 1450 Alexandria, VA 22313-1450	
Dear Sir:	
The references cited on attached Form PTC	0-1449 are being called to the attention
of the Examiner.	
Copies of the cited non-patent and/or foreign Copies of the cited U.S. patents and/or pater Copies of the cited U.S. patents/patent appl accordance with 37 C.F.R. § 1.98(a).	nt applications are enclosed herewith. ication publications are not enclosed in
Copies of the cited references are not en	
§ 1.98(d), because the references were cited by	
Trademark Office in prior application Serial No.	
which is relied upon for an earlier filing date under	
To the best of applicants' belief, the pertine	
are believed to be summarized in the attached E	
although applicants do not necessarily vouch for the	·
Examiner's attention is drawn to the following	(Attorney Ref. No.
Serial No filed	- \ -
Serial No filed	_(Attorney Ref. No)
Submission of the above information is not	intended as an admission that any item
is citable under the statutes or rules to suppor represents analogous art, or that those skilled in	•
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drawn as to the pertinence of the references based on the order in which they are presented. Submission of this statement should not be taken as an indication that a search has been conducted, or that no better art exists.

It is respectfully requested that the cited information be expressly considered during the prosecution of this application and the references made of record therein.

FEES

37 CFR 1.97(b): No fee is believed due in connection with this submission, because the information disclosure statement submitted herewith is satisfied by one of the following conditions ("X" indicates satisfaction): Within three months of the filing date of a national application other than a continued prosecution application under 37 CFR 1.53(d), or Within three months of the date of entry of the national stage as set forth in § 1.491 in an international application, or Before the mailing date of a first Office Action on the merits, or Before the mailing of a first Office action after the filing of a request for continued examination under 37 CFR 1.114. Although no fee is believed due, if any fee is deemed due in connection with this submission, please charge such fee to
Deposit Account 19-1970. 37 CFR 1.97(e): The information disclosure statement transmitted herewith is being filed after all the above conditions (37 CFR 1.97(b)), but before the mailing date of any one of the following conditions: (1) a final action under 37 C.F.R. 1.113, or (2) a notice of allowance under 37 C.F.R. 1.311, or (3) an action that otherwise closes prosecution in the application. This Information Disclosure Statement is accompanied by: A Certification (below) as specified by 37 C.F.R. 1.97(e). Although no fee is believed due, if any fee is deemed due in connection with this submission, please charge such fee to Deposit Account 19-1970. OR Please charge Deposit Account 19-1970 in the amount of \$180.00 for the fee set forth in 37 C.F.R. 1.17(p) for submission of an information disclosure statement. Please credit any overpayment or charge any underpayment to Deposit Account 19-1970.
37 CFR 1.97(d): This Information Disclosure Statement is being submitted after the period specified in 37 CFR 1.97(c). This information Disclosure Statement includes a Certification (below) as specified by 37 C.F.R. 1.97(e) AND Applicants hereby requests consideration of the reference(s) disclosed herein. Please charge Deposit Account 19-1970 in the amount of \$180.00 under 37 C.F.R. 1.17(p). Please credit any overpayment or charge any underpayment to Deposit Account 19-1970. Election to pay the fee should not be taken as an indication that applicant(s) cannot execute a certification.

Certification (37 C.F.R. 1.97(e)) (Applicable only if checked)
The undersigned certifies that: Each item of information contained in this information disclosure statement was first cited in any communication from a foreign patent office in a counterpart foreign application not more than three months prior to the filing of this statement. 37 C.F.R. 1.97(e)(1). A copy of the communication from the foreign patent office is enclosed.
OR
No item of information contained in this information disclosure statement was cited in a communication from a foreign patent office in a counterpart foreign application, and, to the knowledge of the undersigned after making reasonable inquiry, no item of information contained in this Information Disclosure Statement was known to any individual designated in 37 C.F.R. 1.56(c) more than three months prior to the filing of this statement. 37 C.F.R. 1.97(e)(2).
Respectfully submitted,
SHERIDAN ROSS P.C.
By: Jason H. Vick Registration No. 45,285 1560 Broadway, Suite 1200 Denver, Colorado 80202-5141 (303) 863-9700

I A I E IVI AI I E IVA II VIVI E E DE I E I III III I A I I E IVI I I E I E I I I I I I I I I I									ition or Docket Num 13,417	ber
	APPLI	CATION A			umn 2)	SMALL	ENTITY	OR	OTHER SMALL I	
	FOR	NUMBE	BER FILED NUMBER EXTRA RATE(\$) FEE(\$)		1	RATE(\$)	FEE(\$)			
	IC FEE FR 1.16(a), (b), or (c))			J/A	N/A		1	N/A	380	
SEA	RCH FEE FR 1.16(k), (i), or (m))	N	/A	١	J/A	N/A		1	N/A	620
ΞXΑ	MINATION FEE FR 1.16(o), (p), or (q))	N	/A	١	I/A	N/A		1	N/A	250
ГОТ.	AL CLAIMS FR 1.16(i))	1	minus	20= *				OR	x 60 =	0.00
NDE	PENDENT CLAIMS	3 1	minus	3 = *				1	x 250 =	0.00
APF	PLICATION SIZE	sheets of p \$310 (\$15) 50 sheets	oaper, th 5 for sm or fraction	and drawings e e application si all entity) for ea on thereof. See ' CFR 1.16(s).	ze fee due is ch additional					0.00
MUL	TIPLE DEPENDEN	T CLAIM PRE	SENT (3	7 CFR 1.16(j))						0.00
' If th	ne difference in colu	mn 1 is less th	an zero,	enter "0" in colur	nn 2.	TOTAL		1	TOTAL	1250
AMENDMENI A	Total * (37 CFR 1.16(i))	REMAINING AFTER AMENDMENT	Minus	NUMBER PREVIOUSLY PAID FOR	PRESENT EXTRA	RATE(\$) =	ADDITIONAL FEE(\$)	OR	RATE(\$)	ADDITIONAL FEE(\$)
2	Independent * (37 CFR 1.16(h))		Minus	***	=	x =		OR	х =	
AME	Application Size Fee	(37 CFR 1.16(s))			-			1		
ı	FIRST PRESENTATION	ON OF MULTIPL	E DEPEN	IDENT CLAIM (37 C	CFR 1.16(j))			OR		
						TOTAL ADD'L FEE		OR	TOTAL ADD'L FEE	
_		(Column 1) CLAIMS		(Column 2) HIGHEST	(Column 3)			7		
n Z		REMAINING AFTER AMENDMENT		NUMBER PREVIOUSLY PAID FOR	PRESENT EXTRA	RATE(\$)	ADDITIONAL FEE(\$)		RATE(\$)	ADDITIONA FEE(\$)
AMENDMEN	Total * (37 CFR 1.16(i))		Minus	**	=	x =		OR	x =	
	Independent * (37 CFR 1.16(h))		Minus	***	=	x =		OR	x =	
Ž	Application Size Fee	(37 CFR 1.16(s))								
	FIRST PRESENTATION	ON OF MULTIPL	E DEPEN	IDENT CLAIM (37 C	CFR 1.16(j))			OR		
						TOTAL ADD'L FEE		OR	TOTAL ADD'L FEE	
**	If the entry in colur If the "Highest Nur If the "Highest Num	mber Previous ber Previously I	ly Paid F Paid For"	or" IN THIS SPA IN THIS SPACE is	CE is less than 2 s less than 3, ente	nn 3. 20, enter "20".	in column 1	_		



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UNITED STATES DEPARTMENT OF COMMERCE United States Patent and Trademark Office Address: COMMISSIONER FOR PATENTS P.C. Box 1450 Alexandria, Virginia 22313-1450 www.uspio.gov

APPLICATION	FILING or	GRP ART				
NUMBER	371(c) DATE	UNIT	FIL FEE REC'D	ATTY.DOCKET.NO	TOT CLAIMS	IND CLAIMS
13/303 417	11/23/2011	2611	1250	5550-47-CON-5	1	1

62574 Jason H. Vick Sheridan Ross, PC Suite # 1200 1560 Broadway Denver, CO 80202 CONFIRMATION NO. 3035
FILING RECEIPT



Date Mailed: 12/12/2011

Receipt is acknowledged of this non-provisional patent application. The application will be taken up for examination in due course. Applicant will be notified as to the results of the examination. Any correspondence concerning the application must include the following identification information: the U.S. APPLICATION NUMBER, FILING DATE, NAME OF APPLICANT, and TITLE OF INVENTION. Fees transmitted by check or draft are subject to collection. Please verify the accuracy of the data presented on this receipt. If an error is noted on this Filing Receipt, please submit a written request for a Filing Receipt Correction. Please provide a copy of this Filing Receipt with the changes noted thereon. If you received a "Notice to File Missing Parts" for this application, please submit any corrections to this Filing Receipt with your reply to the Notice. When the USPTO processes the reply to the Notice, the USPTO will generate another Filing Receipt incorporating the requested corrections

Applicant(s)

Marcos C. Tzannes, Orinda, CA;

Assignment For Published Patent Application

AWARE, INC., Bedford, MA

Power of Attorney: The patent practitioners associated with Customer Number 62574

Domestic Priority data as claimed by applicant

This application is a CON of 12/783,725 05/20/2010 which is a CON of 12/255,713 10/22/2008 PAT 7,769,104 which is a CON of 11/863,581 09/28/2007 PAT 7,471,721 which is a CON of 11/211,535 08/26/2005 PAT 7,292,627 which is a CON of 09/710,310 11/09/2000 PAT 6,961,369 which claims benefit of 60/164,134 11/09/1999

Foreign Applications (You may be eligible to benefit from the **Patent Prosecution Highway** program at the USPTO. Please see http://www.uspto.gov for more information.)

If Required, Foreign Filing License Granted: 12/06/2011

The country code and number of your priority application, to be used for filing abroad under the Paris Convention, is **US 13/303,417**

Projected Publication Date: 03/22/2012

Non-Publication Request: No Early Publication Request: No

page 1 of 3

Title

SYSTEM AND METHOD FOR SCRAMBLING THE PHASE OF THE CARRIERS IN A MULTICARRIER COMMUNICATIONS SYSTEM

Preliminary Class

375

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APPLICATION NUMBER FILING OR 371(C) DATE FIRST NAMED APPLICANT ATTY. DOCKET NO./TITLE 13/303,417 11/23/2011 Marcos C. Tzannes 5550-47-CON-5

62574 Jason H. Vick Sheridan Ross, PC Suite # 1200 1560 Broadway Denver, CO 80202

CONFIRMATION NO. 3035 POA ACCEPTANCE LETTER



Date Mailed: 12/12/2011

NOTICE OF ACCEPTANCE OF POWER OF ATTORNEY

This is in response to the Power of Attorney filed 11/23/2011.

The Power of Attorney in this application is accepted. Correspondence in this application will be mailed to the above address as provided by 37 CFR 1.33.

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Office of Data Management, Application Assistance Unit (571) 272-4000, or (571) 272-4200, or 1-888-786-0101

Sub	stitute for form	1449A/PTO		Comp	olete if Known	
18.		TION DIO	N OOUDE	Application Number	13/303,417	
			CLOSURE	Filing Date	November 23, 2011	
5	IAIEME	NT BY AP	PLICANI	First Named Inventor	Marcos C. Tzannes	
				Art Unit	2611	
				Examiner Name	Not Yet Assigned	
Sheet	1	of	4	Attorney Docket Number	5550-47-CON-5	

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				Art Unit	2611	
				Examiner Name	Not Yet Assigned	
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) 5	STATEMENT BY APPLICANT			First Named Inventor	Marcos C. Tzannes	
				Art Unit	2611	
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				Filing Date	November 23, 2011	
STATEMENT BY APPLICANT				First Named Inventor	Marcos C. Tzannes	
				Art Unit	2611	
				Examiner Name	Not Yet Assigned	
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- Method and apparatus for reducing the peak power of data sequences.
- ⑤ This application concerns the reduction of the peak power of data sequences, particularly for use in a QAM radio relay system. The peak power reduction leads to fewer problems with non-linear distortion, whether caused by the channel, or by the transmitter power amplifier.

The power reduction is achieved by using a shaping code, which replaces sequences with high power, by sequences with lower power.

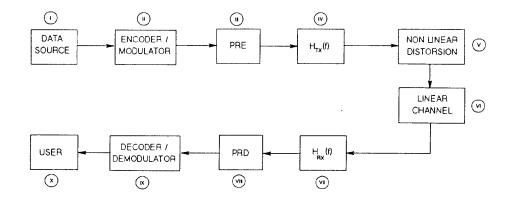


Fig. 1

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(3.10/3.09/3.3.4)

EP 0 584 534 A1

Background of the invention

The present invention relates to a method of reducing the peak power of the signal at the output of the transmit filter of a digital link, e.g. a microwave one. Such reduction allows to minimize the effects of the transmit channel nonlinearity, including in it eventual nonlinearities of the transmit amplifier.

State of the art

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The present digital transmission systems try to obtain high spectral efficiencies through gradually more complex modulation formats. The higher spectral efficiency is counterbalanced by the need of increasing the transmitted power to obtain a prefixed BER (Bit Error Rate: number of wrong bits to total number of bits ratio) value at the receiver. The power delivered by the transmitter generally is limited by the final power amplifier, which has a greatly nonlinear behaviour.

Therefore a serious problem arises with regard to the best exploitation of the nonlinear part of the inputoutput characteristic of the "channel", including in the latter the final amplifier of the transmitter. At present the problem is faced in one of the following ways (see e.g. the papers of G. Karam, H. Sari, "Analysis of predistortion, equalization and ISI cancellation techniques in digital radio systems with nonlinear transmit amplifier", IEEE Transaction on Communications, vol. 37, n. 12, Dec. 1989):

- 1) data predistortion: one tries to modify the constellation used for driving the nonlinear amplifier through a signal such as to obtain the desired constellation at its output;
- 2) analog signal predistortion: a nonlinear circuit having a characteristic opposite to the one of the above-defined "channel", is inserted in the path of the analog signal;
- 3) channel equalization and nonlinear cancellation of the ISI: the receive equalizer tries to cancel the interferences connected with nonlinearity from the present signal sample (through a suitable nonlinear combination of pre- and post-cursors);
- 4) use of "circular" constellations so as to reduce the ratio between the peak power and the average power of the not-filtered signal.

All the above solutions, under special circumstances, can provide unsatisfactory features. In particular the first three ones are not much efficient in the presence of hard limiter characteristic of the transmitter final amplifier; the last one gives rise to gains anyway slight which can be not sufficient in case of reception filter with very narrow band.

Summary of the invention

It is an object of the invention to individuate a base-band system which - at parity of other conditions - reduces the peak power of the filtered signal, i.e. at the input of the nonlinear channel defined above.

It has been found, inter alia, that such reduction is to advantage of radio relay systems links, e.g. allowing the use of smaller antennas or the transmission over longer path sections.

The outstanding features of the invention are set forth in the claims while the various aspects and advantages of the invention will become more apparent from the following description (not limiting).

General solution

The basic idea of the invention is based upon the possibility (other conditions such as minimum distance between transmitted points, average transmitted power, etc. being equal) of avoiding transmission of sequences which a high peak power of the filtered signal is associated with, replacing them with more suitable ones (i.e. with a lower peak power of the filtered signal).

The possibility of carrying out this replacement is given by the increasing of the dimension of the alphabet of the transmitted points. In reception the unwanted sequences, suppressed in transmission, are reconstituted in their original form.

By reducing in this way the peak power of the filtered signal it is possible to exploit in a much more efficient manner the nonlinear characteristic of the above-defined "channel".

Fig. 1 illustrates the schematic block diagram of a generic digital transmission system (blocks I, II, IV, V, VI, VII, IX, X) in which blocks II and VIII, subject of this invention, are inserted. In particular, fig. 1 shows:

- the DATA SOURCE (ref. I) which provides the numeric sequence to be transmitted at its output;
- an ENCODER/MODULATOR block (ref. II) which receives at the input the numeric sequence to be transmitted and carries out the standard encoding operations designed for BER reduction (block, convolutional, Trellis Code Modulation, etc., type encoding) and modulation operations, providing at its

output one of the points of the constellation to be transmitted;

- a PRE block (ref. III), subject of the invention along with block VIII, which eliminates from the transmission the unwanted sequences in terms of peak power of the filtered signal, i.e. of the signal at the output of block IV described below;
- the transmission filter HTx (f) (ref. IV) which provides at its output the analog signal to be transmitted;
- a NONLINEAR DISTORTION block (ref. V) representing an unwanted nonlinear distortion on the signal path. It can be due to the nonlinear characteristic of the final amplifier of the transmitter (as it happens e.g. in microwave links) or, more in general, to a nonlinear behaviour of the information channel;
- the information channel proper (ref. VI) identified as "LINEAR CHANNEL", which outputs a signal constituted by the signal at its input added to and/or combined with disturbances of various kind;
- the reception filter HRx (f) (ref. VII) which receives the signal from the transmit channel and carries out a suitable filtering;
- a PRD block (ref. VIII), subject of the invention along with block III, which reconstitutes the signal in its
 original form containing the unwanted sequences suppressed in transmission by block III;
- a DECODER/DEMODULATOR block (ref. IX) which receives the outgoing signal from block VIII
 demodulates it and carries out the above-mentioned standard decoding operations, providing the user
 with the numeric sequence subject of the transmission;
- the USER (ref. X) which receives the numeric sequence.

In an advantageous and therefore preferred embodiment, blocks PRE (III) and PRD (VIII) in accordance with the invention are realized in the form of digital encoders. As an example, fig. 2 shows a block diagram illustrating how it is possible to realize the PRE in case of a radio relay system transmission using a quadrature amplitude modulation (QAM). Let M be the points of the two-dimensional constellation to be transmitted in the conventional case (hereinafter "standard" constellation) and MR be the redundance points necessary for the encoding (carried out in PRE) subject of the invention; the resulting constellation is composed of (M+MR) points (hereinafter "expanded" constellation).

Typically: 1 < (M + MR)/M < 1.2.

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In fig. 2 there is shown the preferred embodiment of PRE; it includes:

- A delay element T (ref. XIII) which receives as its input the last two-dimensional element of the block
 of N outgoing two-dimensional symbols from XII and outputs it with a delay equal to one channel
 symbol interval. Such output will be indicated as "state" of the machine in the following.
- A map identified as "(M+MR) MAP" (ref. XI) which receives at its input a block of N symbols of the "standard" constellation and provides (M+MR) blocks of N two-dimensional symbols of the "expanded" constellation. Each output block is relative to a particular "state" of the system and represents the best sequence to be transmitted (in the presence of that particular "state" of the machine) in terms of peak power of the filtered signal.
- A multiplexer "MUX" (ref. XII) having (M+MR) inputs and one output which, on the basis of the "state" at the output of block XIII selects (among the M+MR present at its input) the suitable block of N symbols to be provided at the output.

It remains to be defined what is the meaning of "best sequence in terms of peak power of the filtered signal". According to one particular aspect - even if not limiting - of the invention, the calculation is arranged as follows. Let hTX(t) be the impulse response of the transmission filter IV of fig. 1, T the symbol time, $d^{(k)}$ (k = 1,2,..., (M+MR)) the "state" of the system, Ci = (ci, ci+1, ..., cN-1) the generic block of N two-dimensional symbols, the "weight" w of block Ci = (ci, ci+1, ..., cN-1) given the state d(k), can be defined as the quantity:

$$w(d ,Ci) = \max_{-NT \le t < NT} \left| \frac{hTX(to-T)}{hTX(to-T)} \frac{(k)}{d} + \sum_{j=0}^{N-1} \frac{hTX(to+JT)}{hTX(to+JT)} \frac{2}{cj} \right|^2$$
 (1)

then meaning that the best sequences Ci (in terms of peak power of the filtered signal) are those having a lower "weight" $w(d^{(k)}, Ci)$.

The PRD can be realized through a circuit quite similar to the one shown in fig. 2 for PRE; its description in terms of block diagram (being within the reach of those skilled in the art, in the light of what has been set forth hereinbefore) will be omitted for conciseness' sake.

Reference has been made to specific embodiments represented in figs. 1 and 2 for simplicity and illustrative clearness reasons; therefore it is evident that these are susceptible to those variations, modifications, replacements and the like which, being within the reach of those skilled in the art, naturally

EP 0 584 534 A1

fall within the sphere and the spirit of the scope of the invention.

The following possible variants are here mentioned by way of an example:

- in equation (1) a "state" constituted by several two-dimensional symbols could be envisaged;
- blocks XI and XII of fig. 2 could be replaced by a combinatory algebra, thus transforming the structure of PRE into a convolutional one.
- N could be taken great enough to be able to eliminate in fig. 2 the reaction through block XIII thus transforming the structure of PRE into a "block" structure.

Claims

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- 1. Method of transmitting and receiving numerical signals in which:
 - in transmission, data from a numeric or numerized source are modulated, the modulated signal is filtered and transmitted through a nonlinear channel (where the nonlinearity may be due to the nonlinear characteristic of the final amplifier of the transmitter, or more in general to a nonlinear behaviour of the transmit channel proper),
 - in reception the received signal is filtered and demodulated in order to reconstruct the transmitted numeric sequence

characterized in that:

- in transmission, the unwanted sequences in terms of peak power of the filtered signal are eliminated from the modulated signal before filtering and replaced with suitable sequences,
- in reception, the received and filtered signal is restored in its original form (i.e. containing the unwanted sequences suppressed in transmission) and then sent to the demodulator.
- 2. Method according to claim 1, characterized in that:
 - the link is a digital, radio relay system link and uses a quadrature amplitude modulation (QAM),
 - the replacement of said sequences is carried out through a base-band digital encoder.
- 3. Method according to claim 2, characterized in that a "recurring" coding, i.e. using previously transmitted symbols for individuating the symbol to be transmitted, is used.
- 4. Method according to claim 3, characterized in that the individuation of the sequences to be replaced is carried out on the basis of equation (1) or of relations equivalent thereto.
- 5. Method substantially as hereinbefore described and represented.

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- 6. System for implementing the method of the preceding claims, including:
 - in transmission, a data source, an encoder/modulator, a transmit filter and a nonlinear amplifier;
 - in reception, a filter and a decoder/demodulator,

characterized in that:

- in transmission, an encoder for reducing the peak power of the filtered signal is inserted upstream of the transmit filter.
- in reception, a decoder for reducing the peak power of the filtered signal is inserted downstream of the receive filter.
- 7. System according to claim 6, characterized in that the decoder is of "recurrent" type.
 - 8. System according to claim 7, wherein the encoder comprises at least a map and a multiplexer.
- 9. System according to claim 8, characterized in that the map generates the sequences to be transmitted on the basis of equation (1) or of relations equivalent thereto.

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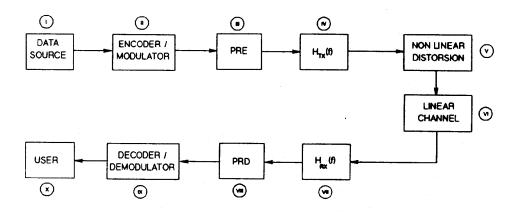


Fig. 1

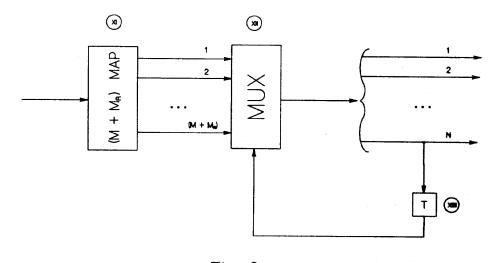


Fig. 2



EUROPEAN SEARCH REPORT

Application Number EP 93 11 1722

Category	Citation of document with indic of relevant passa		Relevant to claim	CLASSIFICATION OF THE APPLICATION (Int.Cl.5)
X	EP-A-0 383 632 (CODEX * abstract; figures 1 * page 2, line 45 - p) -8,10-13 *	1-4,6-9	H04L27/34
X	IEEE JOURNAL ON SELECT COMMUNICATION vol. 7, no. 6 , Augus pages 941 - 958 FORNEY 'Multidimensio part II: Voronoi con * abstract; figures 5 * page 950, left coluright column, paragra	t 1989 , NEW YORK US nal constellations - stellations' -7 * mn, paragraph 4 -	1-4,6-9	
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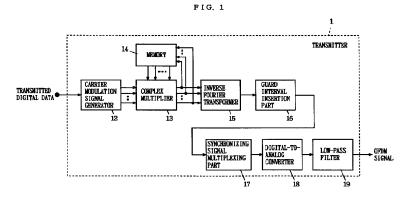
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(54) OFDM system with additional protection against multipath effects

(57) A complex multiplier complex-multiplies a carrier modulation signal group for deciding the phases and amplitudes of a plurality of carriers which are orthogonal to each other on the frequency axis by a complex signal group having a predetermined specific pattern which varies in phase at random. An inverse Fourier transformer performs inverse Fourier transformation on an output of the complex multiplier, for transforming a digital signal which is multiplexed on the frequency axis to an OFDM signal on the time axis. A guard interval insertion part adds front and rear guard intervals to front and rear parts of each symbol of the OFDM signal respectively. The

front and rear guard intervals include data which are identical to those of rear and front end parts of the corresponding symbol respectively. Arithmetic processing which is reverse to that on a transmission side is performed on a receiving side, whereby distortion of received data is removed. Thus, the OFDM signal can be transmitted with no waveform distortion on a data component of each symbol on the frequency axis after Fourier transformation even if a reflected wave is superposed on a direct wave due to a multipath.



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BACKGROUND OF THE INVENTION

Field of the Invention

The present invention relates to an orthogonal frequency division multiplexing (hereinafter referred to as OFDM) transmission method, and more specifically, it relates to a method of transmitting data between a transmission side and a receiving side through a wire or wireless transmission path with an orthogonal frequency division multiplex signal including symbols of prescribed lengths and guard intervals of prescribed lengths which are arranged between the symbols.

1

Description of the Background Art

As well known in the art, an OFDM transmission system is adapted to divide coded data and sort the same into at least hundreds of carriers, for multiplexing and transmitting the data. In relation to digital sound broadcasting for movable terminals or terrestrial digital television broadcasting, communication through an OFDM signal is recently watched with interest. The OFDM signal can transmit a large quantity of data at a high speed while its characteristics are hardly deteriorated by reflected waves even if no waveform equalizer is provided. Further, this signal hardly causes a crossfire to another service since its signal waveform is close to that of a random noise.

A transmission system employing such an OFDM signal is disclosed in "Suitable for Mobile Receiving of OFDM Digital Broadcasting Employing at least Hundreds of Carriers" by Hajime Fukuchi of the Communications Research Laboratory, the Ministry of Posts and Telecommunications of Japan, "Data Compression and Digital Modulation", Nikkei Electronics Books, issued on October 1, 1993, pp. 207 to 222.

Fig. 13 is a block circuit diagram showing the structure of a conventional transmitter 5 for an OFDM signal which is disclosed in the aforementioned literature, and Fig. 14 illustrates the structure of an OFDM signal which is transmitted from the transmitter 5 shown in Fig. 13. Referring to Fig. 13, the transmitter 5 comprises a serial-to-parallel converter 52, an inverse Fourier transformer 53, a parallel-to-serial converter 54, a digital-to-analog converter 55, and a low-pass filter 56. Referring to Fig. 14, (a), (b) and (c) show direct, reflected and composite waves of the OFDM signal respectively, and (d) shows a time window W.

The serial-to-parallel converter 52 of the transmitter 5 is supplied with an input symbol train. The input symbol train is formed by digitally modulated transmission data, and each transmission symbol includes a plurality of data values. The digital modulation is performed by QPSK (quadriphase phase shift keying) modulation or 16 QAM (quadrature amplitude modulation). The serial-to-parallel converter 52 serial-to-parallel converts the input sym-

bol train in every symbol, to obtain a plurality of symbol trains of a lower speed. The degree of parallelism is identical to the number (such as 512, for example, in the range of tens to thousands) of a plurality of carriers, which are orthogonal in phase to each other, employed in the inverse Fourier transformer 53. Due to this operation, the serial-to-parallel converter 52 outputs a group of carrier modulation signals for deciding the amplitudes and phases of the plurality of carriers which are employed in the inverse Fourier transformer 53.

The inverse Fourier transformer 53 allots the carrier modulation signals to the respective carriers which are lined up on the frequency axis in every symbol so that data for one symbol is transformed to a multiplex signal on the frequency axis, and collectively performs inverse Fourier transformation on the signals, thereby transforming the same to a multiplex signal (parallel digital signal in this stage) on the time axis.

The parallel-to-serial converter 54 parallel-to-serial converts the multiplex signal on the time axis, thereby forming a OFDM signal. The digital-to-analog converter 55 converts the OFDM signal to an analog OFDM baseband signal. The low-pass filter 56 limits the band of the OFDM baseband signal, so that no channel-to-channel interference is caused by aliasing.

Following the aforementioned series of operations, the transmitter 5 outputs the OFDM signal including guard intervals Gm and symbols Sm to the transmission path, as shown in Fig. 14. A demodulator (not shown) carries out signal processing which is reverse to that of the modulator 5 on the OFDM signal received through the transmission path, to reproduce an output symbol train which is identical to the input symbol train.

The so-called multipath is caused on the transmission path. Therefore, the receiver receives direct waves of the OFDM signal transmitted from the transmitter and reflected waves which are time-delayed from the direct waves in superposition. If a reflected wave (see (b) in Fig. 14) by the multipath is superposed on a direct wave (see (a) in Fig. 14) in the symbol Sm, for example, an interference part am with the guard interval Gm of the reflected wave is caused on a front end part of the symbol Sm of a composite wave (see (c) in Fig. 14), while an interference part \(\beta m \) with a symbol Sm-1 of the reflected wave is caused on a front end part of the guard interval Gm. At this time, the interference part 6m which is displaced from the time window W exerts no influence on Fourier transformation of the symbol Sm. However, the interference part am is caused in the time window W while the data component of the guard interval Gm is "0", and hence waveform distortion is disadvantageously caused on the data component of each symbol Sm on the frequency axis after the Fourier transformation.

On the other hand, a time delay is caused in the OFDM signal before the same reaches the receiver from the transmitter, due to delay characteristics of the transmission path, deviation in sampling timing resulting from mismatching between clocks of the digital-to-analog converter on the transmission side and an analog-to-dig-

ital converter on the receiving side, and the like. In the receiver, therefore, the time window W must disadvantageously be adjusted on the time axis.

The carrier modulation signals which are outputted from the serial-to-parallel converter 52 may not be out of phase with each other, but may be completely in phase with each other. For example, the carrier modulation signals are completely in phase with each other when a silent state is transmitted in excess of one symbol period in digital sound broadcasting or when a monochromatic picture is transmitted in excess of one symbol period in terrestrial digital television broadcasting. Also in the case of transmitting a sounding state or a multicolor picture, the carrier modulation signals tend to be completely in phase with each other in a digital modulation system such as the QPSK modulation or the 16 QAM, due to a limited number of signal points which are out of phase with each other.

When the carrier modulation signals are completely in phase with each other as described above and these signals are subjected to inverse Fourier transformation, nodes of the respective carriers match with each other on the time axis and addition/increase parts are concentrated to one portion on the time axis, and hence the OFDM signal has an impulsive signal waveform on the time axis, to cause power concentration. Figs. 15(a) to 15(d) show this situation.

Referring to Fig. 15(a), a group of n carrier modulation signals for modulating n carriers which are orthogonal to each other respectively are completely in phase with each other on a complex plane. Referring to Fig. 15(b), the n carriers which are modulated by the n carrier modulation signals shown in Fig. 15(a) are multiplexed on the time axis. When the carrier modulation signals are thus completely in phase with each other, the OFDM signal becomes an impulsive waveform signal. Referring to Fig. 15(c), on the other hand, a group of n carrier modulation signals for modulating n carriers which are orthogonal to each other respectively are at random in phase on a complex plane. Referring to Fig. 15(d), the n carriers which are modulated by the n carrier modulation signals shown in Fig. 15(c) are multiplexed on the time axis. When the carrier modulation signals are thus completely out of phase from each other, the OFDM signal is enenly diffused on the time axis, and becomes a random waveform signal.

As hereinabove described, the OFDM signal has an impulsive waveform to extremely increase the maximum power when the carrier modulation signals are completely in phase with each other, and hence the OFDM signal is disadvantageously readily influenced by nonlinearity of the transmitter, the receiver, a relay amplifier such as a satellite or a CATV included in the transmission path and the like. In this case, the dynamic ranges of the transmitter, the receiver, the relay amplifier and the like may be increased to exert no influences of nonlinearity on the impulsive OFDM signal, while the cost for the transmitter, the receiver, the relay amplifier and the like is disadvantageously increased in this case.

SUMMARY OF THE INVENTION

Accordingly, an object of the present invention is to provide a method of transmitting an OFDM signal which causes no waveform distortion in a data component of each symbol on the frequency axis after Fourier transformation even if a reflected wave is superposed on a direct wave through a multipath, and a transmitter and a receiver therefor.

Another object of the present invention is to provide a method of transmitting an OFDM signal which can readily adjust a time window on the time axis even if a time delay is caused in the OFDM signal before the same reaches a receiving side from a transmission side, and a transmitter and a receiver therefor.

Still another object of the present invention is to provide a method of transmitting an OFDM signal which can reduce an influence of nonlinearity on the OFDM signal in a low-priced structure, and a transmitter and a receiver therefor.

A first aspect of the present invention is directed to a method of transmitting an orthogonal frequency division multiplex signal in every symbol of a prescribed length from a transmission side to a receiving side through a wire or wireless transmission path, and the method comprises:

a first step of transforming a carrier modulation signal group deciding the phases and amplitudes of a plurality of carriers which are orthogonal to each other on the frequency axis to the orthogonal frequency division multiplex signal on the time axis by performing inverse Fourier transformation in every symbol; and

a second step of adding front and rear guard intervals, including data which are identical to those of rear and front end parts of each symbol of the orthogonal frequency division multiplex signal, to front and rear parts of the symbol respectively and transmitting the same to the receiving side.

According to the first aspect of the present invention, as hereinabove described, the front and rear guard intervals including the data which are identical to those of parts of each symbol are added to the front and rear parts of the symbol in transmission of each symbol of the OFDM signal, whereby all data components in a single symbol interval which are lined up on the time axis can be reproduced on the receiving side even if a time window in the Fourier transformation is slightly displaced from the symbol interval of the received signal. Therefore, it is not necessary to correctly coincide the time window with the symbol interval even if a time delay is caused in the OFDM signal before the same reaches the receiving side from the transmission side, whereby the time window can be readily adjusted on the time axis. Even if a symbol interval of a direct wave is superposed with a guard interval of a reflected wave due to a multipath, further, amplitude/phase distortion of each data component appearing on the frequency axis after the Fourier transformation on the receiving side is homogeneous in every symbol. Therefore, such waveform distortion can be readily removed from data components on the frequency axis of one symbol interval on the receiving side by simple arithmetic processing such as multiplication or addition.

In a preferred embodiment of the aforementioned first aspect, the carrier modulation signal group is complex-multiplied by a reference complex signal group on the frequency axis, so that the complex multiplication result is transformed to an OFDM signal and transmitted to the receiving side. On the receiving side, on the other hand, the OFDM signal which is transmitted from the transmission side is transformed to a receiving carrier modulation signal group so that this receiving carrier modulation signal group is complex-divided by the reference complex signal group on the frequency axis. Even if a time delay is caused in the OFDM signal between the transmission side and the receiving side, therefore, modulated data can be obtained on the receiving side with no influence by the time delay.

As to the reference complex signal group for complex-multiplying the carrier modulation signal group, a result of complex multiplication which is carried out in advance of a constant symbol may be employed with respect to each symbol of the carrier modulation signal group.

Alternatively, the reference complex signal group may be prepared from a complex signal group having a predetermined specific pattern with signals which vary in phase at random. In this case, a complex multiplication result which is obtained in a third step is ordinarily transformed to an OFDM signal, while the reference complex signal group is periodically transformed to an OFDM signal. Thus, the absolute reference phases of the respective signals of the carrier modulation signal group are random values, whereby the OFDM signal obtained by the inverse Fourier transformation can be suppressed from time concentration of power. Thus, it is not necessary to increase the dynamic ranges of the transmitter, the receiver and the transmission path but influences exerted by nonlinearity of the transmitter, the receiver and a relay amplifier on the OFDM signal can be reduced through a low-priced structure.

A second aspect of the present invention is directed to a transmitter for an orthogonal frequency division multiplex signal, which is an apparatus for transmitting the orthogonal frequency division multiplex signal to a receiving side in every symbol of a prescribed length through a wire or wireless transmission path, and the transmitter comprises:

a memory part storing a reference complex signal group;

a complex multiplication part, complex-multiplying a carrier modulation signal group deciding the phases and amplitudes of a plurality of carriers which are orthogonal to each other on the frequency axis, by the reference complex signal group stored in the memory part on the frequency axis, for outputting a transmission carrier modulation signal group;

an inverse Fourier transformation part performing

an inverse Fourier operation on the transmission carrier modulation signal group which is outputted from the complex multiplication part in every symbol thereby transforming the transmission carrier modulation signal group to the orthogonal frequency division multiplex signal on the time axis;

a guard interval addition part adding front and rear guard intervals, including data which are identical to those of rear and front end parts of each symbol of the orthogonal frequency division multiplex signal outputted from the inverse Fourier transformation part, to front and rear parts of the symbol respectively; and

a transmission part transmitting the orthogonal frequency division multiplex signal having the added front and rear guard intervals to the receiving side in every symbol.

In a preferred embodiment of the aforementioned second aspect, the memory part stores a complex multiplication result of the complex multiplication part which is precedent to a constant symbol as the reference complex signal group.

In another preferred embodiment of the aforementioned second aspect, the memory part stores a predetermined complex signal group as the reference complex signal group. On the other hand, the complex multiplication part complex-multiplies the carrier modulation signal group by the reference complex signal group which is stored in the memory part on the frequency axis and outputs the result. Further, the inverse Fourier transformation part ordinarily transforms the complex multiplication result which is outputted from the complex multiplication part to an orthogonal frequency division multiplex signal in every symbol, and periodically transforms the reference complex signal group which is outputted from the memory part to an orthogonal frequency division multiplex signal.

According to the aforementioned second aspect, the memory part may hold an output of a pseudo-noise signal generation part generating a pseudo-noise signal or that of a frequency sweep signal generation part generating a frequency sweep signal as the reference complex signal group.

A third aspect of the present invention is directed to a receiver for an orthogonal frequency division multiplex signal, which is an apparatus for receiving the orthogonal frequency division multiplex signal transmitted from a transmission side in every symbol of a prescribed length through a wire or wireless transmission path, and the receiver comprises:

a Fourier transformation part performing a Fourier transformation operation on the orthogonal frequency division multiplex signal on the time axis in every symbol thereby transforming the orthogonal frequency division multiplex signal to a receiving carrier modulation signal group on the frequency axis;

a memory part storing the receiving carrier modulation signal group which is outputted from the Fourier transformation part in every symbol as a receiving reference complex signal group; and 15

a complex division part complex-dividing the receiving carrier modulation signal group which is outputted from the Fourier transformation part by the receiving reference complex signal group which is stored in the memory part on the frequency axis.

A fourth aspect of the present invention is directed to a method of transmitting an orthogonal frequency division multiplex signal from a transmission side to a receiving side in every symbol of a prescribed length through a wire or wireless transmission path, and the method comprises:

- a first step of forming a carrier modulation signal group for deciding the phases and amplitudes of a plurality of carriers which are orthogonal to each other on the frequency axis in every symbol:
- a second step of generating a complex signal group having a predetermined specific pattern with signals varying in phase at random;
- a third step of complex-multiplying the carrier modulation signal group by the complex signal group on the frequency axis in every symbol, thereby randomizing phases of respective signals of the carrier modulation signal group; and
- a fourth step of ordinarily transforming the carrier modulation signal group having the signals which are randomized in phase in the third step to an orthogonal frequency division multiplex signal on the time axis by inverse Fourier transformation in every symbol, and periodically transforming the complex signal group to an orthogonal frequency division multiplex signal by inverse Fourier transformation, for transmitting the same to the receiving side respectively.

The foregoing and other objects, features, aspects and advantages of the present invention will become more apparent from the following detailed description of the present invention when taken in conjunction with the accompanying drawings.

BRIEF DESCRIPTION OF THE DRAWINGS

Fig. 1 is a block diagram showing the structure of a transmitter 1 according to a first embodiment of the present invention:

Fig. 2 is a block diagram showing the structure of a receiver 2 according to the first embodiment of the presentinvention:

Fig. 3 illustrates the structure of an OFDM signal which is transmitted from the transmitter 1 shown in Fig. 1:

Fig. 4 illustrates operations of a memory 14 and a $\,^{50}$ complex multiplier 13 shown in Fig. 1;

Fig. 5 illustrates operations of an envelope wave detector 23 and a synchronous reproducer 24 of the receiver 2 with respect to the OFDM signal which is outputted from the transmitter 1 shown in Fig. 1; Fig. 6 illustrates operations of a memory 26 and a complex divider 27 shown in Fig. 2;

Fig. 7 illustrates comparative results of a simulation for comparing a conventional system and the system

according to the first embodiment with each other in relation to influences exerted by delayed waves by multipaths;

Fig. 8 illustrates results of a simulation for comparing the conventional system and the system according to the first embodiment with each other in relation to influences exerted by time delays through transmission paths etc.;

Fig. 9 is a block diagram showing the structure of a transmitter according to a second embodiment of the present invention;

Fig. 10 illustrates the situation of a complex multiplication of a carrier modulation signal group by a complex signal group in a complex multiplier 13 shown in Fig. 9;

Fig. 11 illustrates operations of a memory 14 and the complex multiplier 13 shown in Fig. 9;

Fig. 12 is a signal structural diagram showing the structure of an OFDM signal which is transmitted from the transmitter shown in Fig. 9;

Fig. 13 is a block diagram showing the structure of a conventional transmitter 5 for an OFDM signal;

Fig. 14 illustrates the structure of the OFDM signal which is transmitted from the transmitter 5 shown in Fig. 13; and

Figs. 15(a) to 15(d) are signal waveform diagrams showing the relations between phase states of carrier modulation signal groups allotted to carriers which are orthogonal to each other and OFDM signals respectively.

DESCRIPTION OF THE PREFERRED EMBODI-MENTS

With reference to the drawings, description is now made on embodiments of a method of transmitting an OFDM signal, and a transmitter and a receiver employed therefor according to the present invention.

Fig. 1 is a block diagram showing a transmitter 1 according to a first embodiment of the present invention, Fig. 2 is a block diagram showing the structure of a receiver 2 according to the first embodiment of the present invention, and Fig. 3 illustrates an exemplary structure of an OFDM signal which is employed in the present invention. Referring to Fig. 3, (a) and (b) show direct and reflected waves of the OFDM signal respectively and (c) and (d) show direct and reflected waves of the OFDM signal causing time delays respectively, while (e) shows a time window W.

The transmitter 1 shown in Fig. 1 and the receiver 2 shown in Fig. 2 are connected with each other through a transmission path (not shown) such as a coaxial cable or an optical fiber cable. The transmitter 1 and the receiver 2 are employed in a digital CATV system, for example. The transmitter 1 is adapted to transmit picture data for multiple channels of a television, for example, to the receiver 2 through an OFDM signal.

Referring to Fig. 1, the transmitter 1 comprises a carrier modulation signal generator 12, a complex multiplier

13, a memory 14, an inverse Fourier transformer 15, a guard interval insertion part 16, a synchronizing signal multiplexing part 17, a digital-to-analog converter 18, and a low-pass filter 19.

The carrier modulation signal generator 12 of the 5 transmitter 1 receives transmitted digital data (bit stream signal) to be transmitted to the receiver 2. The carrier modulation signal generator 12 digital-modulates the inputted transmitted digital data and serial-to-parallel converts the same in every symbol interval, thereby converting the data to a carrier modulation signal group including n (512, for example, in the range of tens to thousands) carrier modulation signals for modulating n carriers which are orthogonal to each other. The digital modulation is performed by QPSK modulation or 16 QAM. The carrier modulation signal group in this stage is similar to that outputted from the serial-to-parallel converter 52 (see Fig. 13) of the conventional transmitter. The carrier modulation signal group which is outputted from the carrier modulation signal generator 12 is supplied to the complex multiplier 13. The memory 14 can store such a carrier modulation signal group D'm outputted from the complex multiplier 13 for one symbol. When a carrier modulation signal group Dm is inputted in the complex multiplier 13, the memory 14 outputs a carrier modulation signal group D'm-1, which is precedent to one symbol, stored therein to the complex multiplier 13 as a prescribed reference complex signal group. The complex multiplier 13 complex-multiplies the inputted transmission signal group Dm by the reference complex signal group D'm-1 which is precedent by one symbol on the frequency axis, thereby forming the following carrier modulation signal group:

$D'm (D'm = Dm \times D'm-1)$

Assuming that Dm[k]real represents the real number part of a k-th (k = 1, 2, ..., n) carrier modulation signal of the carrier modulation signal group, including n carrier modulation signals, which is inputted in the complex multiplier 13 and Dm[k]imag represents the imaginary number part thereof while D'm-1[k]real represents the real number part of the k-th carrier modulation signal which is stored in the memory 14 and D'm-1[k]imag represents the imaginary number part thereof, the complex multiplier 13 carries out multiplication processing as to the real and imaginary number parts of each carrier modulation signal, for outputting:

D'm[k]real = Dm[k]real x D'm-1[k]real

 $D'm[k]imag = Dm[k]imag \times D'm-1[k]imag$

The memory 14 stores the carrier modulation signal D'm (including D'm[k]real and D'm[k]imag) of the real and imaginary numbers outputted from the complex multiplier 13. As shown in Fig. 4, the memory 14 and the complex multiplier 13 repeatedly execute the aforementioned operations.

The inverse Fourier transformer 15 successively allots the respective carrier modulation signals included in the carrier modulation signal group D'm which is outputted from the complex multiplier 13 to the respective carriers which are lined up on the frequency axis in every symbol interval, collectively performs inverse Fourier transformation thereon, and further performs parallel-to-serial conversion, thereby transforming the carrier modulation signal group multiplexed with the respective data components on the frequency axis to an OFDM signal D'mt multiplexed with the respective data components on the time axis.

The guard interval insertion part 16 temporarily stores the digital OFDM signal D'mt which is outputted from the inverse Fourier transformer 15 in its internal buffer in every symbol interval. Then, the guard interval insertion circuit 16 adds front and rear guard intervals Ghm and Gem to front and rear parts of each symbol Sm (see Fig. 3). Time lengths tg1 and tg2 of the front and rear guard intervals Ghm and Gem are prescribed in consideration of time difference between direct and indirect waves due to a multipath caused in the transmission path and time delays resulting from sampling deviation between the digital-to-analog converter 18 of the transmitter 1 and an analog-to-digital converter 22 of the receiver 2. Further, the front and rear guard intervals Ghm and Gem include data D'emt and D'hmt which are identical to those of rear and front end parts Sem and Shm of the corresponding symbol Sm respectively. Thus, the substantial symbol length is extended to tg1 + ts + tg2 . The guard interval insertion part 16 successively outputs the data D'emt, D'm and D'hmt through the front guard interval Ghm, the symbol Sm and the rear quard interval Gem.

The synchronizing signal multiplexing part 17 multiplexes a synchronizing signal on the OFDM signal to which the guard intervals are added on the time axis in every symbol in order to indicate the breakpoint of the symbol, and outputs the signal to the digital-to-analog converter 18. The synchronizing signal is formed by a periodically known nonmodulated carrier, a suppression signal etc. with respect to the OFDM signal, as shown at (a) in Fig. 5, for example.

The digital-to-analog converter 18 converts the OFDM signal of the digital data, to which the guard intervals and the synchronizing signal are added, outputted from the synchronizing signal multiplexing part 17 to an analog OFDM baseband signal. The low-pass filter 19 limits the band of the OFDM baseband signal, so that no channel-to-channel interference is caused by aliasing.

As the result of the aforementioned series of operations, the transmitter 1 outputs the OFDM signal including the guard intervals and the synchronizing signal to the transmission path.

Referring to Fig. 2, the receiver 2 comprises a lowpass filter 21, the analog-to-digital converter 22, an envelope detector 23, a synchronous reproducing part 24, a Fourier transformer 25, a memory 26, a complex divider 27, and a transmission data reproducer 28. The low-pass filter 21 removes unnecessary spectral components of a high-frequency region from the OFDM signal which is received through the transmission path.

In consideration of a time delay Δt caused by the 5 multipath or the delay characteristics of the transmission path, the OFDM signal which is received in the receiver 2 is represented by ZD'mt, where Z represents the signal delay as follows:

$Z = \exp j2\pi fc\Delta t$

The analog-to-digital converter 22 converts data ZD'emt, ZD'mt and ZD'hmt which are included in the front guard interval Ghm, the symbol Sm and the rear guard interval Gem of the analog OFDM signal respectively to those of a digital OFDM signal.

The envelope detector 23 envelope-detects the OFDM signal, thereby outputting an envelope detection signal shown at (b) in Fig. 5 in every symbol. The synchronous reproducing part 24 outputs a reference timing signal shown at (c) in Fig. 5 in every symbol on the basis of the envelope detection signal outputted from the envelope detector 23. This reference timing signal is inputted in the Fourier transformer 25 and the memory 26.

The Fourier transformer 25 observes the OFDM signal which is outputted from the analog-to-digital converter 22 in synchronization with the reference timing signal through the time window W (see (e) in Fig. 3) of the same length as the symbol length ts, thereby extracting only necessary data parts of the respective symbols. The Fourier transformer 25 further performs Fourier transformation operations on the extracted data parts, thereby transforming the OFDM signal on the time axis to a receiving carrier modulation signal group on the frequency axis.

The memory 26 stores the receiving carrier modulation signal group which is outputted from the Fourier transformer 25 for one symbol. When the transmitter 1 transmits data D'm, the memory 26 stores data ZD'm as 40 corresponding data. The data ZD'm is obtained by adding a time delay Z caused by the multipath or the transmission path to the data D'm, as follows:

$$ZD'm = D'm \times expj2\pi fc\Delta t$$

The memory 26 outputs the data ZD'm to the complex divider 27 in synchronization with the reference timing signal. The complex divider 27 establishes synchronization, and then complex-divides data ZD'm+1 of a symbol Sm+1 which is outputted from the Fourier transformer 25 by the data ZD'm held in the memory 26. Namely, the complex divider 27 performs the following operation:

$$ZD'm+1/ZD'm = D'm+1/D'm = Dm+1$$

As shown in Fig. 6, the Fourier transformer 25, the memory 26 and the complex divider 27 repeatedly execute the aforementioned operations.

As hereinabove described, a relative time delay is caused between the direct and reflected waves shown at (a) and (b) in Fig. 3, due to the multipath. Further, specific time delays are caused in the direct and reflected waves, due to the difference in sampling timing between the digital-to-analog converter 18 of the transmitter 1 and the analog-to-digital converter 22 of the receiver 2 (see (c) and (d) in Fig. 3). These time delays are not taken into consideration in the Fourier transformer 25 as to the reference timing signal, and hence positions of the receiving side time window W on the time axis are displaced from the symbol intervals of the received signal, as shown at (e) in Fig. 3.

Even if the time window W is displaced from correct symbol intervals in the Fourier transformer 25 of the receiving side, however, the data observed through the time window W include all data ZD'mt on the time axis which must be originally included in one symbol interval since the front and rear guard intervals Ghm and Gem include the data ZD'emt and ZD'hmt respectively. Therefore, the time delays and superposition of the reflected waves appear as uniform amplitude/phase distortion in every data component on the frequency axis. When the time delays and the characteristics of the reflected waves are uniform, the values of the amplitude/phase distortion in the respective symbol intervals are equal to each other. According to this embodiment, the complex divider 27 complex-divides the data ZD'm+1 of the symbol Sm+1 which is outputted from the Fourier transformer 25 by the data ZD'm held in the memory 26, thereby canceling the data delay Z and obtaining the original carrier modulation signal group Dm+1 with no delay. Namely, the amplitude/phase distortion is canceled by the following operation of the complex divider 27:

ZD'm+1/ZD'm = D'm+1/D'm = Dm+1

Thus, data Dm having no phase/amplitude distortion can be obtained as to each symbol.

According to this embodiment, as hereinabove described, the guard intervals including the data which are identical to those of the front and rear end parts of each symbol are added to the front and rear parts of the symbol respectively for transmitting the data, whereby all data components in one symbol interval which are lined up on the time axis can be reproduced on the receiving side as to both of the direct and reflected waves in the time window W. Therefore, the respective data components appearing on the frequency axis after the Fourier transformation are uniform in amplitude/phase distortion even if the reflected waves are superposed on the direct waves by the multipath to result in superposition of the symbol intervals of the direct waves and the guard intervals of the reflected waves. Therefore, waveform distortion can be readily removed from the receiving carrier modulation signal group on the frequency axis of one symbol interval by executing proper operations (multiplication and division) on the transmission and receiving sides.

According to this embodiment, further, demodulated data can be obtained with no time delay even if a time delay is caused in the OFDM signal between the transmission and receiving sides, by complex-multiplying and complex-dividing the receiving carrier modulation signal group by the prescribed reference complex signal group on the frequency axis. Consequently, it is not necessary to correctly coincide the time window with the symbol interval.

The transmission data reproducer 28 demaps signal points of the receiving carrier modulation signal group Dm which is outputted from the complex divider 27 on a complex plane and decides the signal points, thereby obtaining a receiving digital signal group which is identical in value to the transmission digital signal group of the transmitter 1. As hereinabove described, phase distortion and amplitude distortion are removed from the receiving carrier modulation signal group Dm. Therefore, the transmission data reproducer 28 can correctly and readily determine the original data from the mapping positions on the complex plane.

The inventors have made simulations of comparing the system according to this embodiment with the conventional system with respect to influences exerted by waves delayed by multipaths and those exerted by time axis delays respectively through a calculator. Each simulation was executed on such conditions that the carrier number was 512, only data of a 256-th carrier had an amplitude "1" and a phase "0", and all data of the remaining carriers were "0".

Fig. 7 illustrates the results of the simulation for comparing the system according to this embodiment with the conventional system as to the influences exerted by waves delayed by multipaths. Referring to Fig. 7, (a), (b), (c) and (d) show data distortion states in the case of transforming direct, indirect, composite and composite waves in the conventional system to signals on the frequency axis by Fourier operations respectively. On the other hand, (e), (f), (g) and (h) show data distortion states in the case of converting direct, indirect, composite and composite waves in the system according to this embodiment to signals on the frequency axis by Fourier operations respectively.

In the conventional system, no data is inserted in any guard interval (see α 1 at (b) in Fig. 7), and hence an interference part α2 appears in a time window W of the composite wave (see (c) in Fig. 7). When the composite wave is transformed to a signal on the frequency axis by a Fourier operation in the time window W, therefore, the spectrum of the data of the 256-th carrier is spread and the data of the remaining carriers, which must have originally been "0", are distorted. Thus, the transmission data reproducer 28 readily causes an erroneous determination. Further, the transmission data reproducer 28 also readily causes erroneous determinations as to other carriers. In the system according to this embodiment, on the other hand, data are inserted in the guard intervals and hence no influences are exerted on the data of the remaining carriers.

Fig. 8 illustrates the results of the simulation for comparing the system according to this embodiment with the conventional system as to the influences exerted by time delays caused by transmission paths etc. Referring to Fig. 8, (a) shows a spectrum obtained under such conditions that only the data of the 256-th carrier had an amplitude "1" and a phase "0", and (b) shows a signal waveform in the case of transforming the data at (a) to a signal on the time axis by an inverse Fourier operation. Referring to Fig. 8, further, (c) and (d) show data distortion states in the case of transforming composite and composite waves causing time delays in the conventional system to signals on the frequency axis by Fourier operations respectively. On the other hand, (e) and (f) show data distortion states in the case of transforming composite and composite waves causing time delays in the system according to this embodiment to signals on the frequency axis by Fourier operations respectively.

In the conventional system, no data is inserted in any guard interval (see $\alpha 1$ at (c) in Fig. 8), and hence an interference part $\alpha 2$ appears in a time window W of the composite wave, similarly to the case shown at (c) in Fig. 7. When the composite wave is transformed to a signal on the frequency axis by a Fourier operation in the time window W, therefore, the spectrum of the data of the 256-th carrier is spread and the data of the remaining carriers, which must have originally been "0", are distorted, as shown at (d) in Fig. 8. Thus, the transmission data reproducer 28 readily causes erroneous determinations also as to other carriers. In the system according to this embodiment, on the other hand, data are inserted in the guard intervals and hence no influences are exerted on the data of the remaining carriers.

Fig. 9 is a block diagram showing the structure of a transmitter 3 according to a second embodiment of the present invention. In the transmitter 3 shown in Fig. 9, portions corresponding to those of the transmitter 1 shown in Fig. 1 are denoted by the same reference numerals, to omit redundant description. As to the embodiment shown in Fig. 9, it is to be noted that a memory 14 holds an output of a specific pattern generator 31, i.e., a complex signal group D0 having a predetermined specific pattern with signals which mutually vary in phase at random. Such a complex signal group D0 can be formed by a pseudo-noise signal generator comprising a PN series pseudo-random signal generator for generating a pseudo-random signal which is at a level between zero and 1 and a multiplier for multiplying the pseudorandom signal by 2π for generating a unit vector signal in a phase having a random value in the range of zero to 2π and an amplitude of 1, for example. Alternatively, the complex signal group D0 can be formed by a frequency sweep signal generator for generating a known frequency sweep signal in a phase having a random value in the range of zero to 2π .

A complex multiplier 13 complex-multiplies data Dm of each symbol interval by data D0 on the frequency axis every time data Dm is inputted for forming data D'm (D'm = Dm x D0), thereby randomizing mutual phases

of respective carrier modulation signals included in a carrier modulation signal group to specific patterns.

Fig. 10 illustrates a complex multiplication operation in the complex multiplier 13. In particular, (a) in Fig. 10 shows arrangement of signal points which can be taken 5 by the carrier modulation signals when 16 QAM is employed as a modulation system, (b) shows a unit vector i whose phase varies at random, and (c) shows a carrier modulation signal whose phase is randomized to a specific pattern.

Referring to (a) in Fig. 10, it is assumed that a carrier modulation signal included in a carrier modulation signal group which is allotted to one carrier is arranged at a signal point A on a complex plane. The signal A has a real number part of 3 and an imaginary number part of 1. It is also assumed that the unit vector i has a phase angle of $3\pi/4$ at this time. A carrier modulation signal A' shown at (c) in Fig. 10 is obtained as the result of a complex multiplication. The carrier modulation signal A' has a real number part of - 2.8 and an imaginary number part of 1.4, and takes a signal point which is not present in the arrangement of the 16 QAM. Thus, the phase of the unit vector i varies at random, and hence the complex multiplier 13 outputs a carrier modulation signal group having signals whose phases are mutually randomized to an inverse Fourier transformer 15 even if respective carrier modulation signals included in a carrier modulation signal group which is outputted from a carrier modulation signal generator 12 are in phase with each other.

The complex multiplier 13 repeats such an operation for a prescribed period. Further, the complex multiplier 13 periodically outputs only the data D0. Fig. 11 shows a series of such operations. Assuming that S0 represents a symbol in which the data D0 is inserted, the transmitter 3 periodically outputs the data D0 of the symbol S0 while outputting data Dm of a symbol Sm in other case, as shown in Fig. 12. The inverse Fourier transformer 15 allots the carrier modulation signal group D'm to respective carriers which are lined up on the frequency axis in every symbol, and collectively performs inverse Fourier transformation and parallel-to-serial conversion thereon, thereby converting the same to a digital OFDM signal. Consequently, absolute reference phases of the carrier modulation signal group are at random values in the range of zero to 2π whereby the OFDM signal outputted from the inverse Fourier transformer 15 can be suppressed from power concentration. Thus, it is not necessary to increase the dynamic ranges of the transmitter 3 and a receiver but influences exerted on the OFDM signal by nonlinearity of the transmitter, the receiver, a relay amplifier etc. can be reduced through a low-priced structure. The remaining circuit blocks in the transmitter 3, i.e., those from a guard interval insertion part 16 to a low-pass filter 19, operate similarly to those in the transmitter 1.

The guard interval insertion part 16 inserts a data component D0 which is identical to that of a rear end part of the symbol S0 in a corresponding front guard interval, while inserting a data component which is identical to

that of a front end part of the symbol S0 in a corresponding rear guard interval, similarly to the case of the symbol Sm

When the transmitter 3 shown in Fig. 9 is employed, a receiver of the same structure as the receiver 2 shown in Fig. 2 can basically be employed. In this case, however, a memory 26 of the receiver stores receiving data ZD0 of a reference complex signal group D0 which is stored in the memory 14 of the transmitter 3.

Also in the embodiment shown in Fig. 9, an effect similar to that of the aforementioned first embodiment can be attained. Namely, amplitude/phase distortion of a receiving carrier modulation signal group appearing on the frequency axis after Fourier transformation is entirely uniform even if reflected waves are superposed on direct waves by a multipath and symbol intervals of the direct waves are superposed with guard intervals of the reflected waves, and can be removed by simple operations (multiplication and division). Further, demodulated data can be obtained with no influence by a time delay even if such a time delay is caused in the OFDM signal between the transmission and receiving sides, whereby a time window can be readily adjusted on the time axis.

While the data are transmitted through wire transmission paths in the aforementioned embodiments, the present invention is not restricted to this but data may alternatively be transmitted through a wireless transmission path. While television picture data for multichannels are carried on the respective carriers in the aforementioned embodiments, picture data for one channel may alternatively be time-shared and sequenced in a parallel manner, to be allotted to respective carriers. Further, voice data, text data or the like may be carried on the respective carriers, in place of the picture data. In addition, the present invention may alternatively be carried out in another system such as LAN or WAN, in place of the CATV.

While the reference complex signal group outputted from the memory 14 is periodically inputted in the inverse Fourier transformer 15 through the complex multiplier 13 in the transmitter 3 shown in Fig. 9, the reference complex signal group may alternatively be directly inputted in the inverse Fourier transformer 15.

While the transmitter 3 shown in Fig. 9 employs the complex signal group D0 including signals having a predetermined specific pattern and phases which mutually vary at random as the reference complex signal group to be included in the carrier modulation signal group, further, the reference complex signal group to be included in the carrier modulation signal group may alternatively be formed by a complex signal group including signals having a predetermined specific pattern which are in phase with each other under a situation causing no power concentration in the OFDM signal. Also in this case, amplitude/phase distortion can be removed by simple operations (multiplication and division), similarly to the first embodiment.

Although the present invention has been described and illustrated in detail, it is clearly understood that the

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same is by way of illustration and example only and is not to be taken by way of limitation, the spirit and scope of the present invention being limited only by the terms of the appended claims.

Claims

1. A method of transmitting an orthogonal frequency division multiplex signal from a transmission side to a receiving side in every symbol of a prescribed length through a wire or wireless transmission path, said method comprising:

a first step of transforming a carrier modulation signal group deciding the phases and amplitudes of a plurality of carriers being orthogonal to each other on the frequency axis to said orthogonal frequency division multiplex signal on the time axis by performing inverse Fourier transformation in every symbol; and

a second step of adding front and rear guard 20 intervals, including data being identical to those of rear and front end parts of each symbol of said orthogonal frequency division multiplex signal, to front and rear parts of said symbol respectively and transmitting the same to said receiving side.

2. The method of transmitting an orthogonal frequency division multiplex signal in accordance with claim 1, further comprising a third step of complex-multiplying said carrier modulation signal group by a reference complex signal group on the frequency axis,

said first step being adapted to transform a complex multiplication result being obtained in said third step to said orthogonal frequency division multiplex signal.

- 3. The method of transmitting an orthogonal frequency division multiplex signal in accordance with claim 2, wherein said third step is adapted to complex-multiply each said carrier modulation group by a result of complex multiplication, being carried out in advance of a constant symbol, serving as said reference complex signal group with respect to each said symbol of said carrier modulation signal group.
- 4. The method of transmitting an orthogonal frequency division multiplex signal in accordance with claim 2, further comprising a fourth step of generating a complex signal group having a predetermined specific pattern with signals varying in phase at random,

said third step employs said complex signal group being obtained in said fourth step as said reference complex signal group with respect to each said symbol of said carrier modulation signal group,

said first step ordinarily transforms said complex multiplication result being obtained in said third step to said orthogonal frequency division multiplex signal, while periodically transforming said reference complex signal group to said orthogonal frequency division multiplex signal.

The method of transmitting an orthogonal frequency division multiplex signal in accordance with claim 2, further comprising:

a fifth step of transforming said orthogonal frequency division multiplex signal being transmitted from said transmission side to a receiving carrier modulation signal group corresponding to said carrier modulation signal group in every symbol of said prescribed length, and

a sixth step of complex-dividing said receiving signal group being obtained in said fifth step by a prescribed reference complex signal group on the frequency axis.

6. A transmitter for an orthogonal frequency division multiplex signal, being an apparatus for transmitting said orthogonal frequency division multiplex signal to a receiving side in every symbol of a prescribed length through a wire or wireless transmission path, said transmitter comprising:

memory means storing a reference complex signal group;

complex multiplication means complex-multiplying a carrier modulation signal group deciding the phases and amplitudes of a plurality of carriers being orthogonal to each other on the frequency axis by said reference complex signal group being stored in said memory means on the frequency axis, for outputting a transmission carrier modulation signal group;

inverse Fourier transformation means performing an inverse Fourier operation on said transmission carrier modulation signal group being outputted from said complex multiplication means in every symbol thereby transforming said transmission carrier modulation signal group to said orthogonal frequency division multiplex signal on the time

guard interval addition means adding front and rear guard intervals, including data being identical to those of rear and front end parts of each symbol of said orthogonal frequency division multiplex signal outputted from said inverse Fourier transformation means, to front and rear parts of said symbol respectively: and

transmission means transmitting said orthogonal frequency division multiplex signal having added said front and rear guard intervals to said receiving side in every symbol.

The transmitter for an orthogonal frequency division multiplex signal in accordance with claim 6, wherein said memory means stores a complex multiplication result of said complex multiplication means in advance of a constant symbol as said reference complex signal group.

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 The transmitter for an orthogonal frequency division multiplex signal in accordance with claim 6, wherein said memory means stores a predetermined complex signal group as said reference complex signal group,

said complex multiplication means complexmultiplies said carrier modulation signal group by said reference complex signal group being stored in said memory means on the frequency axis for outputting the same, and

said inverse Fourier transformation means ordinarily transforms a complex multiplication result being outputted from said complex multiplication means to said orthogonal frequency division multiplex signal in every symbol, while periodically transforming said reference complex signal group being outputted from said memory means to said orthogonal frequency division multiplex signal.

- The transmitter for an orthogonal frequency division multiplex signal in accordance with claim 8, wherein said memory means holds an output of pseudonoise signal generation means generating a pseudo-noise signal as said reference complex signal group.
- 10. The transmitter for an orthogonal frequency division multiplex signal in accordance with claim 8, wherein said memory means holds an output of frequency sweep signal generation means generating a frequency sweep signal as said reference complex signal group.
- 11. A receiver for an orthogonal frequency division multiplex signal, being an apparatus for receiving said orthogonal frequency division multiplex signal being transmitted from a transmission side in every symbol of a prescribed length through a wire or wireless transmission path, said receiver comprising:

Fourier transformation means performing a 40 Fourier transformation operation on said orthogonal frequency division multiplex signal on the time axis in every symbol, thereby transforming said orthogonal frequency division multiplex signal to a receiving carrier modulation signal group on the frequency 45 axis:

memory means storing said receiving carrier modulation signal group being outputted from said Fourier transformation means every constant symbol as a receiving reference complex signal group; 50 and

complex division means complex-dividing said receiving carrier modulation signal group being outputted from said Fourier transformation means by said receiving reference complex signal group being stored in said memory means on the frequency axis.

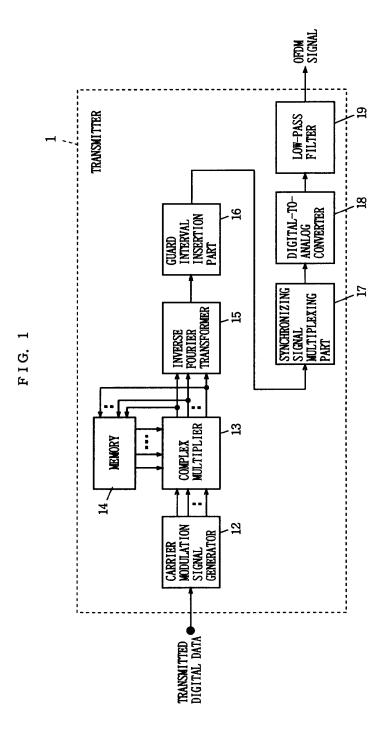
12. A method of transmitting an orthogonal frequency division multiplex signal from a transmission side to a receiving side in every symbol of a prescribed length through a wire or wireless transmission path, said method comprising:

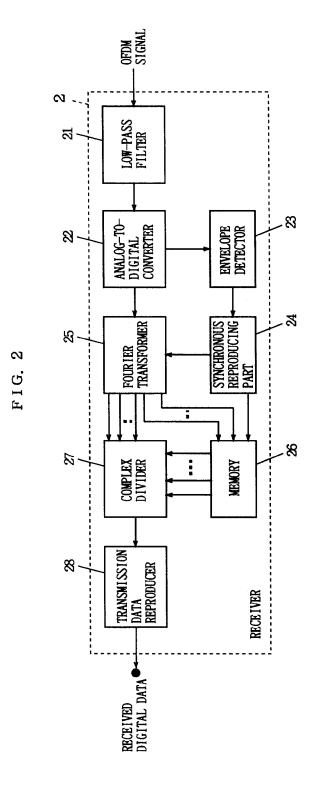
a first step of forming a carrier modulation signal group for deciding the phases and amplitudes of a plurality of carriers being orthogonal to each other in every symbol on the frequency axis;

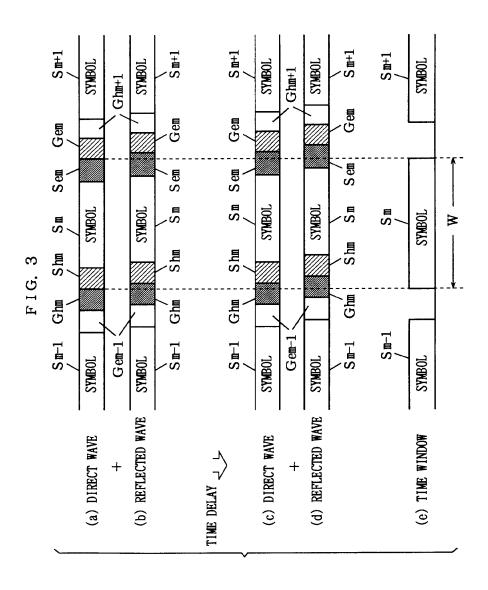
a second step of generating a complex signal group having a predetermined specific pattern with signals varying in phase at random;

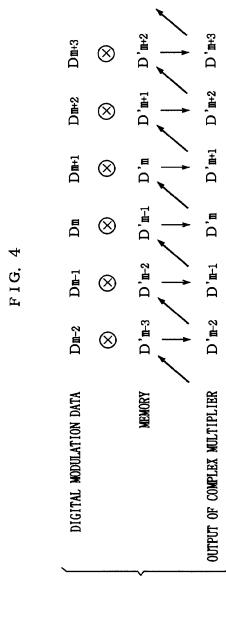
a third step of complex-multiplying said carrier modulation signal group by said complex signal group in every symbol on the frequency axis, thereby randomizing the phases of respective signals of said carrier modulation signal group; and

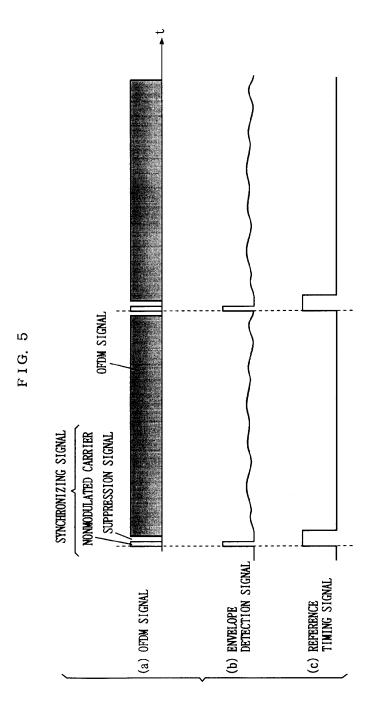
a fourth step of ordinarily transforming said carrier modulation signal group having said signals being randomized in phase in said third step to said orthogonal frequency division multiplex signal on the time axis by performing inverse Fourier transformation in every symbol while periodically transforming said complex signal group to said orthogonal frequency division multiplex signal by inverse Fourier transformation, for transmitting the same to said receiving side respectively.

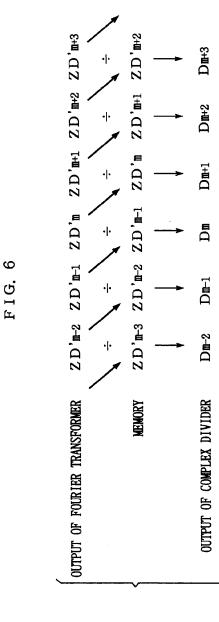












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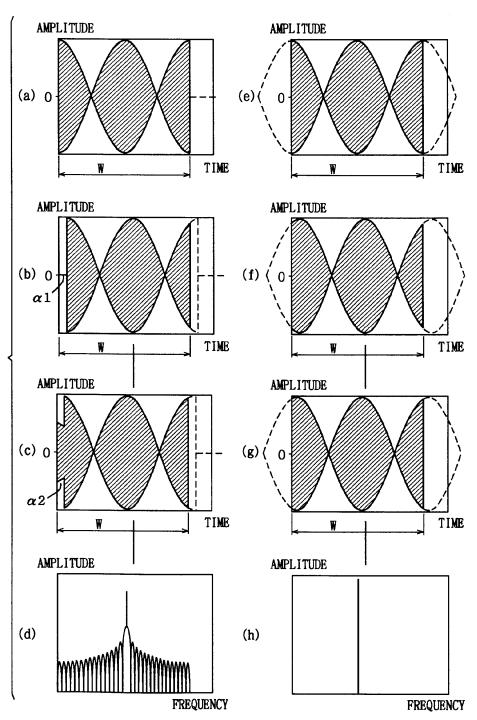
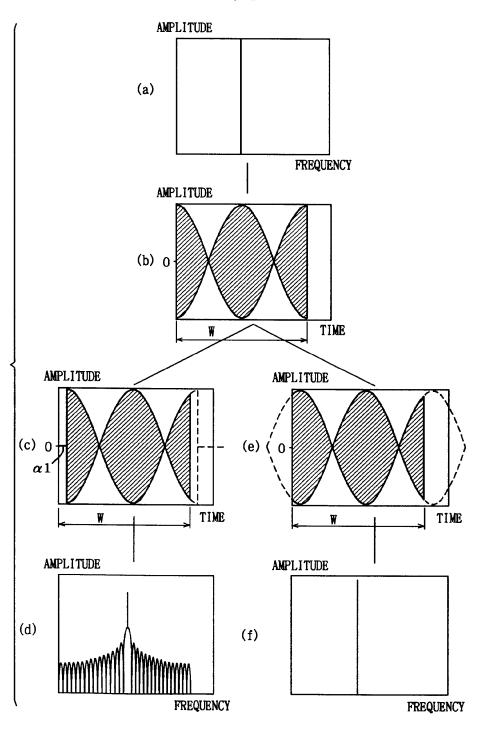
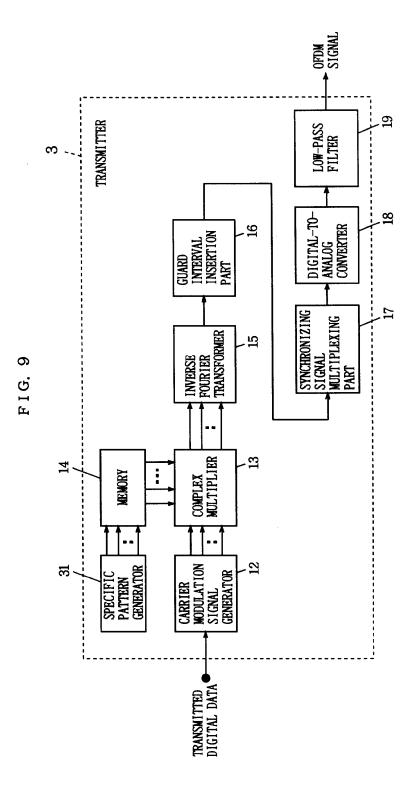
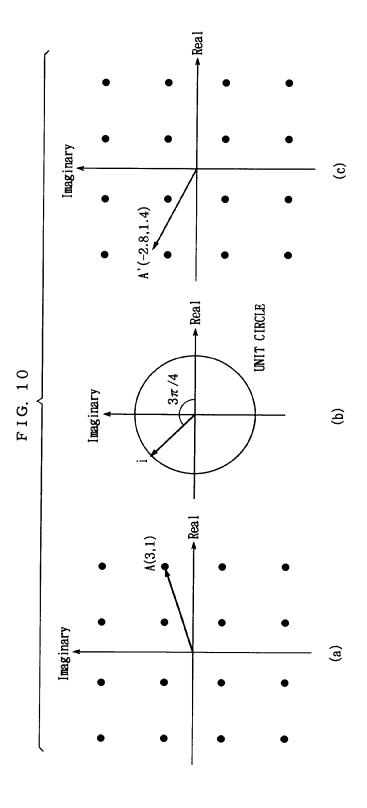
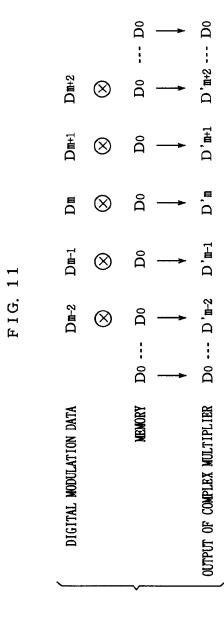


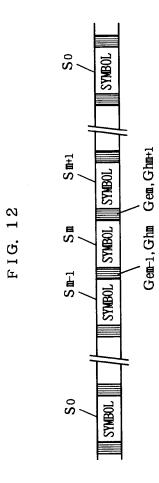
FIG. 8

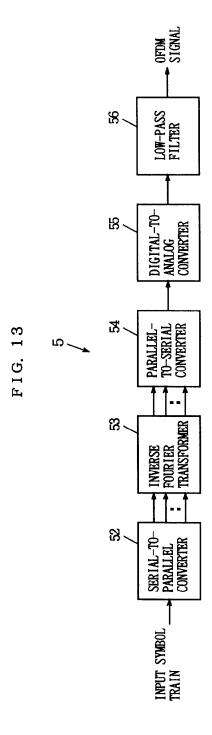


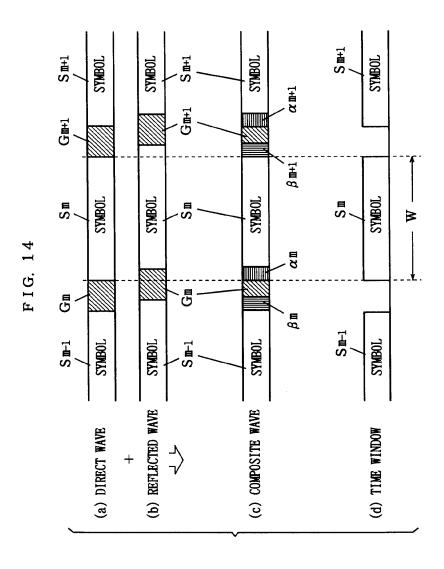


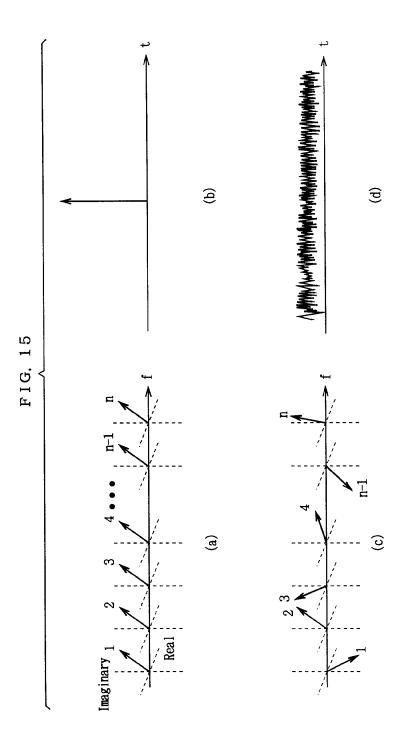












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- (56) Documents Cited

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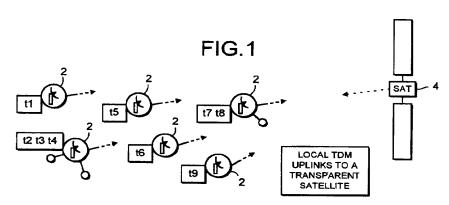
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WO 94/08405 A1

(58) Field of Search

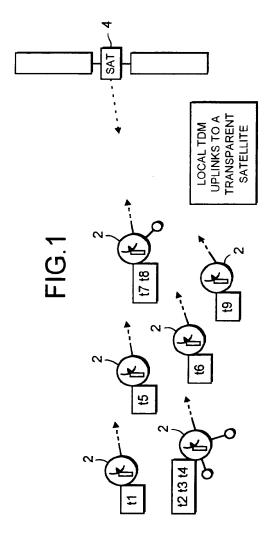
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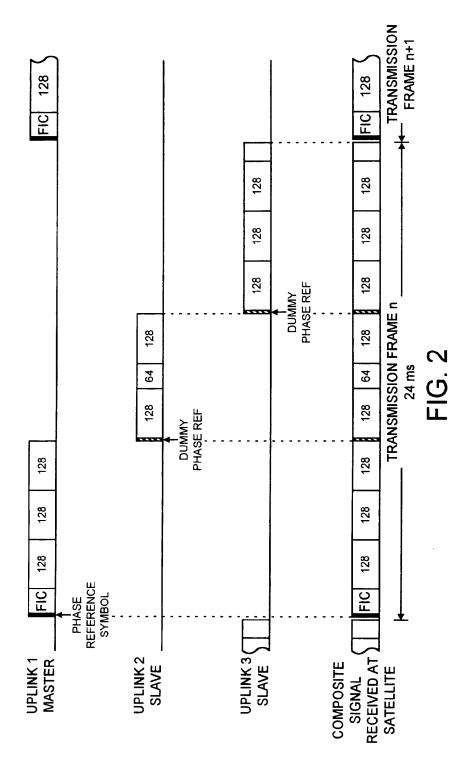
- (54) Abstract Title Digital broadcast systems
- (57) A multi-carrier signal has a regular frame structure and symbol rate and is comprised of contributions from a plurality of different transmitters (2). The contributions from each transmitter are transmitted to a central transmitter (4) in pre-assigned time slots. The received contributions are then re-transmitted as a single signal over a pretermined area of coverage with a dummy symbol inserted at the start of each contribution in the frame for use as a phase reference for demodulating succeeding symbols in that contribution.

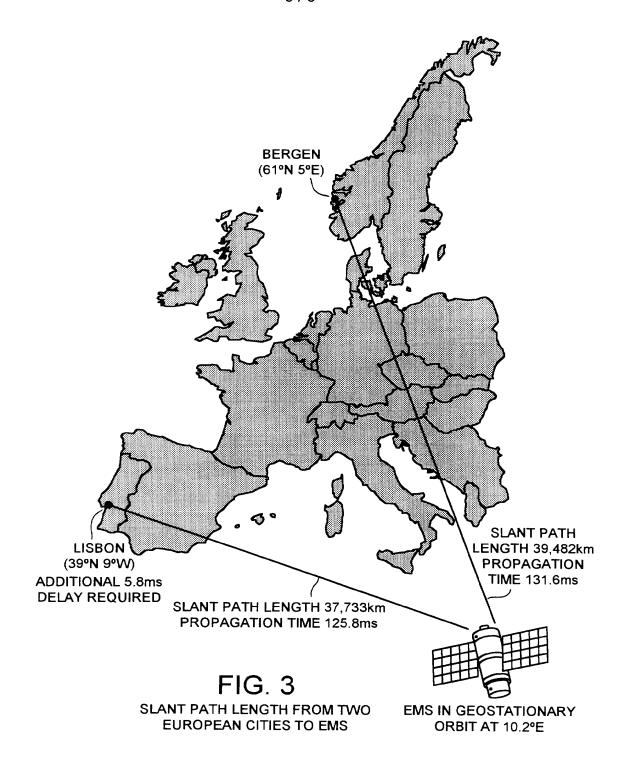


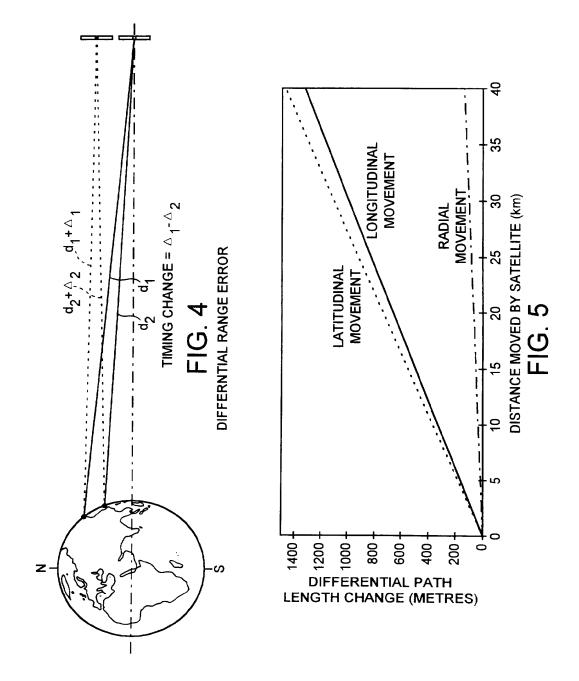
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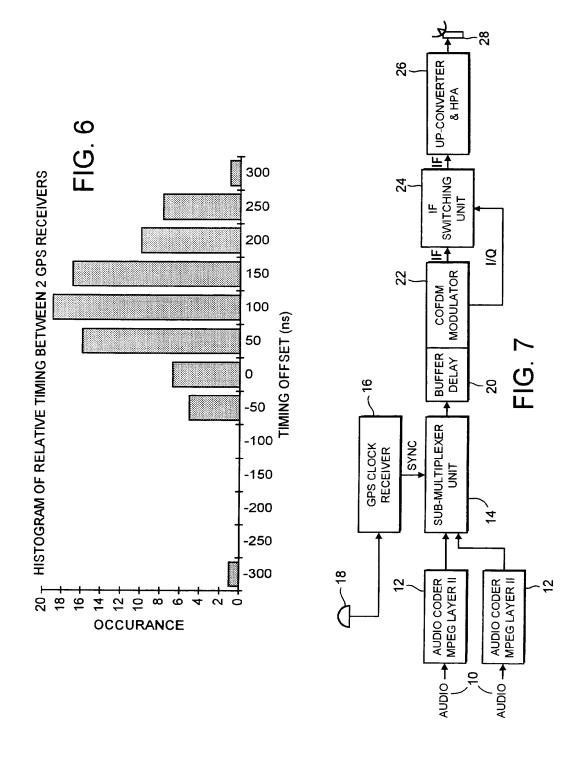
At least one drawing originally filed was informal and the print reproduced here is taken from a later filed formal copy.











DIGITAL BROADCAST SYSTEMS

This invention relates to digital broadcast systems such as digital audio broadcast (DAB) and in particular to a system which enables digital broadcasts from two or more different broadcasters to be combined in a single broad band transmission.

The Eureka-147 DAB system which has been proposed as a practical implementation of digital audio broadcasting operates by using a coded orthogonal frequency division multiplexed (COFDM) system. In this, a large number of carriers are spread over a broad frequency band to carry digital data. Each carrier is modulated with the data so as to carry two bits of data by using quadrature phase shift keying (QPSK). Groups of these carriers are then transformed to the time domain by a Fast Fourier transform to produce what is known as a DAB symbol. A plurality of these symbols are assembled and between them are able to carry data from a large number of channels. The symbols are transmitted together in a DAB frame or multiplex comprising a series of symbols and which commences and terminates with a null symbol for synchronisation. A typical transmission bandwidth of 1.53 MHz can typically accommodate 5 or 6 channels.

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Satellite delivery of digital broadcasts is seen as an attractive option for international broadcasters because it provides coverage of large areas at relatively low costs.

Because a Eureka-147 DAB ensemble carries not just one but several audio channels or other services, several co-operating broadcasters would need to share an ensemble between them. This can be relatively easily accomplished at a national level where there are both national and

local broadcasters by leaving free symbols in national broadcasts into which local broadcasters can insert data. These would typically be combined at a single site and can then be transmitted over the relevant area.

Where it is desired to combine broadcasting over a number of different countries, e.g. UK, France, Germany and Austria, and transmit them as a single DAB ensemble over all of those countries, the combination at a single terrestrial uplink site becomes impractical because of the cost of terrestrial data lines.

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One solution is to use a dedicated satellite with an on-board processor to which all the broadcasters transmit. The satellite then combines all the data and produces a single DAB ensemble from this. The problem with this, of course, is that a dedicated satellite has to be launched in order for the system to become operational.

A preferred embodiment of the present invention provides a system in which a number of different broadcasters each transmit a section of a COFDM ensemble from an uplink site to a conventional satellite in time division multiplex slots. The satellite then amplifies and frequency shifts the received signals before transmitting them over its area of coverage. No on-board processing is involved other than would be used for a conventional radio transmission. Each uplink site would need to be adequately synchronised to the others so that the resulting composite COFDM signal appeared seamless when transmitted by the satellite. This can easily be achieved using the global position system (GPS).

The invention is defined with more precision in the appended claims to which reference should now be made.

The invention will now be described in detail by way of example with reference to the accompanying drawings in which:

Figure 1 shows schematically a number of uplink sites transmitting signals to a conventional satellite in time division multiplex (TDM) slots for retransmission over the satellite's area of coverage in an embodiment of the invention;

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Figure 2 shows schematically the combination of DAB symbols from three different uplink sites in accordance with an embodiment of the invention;

Figure 3 shows schematically the different slant path lengths from two European cities to a satellite in geostationary orbit;

Figure 4 shows the differential range for a satellite in geostationary orbit;

Figure 5 is a graph showing the relationship between differential path length and the distance moved by the satellite;

Figure 6 is a histogram showing relative timings between 2 GPS receivers; and

Figure 7 is a block diagram of an uplink site of the type shown in Figure 1.

It is envisaged that in an embodiment of this invention a multiplexed uplink system would involve two or more uplink sites of the type shown in Figure 1. These can be receiving one or more signals, coding them with COFDM and transmitting them in preassigned time slots to a satellite. Each uplink site provides a portion of the COFDM signal directly to the satellite. The multi carrier nature of the Eureka DAB signal with its frequency and time interleaving means that mapping of a particular bit

pair onto a particular carrier is very complex. Whilst it would be technically possible to identify which carriers are associated with each uplink contribution, it would then be necessary to be able to suppress each carrier on an individual basis. This would be considerably more difficult than switching all of the carriers on and off simultaneously at single symbol boundaries. As will be seen, this is not a severe constraint and greatly simplifies the handover process.

embodying the present invention requires a fairly radical rethink of the requirements of the DAB transmission chain. The TDM system requires complete shutdown of the transmitters RF output at frequent and regular intervals. At present there is no structure to enable the COFDM generator to switch off all the carriers at selective times. Inserting zeros into the multiplex is not the solution since the carriers are phase modulated and this would generate a symbol representing a digital zero. Therefore, TDM operation produces a requirement for a 3-stage control of the COFDM transmitter output, a digital one, a digital nought, and a suppressed carrier. This can be done in two ways.

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In the first method, the symbols which are not going to be transmitted from the specified uplink are filled with dummy data and the RF output of the COFDM generator is switched off for the duration of the other contributions. As the contributing uplink sources only need to switch at a symbol boundary, this option is relatively simple. A small amount of logic is required to count through the symbols of each frame and switch at the appropriate time.

The second method is to configure the multiplexer and COFDM generator internally to switch off the unwanted carriers for the required time. The configuration is controlled from the multiplexing unit and a new interface to the COFDM generator. A new control mechanism would be required if the multiplexer was to be able to control adequately the COFDM generator. This requires access to the software on both devices.

In the TDM uplink arrangement, the transition points between the separate uplink signals as received at the satellite deserve special consideration. Apart from the problems of synchronisation, there is the problem introduced by the use of differential QPSK modulation. The receivers which are proposed for use with the signal decode each symbol in the ensemble with reference to the phase of the previous symbol (except for the first symbol of every transmission frame which is the fixed reference symbol). This is transmitted by uplink station number 1, the master, and is shown in Figure 2.

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The other uplink sites are called slaves. Data uplinked by these slave stations cannot be differentially decoded from the beginning because the previous symbol will originate from a different uplink site and will therefore have no useful phase relationship. Because of this, the first symbol of a slave contribution cannot be differentially decoded to provide any useful data. However, its phase state does then become the reference for the second symbol, thereby allowing the remaining symbols from that uplink contribution to be decoded as normal.

To solve this problem, a dummy phase reference symbol is inserted at the start of each slave contribution as shown in Figure 2. The multiplexer can easily be

configured to insert a dummy service component occupying just a single symbol which it fills with random data or any other data. As the system is differentially modulated, the following symbol will be demodulated with reference to the dummy symbol.

The system of Figure 2 shows three multiplex uplink sites carrying contributions of 128 K-bits/s and 64 K-bits/s as part of a TDM arrangement. The lower line of the diagram shows how the dummy phase reference signals inserted by each slave uplink site become part of the overall composite signal received and retransmitted by the satellite.

Loss of the first symbol of each uplink contribution is not a great problem. In Mode III DAB there is a low data-rate per symbol and this means that only 384 bits are lost for each slave uplink. This amounts to just under 0.7% per symbol and an arrangement using 10 geographically separate uplink sites (i.e. one master and nine slaves) would reduce the user capacity by only 6.25%.

Although a transitional dummy phase reference symbol cannot be used to carry any useful data, it may be used for carrying status information between uplink sites (by using a non-standard receiver).

The composite signal transmitted from the satellite will be the combined result of the several different uplink stations. However, it must not exhibit any artefacts of its TDM origination. Three fundamental parameters which must be kept as constant as possible are:

- 1. synchronisation
- 2. uplink frequency
- 3. power level.

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The handover between uplinks must not create overlaps or gaps in the signal, the power level must be constant throughout the transmission frame, and the frequency for each uplink must be the same so as not to create any discontinuity. That is to say, the final signal reaching the receiver must appear to be the result of a single transmission chain, rather than the combination of several contributing uplinks.

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At the handover point between contributing uplinks, the timing error needs to be accurate to within a fraction of a symbol duration. For Mode III DAB the total symbol duration is 156 microseconds, (which includes a guard interval of 31 μ s). Any "data collision" arising from a mis-aligned uplink would probably cause the loss of some data from both uplinks. In addition, such a data collision would increase the input power to the satellite by 3 dB. Given the finite power capability of a satellite transponder, and the fact that it is likely to be operating close to saturation, this could affect other users of the transponder or even drive the HPA into an overload condition.

A lack of data at the appropriate time could also create problems. In particular, the Eu-147 system uses the null symbol for coarse synchronisation in the time domain, therefore a data gap in the composite signal could be misinterpreted as a null symbol, thereby causing complete synchronisation failure at the receiver, resulting in none of the services on that multiplex being received. Therefore, it is also equally important that a contributing uplink does indeed fill its allocated timeslot.

Various factors must be considered and corrected for to insure that the uplink contributions arrive at the satellite's input antenna at the exact time required.

An uplink site suitable for use in the present invention is shown in figure 7. In this particular example the uplink site is combining two local audio signals for uptransmission to a satellite. Each audio signal is first fed to an MPEG audio coder 12. This compresses the audio data. It is next synchronised in a sub-multiplexor unit 14 which receives a synchronising signal from a global positioning system (GPS) clock receiver 16 which receives the GPS signal via an antenna 18. The multiplexor audio signal is then passed to a buffer delay 20 which feeds them at appropriate intervals to a COFDM modulator 22. This produces a frame of COFDM symbols.

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These symbols are supplied to an IF switching unit 24. This counts through the earth COFDM symbols in the frame in response to a clock signal which is supplied by the COFDM modulator 22 in its I/Q bus. The switching by the IF switching unit 24 makes sure that only symbols containing data relating to the two audio signals 10 are passed to an upconverter and high power amplifier 26 which then sends them to an antenna 28 for transmission to the satellite of Figure 1.

It will thus be appreciated that the system of Figure 1 comprises six uplink units similar to that of Figure 7. Four of these are handling only one audio signal, one is handling two audio signals and a final one is handling three audio signals. Each will be synchronised by its GPS clock receiver unit 16 and thus will insert audio data in symbols at different time periods to those used by other

uplink stations such that at the satellite a complete frame of data will be received.

If a more sophisticated receiver is used, the dummy symbol could carry other information. The first portion could be used as the phase reference. For example, a specific data pattern could be included. This could then be monitored by the various uplink sites to aid synchronisation of uplink contributions.

The dummy symbol could also be used as a data channel to feed back, to the uplink site providing the first contribution for each frame, information to go into the Fast Information Channel (FIC) which the first transmitter compiles and which describes the structure of the frame. Thus, it describes which symbols contain data for each channel and, clearly, which symbols are dummy symbols.

Thus, the data is fed to the transmitter compiling the FIC via the satellite. No land line is required.

Other data which could be included in the dummy symbol are an audio channel for communication between the uplink sites or additional data for various commercial services.

Furthermore, at each uplink site a receiver can be provided to monitor the timing and frequency of the dummy symbol it transmitted to the satellite. This can then be used to adjust the timing and frequency of the signal provided by the transmitter.

Slant Path Length Compensation

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The uplink stations will be located at arbitrary locations on the Earth's surface and will all experience different path lengths to the satellite. In order to create a seamless composite DAB signal the uplink sites with short

path lengths will need compensating delays so that their contributions do not arrive too early.

Given the orbital location of the satellite, and the latitude and longitude of the uplink station, the path length can be readily calculated. Taking a European example, as illustrated in Figure 5:

For a satellite at: 10.2° East
Uplink 1: Lisbon 39°N 9°W

Uplink 2: Bergen 61°N 5°E.

The nominal difference in the slant path range between the two earth stations and the satellite is $1,749~\mathrm{km}$, which corresponds to a delay of $5.83~\mathrm{ms}$.

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This could easily be compensated for by delaying the transmission from the Lisbon uplink site (which is closer to the satellite) by an equal amount. (This then allows the placement of the contributing signal at any point in the DAB transmission frame.)

The maximum possible slant path length would be experienced by an earth station on the very edge of the uplink coverage zone where the elevation angle is lowest. It is generally accepted that a minimum earth station antenna elevation angle of 5 degrees is required, and at such a location this gives a maximum possible slant path length of around 41,130 km (corresponding to a one-way propagation time of 138 ms). On the other hand, the shortest possible slant path distance would be from an earth station exactly at the sub-satellite point at a range of 35786 km, corresponding to a delay of 120 ms. The location of any uplink site can therefore be compensated for using a delay of no more than 18 ms, the

exact figure depending on its geographical location relative to the satellite.

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The BBC COFDM generator (CD2M/44) has a built in compensating delay of up to 4 ms, adjustable in increments of 488 ns, while the Marconi-Eddystone COFDM generator can manage a delay of up to 476 ms, adjustable in steps of approximately 1 μ s.

While the difference in the slant path length is the obvious (and major) consideration in synchronising the uplink stations, there are several other factors which affect the accuracy of the timing of each contribution. Some effects will create a common variation in the propagation delay between all the earth stations and the satellite, causing the whole DAB signal to arrive at the incorrect time. Other effects will cause differential errors which adversely change the synchronisation between the uplinked contribution signals.

Although termed "Geostationary", a satellite in GEO orbit will always have a tendency to wander a little, due to the Earth's gravitational irregularities, the influence of the Sun and Moon and solar pressure. These perturbations in the satellite's intended position complicate the uplinking of a TDM based system. As the satellite wanders about, the path length from the geographically separate contributing uplink sites will obviously vary. The normal satellite station keeping tolerance is usually quoted as +/-0.05° in each plane, corresponding to maintaining the satellite's position within a cube of sides approximately 80 km. This movement can therefore give the calculated slant path length an error of around +/- 40 km.

If this path length variation was identical for every uplink site then each uplink contribution would arrive at

the satellite slightly 'early' or 'late' but would maintain its place in the DAB frame. The whole broadcast signal would then arrive a few microseconds 'late' or 'early' but there would be no overall effect on synchronisation between the uplinks.

But, while the path length change between various uplink sites and the satellite is indeed largely the same, any station keeping error will usually create a small but significant differential change in these path lengths, which means a synchronisation error would be introduced between the various signals arriving at the satellite. This is illustrated in Figure 4 where d_1 and d_2 are the original distances from the uplink sites to the satellite, and Δ_1 and Δ_2 are the changes in distance due to orbital drift. If Δ_1 is then different to Δ_2 , then a synchronisation error will be introduced.

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Satellite station keeping errors can be resolved into three orthogonal planes - latitudinal - i.e. North/South, longitudinal - i.e. East/West, or radial - i.e. towards or away from the Earth. The magnitude of the differential change varies widely depending on the satellite's plane of movement, the location of the uplink sites and the magnitude of the error in the satellite's station keeping.

The maximum possible differential range would be between two uplinks at the extreme (5° elevation) and opposite edges of a global uplink coverage zone, with the satellite moving in the same plane. This would give a differential timing change of 1 $\mu s/km$ of satellite movement. In practice, very few uplink sites operate at these extremes and it is likely that most would be within a couple of thousand miles of each other.

Taking the Bergen/Lisbon/EMS example again, the nominal path length difference was shown to be $1,749~\mathrm{km}$

corresponding to a 5.83 ms fixed delay. Figure 5 shows the differential distance variation between the Bergen and Lisbon uplink paths for variations of the orbital position over the range +/- 0.05° or +/- 40 km in each of the three planes.

For a change in the satellite's latitude, Lisbon, being further south than Bergen, experiences a smaller rate of change of path length than Bergen, and at the extremes the error can be +/- 1344 metres, corresponding to +/- 4.5 μ s (which at 0.06 μ s is a long way short of the theoretical maximum shown above). For a change in the satellite's longitude a similar magnitude of differential error is experienced, while for a radial change in position, both uplink sites experience very similar changes, resulting in little differential error.

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The maximum change in timing would therefore occur when the satellite is at its maximum latitudinal error, and maximum longitudinal error and maximum radial error, combined with two uplink sites located in the same plane as the satellite's positional error. For uplink sites exclusively within Europe and a satellite station-keeping accuracy of +/- 0.05°, this would result in a maximum variation of around +/- 10 μ s, equivalent to +/-3 km. For worldwide uplinking the error could reach +/- 20 km (+/-67 μ s).

Slant path calculations are generally based on the assumption that the Earth is a uniform sphere. In reality it is an irregular ellipsoid, with a polar radius of 6256.74 km, and an equatorial radius of 6278.12 km, meaning the Earth is slightly 'wider' E-W than it is 'tall' N-S. While slant path length calculations generally use an average figure for the radius, this is not accurate enough for the TDM application. In addition,

the 'radius' of the Earth varies along any circumference due to further irregularities in the geodetic sphere. Therefore, if the Earth is incorrectly assumed to be a regular sphere, then the slant path distance may be in error by perhaps +/- 10 km, equivalent to a timing error of +/- 33 μ s.

Several geodetic models have been proposed to approximate the Earth's shape, with GPS for example using "WGS 84". This enables errors due to ellipsoid geometry to be reduced to just a few metres.

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The Earth station's height above sea level can also contribute to a timing error if it is located near the sub-satellite point. Mexico City, the uplink location for our first Eu-147 DAB satellite tests, is at an altitude of around 2 km above sea level.

Each of the contributing uplinks will need to be synchronised to a common time reference. The Global Positioning System (GPS) is a relatively low cost method of global timekeeping and can provide synchronisation to an accuracy of around 1 μ s anywhere in the world. With this application in mind, a pair of GPS based master reference clocks were tested and a histogram produced is shown in Figure 6.

The samples were taken over a period of several weeks, at irregular intervals of at least 15 minutes. As can be seen, there is a distinct fixed offset between the two receivers (an average of 130 ns) but excluding this offset, around 97% of the results show the receivers to be within 175 ns of each other. While the standard GPS specification provides a dithered signal accurate to within +/-340 ns of GPS time/UTC for 95% of the time, the affect of the GPS receiver's flywheel circuitry smooths out the short term phase noise giving a better result.

While the antennas of the two GPS receiver used for the test were located only 0.3 metres apart, the manufacturers claim that similar results would be obtained if the receivers were thousands of miles apart.

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The effect of the ionosphere varies depending mainly on sunspot activity, time of day, and path length through the ionosphere (which in turn depends on the satellite's elevation angle). The error contribution for the downlink path at 1.5 Ghz is likely to be less than 20 metres and will be common to all contributions. Atmospheric refraction on the uplink paths (typically 14 Ghz) is likely to be less than 1 metre (3 ns) and so will have a negligible affect on any particular uplink contribution.

Even a transparent transponder satellite will experience a small throughput delay, due mainly to filtering. This delay will be common to all contributions.

It has been shown above that there are several factors which will influence the accuracy which is achievable from a slant path distance calculation, and these are summarised in the table below. Some factors only cause an overall delay to the composite signal which is of little importance. Others (marked with a *) create a synchronisation error which may need to be compensated for. The figures given are 'typical worse case' examples.

	Parameters:		Distance	Time	
			Error	Error	
	1.	Irregular ellipsoid geometry			
		of Earth	10 km	33 μs	
30	2.	Height of earth station	2 km	6.7 μs	
		a.s.1.			

	3.	Atmospheric refraction	20 me	tres	67	ns
		(downlink @ 1.5 Ghz)				
	4.	Satellite processing delay	1 km		3.3	μs
	5.	Station keeping accuracy	80 km	ι	266	μs
5		of satellite				
	[Di	fferential error due to	6 km*	k	20	μ s*]
	st	ation keeping]				
	6.	Synchronisation clock	300 π	metres'	1	μs*
	*Differential errors			cm	21	μs
	,					

(creating synchronisation errors)

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Therefore, under poor conditions the timing change between two uplink stations in widely separated locations could be double this figure at $42\mu s$.

While several factors cause a delay common across all uplinks this can be compensated for with a fixed delay, but the time-varying differential error due to satellite drift and GPS receiver clock error will always remain and, depending on the uplink location, this could be significant. Using DAB transmission Mode III the guard interval is only 31 μ s, and in a hybrid satellite / terrestrial gap filler system, the erosion of the guard interval due to synchronisation errors would be particularly detrimental.

While the fixed components can all be compensated for by using the programmable internal delay of the COFDM generator, the time varying components may need to be eliminated by some form of closed loop control system based on the composite broadcast signal received at each slave uplink site as discussed earlier.

In a single uplink application the up-converter which mixes the signal to its final uplink frequency need not be

particularly stable as the receiver's AFC is capable of compensating for some error. However, in the COFDM uplink multiplexing system, the receiver's AFC and phase reference circuitry operate only on the first symbol of the DAB frame, and therefore only "tune in" to the master station. Switching to a different signal (i.e. a slave contribution) part way through the frame means a step change in the frequency, and any frequency difference gives rise to a loss of ruggedness of the signal. Therefore, each uplink site must employ a highly stable up-converter. The fact that Eu-147 uses differential coding is of benefit here, as it is the phase change between symbols which is important rather than absolute phase. A frequency reference with a short term (1000 seconds) frequency accuracy of <5 in 1010 is typically available from GPS clock receives which could assist in frequency matching of all slave stations.

Doppler Shift

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Geostationary satellites do not normally create any significant doppler shift of their own due to their fixed orbit (but a mobile terrestrial receiver will experience some doppler shift due to its own velocity unless the satellite is directly overhead). However, doppler shift may be a problem during a repositioning manoeuvre (when compensating for orbital drift), when the satellite may have to move many kilometres in a short period of time.

The frequency shift is caused by two components. The frequency of the uplink transmission (typically at Ku band, 14 Ghz) will appear to be slightly altered, while the frequency of the downlink (broadcast signal) will also change, and in the same direction, compounding the problem. However, because doppler shift is proportional

to frequency, the uplink accounts for around 90% of any frequency change. A fixed frequency error throughout the transmission frame is not a problem as it can be tracked by the AFC circuitry in the consumer's receiver. But in an uplink multiplexing arrangement, the doppler could create a step change in frequency part way through the frame, thereby degrading the quality of the slave contributions.

As the satellite undergoes its repositioning manoeuvre, each uplink signal may experience a different doppler shift, the magnitude of which will vary with the direction of movement of the satellite. The difference between the frequency shifts of the transmissions from the individual uplink sites depends on their geographical separation (in a similar way to the change in time synchronisation with satellite movement).

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Again, the frequency change is dependant on the satellite's velocity (i.e. speed and direction) and the geographical location of the uplink sites. The worse case situation would be between two uplinks at the extreme (5° elevation) and opposite edges of a global uplink coverage zone, with the satellite moving in the same plane. This could create a frequency step of approximately 15V Hz, where V is the velocity in metres/sec, (however this is a rather extreme and unlikely case). Monitoring the frequency transmitted by the satellite at each uplink site enables automatic feedback control of the uplink transmission to be achieved.

Repositioning is only likely to occur every few weeks and it may be possible to request that it happens at a convenient time of the night when audience figures are low (e.g. 04.00 am).

For a power limited system such as this where the link margin may well be just 2 dB, it is vital that the downlink power budget is maximised, and so the satellite must operate at its optimum power output. This requires that the power level of each contribution to the COFDM transmission frame should be matched to within a fraction of a dB when it arrives at the satellite's input antenna. Gain compensation for incorrect uplink power levels will not be possible at the satellite, and so each uplink site will have the responsibility of ensuring that its own power level matches that of the master station.

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The signal levels received at the satellite will depend on several factors - nominal uplink power setting, amplifier efficiency, transmitting antenna misalignment, equipment ageing, satellite receiving antenna gain variation with direction, spreading loss (due to the geographical location of the uplink site). In addition to these "fixed" variables the effect of atmospheric attenuation, and in particular the affects of local rain can change the effective uplink power level by 1 or 2 dB in only a few seconds.

The simplest way of achieving a constant envelope would be to monitor the broadcast signal at each slave site, and adjust the local uplink power as required. This would then take into account all the above variables and can be done automatically with a feedback loop.

MULTIPLEX CONFIGURATION AND THE FAST INFORMATION CHANNEL

In a normal single transmission chain system, the multiplex can be reconfigured dynamically, with the corresponding Multiplex Configuration Information (MCI) being signalled in the Fast Information Channel (FIC). In the TDM uplinking system it is not possible to time

multiplex the FIC data and so the master uplink station alone would provide the FIC and hence the MCI. This leads to some limitations in the way the multiplex can be reconfigured. For the multiplex to operate correctly it is essential that the data supplied by the MCI matches the actual configuration transmitted by each of the slave uplink sites.

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The simplest method is obviously for all parties to agree on a semi-permanent multiplex configuration. The MCI will therefore only need to be changed on the rare occasion when a radical reconfiguration is required, and a suitable scheme could be developed to ensure that all parties complied with the pre-agreed changes.

Where a multiplex reconfiguration is limited to an internal change at one uplink site only, so that the capacity transmitted from that site (i.e. the total number of symbols) remained constant, only the master and that particular slave site need to make any changes. However, a multiplex reconfiguration may require a change in the total capacity contributed by a particular site, and this would involve notifying the other affected sites of the impending change.

When the total number of symbols per frame transmitted by an uplink is to change, a complication arises. The multiplex reconfiguration is not an instant event due to the affect of the time interleaving process, and to comply thoroughly with the Eu-147 specification, would require that some of the data would continue to originate from the first uplink even after the second uplink had started to contribute to its newly acquired symbol. The mapping of bits onto carriers and the necessary switching is extremely complex and while such a

scheme would not be impossible to implement, the benefits would be perhaps marginal.

It is worth noting that while terrestrial DAB will experience roughly the same change in demand across all services through the day, for satellite DAB the situation is different. The different time zones covered by a single beam could mean that a particular service aimed primarily at the eastern edge of its coverage may require a larger proportion of the multiplex at the peak listening time of the day, and a few hours later may wish to relinquish some of its capacity to an uplink site primarily serving the western edge of the downlink beam as this region approaches its own peak listening time.

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One of the disadvantages with any TDMA like scheme is that the transmission equipment must be rated for the peak power levels, even though the average power output may be relatively low. For example, the normal RF power requirement for a DAB uplink, supplying a full multiplex, is typically around 10 Watts. However, the amplifier would need to be backed off by several dB from saturation (to prevent non-linear distortion), and so will need to be rated at around 30 Watts. A single uplink of 128 kbits/s contributing to the 1.152 Mbits/s DAB multiplex will only be operating at 11% duty cycle - in this case with an average power of 1.1 Watts but even so the amplifier used must still be rated at 30 Watts.

For each of the specified DAB operating modes, the carrier spacing is approximately proportional to the transmitting frequency. This means that the affects of oscillator phase noise and doppler shift, which scale with frequency, also remain constant. While for DAB Mode III the carrier spacing of 8 kHz is adequate for the transmitting frequency of around 1.5 Ghz, the uplink

frequency is likely to be several times greater than this, with most uplinks operating at around 6 Ghz (C-band) and 14 Ghz (Ku-band). Any phase noise in the up-converter therefore contributes to a degradation of the DAB signal, and so this component must be carefully chosen.

With any time multiplexed system it is vital that every contributing source is operating correctly synchronised so that it only transmits during its allocated period, otherwise errors will occur. It was pointed out in the section describing timing accuracy that a data collision may not only cause a data loss, and in severe cases may also cause amplifier overload or a reduction in available power for other users of the transponder.

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In particular, the first few data bits of every MPEG audio frame carry the very important MPEG header bits used for audio frame synchronisation. This data is mapped into the first symbol of the DAB audio frame, and so even a one symbol overlap due to an incorrect configuration may cause a complete loss of audio for the second contribution.

(Although the ETSI standard uses a 16-bit time interleaving process to shuffle the data around between frames, it does not change the relative position of the data within the frame, making the data particularly sensitive to frame rate effects).

In the Eu-147 system, the null symbol is essential for coarse synchronisation in the time domain and so if a slave uplink fault condition creates a gap, this can be misinterpreted as a null symbol, thereby preventing the receiver from acquiring synchronisation, and therefore resulting in *non* of the services on that multiplex being received.

In the event of a slave uplink being unable to provide a correctly timed signal at the correct frequency and with an appropriate power level it would be wise for it to drop out immediately, and be replaced by the master uplink for the duration of the fault. Therefore the master uplink station needs the flexibility to allow it to cover for fault conditions at any of the slave sites.

The cost of the additional equipment required to implement a time division multiplexed uplink, as described is relatively small. On the top of the usual equipment required for a 'hub' earth station (multiplexer, COFDM generator, upconverter and power amplifier), the only two extra pieces of equipment required for TDM operation are the GPS master clock receiver, costing around £2,000 and an RF switching unit, which if manufactured commercially would cost approximately £2,000.

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CLAIMS

- 1. A method for transmitting a multi-carrier signal having a regular frame structure and symbol rate comprised of contributions from a plurality of different transmitters comprising the steps of:
- a) transmitting the contributions from each transmitter to a central transmitter in preassigned time slots;
- b) retransmitting the thus received contributions as a single signal over a predetermined area of coverage; and
 - c) inserting at the start of each contribution a dummy symbol for use as a phase reference for demodulating succeeding symbols in that contribution.
- 2. A method according to claim 1 in which the central transmitter comprises a satellite in geostationary orbit and the plurality of transmitters comprise earth based transmitters.
- 3. A method according to claim 1 in which the
 central transmitter comprises a stratospheric platform in
 geostationary orbit and the plurality of transmitters
 comprise Earth based transmitters.
 - 4. A method according to claim 2 or 3 including the step of providing a timing reference signal to each earth based transmitter.

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5. A method according to claim 4 in which the step of providing a timing reference comprises detecting a global timing signal transmitted by a global positioning system (GPS).

- 6. A method according to claim 2 or 3 comprising the step of providing common frequency reference signals to each Earth based transmitter.
- 7. A method according to any preceding claim including the step of delaying transmission of signals from each Earth station to the satellite in dependence on the position on the Earth's surface of each Earth station.
- 8. A method according to claim 7 including the step of monitoring at each Earth station the COFDM signal from the satellite and adjusting the delay applied to transmissions from the Earth station to compensate for any timing errors caused by other factors.
- 9. A method according to claim 8 in which timing errors are caused by the relative position and velocity of the satellite or stratospheric platform.
 - 10. A method according to any preceding claim including the step of monitoring at each transmitter the timing and frequency of the contribution supplied by that transmitter after re-transmission by the central transmitter, and adjusting the timing and frequency of the signal to be transmitted in dependence on the received signal.
- 11. A method according to claim 10 in which the dummy symbol includes a predetermined pattern of data

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which are used for monitoring the timing and frequency of signals received at each transmitter.

- 12. A method according to any preceding claim in which at least part of the dummy symbol is used to transmit data to dedicated receivers.
- 13. A method according to claim 12 in which the data for dedicated receivers is used as data for voice communication channel between Earth stations.
- 14. A method according to any preceding claim in
 which at least part of the dummy symbol is used as a data
 channel to supply data to the transmitter providing the
 first contribution in each frame of data for inclusion in
 an information signal defining the structure of the frame.





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Other: NONE

Documents considered to be relevant:

Сатедогу	Identity of document and relevant passage		
A	GB 2313527 A	MITSUBISHI	None
A	EP 0683576 A1	HITACHI	None
A	WO 94/08405 A1	MOTOROLA	None
A	US 4574379	AT&T	None

Document indicating lack of novelty or inventive step
 Document indicating lack of inventive step if combined with one or more other documents of same category.

Member of the same patent family

A Document indicating technological background and/or state of the art.
 P Document published on or after the declared priority date but before the filing date of this invention.

E Patent document published on or after, but with priority date earlier than, the filing date of this application.

1/1 ページ

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(54) TRANSMISSION METHOD FOR OFDM MODULATION SIGNAL, OFDM TRANSMITTER AND RECEIVER

(57)Abstract:

PROBLEM TO BE SOLVED: To prevent production of distortion by spreading a reference phase of each carrier of the frequency orthogonal division multiplex (OFDM) system and changing the amplitude of each carrier of the OFDM without giving effect on transmitted information so as to suppress a peak level of signals.

SOLUTION: After multiplying a complex code series ejsk (Sk=pk2, p is an optional real number not being zero, 0≤k≤N, N is a total carrier number) with an input coded signal, inverse FFT is applied to the product to generate an OFDM modulation signal and it is transmitted. At a receiver side, a complex code series ejsk (Sk is the same as above) is multiplied with a signal resulting from FFT processing to a received signal and an OFDM demodulation signal is obtained. The information relating to the ejsk required for demodulation is included in the input coded information, or sent in advance from the transmitter side to the receiver side through other transmission line. Thus, the reference phases of each carrier of the OFDM are hardly arranged and the level of transmission signals is suppressed and the resulting signal is sent, then an operating point of amplifiers is set higher.

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CLAIMS

[Claim(s)]

[Claim 1]In the transmitting side, it is a complex code sequence to an input encoded signal. [External Character 1] $_{\rm e}$ i S $_{\rm t}$

(— the signal which they generated the OFDM modulation signal here and transmitted to it as reverse FFT of S_k =pk², the arbitrary real numbers whose p is not zero, $0 \le k \le N$, and the N was carried out after they carried out the multiplication of total number of careers) here, and carried out FFT of the input signal in the receiver — a complex code sequence [External Character 2] e^{-i} S_k

A transmission method of an OFDM modulation signal carrying out the multiplication of $(S_k$ is the same here to said S_k), and acquiring an OFDM demodulation signal.

[Claim 2]Said complex code sequence required for an OFDM recovery [in / on a transmission method of the OFDM signal according to claim 1, and / a receiver] [External Character 3]

the arbitrary real numbers $S_k = pk^2$ and whose p are not zero here. A transmission method of an OFDM modulation signal, wherein $0 \le k \le N$ and N include the information about the total number of careers in said input encoded signal or transmit it to a receiver beforehand from the transmitting side in transmission lines other than the transmission line for OFDM transmission concerned

[Claim 3]It is a complex code sequence to an input encoded signal at least. [External Character 4] $_{\rm e}$; S $_{\rm t}$

(— the OFDM sending set which $S_k = pk^2$, the arbitrary real numbers whose p is not zero, 0 $\leq k \leq N$, and N are provided with the means which carries out the multiplication of total number of careers) here, and is characterized by things.

[Claim 4]It is a complex code sequence to a signal which carried out FFT of the input signal at least. [External Character 5] $_{\rm e}$ – $_{\rm i}$ S $_{\rm x}$

(— the OFDM receiving set which $S_k = pk^2$, the arbitrary real numbers whose p is not zero, 0 $\leq k \leq N$, and N are provided with the means which carries out the multiplication of total number of careers) here, and is characterized by things.

[Claim 5]In a transmission method of an OFDM modulation signal which generates a modulating signal of either BPSKOFDM and a QPSKOFDM modulating signal, and is transmitted, After carrying out the multiplication of two or more constants which make equal amplitude of positive [of a phase which said one of modulating signals can take], and a negative ingredient in an

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amplitude peak period of one of said modulating signals according to a value of an input encoded signal, respectively, A transmission method of an OFDM modulation signal characterized by carrying out reverse FFT, generating an OFDM modulation signal, and making it transmit. [Claim 6]In an OFDM sending set which generates a modulating signal of either BPSKOFDM and a QPSKOFDM modulating signal, and transmits, An OFDM sending set which is provided with a means which carries out the multiplication of two or more constants which make equal amplitude of positive [of a phase which said one of modulating signals can take at least], and a negative ingredient in an amplitude peak period of one of said modulating signals according to a value of an input encoded signal, respectively, and is characterized by things.

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DETAILED DESCRIPTION

[Detailed Description of the Invention] [0001]

[Field of the Invention] In a broadcasting satellite, in order to use the generating electric power by a solar cell, the output of a relay amplifier has restriction, but. This invention, A transmission method and an OFDM sending set of a frequency rectangular cross division multiplex (OFDM:Orthogonal Frequency Division Multiplexing) modulating signal suitable for using it for digital broadcasting in such a satellite system (as opposed to a ground system), etc., It is related with a receiving set.

[0002]

[Problem(s) to be Solved by the Invention] Conventionally, there are phase modulations, such as BPSK and QPSK, in the modulation method of each career of OFDM. In these modulation methods, the reference phase of each career by which multiplex was carried out is constant, and, in the case of BPSK, in the case of a binary and QPSK, the phase which each modulating signal can take is restricted with four value. Therefore, in the phase of each career, by this method, the peak of amplitude may occur on a set or the OFDM time base signal which becomes empty.

[0003] For example, in the relay amplifier for broadcast, while generating electric power and amplifier efficiency have restriction, in order to secure the rate of a service period, and the rate of a place, it is used near the saturation region. In order to secure the rate of a service period, and the rate of a place also in digital broadcasting using an OFDM modulation method, to take the high operating point of an amplifier is desired. However, it is one side, and if the high operating point is taken in this way, in the amplitude peak of an OFDM modulation signal, it will become easy to generate distortion.

[0004]In the situation which has restriction in the above generating electric power and amplifier efficiency, the purpose of this invention is to suppress the amplitude peak of an OFDM signal and to perform little transmission in the high operating point.

[0005]

[Means for Solving the Problem] It is going to control an amplitude peak of a signal by diffusing a reference phase of OFDM each career, or changing amplitude of OFDM each career, without affecting information which should be transmitted in this invention, in order to attain the above—mentioned purpose. In order to make diffusion of these reference phases thru/or change of amplitude perform, in this invention, the multiplication of the specific signal (S) is carried out so that an input encoded signal may not be affected at transmitted data (a case where it amends by a receiver so that it may not be affected is included), OFDM modulation is performed based on it, and each career is transmitted.

[0006]When carrying out signal (S) multiplication and diffusing a reference phase of each career now, a phase of each career becomes difficult to gather and can be transmitted by suppressing a peak of amplitude. In this case, in a receiver, the multiplication of the signal (S*) corresponding to a signal (S) which carried out multiplication at the above-mentioned transmitting side is carried out to an OFDM demodulation signal, and right information is restored.

[0007]When each career does not have information in amplitude directions, such as BPSK and QPSK, in carrying out the multiplication of the signal (S) at the transmitting side, transmission

which suppressed a signal peak is performed by choosing a signal (S) which carries out multiplication so that amplitude of a career with which phases produce an amplitude peak together, and a career with an ingredient of an opposite phase may be enlarged — things can be carried out. Here, since multiplication of a signal (S) can be performed only by calculation by a discrete time, it can respond flexibly with software.

[0008]That is, a transmission method of this invention OFDM modulation signal is a complex code sequence to an input encoded signal in the transmitting side. [External Character 6]

(— a signal which they generated an OFDM modulation signal here and transmitted to it as reverse FFT of S_k =pk², the arbitrary real numbers whose p is not zero, 0 <=k<=N, and the N was carried out after they carried out the multiplication of total number of careers) here, and carried out FFT of the input signal in a receiver — a complex code sequence [External Character 7] e — i S_x

The multiplication of (S_k) is the same here to said S_k) is carried out, and the OFDM demodulation signal was acquired.

[0009]Said complex code sequence which needs the transmission method of this invention OFDM modulation signal for the OFDM recovery in a receiver [External Character 8] $_{\rm e}$ $_{\rm r}$ $_{\rm s}$ $_{\rm s}$

the arbitrary real numbers $S_k = pk^2$ and whose p are not zero here. $0 \le k \le N$ and N include the information about the total number of careers in said input encoded signal, or transmitted it to the receiver beforehand from the transmitting side in transmission lines other than the transmission line for OFDM transmission concerned.

[0010]this invention OFDM sending set is a complex code sequence to an input encoded signal at least. [External Character 9] $_{\rm e}$ $_{\rm i}$ S $_{\rm c}$

 $(S_k=pk^2,$ the arbitrary real numbers whose p is not zero, $0 \le k \le N$, and N are provided with the means which carries out the multiplication of total number of careers) here [0011]this invention OFDM receiving set is a complex code sequence to the signal which carried out FFT of the input signal at least. [External Character 10] $_{R} = 3.5$ x

 $(S_{L}=pk^{2}, the arbitrary real numbers whose p is not zero, 0 <=k<=N, and N are provided with the$ means which carries out the multiplication of total number of careers) here [0012]A transmission method of this invention OFDM modulation signal, In a transmission method of an OFDM modulation signal which generates a modulating signal of either BPSKOFDM and a QPSKOFDM modulating signal, and is transmitted, After carrying out the multiplication of two or more constants which make equal amplitude of positive ∫ of a phase which said one of modulating signals can take], and a negative ingredient in an amplitude peak period of one of said modulating signals according to a value of an input encoded signal, respectively, Reverse FFT is carried out, an OFDM modulation signal is generated, and it was made to transmit. [0013]In an OFDM sending set which this invention OFDM sending set generates a modulating signal of either BPSKOFDM and a QPSKOFDM modulating signal, and transmits, It has a means which carries out the multiplication of two or more constants which make equal amplitude of positive [of a phase which said one of modulating signals can take at least], and a negative ingredient in an amplitude peak period of one of said modulating signals according to a value of an input encoded signal, respectively. [0014]

[Embodiment of the Invention] With reference to an accompanying drawing, this invention is explained in detail based on an embodiment of the invention below. <u>Drawing 1 is a transmission</u>

code series (input encoded signal). [External Character 11] X k **** - To the signal which carried out the parallel conversion, it is an example of a numerals system. [External Character 12] S k (book specification preceding paragraph shows one embodiment of the OFDM sending set by this invention which only carries out the multiplication of (having expressed with S)), and is transmitted. This embodiment is an embodiment which diffuses a phase to BPSKOFDM and controls a signal peak. [0015]In drawing 1, it is a code sequence as an input encoded signal. [External Character 13] Хĸ Direct in the serial/parallel conversion machine 1 - Carry out a parallel conversion and it is made a parallel signal. Furthermore, an OFDM modulation signal is acquired from an output terminal of the converter 3 via reverse FFT circuit (Invers Fast Fourier Transform circuit) 2 and the parallel serial change machine 3. It is a process of the usual OFDM modulation signal generation so far (however, when there is no multiplier 4 in drawing 1). [0016]On the other hand, a code sequence which the multiplier 4 is inserted, respectively between each parallel line between the serial/parallel conversion machine 1 and the reverse FFT (IFFT) circuit 2, and is an input encoded signal as this invention is shown in drawing 1 [External Character 14] Хĸ Code sequence [External Character 15] S_k Multiplication is performed in between and the multiplication result is supplied to reverse FFT circuit 2. The composition of drawing 1 is [in / are an example and / this invention] a code sequence. [External Character 16] Хĸ It is alike and a code sequence. [External Character 17] Sk What is necessary is just the composition by which multiplication is carried out to ******, and it is not necessary to follow circuit arrangement shown in drawing 1. [0017] Here, it is an input code sequence. [External Character 18] X k It is a code sequence of ** and the binary (-either 1 or 1 is taken) of N pieces, and is a code sequence. [External Character 19] S_k ******* [External Character 20] Хĸ

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A complex code sequence at least for ** to diffuse a phase [External Character 21]

e i S k

It comes out. S_k is a known series beforehand in a receiver, for example, the time delay of each career is proportional to frequency — as [Equation 1] $S_k = p k^2$

However, the arbitrary real numbers whose p is not zero, $0 \le K \le N$, and N are the total numbers of careers.

In this way, the input code sequence of a binary [External Character 22]

Хk

A compound code sequence for ** and phase diffusion [External Character 23]

Multiplication is mutually carried out by the multiplier 4. The OFDM modulation signal by which phase diffusion was carried out as a result of multiplication [External Character 24]

T k

It is obtained by the output terminal of the ** parallel serial converter 3. [0018] <u>Drawing 2</u> (a) and (b) is a code sequence (complex code sequence) to the BPSKOFDM modulating-signal generate time mentioned above. [External Character 25]

 S_k

An example of a constellation of each career of a modulating signal a time (the usual BPSKOFDM) of not carrying out multiplication and when multiplication is carried out by this invention is shown, respectively. Drawing 3 (a) and (b) shows an example of a signal wave form of an OFDM modulation signal corresponding to them, respectively. It turns out that it is transmitted as a signal (drawing 3 (b)) with which a phase of each career becomes difficult to gather compared with a case where phase diffusion of the time base waveform of a signal by which phase diffusion was carried out is not carried out (drawing 2 (b)), and a peak of amplitude was suppressed from drawing 3.

[0019]A signal which carried out FFT of the input signal with which <u>drawing 4</u> received a signal (that is, phase diffusion was carried out and transmitted) transmitted by an above-mentioned method

[External Character 26]

R k

It is alike and is a complex code sequence. [External Character 27]

(— this specification preceding paragraph — only (S^*) — a table — the bottom —) — carrying out multiplication — a right received code series [External Character 28]

Хĸ

One embodiment of an OFDM receiving set by this invention to restore is shown. [0020]In drawing 4, it is an input signal.

[External Character 29]

R k

Direct in the serial/parallel conversion machine 5 - A parallel conversion is carried out, it is made a parallel signal, and an OFDM recovery is further carried out in FFT circuit 6. It is a complex code sequence to this OFDM demodulation signal to which it restored. [External

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Character 30]
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S k

Each multiplier 7 for carrying out multiplication is arranged between FFT circuit (fast Fourier Transform circuit) 6 and the parallel serial conversion circuit 8.

[0021]Complex code sequence [External Character 31]

 $S_{\mathbf{k}}$

It is a complex code sequence at ** and the transmitting side at the time of OFDM modulation. [External Character 32]

eiSk

The signal with which phase diffusion also of the receiver was carried out since multiplication was carried out and phase diffusion of the modulated wave was carried out will be received, and it is a right received code series as it is. [External Character 33]

X k

Specifically at the code sequence for carrying out reverse correction of it becoming impossible to restore, it is a complex code sequence. [External Character 34] $_{\alpha}$ = $_{i}$ S_{x}

(S_k is a known series beforehand in a receiver, for example, is $S_k=k^2$;, however 0<=K<=N(N: total number of careers)). A code sequence restored eventually [External Character 35]

Χĸ

It is a code sequence of 1 or 1 [-] of k **. [0022]Thereby, it is an input signal. [External Character 36]

R k

FFT is carried out by FFT6 through the ** serial/parallel conversion 5, and it is inputted into the multiplication terminal of the multiplier 7. On the other hand, it is a code sequence of a reverse correction sake about phase diffusion. [External Character 37]

Sk

In *******, it is known beforehand and is inputted into a multiplication terminal of the multiplier 7. In an output multiplier, it is a received code series of a binary (-1, 1). [External Character 38] \dot{X}_k

It ***** and is taken out via the parallel serial converter 8.

[0023] The complex code sequence which is needed above by a receiver for a right OFDM recovery [External Character 39]

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It must be transmitted to a receiver in ** and a certain form. This is a complex code sequence. [External Character 40]

e - i 5 x

The very thing is not transmitted but information which it can reproduce by a receiver should just be sent. As a transmission method, it is made to contain in an input encoded signal at the transmitting side, and transmits in a transmitted symbol, or may transmit in a transmission line different from it.

[0024]A code sequence used in this invention when performing OFDM modulation and a recovery

by drawing 1 and drawing 4, respectively [External Character 41]

 S_k

[External Character 42]

Sk

** -- a transmission code series a peak is still more apt to break off even if this changes a transmission code series for every symbol by a known method at the transmitting side, and transmits by performing phase diffusion and it is made to carry out reverse correction by a receiver, although it was considered as a regularity (it does not change) thing in time [External Character 43]

X k

Being spread is possible.

[0025]In consideration of the case where the amplitude of each career of an OFDM signal has information in the above-mentioned example (actually, in BPSKOFDM of the above-mentioned example, it does not have, but in the case of the multiple value QAMOFDM, it has information), it is the transmitting side and is a complex code sequence. [External Character 44]

 S_k

Since multiplication is carried out, and the topology will be lost if phase diffusion is carried out, in order to recover this, it is a receiver, and it is a complex code sequence. [External Character 45]

Sk

Multiplication was carried out.

[0026]On the other hand, in BPSKOFDM and QPSKOFDM, there is information only in the phase of each career and it does not have information in the amplitude direction of each career. Then, in BPSKOFDM, it is <u>drawing 1</u>, for example. [External Character 46]

 S_k

It carries out and is a transmission code series. [External Character 47]

Χı

case N/(2N $_1$) of ** 1, and a case of 1 [–] — N/(2N $_2$) — the multiplication of the constant shall be carried out to a transmission code series (equivalent to an input encoded signal), respectively It is here, and N is symbol length and N $_1$ and N $_2$ is in a symbol, respectively.

[External Character 48]

X k

It is the number of **1 and -1. By carrying out like this, it is a transmission signal. [External Character 49]

T k

Since a size of an ingredient 1, -1 or positive, and negative becomes equal at a ****** peak period, a peak of amplitude can be suppressed and transmitted like the above-mentioned example. In this case, in a receiver, since it does not have information in an amplitude direction of each career, reverse correction of career diffusion for right decoding is not needed. [0027]

[Effect of the Invention] According to this invention, as explained above, also in the amplifier which has restriction in generating electric power, such as satellite broadcasting, it has the purpose of securing a hour rate and the rate of a place, and even if it makes it operate in the

higher operating point, it becomes possible to carry out little distorted OFDM transmission.
[Translation done.]

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(54) 【発明の名称】 OFDM変調信号の伝送方法およびOFDM送信装置、受信装置

(57)【要約】

(修正有)

【課題】 OFDM変調方式を用いたディジタル放送において、サービス時間率、場所率を確保するためには増幅器の動作点を高くとると、OFDM変調信号のピークにおいて歪みを発生し易くなる。

【解決手段】 送信側においては、入力符号化信号

X k

に複素符号系列

e 1 S k

(ここに、 $S_k = p k^2$, pはゼロでない任意の実数、 $0 \le k \le N$, Nは総キャリア数)を乗算(4)した後逆 F F T(2)するようにしてO F DM変調信号

Τ×

を生成して送信し、受信側においては、受信信号をFFTした信号に複素符号系列

e - : S x

(ここに、 S_{κ} は上記 S_{κ} に同じ)を乗算してOFDM 復調出力信号を得るようにした。

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【特許請求の範囲】

【請求項1】 送信側においては、入力符号化信号に複素符号系列

【外1】

e i S k

(ここに、 $S_1 = p k^2$, pはゼロでない任意の実数、 $0 \le k \le N$, Nは総キャリア数)を乗算した後逆FFT するようにしてOFDM変調信号を生成して送信し、受信側においては、受信信号をFFTした信号に複素符号系列

【外2】

e - : 5 x

(ここに、 S_R は前記 S_R に同じ)を乗算してOFDM 復調信号を得るようにしたことを特徴とするOFDM変 調信号の伝送方法。

【請求項2】 請求項1記載の0FDM信号の伝送方法 において、受信側における0FDM復調のために必要な 前記複素符号系列

【外3】

e -; **5** k

(ここに、 $S_k = p k^2$, pはゼロでない任意の実数、 $0 \le k \le N$, Nは総キャリア数)に関する情報を、前記入力符号化信号に含ませ、または当該OFDM伝送用伝送路以外の伝送路にて送信側から受信側に予め伝送するようにしたことを特徴とするOFDM変調信号の伝送方法。

【請求項3】 少なくとも入力符号化信号に複素符号系列

【外4】

e i S x

(ここに、 $S_k = p k^2$, pはゼロでない任意の実数、 $0 \le k \le N$, Nは総キャリア数)を乗算する手段を具えてなることを特徴とするOFDM送信装置。

【請求項4】 少なくとも受信信号をFFTした信号に 複素符号系列

【外5】

e - i 5 x

(ここに、 $S_k = p k^4$, pはゼロでない任意の実数、 $0 \le k \le N$, Nは総キャリア数)を乗算する手段を具えてなることを特徴とするO F D M受信装置。

【請求項5】 BPSKOFDMおよびQPSKOFD M変調信号のいずれかの変調信号を生成して送信するOFDM変調信号の伝送方法において、前記いずれかの変調信号がとりうる位相の正と負の成分の振幅を、前記いずれかの変調信号の振幅ピーク時において等しくするような複数の定数を入力符号化信号の値に応じてそれぞれ乗算した後、逆FFTしてOFDM変調信号を生成して送信するようにしたことを特徴とするOFDM変調信号の伝送方法。

【請求項6】 BPSKOFDMおよびQPSKOFD 50

M変調信号のいずれかの変調信号を生成して送信するOFDM送信装置において、少なくとも前記いずれかの変調信号がとりうる位相の正と負の成分の振幅を、前記いずれかの変調信号の振幅ピーク時において等しくするような複数の定数を入力符号化信号の値に応じてそれぞれ乗算する手段を具えてなることを特徴とするOFDM送

【発明の詳細な説明】

[0001]

信装置。

【発明の属する技術分野】放送衛星においては、太陽電池による発生電力を使用するため、中継増幅器の出力に制限があるが、本発明は、そのような衛星系(地上系に対して)でのディジタル放送等に使用するのに適している周波数直交分割多重(OFDM:Orthogonal Frequency Division Multiplexing)変調信号の伝送方法およびOFDM送信装置、受信装置に関する。

[0002]

【発明が解決しようとする課題】従来、OFDMの各キャリアの変調方式には、BPSK,QPSKなどの位相変調がある。これらの変調方式では多重された各キャリアの基準位相は一定であり、各変調信号のとりうる位相はBPSKの場合2値、QPSKの場合4値と限られている。従って、この方式では各キャリアの位相がそろいやすく、OFDM時間軸信号上に振幅のピークが発生する場合がある。

【0003】例えば、放送用中継増幅器においては、発生電力、増幅器効率に制限があるなかでサービス時間率、場所率を確保するために飽和領域付近で使用されている。また、OFDM変調方式を用いたディジタル放送においてもサービス時間率、場所率を確保するためには増幅器の動作点を高くとることが望まれる。しかし一方で、このように動作点を高くとると、OFDM変調信号の振幅ピークにおいて歪みを発生し易くなる。

【0004】本発明の目的は、上記のような発生電力、 増幅器効率に制限がある状況において、OFDM信号の 振幅ピークを抑え高い動作点で歪みの少ない伝送を行う ことにある。

[0005]

【課題を解決するための手段】上記目的を達成するた 40 め、本発明においては、伝送すべき情報に影響を与えることなく、OFDM各キャリアの基準位相を拡散させ、または、OFDM各キャリアの振幅を変化させることによって信号の振幅ピークを抑制しようとするものである。これら基準位相の拡散、ないし振幅の変化を行わせるために、本発明では、入力符号化信号に伝送情報に影響を与えないように(受信側で、影響を与えないように補正する場合を含む)特定の信号(S)を乗算し、それをもとにOFDM変調を行い、各キャリアの伝送を行う。

【0006】いま、信号(S)乗算して各キャリアの基

準位相を拡散させる場合、各キャリアの位相はそろいに くくなり、振幅のピークを抑えて伝送を行うことができ る。この場合、受信側において、上記送信側で乗算した 信号(S)に対応した信号(S*)をOFDM復調信号 に乗算して正しい情報を復元するようにする。

【0007】また、各キャリアがBPSK, QPSKな ど振幅方向に情報を持たない場合、送信側で信号(S) を乗算するにあたり、位相がそろって振幅ピークを生じ るキャリアと逆位相の成分を持つキャリアの振幅を大き くするように乗算する信号(S)を選ぶことによって、 信号ピークを抑えた伝送を行うことできる。ここで、信 号(S)の乗算は離散時間での計算だけで行うことがで きるため、ソフトウェアによって柔軟に対応することが

【0008】すなわち、本発明OFDM変調信号の伝送 方法は、送信側においては、入力符号化信号に複素符号 系列

【外6】

e i S x

(ここに、 $S_k = p k^2$, pはゼロでない任意の実数、 $0 \le k \le N$, Nは総キャリア数)を乗算した後逆FFT するようにしてOFDM変調信号を生成して送信し、受 信側においては、受信信号をFFTした信号に複素符号 系列

【外7】

е - і 5 х

(ここに、Sk は前記Sk に同じ)を乗算してOFDM 復調信号を得るようにしたことを特徴とするものであ

【0009】また、本発明OFDM変調信号の伝送方法 30 は、受信側におけるOFDM復調のために必要な前記複 素符号系列

【外8】

е-і **б**к

(ここに、 $S_k = p k^2$, pはゼロでない任意の実数、 $0 \le k \le N$, Nは総キャリア数)に関する情報を、前記 入力符号化信号に含ませ、または当該OFDM伝送用伝 送路以外の伝送路にて送信側から受信側に予め伝送する ようにしたことを特徴とするものである。

【0010】また、本発明OFDM送信装置は、少なく 40 とも入力符号化信号に複素符号系列

【外9】

e i S x

(ここに、 $S_k = p k^2$, pはゼロでない任意の実数、 0≤k≤N, Nは総キャリア数)を乗算する手段を具え てなることを特徴とするものである。

【0011】また、本発明OFDM受信装置は、少なく とも受信信号を F F T した信号に複素符号系列 【外10】

e - i 5 x

(ここに、 $S_k = p k^2$, pはゼロでない任意の実数、 0≤k≤N, Nは総キャリア数)を乗算する手段を具え てなることを特徴とするものである。

【0012】また、本発明OFDM変調信号の伝送方法 は、BPSKOFDMおよびQPSKOFDM変調信号 のいずれかの変調信号を生成して送信するOFDM変調 信号の伝送方法において、前記いずれかの変調信号がと りうる位相の正と負の成分の振幅を、前記いずれかの変 調信号の振幅ピーク時において等しくするような複数の 10 定数を入力符号化信号の値に応じてそれぞれ乗算した 後、逆FFTしてOFDM変調信号を生成して送信する ようにしたことを特徴とするものである。

【0013】また、本発明OFDM送信装置は、BPS KOFDMおよびQPSKOFDM変調信号のいずれか の変調信号を生成して送信するOFDM送信装置におい て、少なくとも前記いずれかの変調信号がとりうる位相 の正と負の成分の振幅を、前記いずれかの変調信号の振 幅ピーク時において等しくするような複数の定数を入力 符号化信号の値に応じてそれぞれ乗算する手段を具えて なることを特徴とするものである。

[0014]

【発明の実施の形態】以下に添付図面を参照し、発明の 実施の形態に基づいて本発明を詳細に説明する。図1 は、送信符号系列(入力符号化信号)

【外11】

Xk

を直一並列変換した信号に、符号系例 【外12】

Sk

(本明細書前段では、単に(S)にて表した)を乗算し て伝送する本発明によるOFDM送信装置の一実施形態 を示している。なお、本実施形態は、BPSKOFDM に位相の拡散を行って信号ピークを抑制する実施形態で ある。

【0015】図1において、入力符号化信号としての符 号系列

【外13】

Xk

をシリアルーパラレル変換器1において直-並列変換し て並列信号にし、さらに逆FFT回路(Invers Fast Fo urier Transform circuit) 2 およびパラレルーシリアル 変化器3を介して変換器3の出力端子からOFDM変調 信号が得られる。ここまでは通常のOFDM変調信号発 生のプロセスである(但し、図1において乗算器4がな い場合)。

【0016】これに対し、本発明においては、図1に示 50 すように、シリアルーパラレル変換器 1 と逆 F F T (I

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(4) 特開平10-84329

FFT)回路2との間の各並列線の間にそれぞれ乗算器4を介挿し、入力符号化信号である符号系列

【外14】

X k

と符号系列

【外15】

S k

との間で乗算を行い、その乗算結果が逆FFT回路2に 10 供給されるようにする。なお、図1の構成は一例であ り、本発明においては、符号系列

【外16】

· X k

に符号系列

【外17】

 S_k

が相互に乗算されるような構成であればよく、図1に示 20 す回路配置に従う必要はない。

【0017】ここで、入力符号系列

【外18】

X k

は、N個の2値(1, -1のいずれかをとる)の符号系列であり、符号系列

【外19】

 S_{k}

は符号系列

【外20】

. X k

の位相を拡散するための複素符号系列

【外21】

e i S x

である。(S₁ は受信側で予め既知の系列であり、例えば、各キャリアの遅れ時間が周波数に比例するように、

【数1】

 $S_k = p k^2$

ただし、pはゼロでない任意の実数、 $0 \le K \le N$, Nは 総キャリア数である。)

こうして、2値の入力符号系列

【外22】

X k

は、位相拡散のための複合符号系列

【外23】

eisk

と乗算器4によって相互に乗算される。乗算の結果、位 相拡散されたOFDM変調信号

【外24】

T k

がパラレルーシリアル変換器3の出力端子に得られる。 【0018】図2(a),(b)は、上述したBPSKOFDM変調信号生成時に符号系列(複素符号系列) 【外25】

Sk

を乗算しない(通常のBPSKOFDM)ときと、本発明によって乗算したときの変調信号の各キャリアのコンスタレーションの一例をそれぞれ示している。また、図3(a),(b)は、それらに対応したOFDM変調信号の信号波形の一例をそれぞれ示している。図2および図3から、位相拡散された信号の時間軸波形は、位相拡散されない場合に比べ各キャリアの位相がそろいにくくなり(図2(b))、振幅のピークが抑えられた信号(図3(b))として伝送されることが分かる。【0019】図4は、上述の方法で送信された(すなわち、位相拡散して送信された)信号を受信した受信信号

を F F T した信号 【外 2 6】

R k

に複素符号系列

【外27】

30

· ·

(本明細書前段では、単に(S^{\star})で表した)を乗算して正しい受信符号系列

【外28】

X k

を復元する本発明によるOFDM受信装置の一実施形態を示している。

【0020】図4において、受信信号

【外29】

Rъ

をシリアルーパラレル変換器5において直一並列変換して並列信号にし、さらにFFT回路6においてOFDM復調する。この復調されたOFDM復調信号に複素符号系列

【外30】

Sk

を乗算するための各乗算器7がFFT回路 (fast Fouri 50 er Transform circuit)6とパラレルーシリアル変換回 路8との間に配置される。

【0021】複素符号系列

【外31】

 S_k

は、送信側でOFDM変調時に複素符号系列 【外32】

eiSx

を乗算し、変調波を位相拡散させたため、受信側でも位 相拡散された信号が受信されてしまい、そのままでは正 10 しい受信符号系列

【外33】

X k

を復元できなくなるのを逆補正するための符号系列で、 具体的には、複素符号系列

【外34】

e - i 5 k

(Sk は受信側で予め既知の系列であり、例えば、Sk $=k^2$;ただし0 $\leq K \leq N$ (N:総キャリア数))であ 20 る。また、最終的に復元される符号系列

【外35】

Χk

はk個の1または-1の符号系列である。

【0022】これにより、受信信号

【外36】

R k

Tされ乗算器7の被乗算端子に入力される。一方、位相 拡散を逆補正ための符号系列

【外37】

 S_k

は受信側において予め既知であり、乗算器7の乗算端子 に入力される。乗算器出力には2値(-1,1)の受信 符号系列

【外38】

X k

が復元され、パラレルーシリアル変換器8を介して取り 出される。

【0023】以上において、正しいOFDM復調のため に受信側で必要となる複素符号系列

【外39】

е-і **5** к

は、何らかのかたちで受信側に伝送されなければならな い。これは複素符号系列

【外40】

e - i 5 x

そのものを伝送するのでなく、それが受信側で再現でき る情報が送られればよい。伝送方法としては、送信側で 入力符号化信号に含ませ、伝送済みのシンボルの中で伝 送しておくか、それとは別の伝送路で伝送してもよい。 【0024】また、図1、図4でそれぞれOFDM変調 および復調を行う際に本発明において使用する符号系列

S k

【外42】

【外41】

 S_k

は時間的に一定(変化しない)ものとしたが、これは、 送信符号系列を送信側で既知の方法でシンボルごとに変 化させ位相拡散を行って伝送し、受信側で逆補正するよ うにしても、なおピークのたちやすい送信符号系列 【外43】

X k

を拡散することが可能である。

【0025】上述例においては、OFDM信号の各キャ リアの振幅が情報をもつ場合を考慮して(実際には、上 述例のBPSKOFDMの場合はもたないが、多値QA MOFDMの場合には情報をもつ)、送信側で複素符号 系列

【外44】

 S_k

はシリアルーパラレル変換5を経てFFT6によりFF 30 を乗算して位相拡散させると、その位相情報が失われる ため、これを回復させるために受信側で複素符号系列 【外45】

 S_k

を乗算した。

【0026】これに対し、BPSKOFDM、QPSK OFDMなどでは各キャリアの位相のみに情報があり、 各キャリアの振幅方向には情報をもたない。そこで、例 えばBPSKOFDMの場合、図1の

40 【外46】

 S_{k}

として、送信符号系列

【外47】

X k

が1の場合N/(2N1)、また、-1の場合N/(2 N2) なる定数をそれぞれ送信符号系列(入力符号化信 号に相当) に乗算するものとする。ここで、Nはシンボ 50 ル長、N₁ , N₂ はそれぞれシンボル中の

(6)

特開平10-84329

10

【外48】



9

の 1 , -1 の個数である。こうすることにより、伝送信号

【外49】

T k

の振幅ピーク時には $1 \, \&mathcal{E} - 1$ あるいは正と負の成分の大きさが等しくなるため、前述例と同様に振幅のピークを 10 抑えて伝送することができる。この場合、受信側では、各キャリアの振幅方向には情報をもたないため、正しい符号復元のためのキャリア拡散の逆補正を必要としない。

[0027]

【発明の効果】以上説明したように、本発明によれば、 衛星放送などの発生電力に制限のある増幅器において も、時間率、場所率を確保する目的をもって、より高い* *動作点で動作させても歪の少ないOFDM伝送をすることが可能となる。

【図面の簡単な説明】

【図1】本発明によるOFDM送信装置の一実施形態を示している。

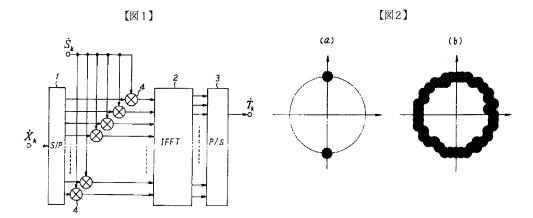
【図2】従来および本発明によるOFDM変調信号の各 キャリアのコンスタレーションの一例を示している。

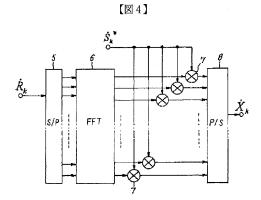
【図3】従来および本発明によるOFDM変調信号の信号波形の一例を示している。

【図4】本発明によるOFDM受信装置の一実施形態を示している。

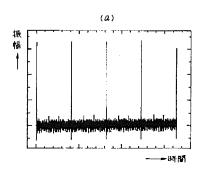
【符号の説明】

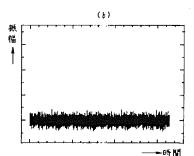
- 1,5 シリアルーパラレル変換器
- 2 逆FFT回路
- 3,8 パラレルーシリアル変換器
- 4, 7 乗算器
- 6 FFT回路













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OUE YUJI

(54) TRANSMISSION METHOD FOR ORTHOGONAL FREQUENCY DIVISION/ MULTIPLEX SIGNAL AND ITS TRANSMITTER AND/RECEIVER

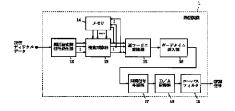
(57) Abstract:

PURPOSE: To effectively remove waveform distortion occurred in a data component on the frequency axis of respective symbols by means of a multipath and the like at the time of transmitting an OFDM signals.

CONSTITUTION: In a transmission device 1, a complex multiplier 13 complex-multiplies a carrier modulation signal group by a complex number signal group which has a previously decided special pattern and in which the phase changes at random. An inverse Fourier transformer 15 executes inverse Fourier transform against the output of the complex multiplier 13, and transforms a digital signal multiplexed on the frequency axis into the OFDM signal of a time axis. A guard time insertion part 16 adds front guard time to

the front parts of the respective symbols of the OFDM signal and rear guard time to rear parts. Data similar to the trailing end part of the corresponding symbol is included in front guard time, and data similar to the front end part of the corresponding symbol is included in rear guard time. The OFDM signals to which front guard time and rear guard time are added are transformed into analog signals and are transmitted to a reception-side. The reception-side executes a processing inverse to a transmission-side and therefore distortion owing to time delay is removed.

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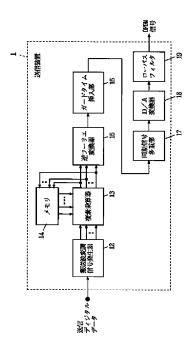
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			最終頁に続く

(54)【発明の名称】 直交周波数分割多重信号の伝送方法ならびにその送信装置および受信装置

(57)【要約】

【課題】 OFDM信号を伝送する際に、マルチパス等によって各シンボルの周波数軸上のデータ成分に生じる 波形歪みを効果的に除去することである。

【解決手段】 送信装置において、複素乗算器13は、 搬送波変調信号群と、予め定められた特定パターンを有 しかつその位相がランダムに変化している複素数信号群 とを複素乗算する。逆フーリエ変換器15は、複素乗算 器13の出力に対して逆フーリエ変換を施し、周波数軸 上で多重されたディジタル信号を、時間軸上のOFDM 信号に変換する。ガードタイム挿入部16は、OFDM 信号の各シンボルの前部に前部ガードタイムを、後部に 後部ガードタイムを付加する。前部ガードタイムには対 応するシンボルの後端部と同じデータが含められ、後部 ガードタイムには対応するシンボルの前端部と同じデー 夕が含められる。前部ガードタイムおよび後部ガードタ イムが付加されたOFDM信号は、アナログ信号に変換 された後、受信側に伝送される。受信側で送信側と逆の 処理を行うことにより、時間遅延による歪みが除去され る。



【特許請求の範囲】

【請求項1】 有線または無線の伝送路を介し、送信側 から受信側に対して、所定長のシンボル毎に直交周波数 分割多重信号を伝送する方法であって、

周波数軸上で互いに直交する複数のキャリアの位相と振 幅とを決定する搬送波変調信号群をシンボル毎に逆フー リエ変換することにより、時間軸上の前記直交周波数分 割多重信号に変換する第1のステップと、

前記直交周波数分割多重信号の各シンボルに対し、その 前部にその後端部と同じデータを含む前部ガードタイム 10 前記逆フーリエ変換手段から出力される前記直交周波数 を付加するとともに、その後部にその前端部と同じデー 夕を含む後部ガードタイムを付加して、前記受信側に送 信する第2のステップとを備える、直交周波数分割多重 信号の伝送方法。

【請求項2】 前記搬送波変調信号群と、基準複素数信 号群とを周波数軸上で複素乗算する第3のステップをさ らに備え、

前記第1のステップは、前記第3のステップで得られた 複素乗算結果を、前記直交周波数分割多重信号に変換す 法。

【請求項3】 前記第3のステップは、前記搬送波変調 信号群の各シンボルについて、その一定シンボル前に複 素乗算した結果を、前記基準複素数信号群として各前記 搬送波変調信号群に複素乗算する、請求項2に記載の直 交周波数分割多重信号の伝送方法。

【請求項4】 予め定められた特定パターンを有し、か つ各信号の位相がランダムに変化している複素数信号群 をシンボル毎に発生する第4のステップをさらに備え、

前記第3のステップは、前記搬送波変調信号群の各シン 30 ボルについて、前記第4のステップで得られた複素数信 号群を、前記基準複素数信号群として使用し、

前記第1のステップは、常時は前記第3のステップで得 られた複素乗算結果を前記直交周波数分割多重信号に変 換し、定期的に前記基準複素数信号群を前記直交周波数 分割多重信号に変換する、請求項2に記載の直交周波数 分割多重信号の伝送方法。

【請求項5】 所定長のシンボル毎に前記送信側から送 信されてきた前記直交周波数分割多重信号を、前記搬送 波変調信号群に対応する受信搬送波変調信号群に変換す 40 る第5のステップと、

前記第5のステップで得られた受信信号群を、所定の基 準複素数信号群により、周波数軸上で複素除算する第6 のステップとを備える、請求項2に記載の直交周波数分 割多重信号の伝送方法。

【請求項6】 有線または無線の伝送路を介し、受信側 に、所定長のシンボル毎に直交周波数分割多重信号を送 信する装置であって、

基準複素数信号群を記憶するメモリ手段と、

周波数軸上で互いに直交する複数のキャリアの位相と振 50 受信搬送波変調信号群を、受信基準複素数信号群として

幅とを決定する搬送波変調信号群と、前記メモリ手段に 記憶された前記基準複素数信号群とを周波数軸上で複素 乗算し、送信搬送波変調信号群を出力する複素乗算手段 ٤.

前記複素乗算手段から出力される送信搬送波変調信号群 に対して、各シンボル毎に逆フーリエ演算を施すことに より、当該送信搬送波変調信号群を、時間軸上の前記直 交周波数分割多重信号に変換する逆フーリエ変換手段

分割多重信号の各シンボルに対し、その前部にその後端 部と同じデータを含む前部ガードタイムを付加するとと もに、その後部にその前端部と同じデータを含む後部ガ ードタイムを付加するガードタイム付加手段と、

前記前部ガードタイムおよび前記後部ガードタイムの付 加された前記直交周波数分割多重信号を、各シンボル毎 に前記受信側に送信する送信手段とを備える、直交周波 数分割多重信号の送信装置。

【請求項7】 前記メモリ手段は、前記複素乗算手段の る、請求項1に記載の直交周波数分割多重信号の伝送方 20 一定シンボル前の複素乗算結果を、前記基準複素数信号 群として記憶する、請求項6に記載の直交周波数分割多 重信号の送信装置。

> 【請求項8】 前記メモリ手段は、予め定められた複素 数信号群を、前記基準複素数信号群として記憶し、

> 前記複素乗算手段は、前記搬送波変調信号群と、前記メ モリ手段に記憶された前記基準複素数信号群とを周波数 軸上で複素乗算して出力し、

前記逆フーリエ変換手段は、常時はシンボル毎に前記複 素乗算手段から出力された複素乗算結果を前記直交周波 数分割多重信号に変換し、定期的に前記メモリ手段から 出力された前記基準複素数信号群を前記直交周波数分割 多重信号に変換する、請求項6に記載の直交周波数分割 多重信号の送信装置。

【請求項9】 前記メモリ手段は、前記基準複素数信号 群として疑似雑音信号を発生する疑似雑音信号発生手段 の出力を保持していることを特徴とする、請求項8に記 載の直交周波数分割多重信号の送信装置。

【請求項10】 前記メモリ手段は、前記基準複素数信 号群として周波数掃引信号を発生する周波数掃引信号発 生手段の出力を保持していることを特徴とする、請求項 8に記載の直交周波数分割多重信号の送信装置。

【請求項11】 有線または無線の伝送路を介し、送信 側から所定長のシンボル毎に送信されてくる直交周波数 分割多重信号を受信する装置であって、

時間軸上の前記直交周波数分割多重信号に対して、シン ボル毎にフーリエ変換演算を施すことにより、当該直交 周波数分割多重信号を、周波数軸上の受信搬送波変調信 号群に変換するフーリエ変換手段と、

前記フーリエ変換手段から一定シンボル毎に出力された

-202-

3

記憶するメモリ手段と、

前記フーリエ変換手段から出力された受信搬送波変調信 号群を、前記メモリ手段に記憶された受信基準複素数信 号群により、周波数軸上で複素除算する複素除算手段と を備える、直交周波数分割多重信号の受信装置。

【請求項12】 有線または無線の伝送路を介し、送信 側から受信側に対して、所定長のシンボル毎に直交周波 数分割多重信号を伝送する方法であって、

周波数軸上で互いに直交する複数のキャリアの位相と振幅とを決定するための搬送波変調信号群をシンボル毎に 10 生成する第1のステップと、

予め定められた特定パターンを有し、かつ各信号の位相 がランダムに変化している複素数信号群を発生する第2 のステップと、

前記搬送波変調信号群と前記複素数信号群とをシンボル 毎に周波数軸上で複素乗算することにより、当該搬送波 変調信号群の各信号の位相をランダム化する第3のステップと、

常時は前記第3のステップで各信号の位相がランダム化された搬送波変調信号群をシンボル毎に逆フーリエ変換 20 して時間軸上の前記直交周波数分割多重信号に変換し、定期的に前記複素数信号群を逆フーリエ変換して前記直交周波数分割多重信号に変換し、それぞれを前記受信側に送信する第4のステップとを備える、直交周波数分割多重信号の伝送方法。

【発明の詳細な説明】

[0001]

【発明の属する技術分野】本発明は、直交周波数分割多重(Orthogonal FrequencyDivision Multiplexing;以下、OFD 30 Mと称す)伝送方法に関し、より特定的には、有線または無線の伝送路を介し、送信側と受信側との間で、所定長のシンボルと当該シンボル間に配置された所定長のガードタイムとを含む直交周波数分割多重信号を用いてデータを伝送する方法に関する。

[0002]

【従来の技術】周知のごとく、OFDM伝送方式は、符号化したデータを分割して、数百以上の搬送波に振り分け、これを多重して伝送する方式である。近年、移動体向けディジタル音声放送や、地上ディジタルテレビ放送 40 等において、OFDM信号を用いた通信が着目されている。なぜならば、OFDM信号は、多量のデータの高速伝送が可能で、波形等価器なしでも反射波による特性劣化が少なく、その信号波形がランダム雑音に近い形となるので、他のサービスに混信妨害を与えにくい等の特質を有しているからである。

【0003】 このようなOFDM信号を用いた伝送方式は、1993年10月1日付け発行のNIKKEI E LECTRONICS BOOKS「データ圧縮とディジタル変調」の第207~222頁において、郵政省、

通信総合研究所の福地一により書かれた「数百以上の搬送波を使うOFDMディジタル放送の移動受信に向く」 に開示されている。

【0004】図13は上記先行文献に開示された従来のOFDM信号の送信装置の構成を示すプロック回路図であり、図14は図13の送信装置から送信されるOFDM信号の構成を示す図である。図13において、送信装置5は、直並列変換器52と、逆フーリエ変換器53と、並直列変換器54と、D/A変換器55と、ローパスフィルタ56とを備える。なお、図14において、(a)はOFDM信号の直接波を示し、(b)はOFDM信号の反射波を示し、(c)はOFDM信号の合成波を示し、(d)は時間窓Wを示している。

【0005】送信装置5の直並列変換器52には、入力 シンボル列が供給されている。入力シンボル列は、ディ ジタル変調された送信データであり、1 伝送シンボル中 には複数のデータ値が含まれている。なお、ディジタル 変調方式としては、QPSK(quadriphase phase shift keying) 変調や、1 6QAM (quadrature amplitude modulation)等が採用される。直並列変換 器52は、入力シンボル列を、1シンボル毎に、直並列 変換して、より低速な複数のシンボル列にする。ここで の並列度は、逆フーリエ変換回路53で使用する複数の 搬送波(相互に位相が直交している)の数(数十~数 千、たとえば512)と同じになる。このような操作に より、直並列変換器52は、逆フーリエ変換回路53で 使用する複数の搬送波のそれぞれの振幅および位相を決 定するための搬送波変調信号群を出力する。

80 【0006】逆フーリエ変換回路53は、搬送波変調信 号群を、1シンボル毎に、周波数軸上に並ぶ各搬送波に 割り当て(これによって、1シンボル分のデータが周波 数軸上で多重された信号となる)、これらに対して一括 的に逆フーリエ変換を施すことにより、時間軸上の多重 信号(この段階では、並列のディジタル信号である)に 変換する。

【0007】並直列変換器54は、時間軸上の多重信号を並直列変換することにより、離散的なOFDM信号を生成する。D/A変換回路55は、離散的なOFDM信号を、アナログのOFDMベースバンド信号に変換する。ローパスフィルタ56は、エイリアシングによるチャネル間干渉が生じないようにするため、OFDMベースバンド信号に帯域制限をかける。

【0008】上記のような一連の操作の結果、送信装置 5は、伝送路に対し、図14に示すようなガードタイム Gm とシンボルSm とを含むOFDM信号を出力する。 図示しない復調装置は、伝送路を介して受信したOFD M信号に対して変調装置 5と逆の信号処理を行い、入力シンボル列と同じ出力シンボル列を再生する。

50 [0009]

【発明が解決しようとする課題】ところで、伝送路上で は、いわゆるマルチパスが発生する。このため、受信装 置側では、送信装置から送信されてきたOFDM信号の 直接波と、直接波から時間遅延した反射波とを重なって 受信する。シンボルSmを例にとると、直接波(図14 (a) 参照) にマルチパスによる反射波(図14(b) 参照) が重なった場合、合成波(図14(c)参照)の シンボルSmの前端部に反射波のガードタイムGmとの 干渉部 am が生じ、ガードタイム Gm の前端部に反射波 のシンボルSm-1 との干渉部βm が生じる。このとき、 干渉部 ß m は、時間窓Wからはずれているため、シンボ ルSmのフーリエ変換には影響を及ぼさない。しかしな がら、干渉部 αm は、時間窓W内に生じ、かつガードタ イムGm のデータ成分が「O」であるため、フーリエ変 換後の各シンボルSm の周波数軸上のデータ成分に波形 歪みを生じるという第1の問題点があった。

【0010】また、伝送路の遅延特性や、送信側のD/ A変換器および受信側のA/D変換器のクロックが一致 していないことに起因してサンプリングのタイミングに ずれが生じる等の理由から、送信装置から受信装置に到 20 達するまでの間に、OFDM信号に時間遅延が発生す る。このため、受信装置では、時間窓Wを時間軸上で調 整する必要があるという第2の問題点もあった。

【0011】また、直並列変換器52から出力される搬 送波変調信号群は、その位相が相互に異なっているだけ でなく、その位相がすべて同一の場合もありうる。例え ば、ディジタル音声放送では無音状態を1シンボル期間 を超えて送信する場合に、地上ディジタルテレビ放送で は一色の映像を1シンボル期間を超えて送信する場合 に、搬送波変調信号群の位相がすべて同一になる。ま た、有音状態を送信する場合や、多色の映像を送信する 場合においても、QPSK変調や、16QAM等のよう なディジタル変調方式では、位相の異なる信号点の配点 数が限られるため、搬送波変調信号群の位相がすべて同 一になりやすい。

【0012】このように、搬送波変調信号群の位相がす べて同一になった場合、この搬送波変調信号群を逆フー リエ変換すると、時間軸上で各搬送波の節が一致し、加 算増加箇所が時間軸上で一箇所に集中するため、時間軸 上のOFDM信号の信号波形がインパルス状になり、電 40 力集中が生じる。この様子を図15に示す。

【0013】図15 (a) は、相互に直交するn本の搬 送波をそれぞれ変調するn個の搬送波変調信号群の複素 平面上での位相がすべて同一の場合を示している。図1 5 (b) は、図15 (a) のn個の搬送波変調信号群で 変調されたn本の搬送波を時間軸上で多重した状態を示 している。このように搬送波変調信号群の位相がすべて 同一の場合には、OFDM信号は、インパルス状の波形 信号になる。なお、図15(c)は、相互に直交するn 本の搬送波をそれぞれ変調するn個の搬送波変調信号群 50 間軸上に並ぶ1シンボル区間内のすべてのデータ成分を

の複素平面上での位相がランダムな場合を示している。 また、図15 (d) は、図15 (c) のn個の搬送波変 調信号群で変調されたn本の搬送波を時間軸上で多重し た状態を示している。このように、搬送波変調信号群の 位相がすべて異なる場合には、OFDM信号は、時間軸 上に平均的に拡散され、ランダム状の波形信号になる。

6

【0014】上記のように、搬送波変調信号群の位相が すべて同一になった場合、OFDM信号がインパルス状 になり、最大電力が極端に大きくなるため、OFDM信 10 号は、送受信装置や伝送路に含まれる中継増幅器(衛星 や CAT V など) 等の非線形性の影響を受けやすくなる という第3の問題点もあった。この場合、OFDM信号 がインパルス状になっても、非線形性の影響を与えない ように、送受信装置や中継増幅器等のダイナミックレン ジを大きくすることも考えられるが、送受信装置や中継 増幅器等が高価になるという別の問題が発生する。

【0015】それ故に、本発明の目的は、マルチパスに より反射波が直接波に重なった場合でも、フーリエ変換 後の各シンボルの周波数軸上のデータ成分に波形歪みを 生じないOFDM信号の伝送方法ならびにその送信装置 および受信装置を提供することである。本発明の他の目 的は、送信側から受信側に到達するまでの間に、OFD M信号に時間遅延が発生しても、時間窓の時間軸上での 調整が容易なOFDM信号の伝送方法ならびにその送信 装置および受信装置を提供することである。本発明のさ らに他の目的は、安価な構成で、OFDM信号に対する 非線形性の影響を軽減したOFDM信号の伝送方法なら びにその送信装置および受信装置を提供することであ

[0016]

【課題を解決するための手段および発明の効果】本発明 の第1の局面は、有線または無線の伝送路を介し、送信 側から受信側に対して、所定長のシンボル毎に直交周波 数分割多重信号を伝送する方法に向けられており、周波 数軸上で互いに直交する複数のキャリアの位相と振幅と を決定する搬送波変調信号群をシンボル毎に逆フーリエ 変換することにより、時間軸上の直交周波数分割多重信 号に変換する第1のステップと、直交周波数分割多重信 号の各シンボルに対し、その前部にその後端部と同じデ ータを含む前部ガードタイムを付加するとともに、その 後部にその前端部と同じデータを含む後部ガードタイム を付加して、受信側に送信する第2のステップとを備え ている。

【0017】上記のように、第1の局面では、OFDM 信号の各シンボルを送信する際に、各シンボルの前部お よび後部に、そのシンボルの一部と同じデータを含む前 部ガードタイムおよび後部ガードタイムを付加するよう にしているので、受信側では、フーリエ変換時における 時間窓が受信信号のシンボル区間から多少ずれても、時

再生することができる。従って、送信側から受信側に到 達するまでの間に、OFDM信号に時間遅延が発生して も、時間窓をシンボル区間に正確に一致させる必要がな くなり、時間窓の時間軸上での調整が容易になる。ま た、マルチパスにより直接波のシンボル区間と反射波の ガードタイムとが重なっても、受信側でフーリエ変換後 の周波数軸上に現れる各データ成分の振幅位相歪みは、 各シンボル間ですべて一様なものとなる。したがって、 簡単な演算処理(乗算、加算等)によって、受信側での それらの波形歪みを除去することが可能となる。

【0018】上記第1の局面において、好ましい実施形 態では、搬送波変調信号群と基準複素数信号群とを周波 数軸上で複素乗算し、この複素乗算結果をOFDM信号 に変換して、受信側に伝送するようにしている。また、 受信側では、送信側から送信されてきたOFDM信号を 受信搬送波変調信号群に変換し、この受信搬送波変調信 号群を、基準複素数信号群により、周波数軸上で複素除 算するようにしている。これによって、送信側と受信側 との間でOFDM信号に時間遅延が発生しても、受信側 20 で時間遅延の影響のない復調データを得ることができ る。

【0019】なお、搬送波変調信号群に複素乗算される 基準複素数信号群としては、搬送波変調信号群の各シン ボルについて、その一定シンボル前に複素乗算した結果 を用いても良い。

【0020】また、予め定められた特定パターンを有 し、かつ各信号の位相がランダムに変化している複素数 信号群を、基準複素数信号群として用いても良い。ただ し、この場合、常時は第3のステップで得られた複素乗 30 算結果がOFDM信号に変換され、定期的に基準複素数 信号群がOFDM信号に変換される。これによって、搬 送波変調信号群の各信号の絶対基準位相がランダムな値 になり、逆フーリエ変換によって得られたOFDM信号 に電力の時間集中がおこるのを抑制できる。従って、送 信装置、受信装置および伝送路のダイナミックレンジを 大きくする必要がなく、安価な構成で、送受信器や中継 増幅器等の非線形性がOFDM信号に与える影響を軽減 することができる。

【0021】本発明の第2の局面は、有線または無線の 40 とを備えている。 伝送路を介し、受信側に、所定長のシンボル毎に直交周 波数分割多重信号を送信する装置に向けられており、基 準複素数信号群を記憶するメモリ手段と、周波数軸上で 互いに直交する複数のキャリアの位相と振幅とを決定す る搬送波変調信号群と、メモリ手段に記憶された基準複 素数信号群とを周波数軸上で複素乗算し、送信搬送波変 調信号群を出力する複素乗算手段と、複素乗算手段から 出力される送信搬送波変調信号群に対して、各シンボル 毎に逆フーリエ演算を施すことにより、当該送信搬送波

換する逆フーリエ変換手段と、逆フーリエ変換手段から 出力される直交周波数分割多重信号の各シンボルに対 し、その前部にその後端部と同じデータを含む前部ガー ドタイムを付加するとともに、その後部にその前端部と

8

同じデータを含む後部ガードタイムを付加するガードタ イム付加手段と、前部ガードタイムおよび後部ガードタ イムの付加された直交周波数分割多重信号を、各シンボ ル毎に受信側に送信する送信手段とを備えている。

【0022】上記第2の局面において、好ましい実施形 1シンボル区間の周波数軸上のデータ成分から、容易に 10 態では、メモリ手段は、複素乗算手段の一定シンボル前 の複素乗算結果を、基準複素数信号群として記憶してい

> 【0023】上記第2の局面において、他の好ましい実 施形態では、メモリ手段は、予め定められた複素数信号 群を、基準複素数信号群として記憶する。また、複素乗 算手段は、搬送波変調信号群と、メモリ手段に記憶され た基準複素数信号群とを周波数軸上で複素乗算して出力 する。さらに、逆フーリエ変換手段は、常時はシンボル 毎に複素乗算手段から出力された複素乗算結果を直交周 波数分割多重信号に変換し、定期的にメモリ手段から出 力された基準複素数信号群を直交周波数分割多重信号に 変換する。

【0024】上記第2の局面において、メモリ手段は、 基準複素数信号群として、疑似雑音信号を発生する疑似 雑音信号発生手段の出力を保持しても良いし、周波数掃 引信号を発生する周波数掃引信号発生手段の出力を保持 しても良い。

【0025】本発明の第3の局面は、有線または無線の 伝送路を介し、送信側から所定長のシンボル毎に送信さ れてくる直交周波数分割多重信号を受信する装置に向け られており、時間軸上の直交周波数分割多重信号に対し て、シンボル毎にフーリエ変換演算を施すことにより、 当該直交周波数分割多重信号を、周波数軸上の受信搬送 波変調信号群に変換するフーリエ変換手段と、フーリエ 変換手段から一定シンボル毎に出力された受信搬送波変 調信号群を、受信基準複素数信号群として記憶するメモ リ手段と、フーリエ変換手段から出力された受信搬送波 変調信号群を、メモリ手段に記憶された受信基準複素数 信号群により、周波数軸上で複素除算する複素除算手段

【0026】本発明の第4の局面は、有線または無線の 伝送路を介し、送信側から受信側に対して、所定長のシ ンボル毎に直交周波数分割多重信号を伝送する方法に向 けられており、周波数軸上で互いに直交する複数のキャ リアの位相と振幅とを決定するための搬送波変調信号群 をシンボル毎に生成する第1のステップと、予め定めら れた特定パターンを有し、かつ各信号の位相がランダム に変化している複素数信号群を発生する第2のステップ と、搬送波変調信号群と複素数信号群とをシンボル毎に 変調信号群を、時間軸上の直交周波数分割多重信号に変 50 周波数軸上で複素乗算することにより、当該搬送波変調 20

信号群の各信号の位相をランダム化する第3のステップ と、常時は第3のステップで各信号の位相がランダム化 された搬送波変調信号群をシンボル毎に逆フーリエ変換 して時間軸上の直交周波数分割多重信号に変換し、定期 的に複素数信号群を逆フーリエ変換して直交周波数分割 多重信号に変換し、それぞれを受信側に送信する第4の ステップとを備えている。

[0027]

【発明の実施の形態】以下、本発明の実施形態に係る〇 置について、図面を参照しながら説明する。

【0028】図1は本発明の第1の実施形態の送信装置 を示すブロック図であり、図2は本発明の第1の実施形 態の受信装置の構成を示すブロック図であり、図3は本 発明で用いるOFDM信号の構成の一例を示す図であ る。なお、図3において、(a)はOFDM信号の直接 波を示し、(b) はOFDM信号の反射波を示し、 (c) は時間遅延が生じた場合のOFDM信号の直接波 を示し、(d)は時間遅延が生じた場合のOFDM信号 の反射波を示し、(e)は時間窓Wを示している。

【0029】図1の送信装置1と、図2の受信装置2と は、同軸ケーブルや、光ファイバケーブル等の伝送路 (図示せず) で接続されている。このような送信装置1 および受信装置2は、たとえばディジタルCATVシス テムにおいて用いられる。送信装置1は、OFDM信号 を用い、受信装置2に対して、たとえばテレビの多チャ ンネル分の映像データを伝送するように構成されてい

【0030】図1において、送信装置1は、搬送波変調 信号発生器12と、複素乗算器13と、メモリ14と、 逆フーリエ変換器15と、ガードタイム挿入部16と、 同期信号多重部17と、D/A変換器18と、ローパス フィルタ19とを備えている。

【0031】送信装置1の搬送波変調信号発生器12に は、受信装置2に送信すべき送信ディジタルデータ(ビ ットストリーム信号) が入力されている。搬送波変調信 号発生器12は、入力された送信ディジタルデータを、 ディジタル変調すると共に、1シンボル区間毎に直並列 変換し、相互に直交するn本(n=数十~数千、たとえ ば512)の搬送波を変調するためのn個の搬送波変調 信号を含む搬送波変調信号群に変換する。なお、ディジ タル変調方式としては、QPSK変調や、16QAM等 が採用される。この段階での搬送波変調信号群は、従来 の直並列変換器52 (図13参照) から出力される搬送 波変調信号群と同様である。搬送波変調信号発生器12 から出力される搬送波変調信号群は、複素乗算器13に 与えられる。メモリ14は、複素乗算器13から出力さ れる搬送波変調信号群D' m を1シンボル分記憶するこ とができる。また、メモリ14は、複素乗算器13に搬

ている1シンボル前の搬送波変調信号群D'm-1 を、所 定の基準複素数信号群として、複素乗算器13に出力す る。複素乗算器13は、入力された送信信号群Dmと、 1シンボル前の基準複素数信号群D'm-1 とを、周波数 軸上で、複素乗算することにより、搬送波変調信号群

10

 $D'm (D'm = Dm \times D'm-1)$ を作成する。

【0032】より具体的に説明すると、複素乗算器13 に入力された搬送波変調信号群(n個の搬送波変調信号 FDM信号の伝送方法ならびその送信装置および受信装 10 を含む)のうち、k(k=1, 2, ..., n)番目の搬送 波変調信号の実数部をDm [k] realとし、その虚 数部をDm [k] imagとし、メモリ14に記憶した k番目の搬送波変調信号の実数部をD'm-1 [k] re alとし、その虚数部をD'm-1 [k] imagとした 場合、複素乗算器13は、各搬送波変調信号の実数部お よび虚数部それぞれについて、乗算処理を行い、

> D'm [k] real=Dm [k] real \times D'm-1 [k] real

 $D'm[k]imag=Dm[k]imag\times D'm-1$ [k] imag

を出力する。メモリ14は、複素乗算器13から出力さ れた実数および虚数の搬送波変調信号D'n (D'n [k] realおよびD'm [k] imagを含む)を 記憶保持する。図4に示すように、メモリ14および複 素乗算器13は、上記のような動作を繰り返し実行す

【0033】逆フーリエ変換器15は、複素乗算器13 から出力される搬送波変調信号群D' m 中のそれぞれの 搬送波変調信号を、シンボル区間毎に、順次周波数軸上 に並ぶ各搬送波に割り当て、これらに対して一括的に逆 フーリエ変換を施し、さらに並直列変換を行うことによ り、周波数軸上で各データ成分が多重された搬送波変調 信号群を、時間軸上で各データ成分が多重されたOFD M信号D'mtに変換する。

【0034】ガードタイム挿入部16は、逆フーリエ変 換器15から出力されるディジタルのOFDM信号D' mtを、各シンボル区間毎に、一旦、内部のバッファに蓄 える。次に、ガードタイム挿入回路16は、各シンボル Sm に対して、その前部に前部ガードタイムGhmを、そ の後部に後部ガードタイムGemを、それぞれ付加する (図3参照)。なお、前部ガードタイムGhmの時間長t g1および後部ガードタイムGemの時間長tg2は、そ れぞれ伝送路で発生するマルチパスによる直接波と間接 波との時間差および送信装置1のD/A変換器18と受 信装置2のA/D変換器22との間のサンプリングのず れによる時間遅延を考慮して定められる。また、前部ガ ードタイムGhmには、対応するシンボルSm の後端部S emと同じデータD'emt が含められ、後部ガードタイム Gemには、対応するシンボルSm の前端部Shmと同じデ 送波変調信号群Dm が入力されたときに、内部に記憶し 50 ーpD hmt が含められる。これにより、実質的なシン

11

ボル長が、 t g 1 + t s + t g 2 に延長されることになる。 ガードタイム挿入部 1 6 は、前部ガードタイムG n、シンボル n 、後部ガードタイム n の n を順次出力する。

【0035】同期信号多重部17は、シンボルの区切りを示すため、シンボル毎に、同期信号を、ガードタイムの付加されたOFDM信号に時間軸上で多重し、D/A変換器18に出力する。同期信号は、たとえば、図5(a)に示すようにOFDM信号に対し、周期的に既知の無変調搬送波と抑圧信号等とから構成する。

【0036】D/A変換器18は、同期信号多重部17から出力される、ガードタイムおよび同期信号が付加されたディジタルデータのOFDM信号を、アナログのOFDMベースバンド信号に変換する。ローバスフィルタ19は、エイリアシングによるチャネル間干渉が生じないようにするため、OFDMベースバンド信号に帯域制限をかける。

【0037】上記のような一連の操作の結果、送信装置 1は、伝送路に対して、ガードタイムおよび同期信号を 含むOFDM信号を出力する。

【0038】図2において、受信装置2は、ローパスフィルタ21と、A/D変換器22と、エンベロープ検波器23と、同期再生部24と、フーリエ変換器25と、メモリ26と、複素除算器27と、送信データ再生器28とを備えている。

【0039】ローパスフィルタ21は、伝送路を介して 受信した〇FDM信号から、不要な高周波域のスペクト ル成分を除去する。

【0040】 ここで、マルチパスや伝送路の遅延特性等による時間遅延 Δ t を考慮し、受信装置2において受信 30したOFDM信号を2D'mtとする。なお、2は、

 $Z = e \times p j 2 \pi f c \Delta t$

であり、信号の遅延分を表している。

【0041】A/D変換器22は、アナログのOFDM信号の前部ガードタイムGhm、シンボルSm、後部ガードタイムGemにそれぞれ含まれるデータZD' ent、ZD' mt、ZD' hmt を、ディジタルのOFDM信号に変換する。

【0042】エンベロープ検波器23は、OFDM信号をエンベロープ検波することにより、図5(b)に示す 40 エンベロープ検波信号を、シンボル毎に出力する。同期再生部24は、エンベロープ検波器23から出力されたエンベロープ検波信号に基づいて、図5(c)に示す基準タイミング信号を、シンボル毎に出力する。この基準タイミング信号は、フーリエ変換器25およびメモリ26に入力される。

【0043】フーリエ変換器25は、基準タイミング信 m+1 を、メモリ26に保持されてい 号に同期して、A/D変換器22から出力されるOFD 複素除算することにより、データの M信号を、シンボル長tsと同じ長さの時間窓W(図3 ルし、遅延の無い元の搬送波変調 (e) 参照)を介して覗くことにより、各シンボルの必 50 る。すなわち、複素除算器27が、

12

要なデータ部分だけを抽出する。また、フーリエ変換器 25は、この抽出されたデータ部分に対して、フーリエ 変換演算を施すことにより、時間軸上のOFDM信号 を、周波数軸上の受信搬送波変調信号群に変換する。

【0044】メモリ26は、フーリエ変換器25から出力される受信搬送波変調信号群を、1シンボル分記憶する。ここで、送信装置1からデータD'mが送られてきた場合、メモリ26には、それに対応するデータとして、データ<math>ZD'mが格納されることになる。データZD'mにマルチパスや伝送路等によって生じた時間遅延分Zを加えたものである。すなわち、ZD'm=D'm×expj 2π fc Δ t

となる。メモリ26は、基準タイミング信号に同期して、データZD'mを複素除算器27に出力する。複素除算器27は、同期を確立した上で、フーリエ変換器25から出力されるシンボルSm+1のデータZD'mによって複素除算する。すなわち、複素除算器27は、

ZD'm+1/ZD'm=D'm+1/D'm=Dm+1 20 の演算を行う。図6に示すように、フーリエ変換器2 5、メモリ26および複素除算器27は、上記のような 動作を繰り返し実行する。

【0045】前述したように、マルチパスに起因して、図3(a)に示す直接波と図3(b)に示す反射波との間に、相対的な時間遅延が生じる。また、送信装置1のD/A変換器18と受信装置2のA/D変換器22とにおけるサンプリングタイミングが異なることに起因して、直接波および反射波にそれぞれ固有の時間遅延が発生する(図3(c)および図3(d)参照)。フーリエ変換器25において、基準タイミング信号は、これらの時間遅延を考慮していないため、図3(e)に示すように、時間軸上における受信側の時間窓Wの位置は、受信信号のシンボル区間からずれている。

【0046】しかしながら、受信側のフーリエ変換器2 5で、時間窓Wが正確なシンボル区間からずれていて も、前部ガードタイムGhmおよび後部ガードタイムGem には、それぞれデータZD'emt およびZD'hmt が含 まれているため、時間窓Wを介して覗いたデータには、 1シンボル区間に本来含まれるべき時間軸上のすべての データZD'mtが含まれていることになる。このため、 この時間遅延および反射波の重なりは、周波数軸上にお いて各データ成分毎に一様な振幅位相歪みとなって現れ る。また、時間遅延および反射波の特性が一様であれ ば、各シンボル区間毎に振幅位相歪みの大きさは等しく なる。本実施形態では、複素除算器27は、フーリエ変 換器 2 5 から出力されたシンボル Sm+1 のデータ ZD' m+1 を、メモリ26に保持されているデータZD'm で 複素除算することにより、データの遅延分2をキャンセ ルし、遅延の無い元の搬送波変調信号群Dm+1 を得てい 1.3

ZD'm+1/ZD'm=D'm+1/D'm=Dm+1の演算を行うことにより、振幅位相歪みは打ち消される こととなり、各シンボルについて、位相・振幅歪みのな いデータDm が得られる。

【0047】以上のように、上記実施形態では、各シン ボルの前後にそのシンボルの後端部および前端部と同じ データを含むガードタイムを付加して送信するようにし ているので、受信側では、時間窓W内に直接波および反 射波の両方について、時間軸上に並ぶ1シンボル区間内 のすべてのデータ成分を再生することができる。このた 10 め、マルチパスにより反射波が直接波に重なり、直接波 のシンポル区間と反射波のガードタイムとが重なって も、フーリエ変換後に周波数軸上に現れる各データ成分 の振幅位相歪みは、すべて一様なものとなる。したがっ て、送信側および受信側で適当な演算処理(乗算、除 算)を実行することで、1シンボル区間の周波数軸上の 受信搬送波変調信号群から、容易に波形歪みを除去する ことができる。

【0048】また、上記実施形態では、送信側と受信側 数軸上で受信搬送波変調信号群を所定の基準複素数信号 群で複素乗算、複素除算することにより、時間遅延のな い復調データを得ることができる。その結果、時間窓を シンボル区間に正確に一致させる必要がなくなる。

【0049】送信データ再生器28は、複素除算器27 から出力された受信搬送波変調信号群Dm の信号点を複 素平面上にマッピングし、信号点を判定することによ り、送信装置1の送信ディジタル信号群と同値の受信デ ィジタル信号群を得る。前述したように、受信搬送波変 調信号群Dm からは、位相歪みや振幅歪みが除去されて 30 いる。したがって、送信データ再生器28は、複素平面 上へのマッピング位置から、正確かつ容易に元のデータ を判定することができる。

【0050】なお、本願発明者は、計算機を使用して、 マルチパスによる遅延波の影響と、時間軸遅延の影響と について、従来のシステムと本実施形態のシステムとを 比較するシミュレーションを行った。なお、このシミュ レーションは、キャリア数が512本、256番目のキ ャリアのデータだけが振幅「1」,位相「0」、他のキ ャリアのデータはすべて「0」を条件として実施され 40 た。

【0051】図7は、マルチパスによる遅延波の影響に ついて、従来のシステムと本実施形態のシステムとを比 較したシミュレーション結果を示す図である。なお、図 7において、(a), (b), (c), (d) は、それ ぞれ、従来のシステムにおける直接波、間接波、合成 波、合成波をフーリエ演算することにより周波数軸上の 信号に変換した場合のデータ歪みを示している。また、 図7において、(e), (f), (g), (h)は、そ れぞれ、本実施形態のシステムにおける直接波、間接 50 定パターンを有し、かつ各信号の位相が相互にランダム

14

波、合成波、合成波をフーリエ演算することにより周波 数軸上の信号に変換した場合のデータ歪みを示してい

【0052】従来のシステムでは、ガードタイムにいか なるデータも挿入されていないため (図7 (b) の α 1 参照)、合成波の時間窓W中に干渉部α2が発生してい る(図7(c)参照)。したがって、合成波を時間窓W でフーリエ演算することにより周波数軸上の信号に変換 すると、図7(d)に示すように、256番目のキャリ アのデータのスペクトルが拡がるとともに、他のキャリ アの本来「0」であったはずのデータに歪みが生じる。 したがって、送信データ再生器28で誤判定が起き易く なる。さらに、他のキャリアについても、送信データ再 生器28で誤判定が起き易くなる。一方、本実施形態の システムでは、ガードタイムにデータが挿入されている ので、他のキャリアのデータに影響を及ぼさない。

【0053】図8は、伝送路等による時間遅延の影響に ついて、従来のシステムと本実施形態のシステムとを比 較したシミュレーション結果を示す図である。図8にお との間で、OFDM信号に時間遅延が発生しても、周波 20 いて、(a) は256番目のキャリアのデータだけが振 幅「1」、位相「0」の場合のスペクトルを示し、

> (b) は (a) のデータを逆フーリエ演算することによ り時間軸上の信号に変換した場合の信号波形を示してい る。また、図8において、(c), (d)は、それぞ れ、従来のシステムにおける時間遅延を生じた合成波、 合成波をフーリエ演算することにより周波数軸上の信号 に変換した場合のデータ歪みを示している。また、図8 において、(e), (f)は、それぞれ、本実施形態の システムにおける時間遅延を生じた合成波、合成波をフ ーリエ演算することにより周波数軸上の信号に変換した 場合のデータ歪みを示している。

> 【0054】従来のシステムでは、ガードタイムにいか なるデータも挿入されていないため(図8(c)の α 1 参照)、図7 (c) の場合と同様に、合成波の時間窓W 中に干渉部 α 2が発生する。したがって、図8(d)に 示すように、合成波を時間窓Wでフーリエ演算すること により周波数軸上の信号に変換すると、256番目のキ ャリアのデータのスペクトルが拡がるとともに、他のキ ャリアの本来「0」であったはずのデータに歪みが生じ る。したがって、他のキャリアについても、送信データ 再生器28で誤判定が起こり易くなる。一方、本実施形 態では、ガードタイムにデータが挿入されているので、 他のキャリアのデータに影響を及ぼさない。

【0055】図9は、本発明の第2の実施形態の送信装 置の構成を示すプロック図である。なお、図9の送信装 置3において、図1の送信装置1の構成と対応する部分 には、同一の参照番号を付し、その説明を省略する。図 9の実施形態で注目すべき点は、メモリ14が、特定パ ターン発生器 3 1 の出力、すなわち、予め定められた特 15

に変化している複素数信号群D0 を保持していることである。このような複素数信号群D0 は、たとえば $0\sim1$ の間のレベルの疑似ランダム信号を発生する PN系列疑似ランダム信号発生器と、この疑似ランダム信号と 2π とを乗算する乗算器とを備え、位相が0から 2π 間でランダムな値を持ち、かつ振幅が1の単位ベクトル信号を生成する疑似雑音信号発生器により形成することができる。また、このような複素数信号群は、位相が0から 2π までのランダムな値を持った既知の周波数掃引信号を発生する、周波数掃引信号発生器により形成することも 10できる。

【0056】複素乗算器13は、各シンボル区間のデータDm が入力される毎に、データDm とデータD0 とを周波数軸上で複素乗算して、データD'm (D'm=Dm×D0) を作成し、搬送波変調信号群中の各搬送波変調信号の相互の位相を特定パターンにランダム化する。

【0057】図10は、複素乗算器13における複素乗算の動作を示す図である。特に、図10(a)は変調方式に16値QAMを用いた場合の搬送波変調信号の取り得る信号点配置を示し、図10(b)は位相がランダム 20に変化する単位ベクトルiを示し、図10(c)は位相を特定パターンにランダム化された搬送波変調信号を示している。

【0058】図10(a)において、今、一つの搬送波に割り当てられる搬送波変調信号群中の一つの搬送波変調信号が、複素平面上の信号点Aに配点されたと仮定する。信号点Aは、その実数部が3、その虚数部が1の大きさを持つ。また、単位ベクトルiは、この時、位相角3 $\pi/4$ を持ったと仮定する。複素乗算の結果、図10(c)に示す搬送波変調信号A,が得られる。搬送波変調信号A,は、実数部が-2.8、虚数部が1.4となり、16値QAMの配置にはない信号点をとることになる。このように、単位ベクトルiの位相がランダムに変化するため、搬送波変調信号発生器12から出力された搬送波変調信号群中の各搬送波変調信号の位相が、たとえ同一であっても、複素乗算器13は、位相が相互にランダム化された搬送波変調信号群を、逆フーリエ変換器15に出力する。

【0059】複素乗算器13は、このような動作を所定の期間繰り返す。また、複素乗算器13は、定期的にデ 40 ータ D 0 だけを出力する。この時の一連の動作を、図11に示す。すなわち、データ D 0 が挿入されるシンボルをS 0 とすると、送信装置3は、図12に示すように、定期的にシンボルS 0 のデータ D 0 を、その他の場合はシンボルS m のデータ D m を出力することになる。逆フーリエ変換器15は、搬送波変調信号群 D m を、シンボル毎に、周波数軸上に並ぶ各搬送波に割り当て、これらに対して一括的に逆フーリエ変換および並直列変換を施すことにより、ディジタルのOFD M 信号に変換する。この結果、機等波変調信号群の確如其準位相が 050

16

から 2π までのランダムな値になり、逆フーリエ変換器15から出力されたOFDM信号に電力集中が起こるのを抑制できる。したがって、送信装置、受信装置のダイナミックレンジを大きくする必要がなく、安価な構成で、OFDM信号への送受信器や中継増幅器等の非線形性からの影響を軽減することができる。送信装置3における他の回路プロック、すなわちガードタイム挿入部150~ローパスフィルタ191は、送信装置10場合と同様に動作する。

【0060】なお、ガードタイム挿入部16は、シンボルSmの場合と同様に、シンボルS0の後端部と同じデータ成分D0を対応する前部ガードタイムに挿入するとともに、シンボルS0の前端部と同じデータ成分を対応する後部ガードタイムに挿入している。

【0061】図9に示す送信装置3を用いた場合、基本的には、図2に示す受信装置2と同じ構成の受信装置を用いることができる。ただし、受信装置のメモリ26には、送信装置3のメモリ14に記憶される基準複素数信号群D0の受信データZD0を記憶させることになる。

【0062】上記した図9の実施形態においても、前述した第1の実施形態と同様の効果が得られる。すなわち、マルチパスにより反射波が直接波に重なり、直接波のシンボル区間と反射波のガードタイムとが重なっても、フーリエ変換後に周波数軸上に現れる受信搬送波変調信号群の振幅位相歪みがすべて一様なものとなり、その除去を簡単な演算処理(乗算、除算)で行える。また、送信側と受信側との間でOFDM信号に時間遅延が発生しても、時間遅延の影響のない復調データを得ることができ、時間窓の時間軸上の調整が容易になる。

【0063】なお、上述の各実施形態は、有線の伝送路を介してデータを伝送するようにしているが、本発明はこれに限定されることなく、無線の伝送路を介してデータを伝送するようにしてもよい。また、上述の各実施形態では、多チャンネル分のテレビの映像データを各搬送波に乗せるようにしたが、1チャンネル分の映像データを時間分割して並列に並び替え、各搬送波に割り当てるようにしてもよい。さらに、映像データに替えて、音声データ、テキストデータ等を各搬送波にのせるようにしてもよい。さらに、CATVに替えて、LAN、WAN等の他のシステムにおいて本発明を実施してもよい。

【0064】さらに、図9の送信装置3では、メモリ14から出力された基準複素数信号群を、定期的に、複素乗算器13を介して逆フーリエ変換器15に入力するようにしたが、基準複素数信号群を、逆フーリエ変換器15に直接入力してもよい。

フーリエ変換器 15は、搬送波変調信号群D' m を、シンボル毎に、周波数軸上に並ぶ各搬送波に割り当て、これらに対して一括的に逆フーリエ変換および並直列変換を施すことにより、ディジタルのOFDM信号に変換する。この結果、搬送波変調信号群の絶対基準位相が、0 50 DM信号に生じる電力集中が生じないような状況下で

は、搬送波変調信号群に含める基準複素数信号群として、予め定められた特定パターンを有し、かつ各信号の位相が相互に同一の複素数信号群を使用しても良い。この場合でも、第1の実施形態と同様、簡単な演算処理(乗算、除算)を行うことで、振幅位相歪みを除去できる。

【図面の簡単な説明】

【図1】本発明の第1の実施形態の送信装置の構成を示すブロック図である。

【図2】本発明の第1の実施形態の受信装置の構成を示 10 すブロック図である。

【図3】図1の送信装置1から送信されるOFDM信号の構成を示す図である。

【図4】図1のメモリ14と、複素乗算器13との動作を示す図である。

【図5】図1の送信装置1から出力されたOFDM信号に対する受信装置2のエンベロープ検波器23と同期再生部24との動作を示す図である。

【図6】図2のメモリ26と、複素除算器27との動作を示す図である。

【図7】マルチパスによる遅延波の影響について、従来のシステムと第1の実施形態のシステムとを比較したシミュレーション結果を示す図である。

【図8】伝送路等による時間遅延の影響について、従来のシステムと第1の実施形態のシステムとを比較したシミュレーション結果を示す図である。

【図9】本発明の第2の実施形態の送信装置の構成を示すブロック図である。

【図10】図9の複素乗算器13における搬送波変調信号群と複素数信号群との複素乗算の様子を示す図である。

18 【図11】図9のメモリ14と複素乗算器13との動作 を示す図である。

【図12】図9のOFDM信号の送信装置から送信されるOFDM信号の構成を示す信号構成図である。

【図13】従来のOFDM信号の送信装置の構成を示す ブロック図である。

【図14】図13の送信装置5から送信されるOFDM信号の構成を示す図である。

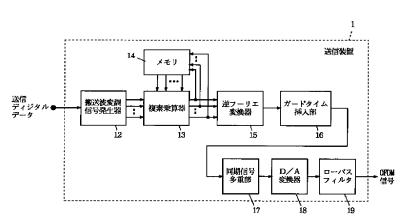
【図15】相互に直交する搬送波に割り当てられた搬送 2 波変調信号群の位相状態とOFDM信号との関係を示す 信号波形図である。

【符号の説明】

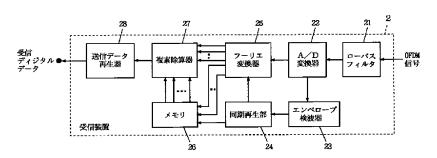
(10)

- 1, 3…送信装置
- 12…搬送波変調信号発生器
- 13…複素乗算器
- 14…メモリ
- 15…逆フーリエ変換器
- 16…ガードタイム挿入部
- 17…同期信号多重部
- 20 18…D/A変換器
 - 19…ローパスフィルタ
 - 31…特定パターン発生器
 - 2…受信装置
 - 21…ローパスフィルタ
 - 2 2 ··· A / D変換器
 - 23…エンベロープ検波器
 - 2 4…同期再生部
 - 25…フーリエ変換器
 - 26…メモリ
- 7 2 7 … 複素除算器
 - 28…送信データ再生器

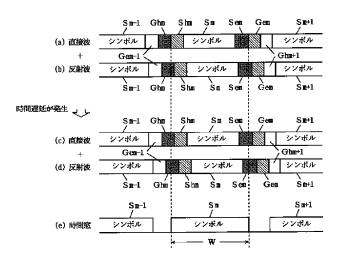
【図1】



[図2]

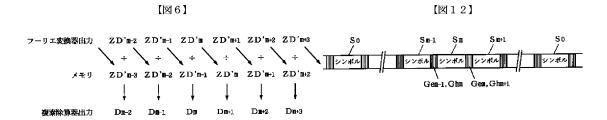


【図3】

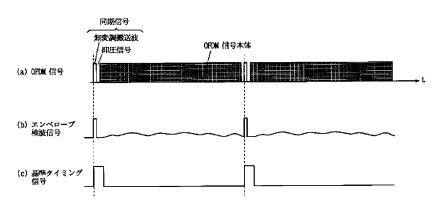


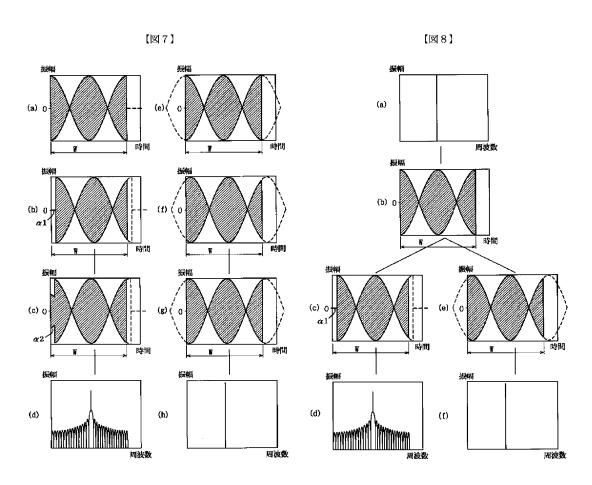
【図4】



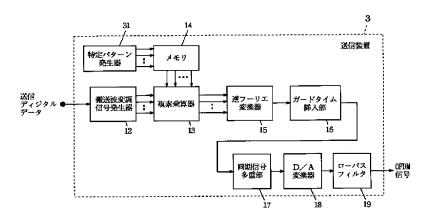




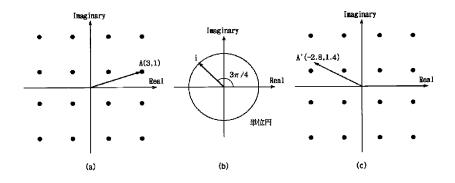




【図9】



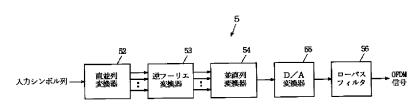
【図10】



【図11】

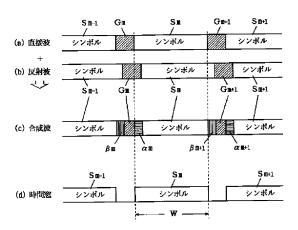


【図13】

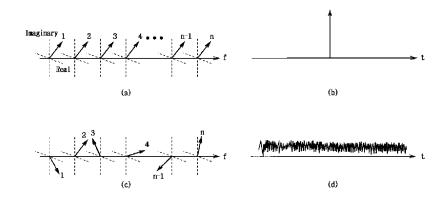


-213-

【図14】



【図15】



フロントページの続き

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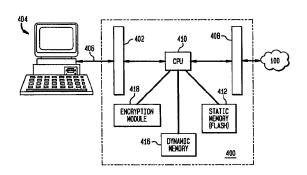
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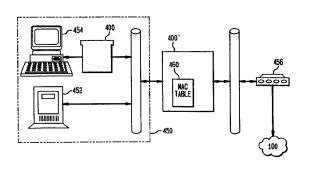
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(54) Title: IMPROVED NETWORK SECURITY DEVICE

(57) Abstract

A network security device is connected between a protected client and a network. The network security device negotiates a session key with any other protected client. Then, all communications between the two clients are encrypted. The inventive device is self-configuring and locks itself to the IP address of its client. Thus, the client cannot change its IP address once set and therefore cannot emulate the IP address of another client. When a packet is transmitted from the protected host, the security device translates the MAC address of the client to its own MAC address before transmitting the packet into the network. Packets addressed to the host, contain the MAC address of the security device. The security device translates its MAC address to the client's MAC address before transmitting the packet to the client.





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IMPROVED NETWORK SECURITY DEVICE

Related Application

This patent claims the benefit of U.S. Provisional Patent Application Serial Number 60/033,995 entitled "Improved Network Security Device", filed on January 3, 1997 for Dr. Aharon Friedman and Dr. Eva Bozoki. This patent application is directed to improvements in the invention described in U.S. Patent Application Serial No. 08/529,497 entitled "Network Security Device" and filed on September 18, 1995. The contents of these two documents are incorporated herein by reference.

Field of the Invention

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The present invention is directed to improvements in a network security device that is connected between a protected computer("the client") and a network and/or a protected local area network (LAN) and a wide area network (WAN) as well as a method for using the network security device.

Background of the Invention

A. Network Architecture

An Internet communications network 100 is depicted in FIG. 1 including five transmit or backbone networks A,B,C,D, and E and three stub networks R, Y, and Z. A "backbone" network is an intermediary network which conveys communicated data from one network to another network. A "stub" network is a terminal or endpoint network from which communicated data may only initially originate or ultimately be received. Each network, such as the stub network R, includes one or more interconnected subnetworks I, J, L, and M. As used herein, the term "subnetwork" refers to a collection of one or more nodes, e.g., (c,w), (d), (a), (b,x,y), (q,v), (r,z), (s,u), (e,f,g),(h,i),(j,k,l),(m,n), and (o,p), interconnected by wires and switches for local internodal communication. Each subnetwork may be a local area network (or "LAN"). Each subnetwork has one or more interconnected nodes which may be host computers ("hosts") u,v,w,x,y,z (indicated by triangles) or routers a,b,c,d,e,f,g,h,i,j,k,l,m,n,o,p,q,r,s (indicated by squares). A host is an endpoint node from which communicated data may initially originate or ultimately

be received. A router is a node which serves solely as an intermediary node between two other nodes; the router receives communicated data from one node and retransmits the data to another node. Collectively, backbone networks, stub networks, subnetworks, and nodes are referred to herein as "Internet systems".

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FIG. 2 shows a block diagram of a host or router node 10. As shown, the node may include a CPU 11, a memory 12 and one or more I/O ports (or network interfaces) 13-1, 13-2,..., 13-N connected to a bus 14. Illustratively, each I/O port 13-1, 13-2,..., 13-N is connected by wires, optical fibers, and/or switches to the I/O port of another node. The I/O ports 13-1, 13-2,..., 13-N are for transmitting communicated data in the form of a bitstream organized into one or more packets to another node and for receiving a packet from another node. If the host 10 is a host computer attached to a subnetwork which is an Ethernet, then the host will have an I/O port which is an Ethernet interface.

A host which initially generates a packet for transmission to another node is called the source node and a host which ultimately receives the packet is called a destination node. Communication is achieved by transferring packets via a sequence of nodes including the source node, zero or more intermediary nodes, and the destination node, in a bucket brigade fashion. For example a packet may be communicated from the node w to the node c, to the node d, to the node b, and to the node x.

An exemplary packet 40 is shown in FIG. 3A having a payload 41 which contains communicated data (i.e., user data) and a header 42 which contains control and/or address information. Typically, the header information is arranged in layers including an IP layer and a physical layer.

The IP layer typically includes an IP source address, an IP destination address, a checksum, and a hop count which indicates a number of hops in a multihop network. A physical layer header includes a MAC (Media Access Control) address (hardware address) of the source and a MAC address of the destination.

The user data may include a TCP (Transfer Control Protocol) packet including TCP headers or a UDP (User Data Protocol) packet including UDP headers. These protocols control among other things, the packetizing of

information to be transmitted, the reassembly of received packets into the originally transmitted information, and the scheduling of transmission and reception of packets (see e.g., D. Commer, "Internetworking With TCP/IP", Vol. 1 (1991); D. Commer and D. Stevens, "Internetworking With TCP/IP", Vol. 2 (1991)).

As seen in FIG. 3B, in an exemplary Internet protocol (IP), each node of the Internet 100 is assigned an Internet address (IP address) which is unique over the entire Internet 100 such as the Internet address 30 for the node y shown in FIG. 3B. See, Information Sciences Institute, RFC 791 "Internet Protocol", September, 1981. The IP addresses are assigned in an hierarchical fashion; the Internet (IP) address 30 of each node contains an address portion 31 indicating the network of the node, an address portion 32 indicating a particular subnetwork of the node, and a host portion 33 which identifies a particular host or router and discriminates between the individual nodes within a particular subnetwork.

In an Internet system 100 which uses the IP protocol, the IP addresses of the source and destination nodes are placed in the packet header 42 (see FIG. 3A) by the source node. A node which receives a packet can identify the source and destination nodes by examining these addresses.

In an Internet system, it is the IP address of a destination that is known, and the physical address (i.e., MAC address) to be placed in the MAC frame header is to be determined. If the destination host is on the same local area subnetwork (and this is easily determined by observing that the network part in both the source and destination IP addresses is the same), then the destination address that is to go into the MAC header destination address field is simply the physical address of the destination host. The MAC destination address may be found by means of the ARP (Address Resolution Protocol) which comprises having the source host broadcast an ARP request packet with the IP address of the destination host and having the destination host respond with its hardware (MAC) address. This MAC address may be placed in the MAC frame (physical layer) headers.

30 B. Encryption Techniques

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Eavesdropping in a network, such as the Internet system 100 of FIG. 1, can be thwarted through the use of a message encryption technique. A message

encryption technique employs an encipherment function which utilizes a number referred to as a session key to encipher data (i.e., message content). Only the pair of hosts in communication with each other have knowledge of the session key, so that only the proper hosts, as paired on a particular conversation, can encrypt and decrypt digital signals. Three examples of encipherment functions are (1) the National Bureau of Standards Data Encryption Standard (DES) (see e.g., National Bureau of Standards, "Data Encryption Standard", FIPS-PUB-45, 1977), (2) Fast Encipherment Algorithm (FEAL)(see e.g., Shimizu and S. Miyaguchi, "FEAL-Fast Data Encipherment Algorithm," Systems and Computers in Japan, Vol. 19, No. 7, 1988 and S. Miyaguchi, "The FEAL Cipher Family", Proceedings of CRYPTO '90, Santa Barbara, Calif., Aug., 1990); and (3) International Data Encryption Algorithm ("IDEA") (see e.g., X. Lai, "On the Design and Security of Block Ciphers," ETH Series in Information Processing, v.1, Konstanz: Hartung - Gorre Verlag 1992). One way to use an encipherment function is the electronic codebook technique. In this technique a plain text message m is encrypted to produce the cipher text message c using the encipherment function f by the formula c = f(m,sk) where sk is a session key. The message c can only be decrypted with the knowledge of the session key sk to obtain the plain text message m = f(c, sk).

Session key agreement between two communicating hosts may be achieved using public key cryptography. (See e.g., U.S. Patent Nos. 5,222,140 and 5,299,263).

Before discussing public key cryptographic techniques, it is useful to provide some background information. Most practical modern cryptography is based on two notorious mathematical problems believed (but not proven) to be hard (i.e., not solvable in polynomial time, on the average). The two problems are known as Factorization and Discrete-Log. The Factorization problem is defined as follows:

Input: N, where N = pq where p and q are large prime numbers

Output: p and/or q.

The Discrete-Log problem is defined as follows:

Input: P,g,y, where $y \equiv g^x \mod P$, and P is a large prime number

Output: x.

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(The Discrete-Log problem can be similarly defined with a composite modulus N = pq).

Based on the Factorization and Discrete-Log problems, some other problems have been defined which correspond to the cracking problems of a cryptographic system.

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One system of such a problem which has previously been exploited in cryptography (see, e.g., H.C. Williams, "A Modification of RSA Public-Key Encryption", IEEE Transactions on Information Theory, Vol. IT-26, No. Nov. 6, 1980) is the Modular Square Root problem, which is defined as follows:

Input: N,y, where $y \equiv x^2 \mod N$, and N = pg, where p and q are large primes Output: x.

Calculating square roots is easy if p and q are known but hard if p and q are not known. When N is composed of two primes, there are in general four square roots mod N. As used herein, $z = \sqrt{-x} \mod N$ is defined to mean that x is the smallest integer whereby $z^2 = x \mod N$.

Another problem is known as the Composite Diffie-Hellman (CDH) problem, which is defined as follows:

Input: N, g, $g^x \mod N$, $g^y \mod N$, where $N \equiv pq$ and p and q are large primes. Output: $g^{xy} \mod N$.

It has been proven mathematically that the Modular Square Root and Composite Diffie-Hellman problems are equally difficult to solve as the above-mentioned factorization problem (see, e.g., M.O. Rabin, "Digitalized Signatures and Public Key Functions as Intractable as Factorization", MIT Laboratory for Computer Science, TR 212, Jan. 1979; Z. Shmuely, "Composite Diffie-Hellman Public Key Generating Schemes Are Hard To Break", Computer Science Department of Technion, Israel, TR 356, Feb. 1985; and K.S. McCurley, "A Key Distribution System Equivalent to Factoring", Journal of Cryptology, Vol. 1, No. 2, 1988, pp. 95-105).

In a typical public-key cryptographic system, each user I has a public key P_i (e.g., a modulus N) and a secret key S_i (e.g., the factors p and q). A message to user I is encrypted using a public operation which makes use of the public key known to everybody (e.g., squaring a number mod N). However, this message is

decrypted using a secret operation (e.g., square root mod N) which makes use of the secret key (e.g., the factors p and q).

C. Network Security Devices

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At present, existing network security products are categorized into two classes: (1) firewalls, such as Janus and ANS, and (2) software products, such as encrypted mail, secured http, one time password, etc.

A firewall is a dedicated computer, usually running a Unix operating system. It acts as a filter for incoming and outgoing communications. The firewall is placed as a router between the local area network (LAN) and the outside world. The decision whether to pass a packet is made based on the source and/or destination IP address, and the TCP port number. Some firewalls also have the ability to encrypt data, provided that both sides of the communication employ the same brand of firewall. Some firewalls have a personal authentication feature.

Software products are based on the premise that the computer on which they are installed are secure, and protection is only needed outside on the network. Thus, such software products can easily be bypassed by breaking into the computer. A typical scheme is when an intruder implants a "Trojan Horse" on a computer which sends him an unencrypted copy of every transaction. Sometimes, it is even done as a delayed action during the off-hours when the computer is not likely to be supervised.

In addition, there are authentication products designed to maintain the integrity of the computer against intrusion. These products are based on the premise that the products are 100% secure. Once the product is compromised, it becomes totally ineffective. Sometimes, careless use by one user may jeopardize all other users of the product.

Firewalls are more effective in maintaining network security. However they are very expensive. Their price range is between \$10,000 and \$50,000, plus the price of the hardware. They require a high level of expertise to install and maintain. The most sophisticated and effective firewalls require a specially trained technician or engineer for their maintenance. The special training cost is up to

\$10,000 per person, and the salary adds \$60,000 to \$120,000 or more per annum to the cost.

Firewalls have to be constantly maintained, modified, and monitored in order to yield reasonable security. They only cover the TCP part of the Internet Protocol and not the UDP part. Thus, they do not provide security to NFS (Network File Services) and many client/server applications.

The firewall is a full service computer which can be logged into for maintenance and monitoring. Thus, it can be broken into. Once a firewall is compromised it loses its effectiveness and becomes a liability rather than a security aid. Firewalls only protect the connection between a LAN and a WAN (Wide Area Network). It does not protect against intrusion into a particular host from within the LAN.

In view of the foregoing, it is an object of the present invention to provide a network security device which overcomes the shortcomings of the prior art network security devices.

It is another object of the present invention to provide a hardware device to provide network security for individual hosts attached to a network.

It is a further object of the present invention to provide a hardware device to provide network security for a local area network connected to a wide area network.

Summary of the Invention

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The present invention provides improvements to the Network Security Device described in U.S. Patent Application Serial Number 08/529,497. These improvements include (1) modifications in the device which adapt it to protect a LAN, (2) improved key generation, (3) an improved key exchange algorithm, and (4) improved packet handling procedures which provide double integrity checks.

A preferred embodiment of the inventive network security device comprises a first network interface connected to a protected client, a second network interface connected to a portion of a network, and a processing circuit connected to both interfaces.

A communication from the protected client goes from the client, to the first interface, to the processing circuit, to the second interface and into the network. Similarly, a communication received from the network goes from the second interface, to the processing circuit, to the first interface and to the protected client.

A preferred embodiment of the present invention has four keys associated with it:

- (1) a static (permanent) private key;
- (2) dynamic (changing) private key;
- (3) a static public key; and
- (4) a dynamic public key.

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In a preferred embodiment, the public keys are exchanged between two network security devices in order to establish a common secret key. The common secret key is the key which is used to encrypt/decrypt all messages between two particular devices. This key should not be transmitted.

The static keys are permanent keys unique to each device. The dynamic keys have a predetermined lifespan and are replaced periodically, such as every 24 hours. Preferably, the static keys are generated using a seed derived from the host's IP address, MAC address of the network interface connected between the protected host and the network security device, and the security device's serial number. Preferably, the dynamic keys are generated using seeds derived from current date and time information.

Packets received from the protected client are encrypted using an encipherment function such as IDEA, FEAL, or DES before being transmitted via the network to a destination. Similarly encrypted packets received from a destination are decrypted. Such encryption and decryption requires a common session key to be possessed jointly by the protected client and the destination (the destination being a protected client of another network security device located someplace else in the network).

The common crypto key (i.e., the common secret key) is obtained using a public key cryptography technique. To aid in the key exchange, the network security device maintains two databases. A static database (SDB) contains

information about secured hosts or nodes in the network. A secured host or node is a host or node that is protected by a network security device. Each entry in the static database contains information about a particular secured host, i.e., the host IP address, time entered in the database, and the host's permanent public key.

A dynamic data base (DDB) contains information about secured and unsecured hosts. Each entry in the dynamic database includes a host's IP address, the time that the host's dynamic key was generated, a flag indicating whether or not the host is secured, a flag indicating whether the host is in transition (i.e., in the middle of a key exchange), and a pointer to a common secret session key.

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The protocol used by the network security device of host i to agree on a common crypto key with a network security device of host j is as follows.

Consider a communication from host i to host j. The communication arrives at the network security device of host i from host i. The network security device checks if host j is in the dynamic database. If host j is in the dynamic database, it is determined if the dynamic database has a common crypto key for communication between host i and host j. If there is such a common session key, the communication from host i is encrypted using the common crypto key and transmitted to host j. If there is no common crypto key, then host i sends the dynamic part of its public key P_i to host j and host j replies by sending the dynamic part of its public key P_i to host i. The exchange of dynamic parts of the public keys may be encrypted using the static part of the public keys, which may be obtained from the static databases at host i and host j. The common crypto key is then calculated according to a Diffie-Hellman technique.

Because the dynamic keys of each network security have a particular lifespan, such as 24 hours, there may be a time difference between times when two device's keys expire. Thus, it is possible that one device's dynamic key may expire before the packet is received. One way to prevent this occurrence is to take into account this time difference. The DDB may correct the time difference between the time the packet was sent and the time the packet is received. Also, the DDB time generation entry indicates to the network security device when the other party's dynamic key expires. Thus, when a communication between the

nodes is initiated, it may be determined whether a new dynamic key exchange is warranted, rather than attempting to use an expired common dynamic key.

Note that this assumes that there is an entry for host j in the static database of host i. If there is not, the exchange of dynamic public keys is preceded by an exchange of static public keys and the forming of a database entry for host j in the static database at host I. Moreover, if there is no entry for host j in the dynamic database of host I, such an entry will be generated before the dynamic key exchange.

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A packet received by the network security device and the connected host is preferably processed in the following manner. The IP and MAC headers from the packet are copied into a new IP packet. The client host's physical address (e.g., the MAC address of the network interface between the client and the network security device) is replaced with the network security device's MAC address (e.g., the MAC address of the network interface between the network security device and the network). The new IP packet includes a proprietary header and proprietary tail. Information about the packet is stored in the proprietary tail, including check sum information. The data and proprietary tail are then encrypted. The proprietary header is then filled in, including check sum information for the encrypted data. This packet is then transmitted into the network.

This processing method provides a double integrity check. The check sum which was calculated after encryption is checked by the receiver before decryption, providing an integrity test of the encrypted data in transit. The check sum in the proprietary tail was calculated before encryption and checked by the received after decryption. This checksum provides a means of strong authentication because the static and dynamic keys used to encrypt the checksum are known only to the two communicating hosts. This check sum will differ if the common secret key is not identical on both sides and also provides an integrity test of the actual data. Note that if the check sum is replaced with a secured hash function, after the packet is encrypted, it becomes a digital signature. It should be noted that encryption takes place at the IP level so that TCP and UDP packets are encoded.

In short, the inventive network security device has a number of significant advantages.

Like a firewall, the inventive network security device is a hardware/software combination in a preferred implementation. However, it is a sealed "box" and cannot be logged into. Hence, it cannot be compromised the way a firewall can. It is much cheaper than a firewall. Thus, each node in the LAN can be equipped with it. This way, it provides protection inside the LAN as well as outside. The network security device works directly at the IP level. It therefore, covers all types of IP protocols and requires no special configuration to different network applications. Thus, the inventive network security device is maintenance free.

Brief Description of the Drawings

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The present invention is described with reference to the following figures:

FIG. 1 schematically illustrates an Internet system.

FIG. 2 schematically illustrates the architecture of a host in the network of FIG. 1.

FIGs. 3A and 3B illustrate the format of a packet transmitted in the network of FIG. 1.

FIG. 4A illustrates a network security device for use with a host in the network of FIG. 1 in accordance with an embodiment of the present invention.

FIG. 4B illustrates a network security device for use with a LAN in accordance with an embodiment of the present invention.

FIG. 5 illustrates an entry in a static database maintained by the network security device of FIG. 4.

FIG. 6 illustrates an entry in a dynamic database maintained by the network security device of FIG. 4.

FIG. 7 is a flow chart illustrating an activation method used by the network security device of FIG. 4.

FIG. 8 is a flowchart illustrating a key exchange method used by the network security device of FIG. 4.

FIG. 9 is a flow chart illustrating an IP packet handling algorithm utilized by the network security device of FIG. 4.

FIG. 10 illustrates an IP packet received from a connected host by the network security device and an IP packet transmitted from the network security device into a network.

FIG. 11 is a flowchart illustrating a method of processing the packets of FIG. 10.

Detailed Description of the Invention

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Overview of the Invention

FIG. 4A schematically illustrates a network security device for protecting a host according to an embodiment of the invention. The security device 400 comprises a first interface 402 which is connected to the client host 404. Specifically, the interface 402 is connected to a network interface in the client host 404 (e.g., an interface 13 of Fig. 2) via a cable or wire 406. The security device 400 comprises a second interface 408 which is connected to a portion of a network 100. Illustratively, the interface 408 is connected to an Ethernet so that the interfaces 402, 408 are Ethernet interfaces such as SMC Elite Ultra Interfaces.

FIG. 4B schematically illustrates a network security device 400' for protecting a LAN according to an embodiment of the invention. As seen in FIG. 4B, a network security device 400' according to the invention is connected between a LAN 450, such as an Ethernet network (including, for example, a file server 452 and a workstation 454), and a router 456 which routes communications between the LAN 450 and a WAN 100, such as the Internet. As discussed in detail below, several modifications are made in the Network Security Device to adapt it for use in protecting a LAN. As also seen in FIG. 4B, network security devices may be arranged in a cascaded topology. Note that workstation 454 is associated with a network security device 400.

Returning to FIG. 4A, a CPU 410 is connected to the interfaces 402, 408. The CPU is, for example, an Intel 486 DX 62-66 or Pentium. Alternatively, the processing circuit may be implemented as one or more ASICs (Application Specific Integrated Circuits) or a combination of ASICs and a CPU. A static memory 412 (e.g., flash EEPROM) is also connected to the CPU 410 and a dynamic memory

416 (e.g., RAM) is connected to the CPU 410. An optional encryption module 418 may be provided to perform encryption and large number arithmetic operations. The encryption unit may be implemented as a programmable logic array. Alternatively, the encryption module 418 may be omitted and its function may be carried out using a software program which is executed by the CPU 410. However, because certain encryption functions are calculation intensive, it may be preferable to separate the encryption functions from other functions of the Network Security Device 400.

The software executed by the CPU 11 preferably has three components: (1) operating system, (2) networking system, and (3) key computation algorithms. The operating system and the networking system may both be part of a Unix-like kernel. The key computation algorithms reside in memory and are signaled into action by the networking system. The operating system is a lobotomized Linux system with all drivers taken out except the RAM, disk, and Ethernet interfaces. The networking system is for communication, key exchange, encryption, configuration, etc. In a preferred embodiment, the key computation software may run independently of the other software. This shifts the computationally intensive task of key computation away from the operating system and networking system.

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The CPU 410 maintains two databases. One database is a static database (SDB) 412 preferably stored in a permanent memory, such as a Flash ROM 412. FIG. 5 illustrates one entry in the SDB 412. The SDB may have an entry for the client host as well as other hosts. As seen in FIG. 5, the static database entry 500 contains permanent information about the network security device 400 and other secured nodes in the network. The static database entry 500 may include the following information about another secured node: the other node's IP address 502, time that this other node was entered into the database 504, the node's permanent public key 506, and a pointer to the static common key shared by the network security device 400 and the other node's device 508. The static database 500 may also contain the IP address and the serial number of the connected host 510.

A second database is a dynamic database (DDB) 416, which may be stored in a volatile memory, such as a RAM. FIG. 6 illustrates one entry in the DDB 416. As seen in FIG. 6, the dynamic database entry 600 contains information about secured and unsecured nodes, i.e., the other node's IP address 602, the time a last packet was sent from that other node 604, a time the other node's dynamic key was generated 606, a pointer to a common secret key shared with that node, time last updated, a secured flag indicating whether the node is secured (e.g., has its own network security device) 602, and a transition flag indicating whether the node is in transition (i.e., in the middle of a key exchange).

Briefly, a preferred embodiment of the present invention operates in the following manner. The interface 402 is put in a promiscuous mode. In this mode, the interface 402 passes all communications from the client host 404 that are sensed on the cable 406 to the CPU 410. The network connection is via the interface 408 which is set to the same IP address as the client 404. The network security device 400 responds to the Address Resolution Protocol by sending its own (rather than the client's) MAC address. This adds a level of security by blocking attempts to bypass the device 400 using the Ethernet protocol.

Received communications are checked to see if they are from a secured host. First, the DDB entry 600 is checked to determine if there is a current dynamic common key shared with the node sending the communication. If yes, this key is used to encrypt and decrypt subsequent packets. If no, if these nodes have communicated previously, a dynamic key exchange is performed. If it is the first time these nodes have communicated, a static key exchange is performed to obtain a static dynamic key. This static key is used to encrypt and decrypt the dynamic key exchange communications.

Activation and Initialization

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In a preferred embodiment, the network security device 400 is a sealed box which cannot be logged into. The network security device 400 senses the IP (and/or MAC) address of the client host 404 and locks itself to it. Once the network security device is locked to the address, the client 404 is prevented by the network security device 400 from changing its IP (and/or MAC) address.

Preferably, before the network security device 400 is placed into service, it is activated. The role of activation is to allow or disallow burning the host's 404 IP address into an entry 500 in the static database 412. As discussed above, the SDB 412 may have an entry 500 for the connected client host.

The network security device's serial number (element 510) and the time of activation (element 504) may also be burned into the static database entry 500. As discussed below, these values may be used to generate a seed for the network security device's static private key.

FIG. 7 is a flowchart 700 illustrating a preferred activation method. First, an "activation packet" containing an activation string in the payload may be sent from a connected computer, such as a host 404, through the network security device 400 (step 702). The packet is received by the device 400, which determines whether it has been activated (step 704). If it has not been activated, the IP address and other information are written into the flash memory (step 706), as described above, and an acknowledgment packet is returned to the computer (step 708). The device 400 may also generate a confirmation message for display on a monitor of the connected computer (step 710). The Address Resolution Protocol (ARP) is the protocol which is used to resolve an IP address into a matching Ethernet machine (MAC) address which is the actual address to which the network interface responds. As discussed above, the inventive network security device uses ARP (Address Resolution Protocol) to configure itself and hide the client host. The manner in which the network security device processes an ARP request is described in related application Serial No. 08/529,497, the contents of this description are incorporated herein by reference.

Key Calculation

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A preferred embodiment of the present invention has four keys associated with it:

- (1) a static (permanent) private key;
- (2) dynamic (changing) private key;
- (3) a static public key; and
 - (4) a dynamic public key.

In a preferred embodiment, the private keys are 128 bits long and are known only to that network security device. In a preferred embodiment, the public keys are 512 bits long and are revealed to others. Public keys, as described above, are exchanged between two network security devices in order to establish a 128 bit long common secret key. The common secret key is the key which is used to encrypt/decrypt all messages between two particular devices. The common secret keys should never be transmitted.

In a preferred embodiment, the keys are generated when the device 400 is turned on. As described in detail below, the static keys are permanent keys unique to each device and the dynamic keys have a predetermined lifespan and are replaced periodically, such as every 24 hours.

Static Keys

Keys are generated using a "seed", or number, which is then processed to generate a key. The seed for a randomly generated static private key for a particular network security device 400 is derived from the device's IP-address, MAC-address, serial number, and a time-stamp. The seed may be determined in the following manner:

seed =
$$IP + MAC_1 + MAC_n + serial + time$$

where:

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20 MAC, is the low four bytes of the device's six byte MAC address;

MAC_h is the high two bytes of the MAC address.

Using this seed, a private key (preferably 128 bits long), is then randomly generated using a random number generator, such as the GNU Multiple Precision library copyrighted by Free Software Foundation Inc. (1996), Boston, Massachusetts, 02111. If the box is non-activated, the seed is the present time, thus it will be different every time the box is turned-on. On the other hand, for activated boxes, the static private key is a property of the box, it will not change by turning the box on/off.

Dynamic Keys

The dynamic private key is randomly generated at predetermined intervals. For example, dynamic keys may be generated every 24 hours. Preferably, the dynamic keys are derived from a random seed obtained from seconds, minutes,

and hours of the present time. The dynamic secret key may be processed from the seed using a random number generator, such as the GNU Multiple Precision library

Public Keys

The static and dynamic public keys are calculated from the private keys according to the equation:

 $X_i \equiv q^{x_i} \pmod{n}$

where:

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X_i is the public key;

10 x_i is the private key;

and q and n prime numbers which are preferably installed in each network security device.

Key Exchanges

The first time a client 404 or LAN 450 sends a message to another network security device, a protocol is executed by which the two devices (i) exchange static public keys (unencrypted), (ii) generate a static common key, and then (iii) exchange dynamic public keys encrypted with their static common key.

8 is a flowchart 800, illustrating the key exchange algorithm.

Consider the case where the host client wants to send a communication to a node in the network whose IP = A. When the communication arrives at the network security device of the host client the dynamic data base 416 (DDB) is checked to determine if there is an entry 600 for node A in the dynamic data base (step 702).

Note that the DDB includes an entry for a "secured" flag 612 and a "transition" flag 614. The secured flag indicates the current security status between the two network security devices. Preferably, the secured flag may be in one of five states:

0 = unsecured

1 = secured

2 = other party was secured, but now does not respond to dynamic key exchange request (i.e., other party has an entry in the SDB 500 but no current entry in the DDB 600)

3 = the device's dynamic key has expired and must renegotiate all dynamic keys

4 = cannot allocate key storage for the other party's key

The transition flag 614 indicates the status of a key exchange. Preferably, the transition flag maybe in one of four states:

0 not in transition

i ≤ N waiting to receive a dynamic public key packet

N + 2 waiting for a dynamic common key calculation

 $-i \ge -N$ waiting for static public key packet

-(N + 2) waiting for a static common key calculation

where N is the maximum number of tries, and i is the actual number of tries. As discussed in detail below, if there is no entry 600 in the DDB 416, the SDB 412 is searched for an entry 500 corresponding to node A.

The database searches return:

15 (i) a transition flag; and

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(ii) and a reference to the entry number in the database.

The "transition" and "secured" flags in the DDB may then be set accordingly. The following table sets out possible outcomes of a DDB/SDB search.

Possible Outcomes of Database Search.

20		Need dynamic key	Need static key have nothing to decrypt with	Not asking for any key	Comments
	Trans	1	-1	0	

	Need dynamic key	Need static key have nothing to decrypt with	Not asking for any key	Comments
Return	O -(i + 1)	O	O -(i + 1) + (i + 1) -(DB size + i + 1)	have nothing to en/decrypt with (no entry in DDB) use st-key to en/decrypt use dyn-key to en/decrypt have st-key in DB, but no dyn-key response (do not encrypt, but use st-key to decrypt)

Where DB size is the number of entries i for node A in the entry number in the dynamic database.

Returning to FIG. 8, if there is an entry for node A in the dynamic data base, a check is made to see if a common dynamic key for node A and the protected client has expired (step 803).

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If, for example, there is an entry for node A and the secured flag = 1, then node A is secured. Thus, the common dynamic key has not expired and the packet is encrypted using the session key and an encipherment function such as IDEA (step 806).

If the common dynamic key has expired, the dynamic data base entry for the node IP=A has a secured flag = 3 and the transition flag is $i \le N$ (step 804) which means a key exchange is taking place.

The exchange of the dynamic parts of the public keys of the host client and the node with IP = A proceeds as follows. The host client (i.e., the source) sends its dynamic public key and IP address to the node with IP = A (the destination) (step 808) and waits for a reply (step 810). The dynamic public key of the host may be encrypted with the static public key of the node with IP = A. The reply is the dynamic public key of the destination node with IP = A. This may be encrypted

with the static public key of the host client. Steps 808 and 810 may be repeated several times, such as three times.

If no reply is received (step 812) from the destination, the source network security device sets the secured flag to 2 and the transition flag to 0 (step 814) in the DDB entry for the destination. If the packet to be encrypted originated from the host (step 816), the packet may be dropped (step 818). If the packet originated from another party, the network security device may try to decrypt the packet using the static private key(step 820).

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If a reply is received, the transition flag for the destination node in the DDB entry 600 of the network security device of the host is set to N+2 (step 822), indicating that the common dynamic key is being calculated. Then a common dynamic (crypto) key for the source and destination is calculated by the network security device of the source (step 824) using, for example, a Diffie-Hellman technique as described above. The common session key is then entered into the DDB entry 600 of the source network security device (step 826) and the transition flag for this DDB entry is marked 0 (step 828) because the transition is complete. The secured flag = 1.

The exchange of dynamic public keys and the calculation of a common crypto key assumes that there is an entry for the destination node with IP=A in the static data base 412 (SDB) of the source network security device and in the dynamic data base 416 of the source network security device. That is, that these two network security devices have communicated before. If these entries do not exist (i.e., these two devices have not previously communicated), they may be created prior to the dynamic public key exchange (steps 708-722 described above).

If there is no DDB entry for node IP = A (step 802), an entry is created (step 830), the secured flag = 0, and the transition flag is marked $-i \ge -N$ (step 832). The SDB 500 is checked to determine if the source network security device has an entry for node IP = A (step 834).

If there is such an entry, proceed with the dynamic key exchange (steps 808-822), the secured flag is set to 1 and the transition flag is set as described above.

If there is no entry for node A in the SDB, then the network security device 400 sends its static public key in a key-packet to node A and drops the original IP-packet (step 836). The device waits a predetermined time, such as five seconds, for a reply (step 838). Steps 836 and 838 may be repeated several, e.g., three times. While waiting for a response, the transition flag is -i≥-N. If a reply is received (step 840), an entry is created in the SDB (step 842), the secured flag = 1 and the transition flag is 0. When the static key is received, the network security device calculates a common static key using its static key and a standard Diffie-Hellman technique. The transition flag is set to -(N+2). Once the static common key is calculated, it is used to encrypt the dynamic key exchange (steps 808-822). The inventive device preferably uses the well-known Diffie-Hellman key exchange protocol.

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If no reply is received, the secured flag = 2 and the transition flag in the entry in the DDB is 0 because the transition off (step 814).

Both the static and the dynamic key request maybe repeated N_{try} times at not less then t_{expire} time intervals (in a preferred embodiment they are set to 8 tries and 2 ms, respectively). Note that entries in the SDB are burned in and are permanent. Entries in the DDB may be volatile, that is, the entries may be overwritten or lost if the device 400 is turned off.

Note that if the second flag for another node is set at either 2, 3, or 4, the network security device will continue to attempt a dynamic key exchange every predetermined period, such as every five minutes.

Expiration of the Dynamic Keys and Synchronization

As indicated above, the dynamic keys have a predetermined lifespan. For example, new dynamic keys may be generated every 24 hours. When the lifespan expires, all of the dynamic common key entries, which were calculated using an expired dynamic key, for other nodes are incorrect. Thus, all secured flags in the DDB are marked as secured = 3. When the dynamic key of a network security device's 24 hours expires, a new dynamic key is generated. The secured flag is then changed back to 1 (or 2) when the next packet (sent to or received from that IP-address) initiates a successful dynamic key exchange.

Because the dynamic keys of each network security device have a particular lifespan, such as 24 hours, there may be a time difference between times when two device's keys expire. For example, if two devices in different time zones are both programmed to generate new dynamic keys at midnight, there may be several hours difference between key expiration times. Thus, it is possible that one device's dynamic key may expire during a communication. Also, because Internet communications are connectionless, that is, the receiving party does not have to be connected to the sending party when the packet is transmitted, one or the other party's dynamic key may have expired before the packet is received.

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One way to prevent this occurrence is to take into account this time difference. Referring back to FIG. 6, the DDB entry 600 contains an entry "time generated" 606, which indicates the time that the other device's dynamic key was generated. This is done by correcting the "time generated" entry by the time difference between the time the packet was sent (the time stamp entry 604 in the DDB entry) and the time the packet is received (the present time).

Also, the time generated 606 entry indicates to the network security device 400 when the other party's dynamic key expires. Thus, when a communication between the nodes is initiated, it may be determined whether a new dynamic key exchange is warranted, rather than attempting to use an expired common dynamic key.

During a dynamic key exchange, one party may have received the other party's dynamic key. The other side, however, may be calculating the common dynamic key and sending dynamic key requests encrypted with the static common key. To avoid having to drop the packet, if a received packet cannot be decrypted with a dynamic key, the device tries to decrypt the packet using the static key. As a result, the packet is dropped only if the packet cannot be decrypted with the static key, that is, if it is an illegal packet.

Receiving a Key Packet

When a network security device receives an IP-packet containing another party's static or dynamic public key (sent either as a reply to a key-request or as an initiation for a key exchange), the public key is extracted from the packet and sent to either the CPU 410 or the encryption module 418 for further processing.

There the shared secret key is calculated from the device's own private key and the other party's just received public key.

As discussed above, these tasks are calculation-intensive, and it may be preferable to provide a separate structure, such as the encryption module 418, so that the throughput of the entire device is not affected.

Packet Processing

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Fig. 9 is a flowchart 900 illustrating a packet handling algorithm utilized by the inventive network security device. Illustratively, the packet arrives with the source address IP = C (step 901). The packet may arrive from the connected host at interface 402 or from the network at interface 408.

First consider the case where the packet arrives from the host at interface 402. If the packet carries an ICMP (Internet Control Message Protocol) or IGMP (Internet Gateway Message Protocol) identification (step 902), the packet is passed to the interface 408 without encryption. However, the source MAC address in the packet is translated to the MAC address of interface 408 (step 904). ICMP and IGMP Packets are not addressed to a destination host. Rather these packets are utilized by intermediate entities in the network, e.g., routers, for various functions. The source IP address is checked to make sure that it is the same as the entry burned into the SDB 412 for the connected host. This prevents an adversary from posing as the connected host to gain access to secure communications. This is called preventing "IP spoofing" and is described in detail in U.S. Patent Application Serial No. 08/529,497. The discussion of preventing IP spoofing is incorporated herein by reference.

If the destination to which the packet is addressed is insecure, the packet is dropped (step 906, 908). The device may be in a secured/unsecured mode (special order). In such case the packet will be sent unchanged.

Next, it is determined if the packet contains a part of a message that has been fragmented (step 910). If the packet contains a fragment, the fragments are collected (step 912) and the message is encrypted (step 914). The encryption takes place using the common session key and an encipherment function. If the encrypted message is too long for the particular LAN (step 916), it is fragmented

(step 918). An encrypted packet is then transmitted to interface 408 for transmission into the network 100 (step 920).

An encrypted packet carries a signature in the protocol part of the IP header. This indicates that the packet is encrypted. The IP address of a packet is not encrypted, otherwise the packet could not be routed through the network.

The case where the packet arrives via the network at interface 408 is now considered. If the packet is an ICMP or IGMP packet (step 940) no decryption is necessary and the packet is sent to the first interface 402 (step 942). If the packet is a key exchange packet (step 944) the packet is processed according to the key exchange protocol (step 946). If the packet is not encrypted (step 948) the packet is dropped (step 950). The device may be in a secured/unsecured mode (special order). In such a case the packet will be sent to the client unchanged. If the packet is encrypted but the network security device does not have the key (step 952), the key exchange protocol is carried out (step 954) and the packet is dropped (step 956). If the key is available in the dynamic data base of the network security device, the packet is decrypted (step 958) and sent to interface 402 (step 960).

For packets received from the network the MAC address of the network security device is translated into the MAC address of the client. For packets received from the protected client, the MAC address of the client is translated into the MAC address of the network security device.

Outgoing Packets

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As discussed above, the network security device 100 receives an IP-packet on the first interface 402, processes it, and sends it onto network 100 via the second interface 408.

FIG. 10 illustrates an IP packet 1010 (IP_{in}) received from host 404, an IP packet (IP_{out}) 1010 prepared by the network security device 400, and an encrypted IP_{out} packet 1030 transmitted by the second interface 408. As seen in FIG. 10, the IP_{in} packet 1000 includes a MAC header 1002, containing the host's 404 MAC address, an IP header 1004, containing the host's IP address, and a payload 1006 containing data. The IP_{out} packet 1010 includes a MAC header 1012 containing the network security device's MAC address, an IP header 1014 containing the

host's 404 IP address, a proprietary header 1016, a payload 1118 containing the data, and a proprietary tail 1020. Preferably, the data in the payload 1118 is compressed and the proprietary tail 1120 includes packet length, protocol fragment, and checksum information. The encrypted IP_{out} packet 1030 preferably has everything after the proprietary header 1016 encrypted, including the compressed data 1018 and the proprietary tail 1020.

FIG. 11 is a flowchart 1100 illustrating the processing of IP_{in} and IP_{out} .

- (a) Packet IP_{in} 1000 is received from the host 404 at the first interface 402 (step 1102).
- 10 (b) The IP and MAC headers are copied from IP_{in} to IP_{out} (step 1104).
 - (c) The destination MAC address in IP_{in} is replaced by the client's MAC-address (step 1106).
 - (d) Skip over the proprietary-header (step 1108).
 - (e) Compress the data from IP_{in} to IP_{out} (step 1110). Preferably, the data is compressed using the LZRW1 compression algorithm.
 - (f) Save original length, protocol, frag-info from IP_{in} into the proprietary tail (step 1112).
 - (g) In the IP_{out} header, set do not frag = off, and set IP_{out} -protocol = 99 (indicating proprietary protocol) (step 1114).
- 20 (h) Calculate the checksum and save it in the proprietary tail 1120(step 1116).
 - (i) Encrypt everything from after the proprietary header until the end of IP_{out} (step 1118).
 - (j) Fill the proprietary-header in IP_{out}; set protocol = 191 (encrypted packet) and calculate the header-checksum (step 1120).

Incoming Packets

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For incoming packets, steps (b)-(j) are performed in reverse order.

Double Integrity Checks

The method illustrated in FIG. 11 provides a double integrity check. The checksum in the proprietary-header on the sender's side is calculated after the encryption and checked on the receiver's side before decryption, thus providing an integrity test of the encrypted data in transit.

The checksum in the proprietary-tail on the sender's side is calculated <u>before</u> encryption and checked on the receiver's side <u>after</u> decryption. This checksum provides strong authentication because the static and dynamic keys used to encrypt the checksum are known only by the two communicating devices. (Strong authentication is where one can prove it knows a secret without revealing the secret.) By using a decrypted checksum that agrees with the packet proves the sender and receiver share the same key.

If the encrypted tail checksum is replaced with a secure hash function, such as the well-known MD5 algorithm, after the packet is encrypted, it becomes a digital signature. Where the tail checksum is encrypted with a static common key, it verifies that the dynamic public key originated from the sender, thus authenticating the sender. When the tail checksum is encrypted with a dynamic common key, it also verifies that the packet originated from the sender, authenticating that the packet originated from the sender.

Modifications for Use With LANS

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Referring again to FIG. 4B, the network security device 400' may be modified to protect a LAN 450 instead of a single host. These modifications are described below. In this illustrative embodiment, the network security device may protect a Class-C LAN having up to 254 clients (i.e., workstations 454, server 452, etc.), but other LAN types, such as Class-A and Class-B, are also contemplated by the invention.

During activation, the IP address burned into the flash memory 412 is the Class-C post of the client LAN's IP address. A default MAC address, such as Oxf may also be burned into the flash memory 412. This default MAC address is used in the static key generation. Recall that the MAC address is used in the static key seed generation.

A LAN-type network security device 400' may build a MAC-table 460 which contains its clients' IP and MAC addresses. This table serves two functions. First, it prevents IP spoofing of any of the LAN device's 400' clients. Thus, if a packet is received on the first interface 402 that does not have an IP or MAC address of one of the nodes in the LAN, that packet is dropped.

Second, it facilitates the delivery of packets to clients connected to the LAN 450. This permits packets to be sent from one protected client to another without the packet appearing at the second interface 408.

In short, a unique network security device has been disclosed. Finally, the above described embodiments of the invention are intended to be illustrative only. Numerous alternative embodiments may be devised by those skilled in the art without departing from the scope of the following claims.

CLAIMS

I claim:

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- A network security device configured to protect at least one particular node, the node having a first physical layer address and an Internet address and which communicates via a network, comprising:
 - a first interface connected to the at least one particular node and having said first physical layer address of the node;
 - a second interface connected to the network and having a second physical layer address, and
- 10 c. a processing circuit connected to said first and second interfaces, said processing circuit:
 - (1) for a packet received at said first interface from said one particular node and the packet having a header containing a source address that is the Internet address of the at least one particular node and said first physical layer address of said one particular node, the circuit configured to:
 - A. replace the first physical layer address contained in the received packet header with the second physical layer address;
 - B. determine a checksum verifying the packet and saving the determined checksum in the packet; and
 - C. encrypting the packet including the checksum, but leaving the Internet address unencrypted and its position in the packet header unchanged;
 - (2) for a packet received at said second interface from said network and the packet having a header containing a destination address that is the Internet address of the at least one particular node and said second physical layer address of said second interface, the circuit configured to:
 - A. decrypt the packet including a received checksum
 - B. determine if the checksum verifies the packet; and

C. replace the second physical layer address contained in the received packet header with said first physical layer address of said at least one particular node before said packet is transmitted to the at least one particular node, and leaving the Internet address unencrypted and its position in the packet header unchanged.

- 2. The network security device of claim 1, wherein the processing circuit is further configured to:
 - a. for a packet received at the first interface:

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- determine a second checksum verifying the encrypted packet;
 and
- (2) save the second checksum in an unencrypted portion of the packet; and
- b. for a packet received at the second interface:
 - (1) determine if the second checksum verifies the encrypted packet.
- 20 3. A method for transmitting a packet into a network comprising the steps of:
 - a. generating a packet having a header containing a first media access control (MAC) address, an IP address of a destination, and user data,
 - in a network security device which does no routing and is connected to said network, translating said first MAC address into a second MAC address of said network security device,
 - c. determining a checksum for the packet and saving the checksum in the packet
 - d. encrypting the user data and the checksum, but not the IP address and retaining as unchanged said IP address and its position in said header, and
 - e. transmitting said packet into said network.

4. The method of claim 3, further comprising the steps of:

- a. determining a second checksum for the packet, including the encrypted user data and checksum;
- b. saving the second checksum in an unencrypted portion of the packet.

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- 5. A network security device connected between: (1) a node having an Internet address and (2) a communication network, the device comprising:
 - (a) a first interface connected to at least one node, the first interface having a first media access control (MAC) address;
- 10 (b) a second interface connected to the communication network and having a second MAC address;
 - (c) a processor connected to the first and second interfaces, the processor configured to:
 - (1) receive a packet from the first interface, the packet having a transport layer header, a network layer header, and the first MAC address; the processor configured to:
 - replace the first MAC address with the second MAC address in the received packet,
 - B. determine a first checksum verifying the received packet and save the first checksum in the packet;
 - C. encrypt the received transport layer header and the first checksum, and to not encrypt the received network layer header; and to transmit the packet to the second interface; and

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(2) receive a packet from the second interface, the packet having an encrypted transport layer header and second checksum, an unencrypted network layer header, and the second MAC address; the processor configured to:

- replace the second MAC address with the first MAC address in the received packet;
- decrypt the packet including the transport layer header and the second checksum; and

- C. to transmit the packet to the first interface.
- 6. A method for generating a secret key for a network security device configured to protect at least one host, the secret key being unique to that network security device, the method comprising the steps of:
 - deriving a seed based on at least one of an Internet protocol (IP) and physical layer address of the at least one host; and
 - b. generating a random number based on the seed.
- 10 7. The method of claim 6, wherein the step of deriving the seed further comprises deriving the seed according to:

$$seed = IP + MAC_I + MAC_h + serial + time$$
 where:

IP = an IP address for the host;

- MAC₁ is a least significant portion of a physical layer address of the host; MAC_h is a most significant portion of host's physical layer address; serial is a serial number of the network security device; and time is a time the seed is derived.
- 20 8. The method of claim 6, wherein the step of generating a random number further comprises supplying the seed to a random number generator and using an output of the random number generator as the secret key.
- 9. A method for synchronizing a key exchange between a first network security device having a first dynamic key and a second network security device having a second dynamic key, said first and second dynamic keys having a predetermined lifespan and in which at least a first dynamic key of the first network security device may expire before being received by the second network security device, the method comprising:
- a. including with the first dynamic key a time stamp indicating a time that the dynamic key was transmitted and a time that the dynamic key was generated;

the second network security device receiving the first dynamic key,
 time stamp, and time the first dynamic key was generated;

- c. maintaining in the second network security device a database containing the received time stamp and time the first dynamic key was generated; and
- d. determining a difference between a time indicated in the received time stamp and a current time; and
- e. correcting the time that the first dynamic key was generated by the determined difference.

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10. A network security device configured to protect a local area network (LAN) having a plurality of nodes, each node having a physical layer address and an Internet address, the LAN being in communication with a second network, the network security device comprising:

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- a first interface connected to the LAN;
- a second interface connected to the second network and having a second physical layer address, and
- c. a processing circuit connected to said first and second interfaces, said processing circuit including a table of physical layer and Internet addresses of each of the plurality of nodes in the LAN;
 - (1) for a packet received at said first interface from one of the plurality of nodes in the LAN and the packet having a header containing a source address that is the Internet address of the one of the plurality of nodes, a physical layer address of the one of the plurality of nodes, and a destination address, the circuit configured to:

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A. determine if the destination address is an Internet address of another node in the LAN;

- if so, transmit the packet to the destination node using the first interface;
- ii. if not, then replace the first physical layer address contained in the received packet header with the

second physical layer address; and encrypt the packet leaving the Internet address unencrypted and its position in the packet header unchanged;

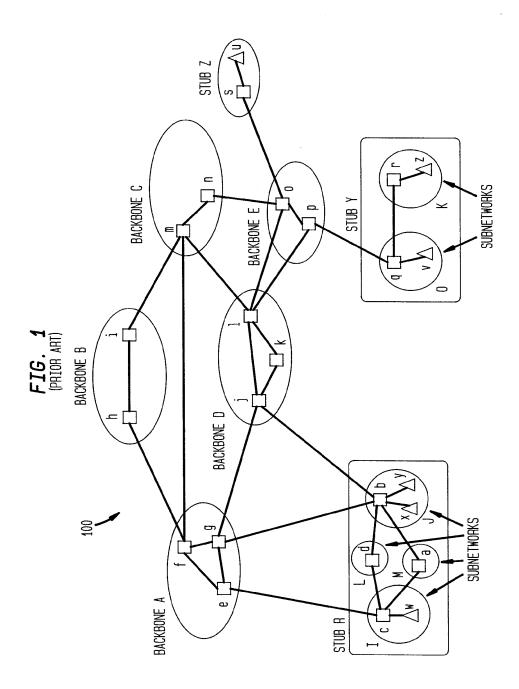
- (2) for a packet received at said second interface from said network and the packet having a header containing a destination address that is the Internet address of one of the plurality of nodes and said second physical layer address of said second interface, the circuit configured to:
 - A. decrypt the packet;

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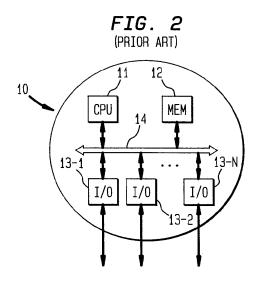
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- B. replace the second physical layer address contained in the received packet header with said physical layer address of said one of the plurality of nodes before the packet is transmitted to the one of the plurality of particular nodes.
- 11. The network security device of claim 10, wherein when a packet is received on the first interface, the processing circuit is further configured to compare at least one of the physical layer address and the Internet address in the received packet with the physical layer and Internet addresses in the table to determine whether the packet originated from one of the plurality of nodes and, if not, dropping the packet.



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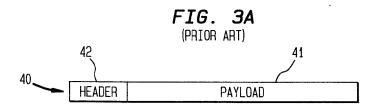
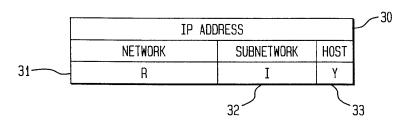


FIG. 3B (PRIOR ART)



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FIG. 4A

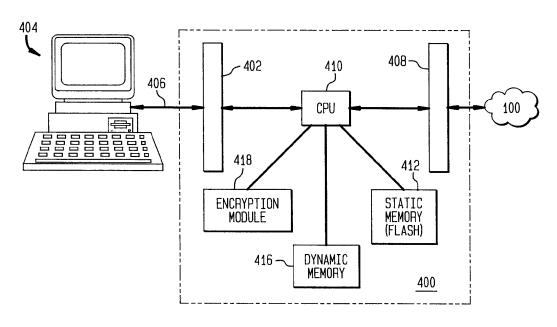
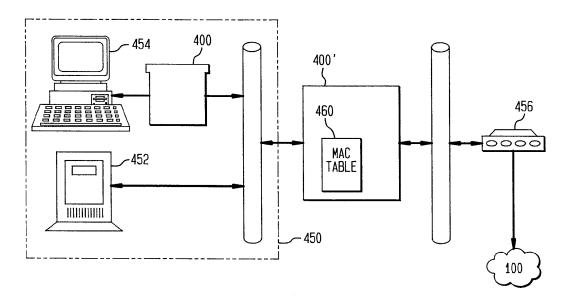
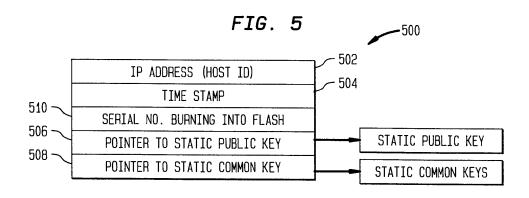


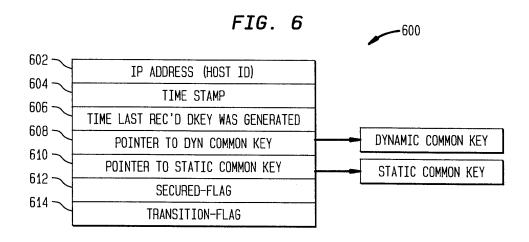
FIG. 4B



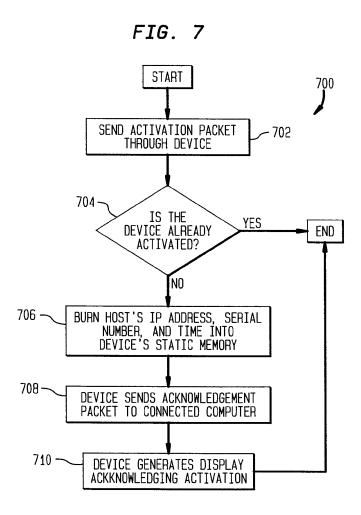
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4/9

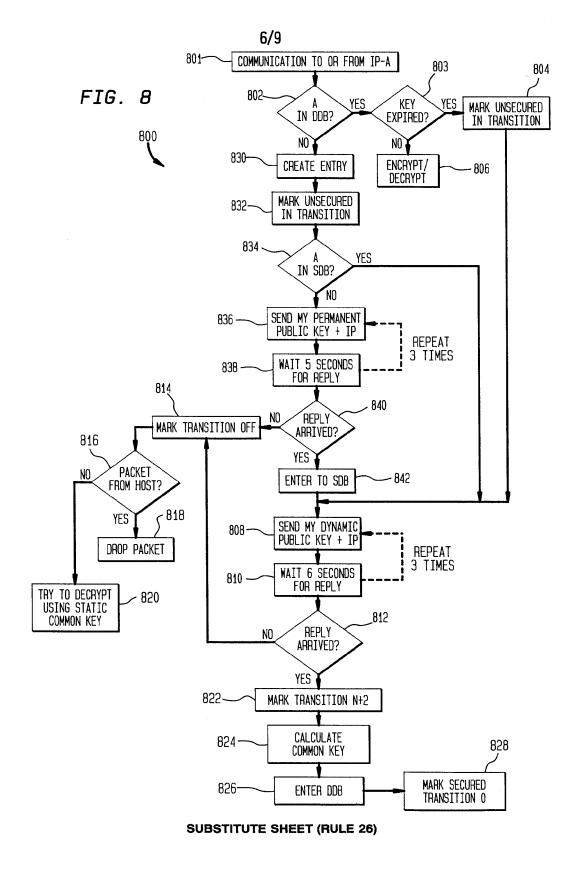




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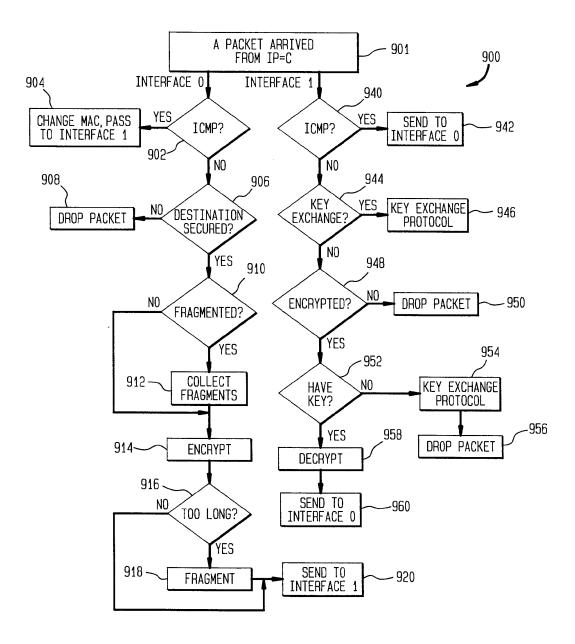


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7/9

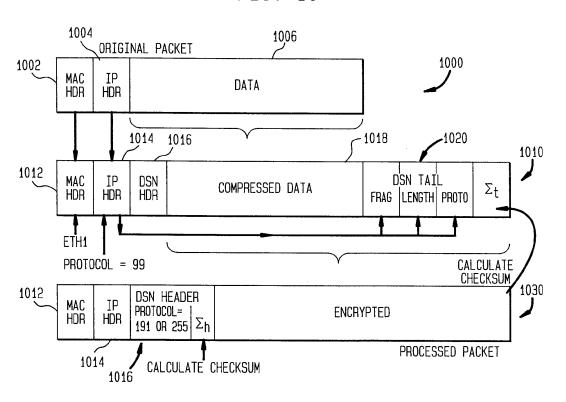
FIG. 9



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8/9

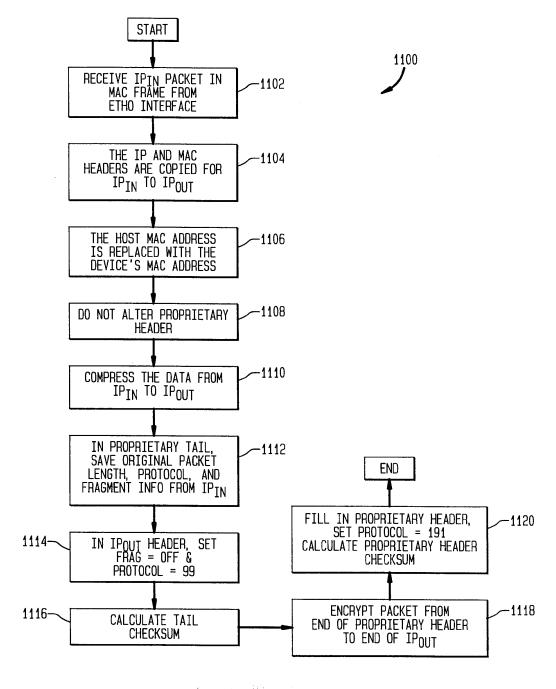
FIG. 10



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9/9

FIG. 11



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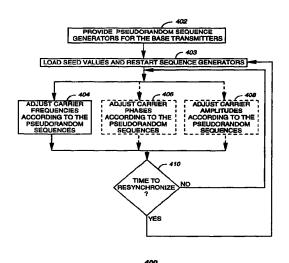
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(57) Abstract

In a messaging system having a plurality of simulcasting base transmitters (104), a plurality of pseudorandom sequence generators (216) are provided (402) for the plurality of base transmitters. The plurality of pseudorandom sequence generators are arranged (403) to ensure that they generate a plurality of pseudorandom sequences having sub-sequences that are different from one another during concurrent transmissions by the plurality of base transmitters. A cancellation-affecting parameter of the plurality of base transmitters is adjusted (404, 406, 408) in accordance with the plurality of pseudorandom sequences during the simulcast transmission from the plurality of base transmitters to limit intervals of carrier cancellation. In addition, a controller (512) controls (1106) at least two transmitters (702) to transmit at least two simulcast signals (902, 904) during a time period. The at least two simulcast signals produce intersymbol interference at a receiver. At least one of the transmitters changes (1108) its output amplitude during a portion of the time period, thereby altering the intersymbol interference during the portion of the time period.



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LIMITING AN INTERVAL OF CARRIER CANCELLATION AND REDUCING ERRORS CAUSED BY INTERSYMBOL INTERFERENCE DURING A SIMULCAST TRANSMISSION

Field of the Invention

This invention relates in general to radio communication systems, and more specifically to a method and apparatus in a messaging system for limiting an interval of carrier cancellation and for reducing errors caused by intersymbol interference during a simulcast transmission.

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Background of the Invention

Radio messaging systems have utilized simulcast transmissions from multiple transmitters for providing radio coverage to large geographic areas. During a simulcast transmission a receiver positioned midway between two transmitters often can receive signals from both transmitters. The resultant instantaneous sum of the two signals depends upon their relative phase, and can be either larger or smaller than either signal alone. For example, if the two signals are substantially equal in amplitude and phase at the receiver, their resultant sum will be about twice the amplitude of either signal alone. If, however, the two signals are substantially equal in amplitude and 180 degrees out of phase, their resultant sum can be so small as to be undetectable by the receiver, due to destructive cancellation of the two signals.

Modern messaging systems utilize forward error correcting codes and bit interleaving to allow messages to be transmitted successfully in the presence of brief fades and noise bursts. Thus, an error-free message can be received even in the presence of intervals of destructive cancellation, provided that the intervals of destructive cancellation are sufficiently brief. To ensure that the intervals of destructive cancellation are sufficiently brief, the prior art messaging systems have employed a technique of permanently offsetting the carrier frequencies of adjacent transmitters with respect to one another by a small, fixed amount, e.g., 15 to 100 Hz.

A problem with the technique of permanently offsetting the carrier frequencies of adjacent transmitters is that it requires additional system

planning and effort in setting up the radio messaging system. Furthermore, the technique can cause difficulties when adding new transmitters to an existing system, because the frequency offsets of many of the existing transmitters may have to be readjusted. In addition, some specific frequency offsets between adjacent transmitters, e.g., 200 Hz, are known to cause a higher word error rate, and should be avoided.

Good simulcast transmission has always required some form of delay equalization or launch time synchronization to ensure that the transmissions from different transmitters begin at the same time. For low speed data, having the transmissions begin at the same time has usually been sufficient. For the high speed data which is becoming more prevalent today, having the transmissions begin at the same time is necessary, but not sufficient. The reason is that differential transmission delay introduced in the air links can become a significant fraction of the symbol period when the symbol rate is high. When differential transmission delay becomes a significant fraction of the symbol period, intersymbol interference can occur when two or more simulcast signals arrive at the receiver with similar amplitudes. Such intersymbol interference can cause a high error rate in the received signal.

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Thus, what is needed is a method and apparatus for limiting the intervals of destructive cancellation during simulcast transmissions. The method and apparatus preferably will limit the intervals of destructive cancellation without utilizing the prior art technique of permanently offsetting the carrier frequencies of adjacent transmitters with respect to one another.

What is further needed is a method and apparatus that can reduce errors caused by intersymbol interference during a simulcast transmission when two or more simulcast signals are received at similar amplitudes with different transmission delays. The method and apparatus preferably will operate without requiring a custom tuning adjustment during installation and system setup.

Summary of the Invention

An aspect of the present invention is a method in a messaging system having a plurality of base transmitters, the method for limiting an interval of carrier cancellation at a reception point during a simulcast

transmission. The method comprises the step of providing a plurality of pseudorandom sequence generators for the plurality of base transmitters, the plurality of pseudorandom sequence generators arranged to ensure that they generate a plurality of pseudorandom sequences having subsequences that are different from one another during concurrent transmissions by the plurality of base transmitters. The method further comprises the step of adjusting a cancellation-affecting parameter of the plurality of base transmitters in accordance with the plurality of pseudorandom sequences during the simulcast transmission from the plurality of base transmitters.

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Another aspect of the present invention is a base transmitter in a messaging system having a plurality of base transmitters, the base transmitter for limiting an interval of carrier cancellation at a reception point during a simulcast transmission. The base transmitter comprises a transmitter element for transmitting a message, and a processing system coupled to the transmitter element for controlling the transmitter element to transmit the message. The base transmitter further comprises an input interface coupled to the processing system for receiving the message; and a pseudorandom sequence generator coupled to the transmitter element, the pseudorandom sequence generator arranged to ensure that it generates a pseudorandom sequence having sub-sequences that are different from those generated in other ones of the plurality of base transmitters during concurrent transmissions by the plurality of base transmitters. The transmitter element is arranged such that the pseudorandom sequence generator adjusts a cancellation-affecting parameter of the transmitter element in accordance with the pseudorandom sequence during the simulcast transmission from the base transmitter.

A third aspect of the present invention is a method in a messaging system having a plurality of base transmitters, the method for limiting an interval of carrier cancellation at a reception point during a simulcast transmission. The method comprises the step of providing a plurality of pseudorandom sequence generators for the plurality of base transmitters, the plurality of pseudorandom sequence generators arranged to generate a plurality of pseudorandom sequences having sub-sequences that have more than a predetermined probability of being different from one another during concurrent transmissions by the plurality of base

transmitters. A parameter of the plurality of pseudorandom sequences is optimized according to a characteristic of a communication protocol utilized by the messaging system. The method further comprises the step of adjusting a cancellation-affecting parameter of the plurality of base transmitters in accordance with the plurality of pseudorandom sequences during the simulcast transmission from the plurality of base transmitters.

A fourth aspect of the present invention is a method in a wireless communication system for reducing errors caused by intersymbol interference in at least two simulcast signals transmitted during a time period. The at least two simulcast signals are received at similar amplitudes and have different transmission delays with respect to one another. The method comprises the steps of transmitting the at least two simulcast signals from a corresponding at least two transmitters, and changing an output amplitude of at least one of the at least two transmitters during a portion of the time period, thereby altering the intersymbol interference during the portion of the time period.

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A fifth aspect of the present invention is a transmitter in a wireless communication system for reducing errors caused by intersymbol interference in at least two simulcast signals transmitted during a time period. The at least two simulcast signals are received by a receiver at similar amplitudes and have different transmission delays with respect to one another. The transmitter comprises a transmitter element for transmitting a first simulcast signal sent simultaneously with at least a second simulcast signal from another transmitter, and a modulator coupled to the transmitter element for changing an output amplitude of the transmitter during a portion of the time period, thereby altering the intersymbol interference at the receiver during the portion of the time period.

A sixth aspect of the present invention is a controller in a wireless communication system for reducing errors caused by intersymbol interference in at least two simulcast signals transmitted during a time period. The at least two simulcast signals are received at similar amplitudes and have different transmission delays with respect to one another. The controller comprises a network interface for receiving a message from a message originator, and a processing system coupled to the network interface for processing the message. The controller further comprises a base station interface coupled to the processing system for

controlling a transmitter to transmit one of the at least two simulcast signals. The processing system is programmed to control the transmitter to change an output amplitude of the transmitter during a portion of the time period, thereby altering the intersymbol interference during the portion of the time period.

Brief Description of the Drawings

- FIG. 1 is an electrical block diagram of a messaging system in accordance with the present invention.
- FIG. 2 is an electrical block diagram of an exemplary inplementation of a base transmitter in accordance with the present invention.
- FIG. 3 is a diagram depicting amplitude and relative phase of two carriers offset in frequency in accordance with the present invention.
- FIG. 4 is a flow chart depicting operation of the messaging system in accordance with the present invention.
- FIG. 5 is an electrical block diagram of an exemplary wireless communication system in accordance with the present invention.
- FIG. 6 is an electrical block diagram of an exemplary controller in accordance with the present invention.
- FIG. 7 is an electrical block diagram of an exemplary base station in accordance with the present invention.
- FIG. 8 is a timing diagram depicting intersymbol interference in a prior art wireless communication system.
- FIG. 9 is a timing diagram depicting reduced intersymbol interference in the wireless communication system in accordance with the present invention.
- FIG. 10 is an exemplary protocol diagram in accordance with the present invention.
- FIG. 11 is a flow diagram depicting operation of the exemplary wireless communication system in accordance with the present invention.

Detailed Description of the Drawings

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Referring to FIG. 1, an electrical block diagram of a messaging system in accordance with the present invention comprises a plurality of

subscriber units 102, which communicate by radio with a fixed portion of the radio system, comprising a plurality of base transmitters 104 and a plurality of controllers 110. The base transmitters 104 are coupled via communication links 106 to the plurality of controllers 110 for control by and communication with the plurality of controllers 110 utilizing wellknown techniques. The controllers 110 are coupled to a home controller 120 via communication links 122, 124, and via a conventional communication network 108 for receiving selective call messages from the home controller 120. The home controller 120 and the controllers 110 preferably communicate by utilizing a well-known protocol, e.g., the Telocator Network Paging Protocol (TNPP), the Wireless Messaging transfer protocol (WMtp™), or the InterPaging Networking Protocol (IPNP). It will be appreciated that, alternatively, the home controller 120 and the controller 110 can be collocated. The home controller 120 is preferably coupled via telephone links 126 to a public switched telephone network 112 (PSTN) for receiving the messages from message originators utilizing, for example, a telephone 114 or a personal computer 116 to originate the messages. It will be appreciated that, alternatively, other types of communication networks, e.g., packet switched networks, local area networks, and the Internet can be utilized as well for transporting originated messages to the home controller 120. The hardware of the home controller 120 is preferably similar to the Wireless Messaging Gateway (WMG™) Administrator! paging terminal, while the hardware of the controllers 110 is preferably similar to that of the RF-Conductor!™ message distributor, both manufactured by Motorola, Inc. of Schaumburg, IL. The hardware of the base transmitters 104 is preferably similar to that of the Nucleus® and RF-Orchestra!® transmitters manufactured by Motorola, Inc. It will be appreciated that other similar hardware can be utilized as well for the home controller 120, the controllers 110, and the base transmitters 104. It will be further appreciated that the present invention can be applied to both one-way and two-way selective call messaging systems.

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The protocol utilized for transmitting the messages between the base transmitters 104 and the subscriber units 102 is preferably similar to Motorola's well-known FLEXTM family of digital selective call signaling protocols. These protocols utilize well-known error detection and error correction techniques and are therefore tolerant to bit errors occurring during transmission,

provided that the bit errors are not too numerous in any one code word. It will be appreciated that other similar messaging protocols can be used as well.

Referring to FIG. 2, an electrical block diagram depicts an exemplary inplementation of the base transmitter 104 in accordance with the present invention. The base transmitter 104 comprises an antenna 204 for emitting a radio signal comprising a message. The base transmitter 104 further comprises a conventional transmitter element 208 coupled to the antenna 204 for transmitting the message, and a processing system 206 coupled to the transmitter element 208 for controlling the transmitter element 208 to transmit the message. The processing system is further coupled to a conventional pseudorandom sequence generator 216, which is also coupled to the transmitter element 208. The pseudorandom sequence generator 216 is arranged to ensure that it generates a pseudorandom sequence having sub-sequences that are different from those generated in other ones of the plurality of base transmitters during concurrent transmissions by the plurality of base transmitters, as described further below. Preferably, the pseudorandom sequence generator 216 is further arranged to provide a pseudorandom sequence identical to that of other base transmitters of the plurality of base transmitters, but initialized, concurrently with the other base transmitters, with a seed value different from that of the other base transmitters. It will be appreciated that, alternatively, the pseudorandom sequence generator 216 can be arranged to provide a pseudorandom sequence that is different from that of other base transmitters 104 of the plurality of base transmitters by, for example, enabling different feedback taps on the pseudorandom sequence generators 216 associated with different base transmitters 104. In addition, the pseudorandom sequence generator 216 preferably has at least a predetermined minimum number of stages, e.g., 20 stages. This preference facilitates allowing the base transmitters 104 to be concurrently initialized with different seed values 226 derived, for example, from the serial number of the base transmitters 104. It also will be appreciated that, alternatively, the pseudorandom sequence generator 216 can be incorporated into the processing system 206, where its functions can be performed in software.

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The transmitter element 208 is arranged such that the pseudorandom sequence generator 216 adjusts a cancellation-affecting parameter of the transmitter element 208 in accordance with the

pseudorandom sequence during a simulcast transmission from the base transmitter 104. More specifically, the transmitter element 208 preferably includes a conventional frequency modulator (not shown) coupled to the pseudorandom sequence generator 216 such that the pseudorandom sequence generator 216 adjusts the carrier frequency of the base transmitter 104 in accordance with the pseudorandom sequence. In the simplest case, the pseudorandom sequence generator 216 cooperates with the transmitter element 208 to adjust the carrier frequency of the base transmitter 104 to one of two levels, e.g., ±50 Hz, about a predetermined nominal carrier frequency. It will be appreciated that, alternatively, the pseudorandom sequence generator 216 and the transmitter element 208 can be arranged to adjust the carrier frequency to one of N predetermined levels in accordance with the pseudorandom sequence, N being an integer greater than unity. It will be further appreciated that, alternatively, the transmitter element 208 can be arranged such that another cancellation-affecting parameter of the base transmitter 104, e.g., the carrier phase or the carrier amplitude, is adjusted in accordance with the pseudorandom sequence, through well-known techniques. It also will be appreciated that, alternatively, the transmitter element 208 can be arranged such that the pseudorandom sequence generator 216 adjusts at least two cancellation-affecting parameters selected from a group of cancellation-affecting parameters consisting of the carrier frequency, the carrier phase, and the carrier amplitude. In addition, it will be appreciated that the pseudorandom sequence may have to be filtered to prevent instantaneous shifts of the cancellation-affecting parameter(s).

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Preferably, the pseudorandom sequence generator 216 is further arranged to optimize a parameter of the plurality of pseudorandom sequences according to a characteristic of the communication protocol utilized by the messaging system, such that the intervals of destructive cancellation will exist only long enough to potentially destroy, i.e., change the value of, less than a predetermined number of bits, e.g., two bits, of a given (interleaved) code word, which will fall within the error correction capability of the protocol. This essentially moves the bit errors around, distributing them randomly such that the forward error correction is very likely to correct all the errors caused by the intervals of destructive cancellation.

For example, consider the FLEX protocol, which uses (32,21) BCH code words interleaved such that there are 5 ms intervals between bits corresponding to the same code word. Each block of interleaved code words lasts 160 ms. Consider the case of no dithering and no frequency offsets and a 1 Hz frequency error between two adjacent FM transmitters. The interval of destructive cancellation may last on the order of 100 ms. which will destroy most code word bits (exceeding the forward error correction capability) in 1 or 2 interleaved blocks of the transmission, yet leaving another 4 or 5 interleaved blocks error free. In this condition little benefit is derived from the forward error correction. Now consider the use of pseudorandom frequency dithering in accordance with the present invention, e.g., \pm 50 Hz about a nominal frequency, with the duration of each dither set to 7.5 ms, for example. An interval of destructive cancellation lasting 7.5 ms and repeating no more frequently than every 160 ms will destroy, on average, 3/4 bit from each code word. If a random phase difference between two signals generates an interval of destructive cancellation with a probability of 0.1 (as derived further below), then, on average, a destructive phase condition will occur 1.6 times per code word in each block, advantageously allowing a greatly increased benefit from forward error correction coding.

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Again referring to FIG. 2, the processing system 206 is further coupled to a conventional clock 202 for generating a timing signal for the base transmitter 104. The accuracy of the timing signal preferably is sufficient to maintain synchronization of the pseudorandom sequence generator 216 within a small time tolerance, e.g., 100 microseconds, between resynchronizations of the pseudorandom sequence generator 216. It will be appreciated that, alternatively, the timing signal can be derived from a Global Positioning Satellite (GPS) receiver. The processing system 206 is also coupled to a conventional input interface 214 for receiving the message via the communication link 106.

The processing system 206 comprises a conventional processor 210 and a conventional memory 212. The memory 212 includes locations for storing messages 222 received through the input interface 214 and, preferably, a pseudorandom sequence seed value 226 derived, for example, from a serial number uniquely assigned by the factory to the base transmitter 104. The memory 212 also includes software elements

for message processing 224 and pseudorandom sequence synchronization 228 in accordance with the present invention.

Referring to FIG. 3, a diagram 300 depicting amplitude (represented by length) and relative phase (θ) of two carriers A1, A2 offset in frequency in accordance with the present invention and received by a receiver at a reception point between two of the base transmitters 104. Assume, for example, that the frequency of the carrier A2 is higher than that of the carrier A1. The result is that the phase of A2 is changing faster with time than that of A1. Periodically, the relative phase θ at the receiver is such that A2 enters the shaded area defined as the zone of destructive cancellation 302. The zone of destructive cancellation 302 preferably is defined, by way of example, to correspond to

 $0.9\pi < \theta < 1.1\pi$.

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As θ traverses 2π for each full revolution, one can conclude that for a fixed, nonzero frequency difference between A1 and A2 the probability that A2 is in the zone of destructive cancellation 302 at a randomly chosen instant of time is P=0.1. For a pseudorandomly varied frequency difference between A1 and A2 in accordance with the present invention the probability that A2 is in the zone of destructive cancellation 302 at a randomly chosen instant of time is also P=0.1.

The instantaneous power at the receiver is

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$$P(t) = (A1 + A2\cos(\theta))^{2} + (A2\sin(\theta))^{2}.$$

If A1 = A2 = 1, the average power is 2.0. At 0.9π and 1.1π the instantaneous power is approximately 0.1. Thus, within the defined zone of destructive cancellation 302 the instantaneous power is approximately 13 dB or more below the average power. While there is no way to prevent the two carriers A1 and A2 from entering the zone of destructive cancellation 302, it is highly desirable to minimize their stay in the zone, as is advantageously accomplished in accordance with the present invention, as described further below. It will be appreciated that, alternatively, other exemplary ranges of θ can be utilized to define the zone of destructive cancellation 302.

FIG. 4 is a flow chart 400 depicting operation of the messaging system in accordance with the present invention. The flow chart 400 begins with providing 402 the pseudorandom sequence generators 216 for the base transmitters 104. After the messaging system is powered up, the processing systems 206 access the seed values 226 corresponding to each of the base transmitters 104. The processing systems 206 then load 403 the seed values 226 and simultaneously restart the pseudorandom sequence generators 216. The restarting of the pseudorandom sequence generators 216 is preferably synchronized by the communication protocol to recur, for example, at the top of each hour. Concurrently restarting the pseudorandom sequence generators 216 periodically in this manner with different seed values 226 advantageously allows identical-sequence pseudorandom sequence generators 216 to be utilized for the base stations, while ensuring that the pseudorandom sequences contain subsequences that are different from one another during concurrent transmissions by the base transmitters 104 (due to the sequences being offset from one another by the different seed values 226). It will be appreciated that, alternatively, the pseudorandom sequence generators 216 can comprise different-sequence pseudorandom sequence generators to ensure that the pseudorandom sequences generated thereby are different from one base transmitter 104 to the next.

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Next, the processing systems 206 preferably begin adjusting the carrier frequencies 404 of the corresponding transmitter elements 208 according to the pseudorandom sequences. Alternatively, the processing systems 206 can adjust another cancellation-affecting parameter, such as the carrier phases 406 and/or the carrier amplitudes 408 in addition to, or in lieu of, adjusting the carrier frequencies. For the case of frequency or phase adjustment, the processing systems 206 preferably are programmed to ensure a uniformly distributed phase between 0 and 2π . The processing systems 206 also check 410 whether it is time to resynchronize the pseudorandom sequence generators 216. If so, the flow returns to step 403. If not, the flow returns to the appropriate ones of the adjusting steps 404, 406 and 408.

Simulations in accordance with the present invention have demonstrated that by continuously adjusting a cancellation-affecting parameter of the base stations in accordance with the present invention, the intervals of destructive carrier cancellation advantageously are

limited in length and are randomly dispersed throughout the interleaved transmission blocks of the communication protocol, thereby substantially reducing the word error rate. The simulations have further demonstrated that no other carrier frequency offsetting technique is needed to meet performance objectives. In addition, by randomizing the starting points of the pseudorandom sequences through the use of seed values derived from a random number source, which can include the base station serial numbers, no additional system planning effort is required for adjusting the base stations relative to one another to limit carrier cancellation.

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FIG. 5 is an electrical block diagram of an exemplary wireless communication system in accordance with the present invention, comprising a fixed portion 502 including a controller 512 and a plurality of base stations 516, the wireless communication system also including a plurality of receivers 522. The base stations 516 preferably communicate with the receivers 522 utilizing conventional radio frequency (RF) signals for sending simulcast transmissions in accordance with the present invention, as will be explained further below. The base stations 516 are coupled by communication links 514 to the controller 512, which controls the base stations 516.

The hardware of the controller 512 is preferably a combination of the Wireless Messaging Gateway (WMGTM) Administrator! paging terminal, and the RF-Conductor!TM message distributor manufactured by Motorola, Inc., and includes software modified in accordance with the present invention. The base stations 516 comprise a transmitter preferably similar to the RF-Orchestra! transmitter, modified in accordance with the present invention, and can include, in two-way wireless communication systems, the RF-Audience!TM receiver manufactured by Motorola, Inc. The receivers 522 are preferably similar to the Advisor GoldTM and PagefinderTM wireless communication units, also manufactured by Motorola, Inc. It will be appreciated that other similar hardware can be utilized as well for the controller 512, the base stations 516, and the receivers 522.

Each of the base stations 516 transmits RF signals to the receivers 522 via an antenna 518. The RF signals transmitted by the base stations 516 to the receivers 522 (outbound messages) comprise selective call addresses identifying the receivers 522, and voice and data messages originated by a caller, as well as commands originated by the controller 512 for adjusting operating parameters of the radio communication system.

The controller 512 preferably is coupled by telephone links 501 to a public switched telephone network (PSTN) 510 for receiving selective call message originations therefrom. Selective call originations comprising voice and data messages from the PSTN 510 can be generated, for example, from a conventional telephone 511 or a conventional computer 517 coupled to the PSTN 510. It will be appreciated that, alternatively, other types of communication networks, e.g., packet switched networks, the Internet, and local area networks, can be utilized as well for transporting originated messages to the controller 512.

The over-the-air protocol utilized for the transmissions is preferably selected from Motorola's well-known FLEXTM family of digital selective call signaling protocols. These protocols utilize well-known error detection and error correction techniques and are therefore tolerant to bit errors occurring during transmission, provided that the bit errors are not too numerous. It will be appreciated that other suitable protocols can be used as well. It will be further appreciated that, while one embodiment for practicing the present invention is a one-way wireless communication system, the present invention is applicable also to a two-way wireless communication system.

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FIG. 6 is an electrical block diagram depicting an exemplary controller 512 in accordance with the present invention. The controller 512 comprises a network interface 618 for receiving a message from a message originator via the telephone links 501. The network interface 618 is coupled to a processing system 610 for controlling and communicating with the network interface 618. The processing system is coupled to a base station interface 604 for controlling and communicating with the base stations 516 via the communication links 514. The processing system 610 is also coupled to a conventional clock 630 for providing a timing signal to the processing system 610. The processing system 610 comprises a conventional computer 612 and a conventional mass medium 614, e.g., a magnetic disk drive, programmed with information and operating software in accordance with the present invention. The mass medium 614 comprises a subscriber database 620, including information about the receivers 522 controlled by the controller 512. The mass medium 614 also includes a message processing element 622 for programming the processing system 610 to process messages for the receivers 522 in a conventional manner. In accordance with the present invention, the mass medium 614 also includes a transmitter

output amplitude control element 624 for programming the processing system to control the transmitter 702 (FIG. 7) of the base stations 516 to change an output amplitude of the transmitter 702 during a portion of a time period during which simulcast transmissions are sent, thereby altering the intersymbol interference during the portion of the time period.

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FIG. 7 is an electrical block diagram of an exemplary base station 516 in accordance with the present invention. The base station 516 comprises the antenna 518 for radiating a signal comprising a message. The antenna 518 is coupled to a transmitter 702 for transmitting the message. The transmitter 702 preferably comprises a conventional frequency shift keyed (FSK) transmitter element 708 for transmitting a first simulcast signal sent simultaneously with at least a second simulcast signal from another transmitter 702 (as coordinated by the controller 512 through well-known techniques). It will be appreciated that, alternatively, other types of transmitter elements for demodulating other types of modulated signals can be utilized as well for the transmitter element 708. The transmitter 702 further comprises a conventional amplitude modulator 703 coupled to the transmitter element 708 for changing an output amplitude of the transmitter 702 during a portion of the time period of the first simulcast signal, thereby altering the intersymbol interference at the receiver during the portion of the time period. The transmitter 702 is coupled to a processing system 706 for processing the message and for controlling the transmitter 702 in accordance with the present invention. A conventional controller interface 714 preferably is also coupled to the processing system 706 for interfacing with the controller 512 via the communication link 514 through well-known techniques. In addition, a conventional clock 707 is coupled to the processing system 706 for providing a timing signal thereto.

The processing system 706 comprises a conventional processor 710 and a conventional memory 712. The memory 712 comprises software elements and other variables for programming the processing system 706 in accordance with the present invention. The memory 712 includes a transmitter control element 722 for controlling the transmitter 702 through well-known techniques. In addition, the memory 712 includes a message processing element 724 for programming the processing system 706 to process the message in a conventional manner. The memory 712

further comprises a transmitter output amplitude control element 726 for cooperating with the modulator 703 to control the output amplitude of the transmitter 702 in accordance with the present invention, as described further below.

FIG. 8 is an exemplary timing diagram 800 depicting intersymbol interference in a prior art wireless communication system. The diagram 800 depicts amplitude versus time of a first signal 802 from a first simulcast transmitter and a second signal 804 from a second simulcast transmitter, the second signal 804 identical to, but delayed with respect to, the first signal 802. When the first and second signals 802, 804 are received by a receiver at nearly the same amplitudes, e.g., less than 4 dB of difference, the received signal 806 can comprise indeterminant areas 808 where the received bit cannot be decoded. When the indeterminant areas 808 occupy more than about 50% of the symbol period (corresponding to a differential delay of 25% of the symbol period), receiver sensitivity begins to be reduced slightly. When the indeterminant areas increase to 100% of the symbol period), receiver sensitivity is reduced to zero.

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FIG. 9 is an exemplary timing diagram 900 depicting reduced intersymbol interference in the wireless communication system in accordance with the present invention. The diagram 900 depicts amplitude versus time of a first signal 902 and a second signal 904. A "nominal" value of the amplitude of the first and second signals is represented by the dashed lines 910. Note that during a portion of the time period of the first and second signals 902, 904, the amplitude is changed above and/or below the nominal value, preferably by adjusting the output amplitude of the transmitter 702 by the modulator 703 under control of the processing system 706, in accordance with the present invention. When the nominal values of the first and second signals 902, 904 would be received by a receiver at nearly the same amplitudes, the advantageous effect of changing the output amplitudes of the first and second signals 904, 904 is demonstrated by the decoded signal 906. Note that the indeterminant areas 908 advantageously are reduced in number compared to the diagram 800. The reason for the reduced number of indeterminant areas 908 is that when the amplitudes of the first and second signals 902, 904 are different by more than about 4 dB, receiver

"capture" causes one of the signals to dominate, and the intersymbol interference goes away.

FIG. 10 is an exemplary protocol diagram 1000 in accordance with the present invention. This protocol is used by the controller 912 to communicate to the base station 916 how the transmitter 702 is to change its output amplitude during simulcast transmissions. The diagram 1000 comprises a synchronization portion 1002 for synchronizing the base station 916 with the communications of the controller 912, using wellknown techniques. The diagram 1000 further comprises a type indication 1004 for indicating the message type, e.g., output amplitude control command. In addition, the diagram 1000 includes a command 1006 for controlling the output amplitude configuration of the base station transmitter 702. This protocol advantageously allows the base station 916 to be reconfigured from time to time with regard to how it changes the output amplitude of its transmissions in accordance with the present invention. As an alternative, the base station 916 can be preprogrammed, either in the field or during manufacture, with fixed instructions as to how the transmitter 702 should change its output amplitude during simulcast transmissions.

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FIG. 11 is a flow diagram 1100 depicting operation of the exemplary wireless communication system in accordance with the present invention. The diagram 1100 preferably begins with the controller 912 communicating with the base station 916 to control 1102 the transmitter 1102 to change the output amplitude of its transmissions during a portion of the time period of each simulcast transmission. Simulations have shown that relatively small changes in the output amplitude, e.g. about ±0.5 dB, can produce a sizable, e.g., two to one, improvement in word error rate. Alternatively, the transmitter 1102 can be arranged 1104 during installation and setup, or during manufacture, to change the output amplitude during a portion of the time period of each simulcast transmission. However the transmitter 1102 is programmed, the controller 912 then controls the base stations 916 to send a simulcast transmission. The transmitter 1102 then performs 1106 according to its programmed instructions for changing the output amplitude. Preferably, the transmitter 1102 changes the output amplitude as a predetermined function of time, in synchronism with the symbols transmitted by the transmitter 1102.

In one embodiment, the transmitter 1102 is arranged such that the output amplitude of a central portion of each symbol does not change, while the output amplitude of non-central portions of the symbol do change. This technique exploits the fact that with normally encountered differential delay characteristics, intersymbol interference occurs primarily in the non-central portions of the symbols. In another embodiment, the transmitted signal comprises an error correcting code that can correct a predetermined number of errors in a code block, and the transmitter 1102 changes the output amplitude according to a pseudorandom sequence having a predetermined number of states, e.g., two states, during a transmission of the code block. In this embodiment, it is preferred that the transmitters 1102 in the wireless communication system utilize pseudorandom sequences that are offset from one another, so that different transmitters 1102 do not adjust their output amplitudes identically at every step of the sequence. In yet another embodiment, the transmitter 1102 is arranged to repeat a change to the output amplitude for a number of symbols, wherein the number of symbols is determined from an encoding characteristic employed by the wireless communication system, e.g., the length of an error correcting code block.

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Regardless which embodiment in accordance with the present invention is used, an overall objective is to reduce errors due to intersymbol interference. When used with an error correcting code, the present invention often can reduce the number of received errors sufficiently to change an uncorrectable number of errors into a correctable number of errors, thereby advantageously salvaging a message which would otherwise have been corrupted.

Thus, it should be clear from the preceding disclosure that the present invention provides a method and apparatus for limiting the intervals of destructive cancellation during simulcast transmissions. The method and apparatus advantageously limits the intervals of destructive cancellation without utilizing the administratively difficult prior art technique of permanently offsetting the carrier frequencies of adjacent transmitters with respect to one another. In addition, the present invention provides a method and apparatus that advantageously reduces errors caused by intersymbol interference during a simulcast transmission when two or more simulcast signals are received at similar amplitudes with different transmission delays. The method and

apparatus operates without requiring a custom tuning adjustment during installation and system setup.

Many modifications and variations of the present invention are possible in light of the above teachings. Thus, it is to be understood that, within the scope of the appended claims, the invention can be practiced other than as described herein above for the exemplary embodiments.

What is claimed is:

CLAIMS

1. A method in a messaging system having a plurality of base transmitters, the method for limiting an interval of carrier cancellation at a reception point during a simulcast transmission, the method comprising the steps of:

providing a plurality of pseudorandom sequence generators for the plurality of base transmitters, the plurality of pseudorandom sequence generators arranged to ensure that they generate a plurality of pseudorandom sequences having sub-sequences that are different from one another during concurrent transmissions by the plurality of base transmitters; and

adjusting a cancellation-affecting parameter of the plurality of base transmitters in accordance with the plurality of pseudorandom sequences during the simulcast transmission from the plurality of base transmitters.

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2. A base transmitter in a messaging system having a plurality of base transmitters, the base transmitter for limiting an interval of carrier cancellation at a reception point during a simulcast transmission, the base transmitter comprising:

a transmitter element for transmitting a message;

a processing system coupled to the transmitter element for controlling the transmitter element to transmit the message;

an input interface coupled to the processing system for receiving the message; and

a pseudorandom sequence generator coupled to the transmitter element, the pseudorandom sequence generator arranged to ensure that it generates a pseudorandom sequence having sub-sequences that are different from those generated in other ones of the plurality of base transmitters during concurrent transmissions by the plurality of base transmitters,

wherein the transmitter element is arranged such that the pseudorandom sequence generator adjusts a cancellation-affecting parameter of the transmitter element in accordance with the pseudorandom sequence during the simulcast transmission from the base transmitter.

3. The base transmitter of claim 2, wherein the transmitter element is further arranged such that the pseudorandom sequence generator adjusts a carrier frequency of the base transmitter.

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- 4. The base transmitter of claim 2, wherein the transmitter element is further arranged such that the pseudorandom sequence generator adjusts a carrier phase of the base transmitter.
- 5. The base transmitter of claim 2, wherein the transmitter element is further arranged such that the pseudorandom sequence generator adjusts a carrier amplitude of the base transmitter.
 - 6. The base transmitter of claim 2, wherein the transmitter element is further arranged such that the pseudorandom sequence generator adjusts at least two cancellation-affecting parameters selected from a group of cancellation-affecting parameters consisting of a carrier frequency, a carrier phase, and a carrier amplitude.
 - 7. The base transmitter of claim 2, wherein the pseudorandom sequence generator is further arranged to optimize a parameter of the plurality of pseudorandom sequences according to a characteristic of a communication protocol utilized by the messaging system.
- 8. The base transmitter of claim 2, wherein the pseudorandom sequence generator is further arranged to provide a pseudorandom sequence identical to that of other base transmitters of the plurality of base transmitters but initialized, concurrently with the other base transmitters, with a seed value different from that of the other base transmitters.
 - 9. The base transmitter of claim 2, wherein the pseudorandom sequence generator is further arranged to provide a pseudorandom sequence that is different from that of other base transmitters of the plurality of base transmitters.

10. The base transmitter of claim 2, wherein the pseudorandom sequence generator has at least a predetermined minimum number of stages.

11. The base transmitter of claim 2,

wherein the transmitter element is further arranged to adjust the cancellation-affecting parameter to one of N predetermined levels in accordance with the pseudorandom sequence, N being an integer greater than unity.

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12. A method in a messaging system having a plurality of base transmitters, the method for limiting an interval of carrier cancellation at a reception point during a simulcast transmission, the method comprising the steps of:

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providing a plurality of pseudorandom sequence generators for the plurality of base transmitters, the plurality of pseudorandom sequence generators arranged to generate a plurality of pseudorandom sequences having sub-sequences that have more than a predetermined probability of being different from one another during concurrent transmissions by the plurality of base transmitters, wherein a parameter of the plurality of pseudorandom sequences is optimized according to a characteristic of a communication protocol utilized by the messaging system; and

adjusting a cancellation-affecting parameter of the plurality of base transmitters in accordance with the plurality of pseudorandom sequences during the simulcast transmission from the plurality of base transmitters.

13. A method in a wireless communication system for reducing errors caused by intersymbol interference in at least two simulcast signals transmitted during a time period, the at least two simulcast signals received at similar amplitudes and having different transmission delays with respect to one another, the method comprising the steps of:

transmitting the at least two simulcast signals from a corresponding at least two transmitters; and

changing an output amplitude of at least one of the at least two transmitters during a portion of the time period, thereby altering the intersymbol interference during the portion of the time period.

- 14. A transmitter in a wireless communication system for reducing errors caused by intersymbol interference in at least two simulcast signals transmitted during a time period, the at least two simulcast signals received by a receiver at similar amplitudes and having different transmission delays with respect to one another, the transmitter comprising:
- a transmitter element for transmitting a first simulcast signal sent simultaneously with at least a second simulcast signal from another transmitter; and
- a modulator coupled to the transmitter element for changing an output amplitude of the transmitter during a portion of the time period, thereby altering the intersymbol interference at the receiver during the portion of the time period.
- 15. The transmitter of claim 14, wherein the modulator is arranged to change the output amplitude in synchronism with a symbol transmitted from the transmitter.

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16. The transmitter of claim 14, wherein the modulator is arranged to change the output amplitude as a predetermined function of time.

17. The transmitter of claim 14,

wherein the modulator is arranged to change the output amplitude in synchronism with a symbol transmitted from the transmitter, such that the output amplitude of a central portion of the symbol does not change, while the output amplitude of non-central portions of the symbol do change.

18. The transmitter of claim 14,

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wherein the first simulcast signal comprises an error correcting code that can correct a predetermined number of errors in a code block, and

wherein the modulator is arranged to change the output amplitude according to a pseudorandom sequence having a predetermined number of states during a transmission of the code block.

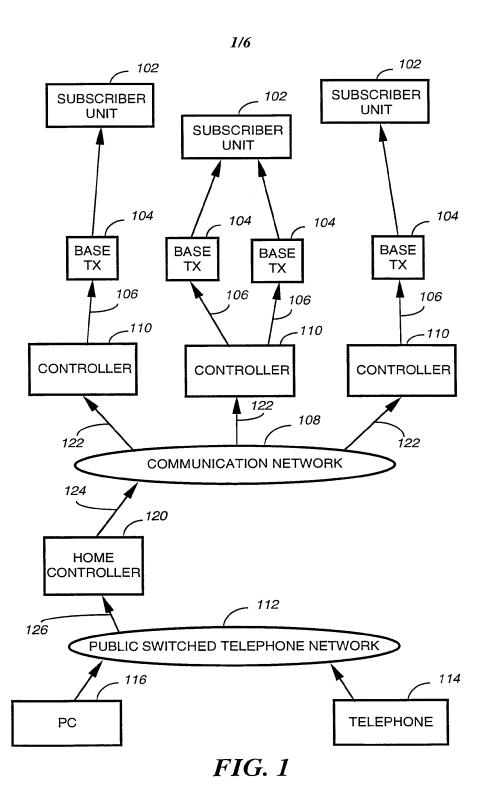
- 19. The transmitter of claim 14, wherein the modulator is arranged to repeat a change to the output amplitude for a number of symbols, wherein the number of symbols is determined from an encoding characteristic employed by the wireless communication system.
- 20. A controller in a wireless communication system for reducing errors caused by intersymbol interference in at least two simulcast signals transmitted during a time period, the at least two simulcast signals received at similar amplitudes and having different transmission delays with respect to one another, the controller comprising:

a network interface for receiving a message from a message originator;

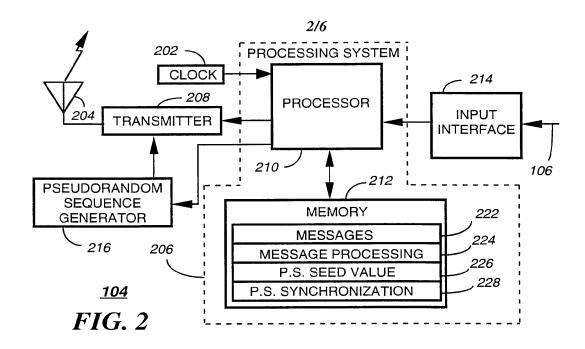
a processing system coupled to the network interface for processing the message; and

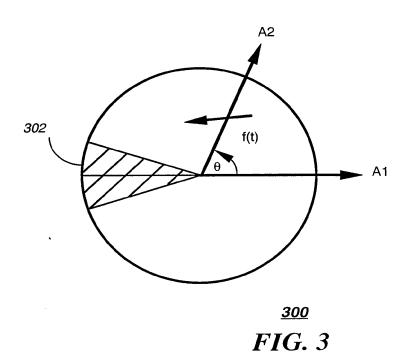
a base station interface coupled to the processing system for controlling a transmitter to transmit one of the at least two simulcast signals,

wherein the processing system is programmed to control the transmitter to change an output amplitude of the transmitter during a portion of the time period, thereby altering the intersymbol interference during the portion of the time period.

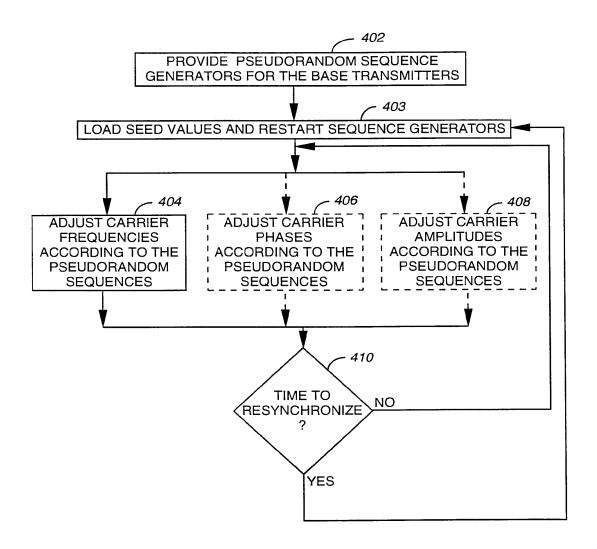


DISH Exhibit 1002 Page 363





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400 FIG. 4

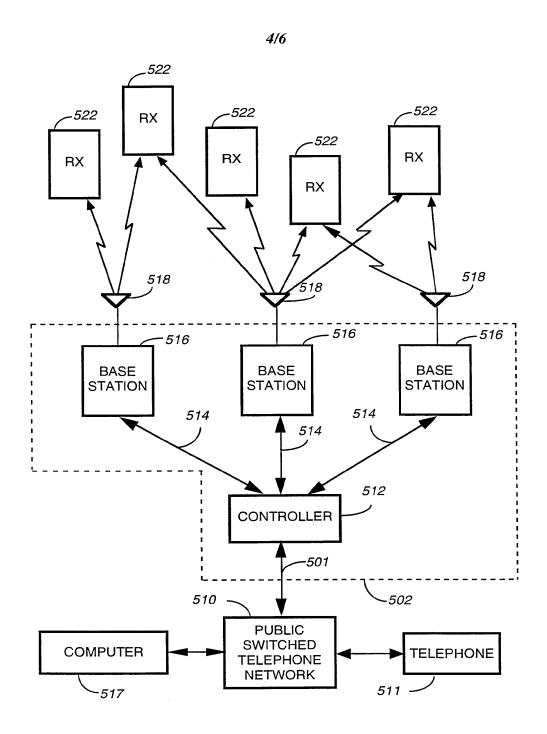
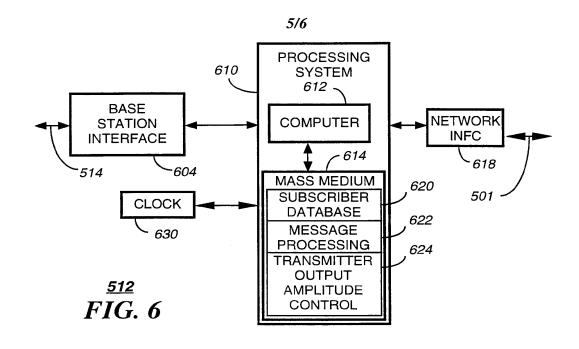
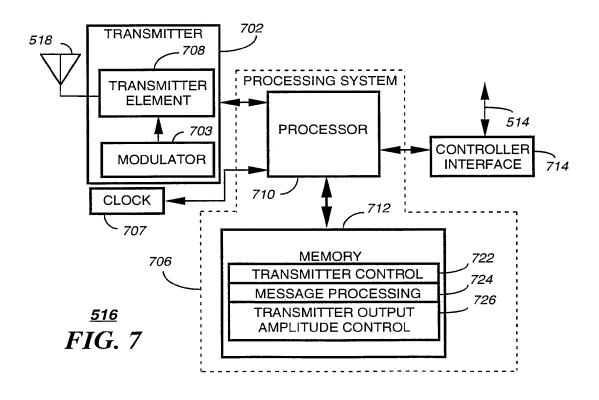
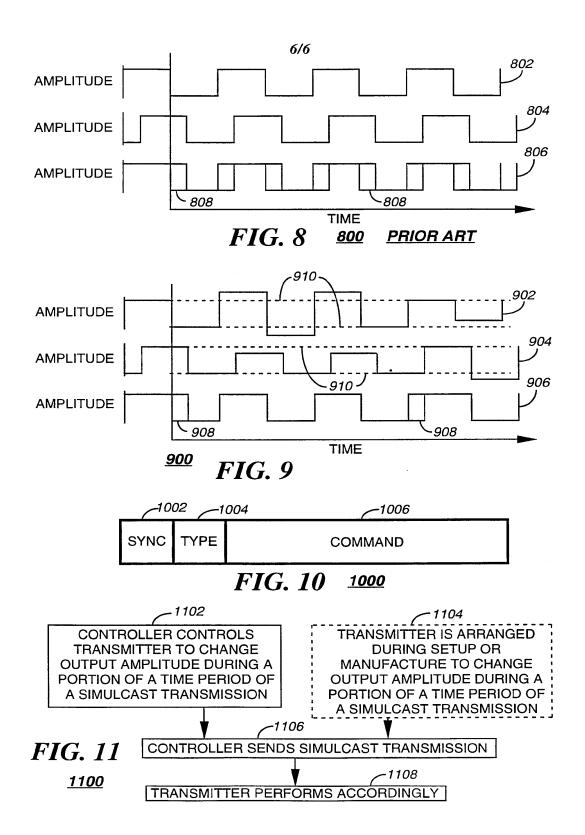


FIG. 5







INTERNATIONAL SEARCH REPORT

International application No. PCT/US98/22801

A. CLA IPC(6) US CL									
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A	US 5,392,452 A (DAVIS) 21 FEBRU	ARY 1995, see abstract	1-20						
A,P	US 5,802,117 A (GHOSH) 01 SEPTI line 35 through column 9, line 12	EMBER 1998, see column 4,	1-20						
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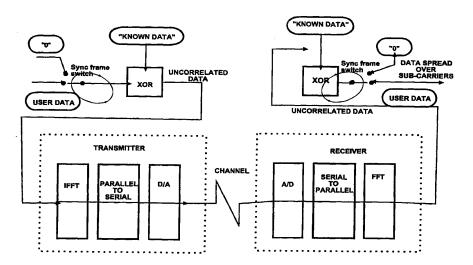
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(57) Abstract

The present invention simplifies known data scramblers by making use of the synchronisation frames, normally used for measuring channel characteristics, as a source of pseudo-random data which can be combined with incoming user data. The present invention has particular application to multi-carrier transmission systems which employ DMT, or OFDM. Many of these transmission systems send known data, usually referred to as synchronisation frames, to measure channel characteristics such as signal to noise ratio. The known data contained in a synchronisation frame is selected to have a suitable statistical distribution, e.g. pseudo-random. In the data scrambler of the present invention, user data bits are combined with the known synchronisation frame data using an exclusive-OR function. This results in a statistically and computationally efficient scrambling of the user data.

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Improvements in, or Relating to, Data Scramblers

The present invention relates to data scramblers and descramblers, suitable for use with a multi-camer transmission system, multi-carrier transmission systems incorporating data scramblers and de-scramblers, and methods for scrambling and descrambling data in multi-carrier transmission systems.

Most telecommunications transmission systems are designed to give optimum performance when uncorrelated data is transmitted over them. Unfortunately, user data is not usually uncorrelated and may, for example, include relatively long strings of binary "0"s, or "1"s. If such data is transmitted over a transmission system intended for uncorrelated data, it can result in saturation, i.e. too large a dynamic range, synchronisation drift, etc.. This problem has long been recognised by telecommunications engineers and the conventional solution is to scramble the incoming user data so that it behaves as though it was uncorrelated data.

Known data scramblers employ an algorithm to combine user data with a random data string, thereby producing an uncorrelated data stream for transmission.

The present invention simplifies known data scramblers by making use of the synchronisation frames, normally used for measuring channel characteristics, as a source of pseudo-random data which can be combined with incoming user data.

The present invention has particular application to multi-camer transmission systems, including copper based transmission systems such as ADSL, VDSL and HDSL which employ DMT, and/or radio based transmission systems employing OFDM. Many of these transmission systems send known data, usually referred to as synchronisation frames, to measure changel characteristics such as signal to noise ratio. The known data contained in a synchronisation frame is selected to have a suitable statistical distribution, e.g. pseudo-random. In a typical DMT

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system, used at the present time, the known synchronisation frame data comprises two bits per sub-carrier. In other words, a predetermined 4-QAM signal is transmitted on each sub-carrier.

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In the present invention, user data bits are combined with the known synchronisation frame data bits, typically the two most significant bits, using an exclusive-OR function. This results in a statistically and computationally efficient scrambling of the user data.

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According to a first aspect of the present invention, there is provided a data scrambler, for use in a multi-carrier transmission system in which synchronisation frame data is periodically transmitted from a transmitter to a receiver to measure transmission channel characteristics, characterised in that combiner means are provided to combine user data with frame synchronisation data.

Said combiner means may have a XOR function.

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Said frame synchronisation data is pseudo random.

Said combiner means may be adapted to combine said user data with the two most significant bits of a synchronisation frame.

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According to a second aspect of the present invention, there is provide a data descrambler, for use in a multi-carrier transmission system in which synchronisation frame data is periodically transmitted from a transmitter to a receiver to measure transmission channel characteristics, and transmitted data is scrambled using a data scrambler as claimed in any of claims 1 to 4, characterised in that combiner means are provided to combine received data with frame synchronisation data.

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Said combiner means may have a XOR function.

Said frame synchronisation data may be pseudo random.

Said combiner means may be adapted to combine said received data with the two most significant bits of a synchronisation frame.

According to a third aspect of the present invention, there is provided a multi-carrier transmission system in which synchronisation frame data is periodically transmitted from a transmitter to a receiver to measure transmission channel characteristics, characterised in that said transmission system incorporates a data scrambler as set forth in any preceding paragraph, connected to said transmitter.

Said receiver may be connected to a data descrambler as set forth in any preceding paragraph.

Said multi-carrier transmission system may employ DMT.

Said multi-carrier transmission system may employ OFDM.

Means may be provided for transmitting frame synchronisation data from said data scrambler to said data descrambler.

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According to a fourth aspect of the present invention, there is provided, in a multi-carrier transmission system in which synchronisation frame data is periodically transmitted from a transmitter to a receiver to measure transmission channel characteristics, a method of scrambling user data prior to transmission, characterised by combining user data with frame synchronisation data.

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User data maybe combined with frame synchronisation data by mean of an XOR function.

Said frame synchronisation data may be pseudo random.

Said user data may be combined with the two most significant bits of a synchronisation frame.

According to a fifth aspect of the present invention, there is provided, in a multi-carrier transmission system in which synchronisation frame data is periodically transmitted from a transmitter to a receiver to measure transmission channel characteristics, a method of descrambling received data which has been scrambled by a scrambling method as set forth in preceding, characterised by combining received data with frame synchronisation data.

Received data may be combined with frame synchronisation data using an XOR function.

Said frame synchronisation data may be pseudo random.

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Said received data may be combined with the two most significant bits of a synchronisation frame.

Said multi-carrier transmission system may employ DMT.

Said multi-carrier transmission system may employ OFDM.

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Embodiments of the invention will now be described, by way of example, with reference to the accompanying drawing, in which:

Figure 1 illustrates data scramblers and descramblers, according to the present invention, in a multi-carrier transmission system

In order to facilitate an understanding of the present invention a glossary of terms used in the description of the present invention is provided below:

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A/D:

Analogue to Digital

ADSL:

Asynchronous Digital Subscriber Line

D/A:

Digital to Analogue

- 5 -

DMT:

Digital Multi Tone

FFT:

Fast Fourier Transform

HDSL:

High bit rate Digital Subscriber Line

IFFT:

Inverse Fast Fourier Transform

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OFDM:

Orthogonal Frequency Division Multiplex

QAM:

Quadrature Amplitude Modulation

VDSL:

Very high bit rate Digital Subscriber Line

XOR:

Exclusive OR

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Figure 1 shows a transmitter and receiver, in a multi-carrier transmission system, linked by a communications channel. The communications channel may be a copper pair (VDSL etc.), or a radio channel (OFDM). Incoming user data, intended for transmission over the communications channel, is passed via a sync frame switch, to a XOR gate. The sync frame switch permits one of the inputs to the XOR gate to be switched between user data and a string of "0"s. The second input to the XOR gate receives the known synchronisation frame data. When the string of "0"s is passed to the XOR gate, the output from the XOR gate is the synchronisation data, i.e. the "known data" appears at the output of the XOR gate.

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The incoming user data will almost certainly be far from random, i.e. it will be highly correlated. The incoming user data is combined with the "known data" in the XOR gate. The "known data" is pseudo random, i.e. uncorrelated. The output from the XOR gate will, therefore, also be uncorrelated, i.e. will itself be pseudo-random. This data has the necessary properties to permit good transmission over the transmission channel.

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The scrambled data is then passed to the receiver where it is first

processed by an Inverse Fast Fourier Transform unit, IFFT, converted from parallel form to serial form, passed to a digital to analogue convertor, D/A, prior to QAM modulation and transmission over the channel. Details of the multiplexing techniques and modulation techniques used in multi-carrier transmission systems will be familiar to those skilled in the art and are not described in detail in this patent specification.

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The signal received from the transmission channel is demodulated and demultiplexed in the receiver by, inter alia, an A/D convertor, a serial to parallel convertor, and a fast Fourier transform unit FFT. The received data is, of course, scrambled. The received scrambled data is passed to the XOR gate, where it is combined with the "known data", i.e. the same data that was mixed into the signal in the transmitter. The output from the XOR gate will contain the user data, or a string of "0"s depending on the setting of the sync frame switch in the transmitter. The sync frame switch in the receiver is used for synchronisation purposes, i.e. when the receiver is properly synchronised with the transmitter, and a sync frame is transmitted, rather than user data, the output from the XOR gate will be a string of "0"s. Details of transmitter and receiver synchronisation in multi-carrier systems will be well known to those skilled in the art.

It should, however, be noted that synchronisation frame data is stored in both the transmitter and receiver, so the receiver always has prior knowledge of the "known data" used by the transmitter.

In summary, the present invention scrambles user data by mixing that data with known data normally used in a synchronisation frame, typically the two most significant bits of the synchronisation frame data, using an exclusive-OR function. This results in both statistically and computationally efficient scrambling. Descrambling is achieved by the reverse process, i.e. combining the received scrambled data with the same known data used for scrambling in an exclusive-OR function.

The present invention results in a much improved statistical distribution of modulated sub-carriers, in a multi-carrier transmission system, compared to the

case where no scrambling is used for correlated, or null data situations.

As synchronisation data must be present in a multi-carrier receiver and transmitter for use in the synchronisation process, the scrambling technique of the present invention does not increase system complexity.

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Transmission of the known data is very simple because it only needs to be combined with a string of "0"s.

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The scrambler of the present invention can be used in all transmission systems that measure channel characteristics by sending known data from transmitter to receiver and use OFDM, DMT, or related multiplexing techniques to spread out the transmitted data over a number of sub-carriers, i.e. multi-carrier transmission techniques.

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CLAIMS

- 1. A data scrambler, for use in a multi-carrier transmission system in which synchronisation frame data is periodically transmitted from a transmitter to a receiver to measure transmission channel characteristics, characterised in that combiner means are provided to combine user data with frame synchronisation data.
- 2. A data scrambler, as claimed in claim 1, characterised in that said combiner means has a XOR function.
- 3. A data scrambler, as claimed in either claim 1, or claim 2, characterised in that said frame synchronisation data is pseudo random.
- 4. A data scrambler, as claimed in any previous claim, characterised in that said combiner means is adapted to combine said user data with the two most significant bits of a synchronisation frame.
- 5. A data descrambler, for use in a multi-carrier transmission system in which synchronisation frame data is periodically transmitted from a transmitter to a receiver to measure transmission channel characteristics, and transmitted data is scrambled using a data scrambler as claimed in any of claims 1 to 4, characterised in that combiner means are provided to combine received data with frame synchronisation data.
- 6. A data descrambler, as claimed in claim 5, characterised in that said combiner means has a XOR function.
 - 7. A data descrambler, as claimed in either claim 5, or claim 6, characterised in that said frame synchronisation data is pseudo random.
- 8. A data descrambler, as claimed in any of claims 5 to 7, characterised in that said combiner means is adapted to combine said received data with the two

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most significant bits of a synchronisation frame.

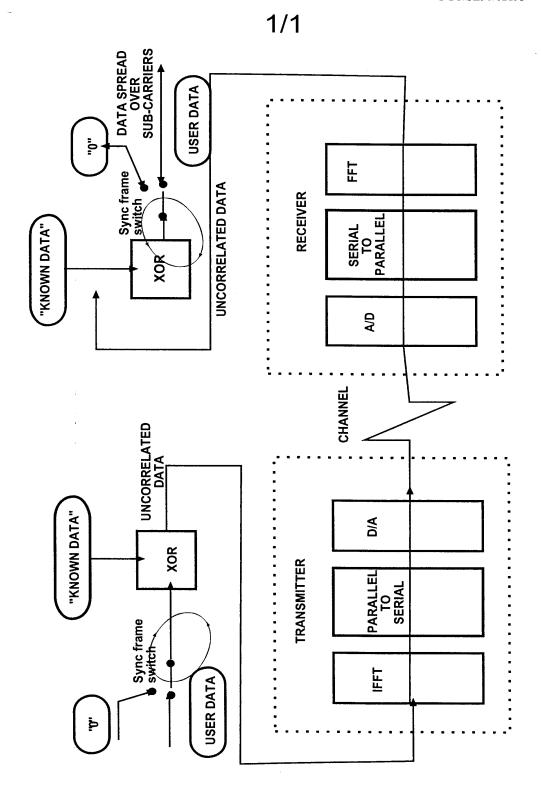
- 9. A multi-carrier transmission system in which synchronisation frame data is periodically transmitted from a transmitter to a receiver to measure transmission channel characteristics, characterised in that said transmission system incorporates a data scrambler as claimed in any of claims 1 to 4, connected to said transmitter.
- 10. A multi-camer transmission system, as claimed in claim 9, characterised in that said receiver is connected to a data descrambler as claimed in any of claims 5 to 8.
- A multi-camer transmission system, as claimed in claim 10, characterised in that said multi-carrier transmission system employs DMT.
 - 12. A multi-camer transmission system, as claimed in claim 10, characterised in that said multi-carrier transmission system employs OFDM.
 - 13. A multi-carrier transmission system, as claimed in any of claims 10 to 12, characterised in that means are provided for transmitting frame synchronisation data from said data scrambler to said data descrambler.
 - 14. In a multi-carrier transmission system in which synchronisation frame data is periodically transmitted from a transmitter to a receiver to measure transmission channel characteristics, a method of scrambling user data prior to transmission, characterised by combining user data with frame synchronisation data.
 - 15. A method, as claimed in claim 14, characterised by combining user data with frame synchronisation data by mean of an XOR function.
 - 16. A method, as claimed in either claim 14, or claim 15, characterised by said frame synchronisation data being pseudo random.
- 25 17. A method, as claimed in any of claims 14 to 16, characterised by combining

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said user data with the two most significant bits of a synchronisation frame.

- 18. In a multi-carrier transmission system in which synchronisation frame data is periodically transmitted from a transmitter to a receiver to measure transmission channel characteristics, a method of descrambling received data which has been scrambled by the method claimed in any of claims 14 to 17, characterised by combining received data with frame synchronisation data.
- 19. A method, as claimed in claim 18, characterised by combining received data with frame synchronisation data using an XOR function.
- 20. A method, as claimed in either claim 18, or claim 19, characterised by said frame synchronisation data being pseudo random.
- 21. A method, as claimed in any of claims 18 to 20, characterised by combining said received data with the two most significant bits of a synchronisation frame.
- 22. A method, as claimed in any of claims 14 to 21, characterised by said multicarrier transmission system employing DMT.
- 23. A method, as claimed in any of claims 14 to 21, characterised by said multicarrier transmission system employing OFDM.



Annex to Form PCT/ISA/206 CAMMUNICATION RELATING TO THE RESULTS OF THE PARTIAL INTERNATIONAL SEARCH

International Application No PCT/US 00/30958

1. The present communication is an Annex to the invitation to pay additional fees (Form PCT/ISA/206). It shows the results of the international search established on the parts of the international application which relate to the invention first mentioned in claims Nos.:

1-13, 20-30, 37, 38 2. This communication is not the international search report which will be established according to Article 18 and Rule 43,

3.If the applicant does not pay any additional search fees, the information appearing in this communication will be considered as the result of the international search and will be included as such in the international search report.

4.If the applicant pays additional fees, the international search report will contain both the information appearing in this communication and the results of the international search on other parts of the international application for which such fees will have been paid.

	ENTS CONSIDERED TO BE RELEVANT		1
Category °	Citation of document, with indication, where appropriate, of t	he relevant passages	Relevant to daim No.
X	EP 0 719 004 A (MATSUSHITA EL LTD) 26 June 1996 (1996-06-26		1-3, 6-12,20, 21, 24-30, 37,38
	column 3, line 28 - line 46		ĺ
Υ	column 14, line 39 - line 55 column 16, line 4 - line 6		13
Y	WO 99 29078 A (BAHLENBERG GUN ;HAAKANSSON SIWERT (SE); LJUN MARIE (SE) 10 June 1999 (1999 page 5, line 18 - line 23	GGREN LIS	13
X	BAUML R W ET AL: "REDUCING TO PEAK-TO-AVERAGE POWER RATIO O MULTICARRIER MODULATIONBY SEL MAPPING" ELECTRONICS LETTERS,GB,IEE ST vol. 32, no. 22, 24 October 1996 (1996-10-24), 2056-2057, XP000643915 ISSN: 0013-5194 page 2056, right-hand column, 7,8 -page 2057, left-hand column, paragraph 3	F ECTED EVENAGE, pages paragraphs	1,2,4,5, 20,22, 23,37,38
A	WO 98 32065 A (FORTRESS TECHNO 23 July 1998 (1998-07-23) page 16, line 30 -page 17, lin		8-12, 26-30
		-/	
X Furt	ner documents are listed in the continuation of box C.	X Patent family members are listed in	n annex.
"A" docume consid	tegories of cited documents : ent defining the general state of theart which is not ered to be of particular relevance	*T* later document published after theinten or priority date and not in conflict with cited to understand the principle or the invention	theapplication but
filing d L" docume which citation O" docume other r P" docume	nt which may throw doubts on priority claim(s) or is cited to establish the publication date of another n or other special reason (as specified) ant referring to an oral disclosure, use, exhibition or	"X" document of particular relevance; the cannot be considered novel or cannot involve an inventive step when the do "Y" document of particular relevance; the cannot be considered to involve an indocument is combined with one or more ments, such combination being obvious in the art. "&" document member of the same patent	be considered to cument is taken alone laimedinvention ventive step when the ore othersuch docu- us to a person skilled

Form PCT/ISA/206 (Annex, first sheet) (July 1992)

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Annex to Form PCT/ISA/206 COMMUNICATION RELATING TO THE RESULTS OF THE PARTIAL INTERNATIONAL SEARCH

International Application No
PCT/US 00/30958

0.104 11	OF THE PARTIAL INTERNATIONAL SEATON	FC1/U3 00/30956			
C.(Continu	ation) DOCUMENTS CONSIDERED TO BE RELEVANT Citation of document, with indication, where appropriate, of the relevant passages		Relevant to claim No.		
Jalegory	onation of document, man monotonymics appropriately of the release passages				
Α	WO 99 22463 A (MOTOROLA INC) 6 May 1999 (1999-05-06) page 7, line 29 - line 32		8-12, 26-30		
			·		
-					

Form PCT/ISA/206 (Annex, continuation sheet)(July 1992)

1

Patent Family Annex

Information on patent family members

International Application No PCT/US 00/30958

Patent document cited in search repo		Publication date		Patent family member(s)	Publication date
EP 0719004	Α .	26-06-1996	JP US	8321820 A 5682376 A	03-12-1996 28-10-1997
WO 9929078	Α	10-06-1999	SE	9704497 A	04-06-1999
WO 9832065	Α	23-07-1998	AU EP	5811598 A 0951767 A	07-08-1998 27-10-1999
W0 9922463	Α	06-05-1999	US US	6044276 A 6061574 A	28-03-2000 09-05-2000

Form PCT/ISA/206 (patent family annex) (July 1992)

PATENT COOPERATION TREATY

PCT

MIPO O 6 MAR 2002 INTERNATIONAL PRELIMINARY EXAMINATION REPORT

(PCT Article 36 and Rule 70)

Applicant's	s or age	ent's file reference	1		
081513-49			FOR FURTHER A		See Notification of Transmittal of International Preliminary Examination Report (Form PCT/IPEA/416)
Internation	nal appi	lication No.	International filing date	(day/month/yea	ar) Priority date (day/month/year)
PCT/US	00/30	958	09/11/2000		09/11/1999
H04L27		ent Classification (IPC) or na	itional classification and IF	PC	
Applicant AWARE	, INC.				
1. This and i	interna s trans	ational preliminary exami smitted to the applicant a	nation report has been coording to Article 36.	prepared by	this International Preliminary Examining Authority
2. This	REPO	PRT consists of a total of	6 sheets, including thi	s cover sheet	t.
l k	een a	port is also accompanied mended and are the bas ule 70.16 and Section 60	is for this report and/oı	r sheets conta	escription, claims and/or drawings which have aining rectifications made before this Authority under the PCT).
Thes	e anne	exes consist of a total of	sheets.		
		•		,	•
· · · · · ·					
3. This	report	contains indications relat	ing to the following ite	ms:	
ı	\boxtimes	Basis of the report			
11		Priority		1	
HI	\boxtimes	Non-establishment of or	oinion with regard to no	ovelty, inventi	ive step and industrial applicability
IV		Lack of unity of invention			
٧	\boxtimes	Reasoned statement un citations and explanation	der Article 35(2) with r ns suporting such state	egard to nove ement	elty, inventive step or industrial applicability;
VI		Certain documents cite			
VII		Certain defects in the in			ì
VIII		Certain observations on	the international application	cation	,
Date of sub	missio	n of the demand		Date of comp	pletion of this report
06/06/20	01	_		04.03.2002	
	examir	address of the international ning authority:		Authorized of	fficer State of the state of th
European Patent Office D-80298 Munich				Pajatakis, I	E (Gauss

Telephone No. +49 89 2399 8898

Form PCT/IPEA/409 (cover sheet) (January 1994)

Fax: +49 89 2399 - 4465

Tel. +49 89 2399 - 0 Tx: 523656 epmu d

INTERNATIONAL PRELIMINARY EXAMINATION REPORT

International application No. PCT/US00/30958

I.	Basis of the report								
1.	 With regard to the elements of the international application (Replacement sheets which have been furnished to the receiving Office in response to an invitation under Article 14 are referred to in this report as "originally filed" and are not annexed to this report since they do not contain amendments (Rules 70.16 and 70.17)): Description, pages: 								
	1-1	7	as originally filed						
	Cla	ims, No.:							
	1-3	9	as originally filed						
	Dra	wings, sheets:							
	1/2	-2/2	as originally filed						
2.	With lanç	h regard to the lang guage in which the i	juage, all the elements marked above were available or furnished to this Authority in the international application was filed, unless otherwise indicated under this item.						
	The	ese elements were a	available or furnished to this Authority in the following language: , which is:						
		the language of a	translation furnished for the purposes of the international search (under Rule 23.1(b)).						
		the language of pu	blication of the international application (under Rule 48.3(b)).						
		the language of a f 55.2 and/or 55.3).	translation furnished for the purposes of international preliminary examination (under Rule						
3.	With inte	n regard to any nuc rnational preliminar	leotide and/or amino acid sequence disclosed in the international application, the y examination was carried out on the basis of the sequence listing:						
		contained in the in	ternational application in written form.						
		filed together with	the international application in computer readable form.						
		furnished subsequ	ently to this Authority in written form.						
		furnished subsequ	ently to this Authority in computer readable form.						
☐ The statement that the subsequently furnished written sequence listing does not go beyond the disclosu the international application as filed has been furnished.									
		The statement that listing has been ful	the information recorded in computer readable form is identical to the written sequence nished.						
4.	The	amendments have	resulted in the cancellation of:						
		the description,	pages:						
		the claims,	Nos.:						

Form PCT/IPEA/409 (Boxes I-VIII, Sheet 1) (July 1998)

INTERNATIONAL PRELIMINARY EXAMINATION REPORT

International application No. PCT/US00/30958

		the drawings,	sheets:								
5.		This report has been considered to go bey	establish ond the c	ned as if (disclosure	some of) as filed	the ame (Rule 70	ndments ha	ad not bee	n made, si	nce they ha	ıve been
		(Any replacement sh report.)	eet conta	ining suc	h amend	lments m	ust be refe	rred to una	ler item 1 a	and annexe	d to this
6.	Add	ditional observations, i	f necessa	ıry:	٠						
111.	Nor	n-establishment of o	pinion wi	th regard	i to nove	elty, inve	ntive step	and indus	strial appli	cability	
1.	The obv	questions whether th	e claimed ally applic	l inventior able hav	n appear e not be	s to be n en exami	ovel, to invo	olve an inv	entive step	(to be non	-
		the entire international	al applicat	tion.							
	×	claims Nos. 20-36.									
be	caus	se:									
		the said international not require an interna	applicational pre	on, or the eliminary e	said clai examinat	ms Nos. ion (<i>spe</i> e	relate to th	e following	g subject m	atter which	does
		the description, claim that no meaningful op	s or draw oinion cou	ings (<i>indi</i> ld be forn	<i>cate pan</i> ned (<i>spe</i>	ticular ele cify):	ements belo	סישי) or said	claims Nos	s. are so ui	nclear
	×	the claims, or said cla	ims Nos. ed.	20-36 are	e so inac	lequately	supported	by the des	scription the	at no meani	ngful
		no international searc	h report h	nas been	establish	ed for th	e said clain	ns Nos			
2.	and/	eaningful international or amino acid sequen uctions:	prelimina ce listing	ary exami to comply	nation ca	innot be	carried out d provided t	due to the for in Anne	failure of the	ne nucleotic Administrat	le ive
	_	the second second								1	
		the written form has n					-				
	_	the computer readable	e ioiiii ila	s not bee	n iumsii	ed or do	es not comp	piy with the	e standard.	1	
V.	Reas citat	soned statement und ions and explanation	ler Article ns suppo	e 35(2) w rting suc	ith rega h stater	d to nov	elty, inven	itive step	or industri	ial applicat	oility;
		ement	••	•					Ť		
	Nove	eity (N)	Yes:	Claims	4,5,7-1	3,39					

Form PCT/IPEA/409 (Boxes I-VIII, Sheet 2) (July 1998)

INTERNATIONAL PRELIMINARY EXAMINATION REPORT

International application No. PCT/US00/30958

No: Claims 1-3,6,19,37,38

Inventive step (IS) Yes: Claims 39

No: Claims 1-19,37,38

Industrial applicability (IA) Yes: Claims 1-19,37-39

No: Claims

2. Citations and explanations see separate sheet

Re Item III

Non-establishment of opinion with regard to novelty, inventive step and industrial applicability

- 1. According to the description (page 3, lines 10-16, page 4, lines 4-10) the phase characteristics of the modulated carrier signals are scrambled by combining the phase shift computed for each carrier signal with the phase characteristic of that carrier signal. As this essential feature is missing from Claims 20-36, their scope comprises embodiments in which phase scrambling is carried out without the above feature which are not supported by the description, see also Guidelines 111, 4.3.
- 2. The application comprises multiple independent claims of the same category and does not therefore meet the requirement of conciseness, Article 6.

Re Item V

Reasoned statement under Article 35(2) with regard to novelty, inventive step or industrial applicability; citations and explanations supporting such statement

- The subject-matter of Claim 1 is not new, Article 33(2) 1.
 - D1 = BAUML R W ET AL: 'REDUCING THE PEAK-TO-AVERAGE POWER RATIO OF MULTICARRIER MODULATIONBY SELECTED MAPPING' ELECTRONICS LETTERS, GB, IEE STEVENAGE, vol. 32, no. 22, 24 October 1996 (1996-10-24), pages 2056-2057, XP000643915 ISSN: 0013-5194 discloses a method for scrambling the phase characteristics of the carrier signals in a multicarrier modulation system. The method comprises associating each carrier signal $V(\mu)$ with a value ϕ_u determined independently of any input bit value (page 2056, right col., last but one paragraph). A phase shift $e^{i\phi\mu}$ is computed for each carrier signal and combined with the phase characteristic of that carrier signal so as to substantially scramble the phase characteristics of the plurality of the carrier signals (page 2056, right col., equation 4).
- All features of Claim 1 are also known from D2 = EP-A-0 719 004 (col. 14, line 39 2. - col. 15, fig. 9).

Form PCT/Separate Sheet/409 (Sheet 1) (EPO-April 1997)

3.

- 4. The additional features of the dependent claims do not add anything new or inventive to the above-mentioned independent claims because these features are either known from the above prior art (reduced peak-to-average power ratio, varying value with each carrier, pseudo-random pattern) or common measures (using symbol and frame counts).

The above finding also applies to Claim 37 which corresponds to Claim 1.

- The subject-matter of Claim 39 is new and involves an inventive step, Article 33(2)(3).
- 5.1 Claim 39 relates to a method for communicating in a multicarrier system comprising receiving a transmission signal comprising DMT symbols each having a bit-value pattern.
 - Such a method is known from D3 = EP-A-0 584 534.
- 5.2 The underlying problem is avoiding demodulation errors due to clipping of transmitted signals as result of non linear distortions.
 - This problem is solved by comparing received DMT symbols with a bit-value pattern. DMT symbols matching the bit-value pattern are discarded.
- 5.3 Neither the problem not the solution is suggested by the prior art relevant for this invention.
 - In D3 high power bit-value patterns, which could cause nonlinear distortions, are not clipped but replaced with different patterns using an enhanced alphabet. At the receiver those patterns are restored to the original patterns. Thus, D3 leads on a different way which is recovering high power patterns rather than discarding them.
 - GB-A-2 330 49 does not relate to treating bit patterns which can cause non linear distortions. A dummy symbol is inserted at the start of each frame in order to provide a phase reference for differential encoding.

Form PCT/Separate Sheet/409 (Sheet 2) (EPO-April 1997)

PATENT COOPERATION TREATY

From the INTERNATIONAL PRELIMINARY EXAMINING AUTHORITY PCT VICK, Jason H. Nixon Peabody LLP 8180 Greensboro Drive, Suite 800 WRITTEN OPINION McLean, Virginia 22102 **ETATS-UNIS D'AMERIQUE** (PCT Rule 66) Date of mailing (day/month/year) 18.12.2001 **REPLY DUE** within 1 month(s) and 15 days Applicant's or agent's file reference from the above date of mailing 081513-49 International application No. International filing date (day/month/year) Priority date (day/month/year) PCT/US00/30958 09/11/2000 09/11/1999 International Patent Classification (IPC) or both national classification and IPC H04L27/26 Applicant AWARE, INC. This written opinion is the first drawn up by this International Preliminary Examining Authority. 2. This opinion contains indications relating to the following items: Basis of the opinion ☐ Priority П Non-establishment of opinion with regard to novelty, inventive step and industrial applicability ☐ Lack of unity of invention Reasoned statement under Rule 66.2(a)(ii) with regard to novelty, inventive step or industrial applicability;

3. The applicant is hereby **invited to reply** to this opinion.

☐ Certain document cited

When? See the time limit indicated above. The applicant may, before the expiration of that time limit,

request this Authority to grant an extension, see Rule 66.2(d).

citations and explanations supporting such statement

☐ Certain defects in the international application

☐ Certain observations on the international application

How? By submitting a written reply, accompanied, where appropriate, by amendments, according to Rule 66.3.

For the form and the language of the amendments, see Rules 66.8 and 66.9.

Also: For an additional opportunity to submit amendments, see Rule 66.4.

For the examiner's obligation to consider amendments and/or arguments, see Rule 66.4 bis.

For an informal communication with the examiner, see Rule 66.6.

If no reply is filed, the international preliminary examination report will be established on the basis of this opinion.

 The final date by which the international preliminary examination report must be established according to Rule 69.2 is: 09/03/2002.

Name and mailing address of the international preliminary examining authority:

V١

VII

European Patent Office D-80298 Munich

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Fax: +49 89 2399 - 4465

Authorized officer / Examiner

Pajatakis, E

Formalities officer (incl. extension of time limits)

Barrio Baranano, A

Telephone No. +49 89 2399 8621



Form PCT/IPEA/408 (cover sheet) (January 1994)

Re Item III

Non-establishment of opinion with regard to novelty, inventive step and industrial applicability

- 1. According to the description (page 3, lines 10-16, page 4, lines 4-10) the phase characteristics of the modulated carrier signals are scrambled by combining the phase shift computed for each carrier signal with the phase characteristic of that carrier signal. As this essential feature is missing from Claims 20-36, their scope comprises embodiments in which phase scrambling is carried out without the above feature which are not supported by the description, see also Guidelines III, 4.3.
- 2. To meet the requirement of conciseness, Article 6, a single independent claim in each category should be filed for the first invention.

Re Item V

Reasoned statement under Rule 66.2(a)(ii) with regard to novelty, inventive step or industrial applicability; citations and explanations supporting such statement

- 1. The subject-matter of Claim 1 is not new, Article 33(2)
 - **D1** = BAUML R W ET AL: 'REDUCING THE PEAK-TO-AVERAGE POWER RATIO OF MULTICARRIER MODULATIONBY SELECTED MAPPING' ELECTRONICS LETTERS, GB, IEE STEVENAGE, vol. 32, no. 22, 24 October 1996 (1996-10-24), pages 2056-2057, **XP000643915** ISSN: 0013-5194 discloses a method for scrambling the phase characteristics of the carrier signals in a multicarrier modulation system. The method comprises associating each carrier signal **V(μ)** with a value ϕ_{μ} determined independently of any input bit value (page 2056, right col., last but one paragraph). A phase shift $e^{i\phi_{\mu}}$ is computed for each carrier signal and combined with the phase characteristic of that carrier signal so as to substantially scramble the phase characteristics of the plurality of the carrier signals (page 2056, right col., equation 4).
- 2. All features of Claim 1 are also known from **D2 = EP-A-0 719 004** (col. 14, line 39 col. 15, fig. 9).

- 3. The above finding also applies to Claim 37 which corresponds to Claim 1.
- 4. The additional features of the dependent claims do not add anything new or inventive to the above-mentioned independent claims because these features are either known from the above prior art (reduced peak-to-average power ratio, varying value with each carrier, pseudo-random pattern) or common measures (using symbol and frame counts).

ı.	Das	sis of the opinion						
1.	With regard to the elements of the international application (Replacement sheets which have been furnished to the receiving Office in response to an invitation under Article 14 are referred to in this opinion as "originally filed",							
	Des	scription, pages:						
	1-1	7	as originally filed					
	Cla	ims, No.:						
	1-3	9	as originally filed					
	Dra	wings, sheets:						
	1/2-	-2/2	as originally filed					
2.	. With regard to the language , all the elements marked above were available or furnished to this Authority in the language in which the international application was filed, unless otherwise indicated under this item.							
	These elements were available or furnished to this Authority in the following language: , which is:							
		the language of a	translation furnished for the purposes of the international search (under Rule 23.1(b)).					
		the language of p	ublication of the international application (under Rule 48.3(b)).					
		the language of a 55.2 and/or 55.3).	translation furnished for the purposes of international preliminary examination (under Rule					
3.	With regard to any nucleotide and/or amino acid sequence disclosed in the international application, the international preliminary examination was carried out on the basis of the sequence listing:							
		contained in the ir	nternational application in written form.					
		filed together with	the international application in computer readable form.					
		furnished subsequ	uently to this Authority in written form.					
		furnished subsequ	uently to this Authority in computer readable form.					
			at the subsequently furnished written sequence listing does not go beyond the disclosure in application as filed has been furnished.					
		The statement that listing has been for	at the information recorded in computer readable form is identical to the written sequence urnished.					
4.	The	amendments have	e resulted in the cancellation of:					

Form PCT/IPEA/408 (Boxes I-VIII, Sheet 1) (July 1998)

pages:

Nos.:

 \square the description,

 \square the claims,

W	RIT	TEN OPINION			International application No.	PCT/US00/30958
		the drawings,	sheets:			
5.				s if (some of) the ar sure as filed (Rule	mendments had not been made, 70.2(c)):	since they have been
		(Any replacement report.)	sheet containing	ı such amendments	must be referred to under item	1 and annexed to this
6.	Add	ditional observation	s, if necessary:			
III.	No	n-establishment o	f opinion with re	egard to novelty, ir	nventive step and industrial ap	plicability
1.					e novel, to involve an inventive s I will not be examined in respect	
		the entire internat	ional application,			
	\boxtimes	claims Nos. 20-36	5,			
be	cau	se:				
				r the said claims No nary examination (s	os. relate to the following subject pecify):	t matter which does
				(indicate particular e formed (specify):	elements below) or said claims	Nos. are so unclear
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		no international se	earch report has I	oeen established fo	r the said claims Nos	
2.					nucleotide and/or amino acid sed Iministrative Instructions:	quence listing to
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٧.				(a)(ii) with regard g such statement	to novelty, inventive step or in	dustrial applicability;
1.		tement velty (N)	Claims	1-3,6,19,37,38		

Form PCT/IPEA/408 (Boxes I-VIII, Sheet 2) (July 1998)

Inventive step (IS) Claims 1-19,37,38

Industrial applicability (IA) Claims

2. Citations and explanations see separate sheet

Electronic Acknowledgement Receipt			
EFS ID:	11577639		
Application Number:	13303417		
International Application Number:			
Confirmation Number:	3035		
Title of Invention:	SYSTEM AND METHOD FOR SCRAMBLING THE PHASE OF THE CARRIERS IN A MULTICARRIER COMMUNICATIONS SYSTEM		
First Named Inventor/Applicant Name:	Marcos C. Tzannes		
Customer Number:	62574		
Filer:	Jason Vick/Amy Duarte		
Filer Authorized By:	Jason Vick		
Attorney Docket Number:	5550-47-CON-5		
Receipt Date:	08-DEC-2011		
Filing Date:	23-NOV-2011		
Time Stamp:	18:46:05		
Application Type:	Utility under 35 USC 111(a)		

Payment information:

Submitted wi	th Payment	no			
File Listin	g:				
Document Number	Document Description	File Name	File Size(Bytes)/ Message Digest	Multi Part /.zip	Pages (if appl.)
1		IDS_01.pdf	727047 da7fca8dfc618330f1ba0b043d7ab8499bec c6c4	yes	7

	Multipart Description/PDF files in .zip description					
	Document Description		Start	Eı	nd	
	Transmittal Letter		1	3		
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Warnings:						
Information:						
2	Foreign Reference	EP0584534A1.pdf	374865	no	7	
			daa67e969afd848af68a2172d85f9751da31 5388			
Warnings:						
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3	Foreign Reference	EP0719004A2.pdf	1255520	no	26	
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4	Foreign Reference	GB2330491A.pdf	1179859	. no	33	
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5	Foreign Reference	JP_H10_1998084329.pdf	1648963	no	18	
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6	Foreign Reference	JP8321820A.pdf	1472960	no	15	
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7	Foreign Reference	WO9832065A2.pdf	1705757	no	44	
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8	Foreign Reference	WO9922463A1.pdf	1510884	no	32	
			6ace37c2cf40c308a83d67456dbe4f0030fa c76b			
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9	Foreign Reference	WO99029078.pdf	457552 b4a959a6b0a559f2d90c59bc2c5159e9665	no	13
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10	Non Patent Literature	Bauml_Reducing.pdf		no	2
			2870f9e075f40e668b3949db807f17ca765a 9ae1		
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11	N. B. H.	HENKEL_Analog_Codes_for_Pe	350745		_
11	Non Patent Literature	ak-to- Average_Ratio_Reduction.pdf	2c1b320e0b94c6d90f5c272e2987ed97d0e af570	no	5
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12	Non Patent Literature	NARAHASHI_New_phasing_sc heme_of_N_multiple_carriers.		no	2
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13	Non Patent Literature	TELLADO_Revisiting_DMTs_pe	972642	no	
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14	Non Patent Literature	ique_for_reducing_peak-to-	4933268ef1c93386e1aacd07f8691a302292	no	5
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16	Non Patent Literature	VAN_EETVELT_Peak_to_averag e_power_reduction.pdf	13b49ed5b534c38620bc3af3fd93c2ac1ea4	no	2
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17	Non Patent Literature	5550-47-PCT_Annex.pdf	/8/35	no no	3
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18	Non Patent Literature	ent Literature 5550-47-PCT_Search_Report. pdf		no	6
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19	Non Patent Literature	5550-47-PCT_IPER.pdf		no	6
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20	Non Patent Literature	5550-47- PCT_Written_Opinion_2001-12	184614	no	6
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21	Non Patent Literature	5550-47-PJP_OA_3-3-08.pdf	51dea09ec4c4130a741e3d331e1e8cfa31a6	no	4
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22	Non Patent Literature	5550-47-PJP_OA_11-4-08.pdf		no	3
			46159ab02ab5c03caf5c5b33b9a18e570e1 7f14f		
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23	Non Patent Literature	5550-47-	281029	no	3
	Trong atom Electrical	PKR_Rejection_2006-11-22.pdf	f95ea0a061977b8a5f939acce4c1eaff1c68f2 bf	0	
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24	Non Patent Literature	5550-47_OA_2004-05-04.pdf	09508213fd8563650b43804c9bea1773825	no 5	12
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32	Non Patent Literature	CON-4_NOA_2011-11-17.pdf	bf44ba2299bea4e92ed40cdcac657147e4b 9afbb	no	9
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31	Non Patent Literature	pdf	45e2b2e22f073d4802c5c23182b770d2919 bfc83	no	10
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		5550-47-CON-3_OA_10-15-09.	573466		
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29	Non Patent Literature	CON-2_NOA_10-08-08.pdf	e92225f87acb5d1667ebd6aabf9cd1bec5d 24fcc	no	5
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28	Non Patent Literature	5550-47-CON-2_OA_2-6-08.pdf	bb3a2d806d1e7caaeda7cd7eade83670b6f 23749	no	10
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27	Non Patent Literature	DIV_NOA_2011-10-17.pdf	ae238c9e3df0d089fef942aeb6fa541fb4800 38f	no	11
27	New Determination	5550-47-CON-	448359		11

This Acknowledgement Receipt evidences receipt on the noted date by the USPTO of the indicated documents, characterized by the applicant, and including page counts, where applicable. It serves as evidence of receipt similar to a Post Card, as described in MPEP 503.

New Applications Under 35 U.S.C. 111

If a new application is being filed and the application includes the necessary components for a filing date (see 37 CFR 1.53(b)-(d) and MPEP 506), a Filing Receipt (37 CFR 1.54) will be issued in due course and the date shown on this Acknowledgement Receipt will establish the filing date of the application.

National Stage of an International Application under 35 U.S.C. 371

If a timely submission to enter the national stage of an international application is compliant with the conditions of 35 U.S.C. 371 and other applicable requirements a Form PCT/DO/EO/903 indicating acceptance of the application as a national stage submission under 35 U.S.C. 371 will be issued in addition to the Filing Receipt, in due course.

New International Application Filed with the USPTO as a Receiving Office

If a new international application is being filed and the international application includes the necessary components for an international filing date (see PCT Article 11 and MPEP 1810), a Notification of the International Application Number and of the International Filing Date (Form PCT/RO/105) will be issued in due course, subject to prescriptions concerning national security, and the date shown on this Acknowledgement Receipt will establish the international filing date of the application.

IN THE UNITED STATES PATENT AND TRADEMARK OFFICE

In Re t	the Application of:)	Group Art	Unit: 26	511		
	TZANNES, Marcos C.)	Confirmati	ion No.:	303	5	
Serial	No.: 13/303,417)	Examiner:	Not Yet	t Ass	igned	
Filed:	November 23, 2011)					
Atty. F	File No.: 5550-47-CON-5)	INFORM	ATION STATEN			<u>RE</u>
Phase	d: "System and Method for Scrambling the of the Carriers in a Multicarrier funications System")		ronically			
P.O. B	nissioner for Patents ox 1450 ndria, VA 22313-1450						
Dear S	ir:						
	The references cited on attached Form PTC	0-14	49 are being	g called t	to the	e attenti	on
of the	Examiner.						
\boxtimes	Copies of the cited non-patent and/or foreign	n ref	erences are	enclosed	here	with.	
	Copies of the cited U.S. patents and/or pater	ıt ap	plications ar	e enclose	ed he	rewith.	
\boxtimes	Copies of the cited U.S. patents/patent appl	icati	on publicati	ons are r	not e	nclosed	in
accord	ance with 37 C.F.R. § 1.98(a).						
	Copies of the cited references are not en	clos	ed, in accor	rdance w	vith :	37 C.F.	R.
	(d), because the references were cited by					atent a	nd
	mark Office in prior application Serial No						,
	is relied upon for an earlier filing date under						
	To the best of applicants' belief, the pertine						
	lieved to be summarized in the attached E	_				e figur	es,
-	gh applicants do not necessarily vouch for the		•		on,		
\boxtimes	Examiner's attention is drawn to the following	_			_		
	• Serial No. 09/710,310 filed 11/09/ (Attorney's Ref. No. 5550-47)	/00	now U.S.	Patent 1	No.	6,961,3	69
	 Serial No. 11/211,535 filed 08/26/ (Attorney's Ref. No. 5550-47-CON) 	/05	now U.S.	Patent 1	No.	7,292,6	27
	 Serial No. 11/860,080 filed 09/24/ (Attorney's Ref. No. 5550-47-CON-DI 		now U.S.	Patent 1	No.	8,073,0	41

- Serial No. 11/863,581 filed 09/28/07 now U.S. Patent No. 7,471,721 (Attorney's Ref. No. 5550-47-CON-2)
- Serial No. 12/255,713 filed 10/22/08 now U.S. Patent No. 7,769,104 (Attorney's Ref. No. 5550-47-CON-3)
- Serial No. 12/783,725 filed 05/20/10 (Attorney's Ref. No. 5550-47-CON-4)
- Serial No. 13/284,549 filed 10/28/11 (Attorney's Ref. No. 5550-47-CON-DIV-CON)

Other:	
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Submission of the above information is not intended as an admission that any item is citable under the statutes or rules to support a rejection, that any item disclosed represents analogous art, or that those skilled in the art would refer to or recognize the pertinence of any reference without the benefit of hindsight, nor should an inference be drawn as to the pertinence of the references based on the order in which they are presented. Submission of this statement should not be taken as an indication that a search has been conducted, or that no better art exists.

It is respectfully requested that the cited information be expressly considered during the prosecution of this application and the references made of record therein.

FEES

37 CFR 1.97(b): No fee is believed due in connection with this submission, because the information disclosure statement submitted herewith is satisfied by one of the following conditions ("X" indicates satisfaction): Within three months of the filing date of a national application other than a continued prosecution application under 37 CFR 1.53(d), or
Within three months of the date of entry of the national stage as set forth in § 1.491 in an international application, or
Before the mailing date of a first Office Action on the merits, or
Before the mailing of a first Office action after the filing of a request for continued examination under 37 CFR 1.114.
Although no fee is believed due, if any fee is deemed due in connection with this submission, please charge such fee to Deposit Account 19-1970.
37 CFR 1.97(c): The information disclosure statement transmitted herewith is being filed after all the above conditions (37 CFR 1.97(b)), but before the mailing date of any one of the following conditions: (1) a final action under 37 C.F.R. 1.113, or (2) a notice of allowance under 37 C.F.R. 1.311, or (3) an action that otherwise closes prosecution in the application. This Information Disclosure Statement is accompanied by: A Certification (below) as specified by 37 C.F.R. 1.97(e). Although no fee is believed due, if any fee is deemed due in connection with this submission, please charge such fee to Deposit Account 19-1970. OR Please charge Deposit Account 19-1970 in the amount of \$180.00 for the fee set forth in 37 C.F.R. 1.17(p) for submission of an information disclosure statement. Please credit any overpayment or charge any underpayment to Deposit Account 19-1970.
37 CFR 1.97(d): This Information Disclosure Statement is being submitted after the period specified in 37 CFR 1.97(c). This information Disclosure Statement includes a Certification (below) as specified by 37 C.F.R. 1.97(e) AND Applicants hereby requests consideration of the reference(s) disclosed herein. Please charge Deposit Account 19-1970 in the amount of \$180.00 under 37 C.F.R. 1.17(p). Please credit any overpayment or charge any underpayment to Deposit Account 19-1970. Election to pay the fee should not be taken as an indication that applicant(s) cannot execute a certification.

Certification (37 C (Applicable only	
first cited in any communication from application not more than three mont 1.97(e)(1).	ed in this information disclosure statement was a foreign patent office in a counterpart foreign hs prior to the filing of this statement. 37 C.F.R. ation from the foreign patent office is enclosed.
OR	
cited in a communication from a fore application, and, to the knowledge of inquiry, no item of information conta	I in this information disclosure statement was ign patent office in a counterpart foreign the undersigned after making reasonable ined in this Information Disclosure Statement and in 37 C.F.R. 1.56(c) more than three months of C.F.R. 1.97(e)(2).
Respec	etfully submitted,
SHER	IDAN ROSS P.C.
By:	Jasen H. Vick Registration No. 45,285 1560 Broadway, Suite 1200 Denver, Colorado 80202-5141 (303) 863-9700

IN THE UNITED STATES PATENT AND TRADEMARK OFFICE

In Re the Application of: Marcos C. Tzannes) Group Art Unit: 2611
Application No.:	Examiner:
Filed: Herewith	Confirmation No.:
Atty. File No.: 5550-47-CON-5)

For: SYSTEM AND METHOD FOR SCRAMBLING THE PHASE OF THE CARRIERS IN A MULTICARRIER COMMUNICATIONS SYSTEM

Commissioner for Patents P.O. Box 1450 Alexandria, VA 22313

PRELIMINARY AMENDMENT

Dear Sir:

Prior to the initial review of the above-identified patent application by the Examiner, please enter the following Preliminary Amendment. Although Applicants do not believe that any fees are due based upon the filing of this Preliminary Amendment, please charge any such fees to Deposit Account 19-1970.

Please amend the above-identified patent application as follows:

Amendments to the Specification begin on page 2 of this paper.

Amendments to the Claims are shown in the listing of claims which begin on page 3 of this paper.

Amendments to the Drawings begin on page 4 of this paper and include both an attached replacement sheet and an annotated sheet showing changes.

Remarks begin on page 5 of this paper.

1

AMENDMENTS TO THE SPECIFICATION

Submitted herewith is a marked-up and clean version of a substitute specification. No new matter is believed to have been added therein.

2

AMENDMENTS TO THE CLAIMS

This listing of claims will replace all prior versions, and listings, of claims in the application.

Listing of Claims:

1. (Original) In a multicarrier modulation system including a first transceiver in communication with a second transceiver using a transmission signal having a plurality of carrier signals for modulating an input bit stream, each carrier signal having a phase characteristic associated with the input bit stream, a method for scrambling the phase characteristics of the carrier signals comprising:

associating each carrier signal with a value determined independently of any input bit value carried by that carrier signal;

computing a phase shift for each carrier signal based on the value associated with that carrier signal; and

combining the phase shift computed for each carrier signal with the phase characteristic of that carrier signal so as to substantially scramble the phase characteristics of the plurality of carrier signals.

2.-20. (Cancelled)

3

AMENDMENTS TO THE DRAWINGS:

The attached drawing sheet(s) include(s) changes to Figure 1. This sheet, which includes Figure 1 replaces the original Figure 1 sheet.

REMARKS/ARGUMENTS

By this amendment, claims 2-20 are canceled without prejudice or disclaimer. Additionally, the specification and figures have been amended in a manner consistent with the parent application.

Applicant requests examination on the merits.

Applicant believes that the pending claims are in condition for allowance and such disposition is respectfully requested. In the event that a telephone conversation would further prosecution and/or expedite allowance, the Examiner is invited to contact the undersigned.

The Commissioner is hereby authorized to charge to Deposit Account No. 19-1970 any fees under 37 C.F.R. §§ 1.16 and 1.17 that may be required by this paper and to credit any overpayment to that Account. If any extension of time is required in connection with the filing of this paper and has not been separately requested, such extension is hereby Petitioned.

Respectfully submitted,

SHERIDAN ROSS P.C.

By:

Jason H. Vick

Registration No. 45,285 1560 Broadway, Suite 1200 Denver, Colorado 80202-5141

(303) 863-9700

Date: 27 NN 11

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A System and Method for Scrambling the Phase of the Carriers in a Multicarrier Communications System

Related Application

This application claims the benefit of the filing date of copending U.S. Provisional Application, Serial No. 60/164,134, filed November 9, 1999, entitled "A Method For Randomizing The Phase Of The Carriers In A Multicarrier Communications System To Reduce The Peak To Average Power Ratio Of The Transmitted Signal," the entirety of which provisional application is incorporated by reference herein.

Field of the Invention

This invention relates to communications systems using multicarrier modulation. More particularly, the invention relates to multicarrier communications systems that lower the peak-to-average power ratio (PAR) of transmitted signals.

Background of the Invention

In a conventional multicarrier communications system, transmitters communicate over a communication channel using multicarrier modulation or Discrete Multitone Modulation (DMT). Carrier signals (carriers) or sub-channels spaced within a usable frequency band of the communication channel are modulated at a symbol (i.e., block) transmission rate of the system. An input signal, which includes input data bits, is sent to a DMT transmitter, such as a DMT modem. The DMT transmitter typically modulates the phase characteristic, or phase, and amplitude of the

carrier signals using an Inverse Fast Fourier Transform (IFFT) to generate a time domain signal, or transmission signal, that represents the input signal. The DMT transmitter transmits the transmission signal, which is a linear combination of the multiple carriers, to a DMT receiver over the communication channel.

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The phase and amplitude of the carrier signals of DMT transmission signal can be considered random because the phase and amplitude result from the modulation of an arbitrary sequence of input data bits comprising the transmitted information. Therefore, under the condition that the modulated data bit stream is random, the DMT transmission signal can be approximated as having a Gaussian probability distribution. A bit scrambler is often used in the DMT transmitter to scramble the input data bits before the bits are modulated to assure that the transmitted data bits are random and, consequently, that the modulation of those bits produces a DMT transmission signal with a Gaussian probability distribution.

With an appropriate allocation of transmit power levels to the carriers or sub-channels, such a system provides a desirable performance. Further, generating a transmission signal with a Gaussian probability distribution is important in order to transmit a transmission signal with a low peak-to-average ratio (PAR), or peak-to-average power ratio. The PAR of a transmission signal is the ratio of the instantaneous peak value (i.e., maximum magnitude) of a signal parameter (e.g., voltage, current, phase, frequency, power) to the time-averaged value of the signal parameter. In DMT systems, the PAR of the transmitted signal is determined by the probability of the random transmission signal reaching a certain peak voltage during the time interval required for a certain number of symbols. An example of the PAR of a transmission signal transmitted from a DMT transmitter is 14.5 dB, which is equivalent to having a 1E-7 probability of clipping. The PAR of a

transmission signal transmitted and received in a DMT communication system is an important consideration in the design of the DMT communication system because the PAR of a signal affects the communication system's total power consumption and component linearity requirements of the system.

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If the phase of the modulated carriers is not random, then the PAR can increase greatly. Examples of cases where the phases of the modulated carrier signals are not random are when bit scramblers are not used, multiple carrier signals are used to modulate the same input data bits, and the constellation maps, which are mappings of input data bits to the phase of a carrier signal, used for modulation are not random enough (i.e., a zero value for a data bit corresponds to a 90 degree phase characteristic of the DMT carrier signal and a one value for a data bit corresponds to a –90 degree phase characteristic of the DMT carrier signal). An increased PAR can result in a system with high power consumption and/or with high probability of clipping the transmission signal. Thus, there remains a need for a system and method that can effectively scramble the phase of the modulated carrier signals in order to provide a low PAR for the transmission signal.

Summary of the Invention

The present invention features a system and method that scrambles the phase characteristics of the modulated carrier signals in a transmission signal. In one aspect, a value is associated with each carrier signal. A phase shift is computed for each carrier signal based on the value associated with that carrier signal. The value is determined independently of any input bit value carried by that carrier signal. The phase shift computed for each carrier signal is combined with the phase characteristic of that carrier signal to substantially scramble the phase characteristics of the carrier signals.

In one embodiment, the input bit stream is modulated onto the carrier signals having the substantially scrambled phase characteristic to produce a transmission signal with a reduced peak-to-average power ratio (PAR). The value is derived from a predetermined parameter, such as a random number generator, a carrier number, a DMT symbol count, a superframe count, and a hyperframe count. In another embodiment, a predetermined transmission signal is transmitted when the amplitude of the transmission signal exceeds a certain level.

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In another aspect, the invention features a method wherein a value is associated with each carrier signal. The value is determined independently of any input bit value carried by that carrier signal. A phase shift for each carrier signal is computed based on the value associated with that carrier signal. The transmission signal is demodulated using the phase shift computed for each carrier signal.

In another aspect, the invention features a system comprising a phase scrambler that computes a phase shift for each carrier signal based on a value associated with that carrier signal. The phase scrambler also combines the phase shift computed for each carrier signal with the phase characteristic of that carrier signal to substantially scramble the phase characteristic of the carrier signals. In one embodiment, a modulator, in communication with the phase scrambler, modulates bits of an input signal onto the carrier signals having the substantially scrambled phase characteristics to produce a transmission signal with a reduced PAR.

Description of the Drawings

The invention is pointed out with particularity in the appended claims. The advantages of the invention described above, as well as further advantages of the invention, may be better understood

by reference to the following description taken in conjunction with the accompanying drawings, in which:

Fig. 1 is a block diagram of an embodiment of a digital subscriber line communications system including a DMT (discrete multitone modulation) transceiver, in communication with a remote transceiver, having a phase scrambler for substantially scrambling the phase characteristics of carrier signals; and

Fig. 2 is a flow diagram of an embodiment of a process for scrambling the phase characteristics of the carrier signals in a transmission signal.

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Detailed Description

Fig. 1 shows a digital subscriber line (DSL) communication system 2 including a discrete multitone (DMT) transceiver 10 in communication with a remote transceiver 14 over a communication channel 18 using a transmission signal 38 having a plurality of carrier signals. The DMT transceiver 10 includes a DMT transmitter 22 and a DMT receiver 26. The remote transceiver 14 includes a transmitter 30 and a receiver 34. Although described with respect to discrete multitone modulation, the principles of the invention apply also to other types of multicarrier modulation, such as, but not limited to, orthogonally multiplexed quadrature amplitude modulation (OQAM), discrete wavelet multitone (DWMT) modulation, and orthogonal frequency division multiplexing (OFDM).

The communication channel 18 provides a downstream transmission path from the DMT transmitter 22 to the remote receiver 34, and an upstream transmission path from the remote transmitter 30 to the DMT receiver 26. In one embodiment, the communication channel 18 is a pair of twisted wires of a telephone subscriber line. In other embodiments, the communication channel 18 can be a fiber optic wire, a quad cable, consisting of two pairs of twisted wires, or a quad cable

that is one of a star quad cable, a Dieselhorst-Martin quad cable, and the like. In a wireless communication system wherein the transceivers 10, 14 are wireless modems, the communication channel 18 is the air through which the transmission signal 38 travels between the transceivers 10, 14.

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By way of example, the DMT transmitter 22 shown in Fig. 1 includes a quadrature amplitude modulation (QAM) encoder 42, a modulator 46, a bit allocation table (BAT) 44, and a phase scrambler 66. The DMT transmitter 22 can also include a bit scrambler 74, as described further below. The remote transmitter 30 of the remote transceiver 14 comprises equivalent components as the DMT transmitter 22. Although this embodiment specifies a detailed description of the DMT transmitter 22, the inventive concepts apply also to the receivers 34, 36 which have similar components to that of the DMT transmitter 22, but perform inverse functions in a reverse order.

The QAM encoder 42 has a single input for receiving an input serial data bit stream 54 and multiple parallel outputs to transmit QAM symbols 58 generated by the QAM encoder 42 from the bit stream 54. In general, the QAM encoder 42 maps the input serial bit-stream 54 in the time domain into parallel QAM symbols 58 in the frequency domain. In particular, the QAM encoder 42 maps the input serial data bit stream 54 into N parallel quadrature amplitude modulation (QAM) constellation points 58, or QAM symbols 58, where N represents the number of carrier signals generated by the modulator 46. The BAT 44 is in communication with the QAM encoder 42 to specify the number of bits carried by each carrier signal. The QAM symbols 58 represent the amplitude and the phase characteristic of each carrier signal.

The modulator 46 provides functionality associated with the DMT modulation and transforms the QAM symbols 58 into DMT symbols 70 each comprised of a plurality of time-

domain samples. The modulator 46 modulates each carrier signal with a different QAM symbol 58. As a result of this modulation, carrier signals have phase and amplitude characteristics based on the QAM symbol 58 and therefore based on the input-bit stream 54. In particular, the modulator 46 uses an inverse fast Fourier transform (IFFT) to change the QAM symbols 58 into a transmission signal 38 comprised of a sequence of DMT symbols 70. The modulator 46 changes the QAM symbols 58 into DMT symbols 70 through modulation of the carrier signals. In another embodiment, the modulator 46 uses the inverse discrete Fourier transform (IDFT) to change the QAM symbols 58 into DMT symbols 70. In one embodiment, a pilot tone is included in the transmission signal 38 to provide a reference signal for coherent demodulation of the carrier signals in the remote receiver 34 during reception of the transmission signal 38.

The modulator 46 also includes a phase scrambler 66 that combines a phase shift computed for each QAM-modulated carrier signal with the phase characteristic of that carrier signal.

Combining phase shifts with phase characteristics, in accordance with the principles of the invention, substantially scrambles the phase characteristics of the carrier signals in the transmission signal 38. By scrambling the phase characteristics of the carrier signals, the resulting transmission signal 38 has a substantially minimized peak-to-average (PAR) power ratio. The phase scrambler 66 can be part of or external to the modulator 46. Other embodiments of the phase scrambler 66 include, but are not limited to, a software program that is stored in local memory and is executed on the modulator 46, a digital signal processor (DSP) capable of performing mathematical functions and algorithms, and the like. The remote receiver 34 similarly includes a phase scrambler 66' for use when demodulating carrier signals that have had their phase characteristics adjusted by the phase scrambler 66 of the DMT transceiver 10.

To compute a phase shift for each carrier signal, the phase scrambler 66 associates one or more values with that carrier signal. The phase scrambler 66 determines each value for a carrier signal independently of the QAM symbols 58, and, therefore, independently of the bit value(s) modulated onto the carrier signal. The actual value(s) that the phase scrambler 66 associates with each carrier signal can be derived from one or more predefined parameters, such as a pseudo-random number generator (pseudo-RNG), a DMT carrier number, a DMT symbol count, a DMT superframe count, a DMT hyperframe count, and the like, as described in more detail below. Irrespective of the technique used to produce each value, the same technique is used by the DMT transmitter 22 and the remote receiver 34 so that the value associated with a given carrier signal is known at both ends of the communication channel 18.

The phase scrambler 66 then solves a predetermined equation to compute a phase shift for the carrier signal, using the value(s) associated with that carrier signal as input that effects the output of the equation. Any equation suitable for computing phase shifts can be used to compute the phase shifts. When the equation is independent of the bit values of the input serial bit stream 54, the computed phase shifts are also independent of such bit values.

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In one embodiment (shown in phantom), the DMT transmitter 22 includes a bit scrambler 74, which receives the input serial bit stream 54 and outputs data bits 76 that are substantially scrambled. The substantially scrambled bits 76 are then passed to the QAM encoder 42. When the bit scrambler 74 is included in the DMT transmitter 22, the operation of the phase scrambler 66 further assures that the transmission signal 38 has a Gaussian probability distribution and, therefore, a substantially minimized PAR.

Fig. 2 shows embodiments of a process used by the DMT transmitter 22 for adjusting the phase characteristic of each carrier signal and combining these carrier signals to produce the transmission signal 38. The DMT transmitter 22 generates (step 100) a value that is associated with a carrier signal. Because the value is being used to alter the phase characteristics of the carrier signal, both the DMT transmitter 22 and the remote receiver 34 must recognize the value as being associated with the carrier signal. Either the DMT transmitter 22 and the remote receiver 34 independently derive the associated value, or one informs the other of the associated value. For example, in one embodiment the DMT transmitter 22 can derive the value from a pseudo-RNG and then transmit the generated value to the remote receiver 34. In another embodiment, the remote receiver 34 similarly derives the value from the same pseudo-RNG and the same seed as used by the transmitter (i.e., the transmitter pseudo-RNG produces the same series of random numbers as the receiver pseudo-RNG).

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As another example, the DMT transmitter 22 and the remote receiver 34 can each maintain a symbol counter for counting DMT symbols. The DMT transmitter 22 increments its symbol counter upon transmitting a DMT symbol; the remote receiver 34 upon receipt. Thus, when the DMT transmitter 22 and the remote receiver 34 both use the symbol count as a value for computing phase shifts, both the DMT transmitter 22 and remote receiver 34 "know" that the value is associated with a particular DMT symbol and with each carrier signal of that DMT symbol.

Values can also be derived from other types of predefined parameters. For example, if the predefined parameter is the DMT carrier number, then the value associated with a particular carrier signal is the carrier number of that signal within the DMT symbol. The number of a carrier signal represents the location of the frequency of the carrier signal relative to the frequency of other carrier

signals within a DMT symbol. For example, in one embodiment the DSL communication system 2 provides 256 carrier signals, each separated by a frequency of 4.3125 kHz and spanning the frequency bandwidth from 0 kHz to 1104 kHz. The DMT transmitter 22 numbers the carrier signals from 0 to 255. Therefore, "DMT carrier number 50" represents the 51st DMT carrier signal which is located at the frequency of 215.625 kHz (i.e., 51 x 4.3125 kHz).

Again, the DMT transmitter 22 and the remote receiver 34 can know the value that is associated with the carrier signal because both the DMT transmitter 22 and the remote receiver 34 use the same predefined parameter (here, the DMT carrier number) to make the value-carrier signal association. In other embodiments (as exemplified above with the transmitter pseudo-RNG), the DMT transmitter 22 can transmit the value to the remote receiver 34 (or vice versa) over the communication channel 18.

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In other embodiments, other predefined parameters can be used in conjunction with the symbol count. One example of such a predefined parameter is the superframe count that increments by one every 69 DMT symbols. One exemplary implementation that achieves the superframe counter is to perform a modulo 68 operation on the symbol count. As another example, the DMT transmitter 22 can maintain a hyperframe counter for counting hyperframes. An exemplary implementation of the hyperframe count is to perform a modulo 255 operation on the superframe count. Thus, the hyperframe count increments by one each time the superframe count reaches 255.

Accordingly, it is seen that some predefined parameters produce values that vary from carrier signal to carrier signal. For example, when the predefined parameter is the DMT carrier number, values vary based on the frequency of the carrier signal. As another example, the pseudo-RNG generates a new random value for each carrier signal.

Other predefined parameters produce values that vary from DMT symbol 70 to DMT symbol 70. For example, when the predefined parameter is the symbol count, the superframe count, or hyperframe count, values vary based on the numerical position of the DMT symbol 70 within a sequence of symbols, superframes, or hyperframes. Predefined parameters such as the pseudo-RNG, symbol count, superframe count, and superframe can also be understood to be parameters that vary values over time. Any one or combination of the predefined parameters can provide values for input to the equation that computes a phase shift for a given carrier signal.

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In one embodiment, the phase scrambling is used to avoid clipping of the transmission signal 38 on a DMT symbol 70 by DMT symbol 70 basis. In this embodiment, the DMT transmitter 22 uses a value based on a predefined parameter that varies over time, such as the symbol count, to compute the phase shift. It is to be understood that other types of predefined parameters that vary the values associated with carrier signals can be used to practice the principles of the invention. As described above, the transceivers 10, 14 may communicate (step 110) the values to synchronize their use in modulating and demodulating the carrier signals.

The DMT transmitter 22 then computes (step 115) the phase shift that is used to adjust the phase characteristic of each carrier signal. The amount of the phase shift combined with the phase characteristic of each QAM-modulated carrier signal depends upon the equation used and the one or more values associated with that carrier signal.

The DMT transmitter 22 then combines (step 120) the phase shift computed for each carrier signal with the phase characteristic of that carrier signal. By scrambling the phase characteristics of the carrier signals, the phase scrambler 66 reduces (with respect to unscrambled phase characteristics) the combined PAR of the plurality of carrier signals and, consequently, the

transmission signal 38. The following three phase shifting examples, PS #1 – PS #3, illustrate methods used by the phase scrambler 66 to combine a computed phase shift to the phase characteristic of each carrier signal.

Phase Shifting Example #1

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Phase shifting example #1 (PS #1) corresponds to adjusting the phase characteristic of the QAM-modulated carrier signal associated with a carrier number N by $N \times \frac{\pi}{3}$, modulo (mod) 2π . In this example, a carrier signal having a carrier number N equal to 50 has a phase shift added to the phase characteristic of that carrier signal equal to $50 \times \frac{\pi}{3} \pmod{2\pi} = \frac{2}{3}\pi$. The carrier signal with a carrier number N equal to 51 has a phase shift added to the phase characteristic of that carrier signal equal to $51 \times \frac{\pi}{3} \pmod{2\pi} = \pi$. The carrier signal with a carrier number N equal to 0 has no phase shift added to the phase characteristic of that carrier signal.

Phase Shifting Example #2

Phase shifting example #2 (PS #2) corresponds to adjusting the phase characteristic of the QAM-modulated carrier signal associated with a carrier number N by $(N+M) \times \frac{\pi}{4}$, mod 2π , where M is the symbol count. In this example, a carrier signal having a carrier number N equal to 50 on DMT symbol count M equal to 8 has a phase shift added to the phase characteristic of that carrier signal equal to $(50+8) \times \frac{\pi}{4} (\text{mod } 2\pi) = \frac{\pi}{2}$. The carrier signal with the same carrier number N equal to 50 on the next DMT symbol count M equal to 9 has a phase shift added to the phase characteristic of that carrier signal equal to $(50+9) \times \frac{\pi}{4} (\text{mod } 2\pi) = \frac{3\pi}{4}$.

Phase Shifting Example #3

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Phase shifting example #3 (PS #3) corresponds to adjusting the phase characteristic of the QAM-modulated carrier signal associated with a carrier number N by $(X_N) \times \frac{\pi}{6}$, mod 2π , where X_N is an array of N pseudo-random numbers. In this example, a carrier signal having a carrier number N equal to 5 and X_N equal to [3, 8, 1, 4, 9, 5, ...] has a phase shift added to the phase characteristic of the carrier signal that is equal to $(9) \times \frac{\pi}{6} \pmod{2\pi} = \frac{\pi}{3}$. (Note that 9 is the 5th value in X_N .) The carrier signal with a carrier number N equal to 6 has a phase shift added to the phase characteristic of the carrier signal equal to $(5) \times \frac{\pi}{6} \pmod{2\pi} = \frac{5\pi}{3}$.

It is to be understood that additional and/or different phase shifting techniques can be used by the phase scrambler 66, and that PS #1, #2, and #3 are merely illustrative examples of the principles of the invention. The DMT transmitter 22 then combines (step 130) the carrier signals to form the transmission signal 38. If the transmission signal is not clipped, as described below, the DMT transmitter 22 consequently transmits (step 160) the transmission signal 38 to the remote receiver 34.

Clipping of Transmission Signals

A transmission signal 38 that has high peak values of voltage (i.e., a high PAR) can induce non-linear distortion in the DMT transmitter 22 and the communication channel 18. One form of this non-linear distortion of the transmission signal 38 that may occur is the limitation of the amplitude of the transmission signal 38 (i.e., clipping). For example, a particular DMT symbol 70 clips in the time domain when one or more time domain samples in that DMT symbol 70 are larger than the maximum allowed digital value for the DMT symbols 70. In multicarrier communication

systems when clipping occurs, the transmission signal 38 does not accurately represent the input serial data bit signal 54.

In one embodiment, the DSL communication system 2 avoids the clipping of the transmission signal 38 on a DMT symbol 70 by DMT symbol 70 basis. The DMT transmitter 22 detects (step 140) the clipping of the transmission signal 38. If a particular DMT symbol 70 clips in the time domain to produce a clipped transmission signal 38, the DMT transmitter 22 substitutes (step 150) a predefined transmission signal 78 for the clipped transmission signal 38.

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The predefined transmission signal 78 has the same duration as a DMT symbol 70 (e.g., 250 ms) in order to maintain symbol timing between the DMT transmitter 22 and the remote receiver 34. The predefined transmission signal 78 is not based on (i.e., independent of) the modulated input data bit stream 54; it is a bit value pattern that is recognized by the remote receiver 34 as a substituted signal. In one embodiment, the predefined transmission signal 78 is a known pseudo-random sequence pattern that is easily detected by the remote receiver 34. In another embodiment, the predefined transmission signal 78 is an "all zeros" signal, which is a zero voltage signal produced at the DMT transmitter 22 output (i.e., zero volts modulated on all the carrier signals). In addition to easy detection by the remote receiver 34, the zero voltage signal reduces the power consumption of the DMT transmitter 22 when delivered by the DMT transmitter 22. Further, a pilot tone is included in the predefined transmission signal 78 to provide a reference signal for coherent demodulation of the carrier signals in the remote receiver 34 during reception of the predefined transmission signal 78.

After the remote receiver 34 receives the transmission signal 38, the remote receiver 34 determines if the transmission signal 38 is equivalent to the predefined transmission signal 78. In

one embodiment, when the remote receiver 34 identifies the predefined transmission signal 78, the remote receiver 34 ignores (i.e., discards) the predefined transmission signal 78.

Following the transmission of the predefined transmission signal 78, the phase scrambler 66 shifts (step 120) the phase characteristic of the QAM-modulated carrier signals (based on one of the predefined parameters that varies over time). For example, consider that a set of QAM symbols 58 produces a DMT symbol 70 comprising a plurality of time domain samples, and that one of the time domain samples is larger than the maximum allowed digital value for the DMT symbol 70. Therefore, because the transmission signal 38 would be clipped when sent to the remote receiver 34, the DMT transmitter 22 sends the predefined transmission signal 78 instead.

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After transmission of the predefined transmission signal 78, the DMT transmitter 22 again attempts to send the same bit values that produced the clipped transmission signal 38 in a subsequent DMT symbol 70'. Because the generation of phase shifts in this embodiment is based on values that vary over time, the phase shifts computed for the subsequent DMT symbol 70' are different than those that were previously computed for the DMT symbol 70 with the clipped time domain sample. These different phase shifts are combined to the phase characteristics of the modulated carrier signals to produce carrier signals of the subsequent DMT symbol 70' with different phase characteristics than the carrier signals of the DMT symbol 70 with the clipped time domain sample.

DMT communication systems 2 infrequently produce transmission signals 38 that clip (e.g., approximately one clip every 10⁷ time domain samples 70). However, if the subsequent DMT symbol 70' includes a time domain sample that clips, then the predefined transmission signal 78 is again transmitted (step 150) to the remote receiver 34 instead of the clipped transmission signal 38. The clipping time domain sample may be on the same or on a different carrier signal than the

previously clipped DMT symbol 70. The DMT transmitter 22 repeats the transmission of the predefined transmission signal 78 until the DMT transmitter 22 produces a subsequent DMT symbol 70' that is not clipped. When the DMT transmitter 22 produces a DMT symbol 70' that is not clipped, the DTM transmitter 22 transmits (step 160) the transmission signal 38 to the remote receiver 34. The probability of a DMT symbol 70 producing a transmission signal 38 that clips in the time domain depends on the PAR of the transmission signal 38.

For example, the following phase shifting example, PST #4, illustrates the method used by the phase scrambler 66 to combine a different phase shift to the phase characteristic of each carrier signal to avoid the clipping of the transmission signal 38.

Phase Shifting Example #4

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Phase shifting example #4 (PS #4) corresponds to adjusting the phase characteristic of the carrier signal associated with a carrier number N by $\frac{\pi}{3} \times (M+N)$, mod 2π , where M is the DMT symbol count. In this example, if the DMT symbol 70 clips when the DMT symbol count M equals 5, the predefined transmission signal 78 is transmitted instead of the current clipped transmission signal 38. On the following DMT symbol period, the DMT count M equals 6, thereby causing a different set of time domain samples to be generated for the subsequent DMT symbol 70', although the QAM symbols 58 used to produce both DMT symbols 70, 70' are the same.

If this different set of time domain samples (and consequently the transmission signal 38) is not clipped, the DMT transmitter 22 sends the transmission signal 38. If one of the time domain samples in the different set of time domain samples 70 (and consequently the transmission signal 38) is clipped, then the DMT transmitter 22 sends the predefined transmission signal 78 again. The process continues until a DMT symbol 70 is produced without a time domain sample 70 that is

clipped. In one embodiment, the transmitter 22 stops attempting to produce a non-clipped DMT symbol 70' for the particular set of QAM symbols 58 after generating a predetermined number of clipped DMT symbols 70'. At that moment, the transmitter 22 can transmit the most recently produced clipped DMT symbol 70' or the predetermined transmission signal 78.

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The PAR of the DSL communication system 2 is reduced because the predefined transmission signal 78 is sent instead of the transmission signal 38 when the DMT symbol 70 clips. For example, a DMT communication system 2 that normally has a clipping probability of 10^{-7} for the time domain transmission signal 38 can therefore operate with a 10^{-5} probability of clipping and a lower PAR equal to 12.8 dB (as compared to 14.5 dB). When operating at a 10^{-5} probability of clipping, assuming a DMT symbol 70 has 512 time-domain samples 70, the DMT transmitter 22 experiences one clipped DMT symbol 70 out of every $\frac{10^5}{512}$, or 195 DMT symbols 70. This results in the predefined (non-data carrying) transmission signal 78 being transmitted, on average, once every 195 DMT symbols. Although increasing the probability of clipping to 10^{-5} results in approximately a 0.5% (1/195) decrease in throughput, the PAR of the transmission signal 38 is reduced by 1.7 dB, which reduces transmitter complexity in the form of power consumption and component linearity.

While the invention has been shown and described with reference to specific preferred embodiments, it should be understood by those skilled in the art that various changes in form and detail may be made therein without departing from the spirit and scope of the invention as defined by the following claims. For example, although the specification uses DSL to describe the invention, it is to be understood that various form of DSL can be used, e.g., ADSL, VDSL, SDSL, HDSL, HDSL2, or SHDSL. It is also to be understood that the principles of the invention apply to

various types of applications transported over DSL systems (e.g., telecommuting, video conferencing, high speed Internet access, video-on demand).

What is Claimed:

1. In a multicarrier modulation system including a first transceiver in communication with a second transceiver using a transmission signal having a plurality of carrier signals for modulating an input bit stream, each carrier signal having a phase characteristic associated with the input bit stream, a method for scrambling the phase characteristics of the carrier signals comprising:

associating each carrier signal with a value determined independently of any input bit value carried by that carrier signal;

computing a phase shift for each carrier signal based on the value associated with that carrier signal; and

combining the phase shift computed for each carrier signal with the phase characteristic of that carrier signal so as to substantially scramble the phase characteristics of the plurality of carrier signals.

- 2. The method of claim 1 further comprising modulating bits of the input bit stream onto the carrier signals having the substantially scrambled phase characteristics to produce a transmission signal with a reduced peak-to-average power ratio (PAR).
- 3. The method of claim 1 further comprising independently deriving the value associated with each carrier signal at each transceiver.
- 4. The method of claim 1 further comprising transmitting the value associated with each carrier signal from one transceiver to the other transceiver.

- 5. The method of claim 1 further comprising maintaining synchronization between the transceivers using the value associated with each carrier signal.
 - 6. The method of claim 1 wherein the value varies with each carrier signal.
 - 7. The method of claim 1 wherein the value varies with each DMT symbol.
 - 8. The method of claim 1 wherein the value is derived from a predetermined parameter.
 - 9. The method of claim 8 wherein the predefined parameter is a carrier number.
 - 10. The method of claim 8 wherein the predefined parameter is a symbol count.
 - 11. The method of claim 8 wherein the predefined parameter is a hyperframe count.
 - 12. The method of claim 8 wherein the predefined parameter is a superframe count.
 - 13. The method of claim 1 further comprising scrambling the bits of the input bit stream.
- 14. The method of claim 1 further comprising transmitting a predetermined transmission signal when the amplitude of the transmission signal exceeds a certain level.
- 15. The method of claim 14 wherein the predetermined transmission signal comprises a predetermined pattern of bits.

- 16. The method of claim 14 wherein the predetermined transmission signal comprises a pilot tone.
- 17. The method of claim 16 wherein the pilot tone is used to maintain timing synchronization between the first transceiver and the second transceiver.
- 18. The method of claim 15 wherein each bit value in the predetermined pattern of bits is a zero value.
- 19. The method of claim 15 wherein the predetermined pattern of bits is a pseudorandom sequence pattern.
- 20. In a multicarrier modulation system including a first transceiver in communication with a second transceiver using a transmission signal having a plurality of carrier signals for modulating an input bit stream, each carrier signal having a phase characteristic with the input bit stream, a method for scrambling the phase characteristics of the carrier signals comprising:

associating each carrier signal with a value determined independently of any input bit value carried by that carrier signal;

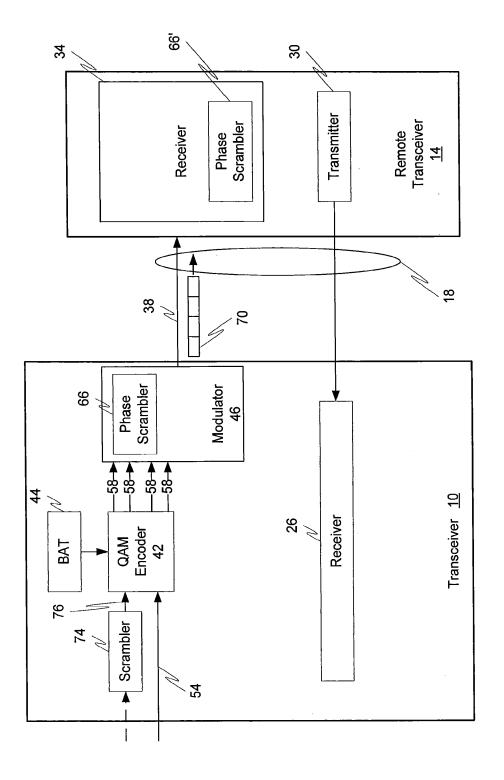
computing a phase shift for each carrier signal based on the value associated with that carrier signal; and

demodulating the transmission signal using the phase shift computed for each carrier signal.

Abstract

A system and method that scrambles the phase characteristic of a carrier signal are described. The scrambling of the phase characteristic of each carrier signal includes associating a value with each carrier signal and computing a phase shift for each carrier signal based on the value associated with that carrier signal. The value is determined independently of any input bit value carried by that carrier signal. The phase shift computed for each carrier signal is combined with the phase characteristic of that carrier signal so as to substantially scramble the phase characteristic of the carrier signals. Bits of an input signal are modulated onto the carrier signals having the substantially scrambled phase characteristic to produce a transmission signal with a reduced PAR.

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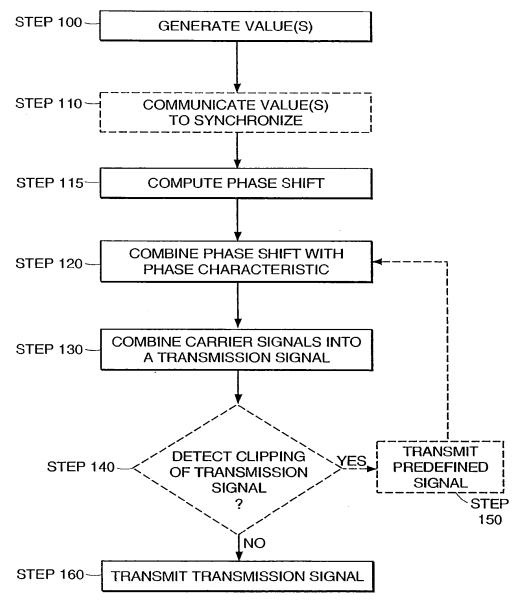


FIG. 2

A System and Method for Scrambling the Phase of the Carriers in a Multicarrier Communications System

Related Application

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This application is a Continuation of U.S. Patent Application No. 12/783,725, filed May 20, 2010, which is a Continuation of U.S. Patent Application No. 12/255,713, filed October 22, 2008, now U.S. Patent No. 7,769,104, which is a Continuation of U.S. Patent Application No. 11/863,581, filed September 28, 2007, now U.S. Patent No. 7,471,721, which is a Continuation of U.S. Application No. 11/211,535, filed August 26, 2005, now U.S. Patent No. 7,292,627, which is a Continuation of U.S. Patent Application No. 09/710,310, filed November 9, 2000, now U.S. Patent No. 6,961,369, whichThis application claims the benefit of the filing date of copending U.S. Provisional Application, Serial No. 60/164,134, filed November 9, 1999, entitled "A Method For Randomizing The Phase Of The Carriers In A Multicarrier Communications System To Reduce The Peak To Average Power Ratio Of The Transmitted Signal," each of which are incorporated herein by reference in their entiretythe entirety of which provisional application is incorporated by reference herein.

Field of the Invention

This invention relates to communications systems using multicarrier modulation.

More particularly, the invention relates to multicarrier communications systems that lower the peak-to-average power ratio (PAR) of transmitted signals.

Background of the Invention

In a conventional multicarrier communications system, transmitters communicate over a communication channel using multicarrier modulation or Discrete Multitone Modulation (DMT). Carrier signals (carriers) or sub-channels spaced within a usable frequency band of the communication channel are modulated at a symbol (i.e., block) transmission rate of the system. An input signal, which includes input data bits, is sent to a DMT transmitter, such as a DMT modem. The DMT transmitter typically modulates the phase characteristic, or phase, and amplitude of the carrier signals using an Inverse Fast Fourier Transform (IFFT) to generate a time domain signal, or transmission signal, that

represents the input signal. The DMT transmitter transmits the transmission signal, which is a linear combination of the multiple carriers, to a DMT receiver over the communication channel.

The phase and amplitude of the carrier signals of DMT transmission signal can be considered random because the phase and amplitude result from the modulation of an arbitrary sequence of input data bits comprising the transmitted information. Therefore, under the condition that the modulated data bit stream is random, the DMT transmission signal can be approximated as having a Gaussian probability distribution. A bit scrambler is often used in the DMT transmitter to scramble the input data bits before the bits are modulated to assure that the transmitted data bits are random and, consequently, that the modulation of those bits produces a DMT transmission signal with a Gaussian probability distribution.

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With an appropriate allocation of transmit power levels to the carriers or subchannels, such a system provides a desirable performance. Further, generating a transmission signal with a Gaussian probability distribution is important in order to transmit a transmission signal with a low peak-to-average ratio (PAR), or peak-to-average power ratio. The PAR of a transmission signal is the ratio of the instantaneous peak value (i.e., maximum magnitude) of a signal parameter (e.g., voltage, current, phase, frequency, power) to the time-averaged value of the signal parameter. In DMT systems, the PAR of the transmitted signal is determined by the probability of the random transmission signal reaching a certain peak voltage during the time interval required for a certain number of symbols. An example of the PAR of a transmission signal transmitted from a DMT transmitter is 14.5 dB, which is equivalent to having a 1E-7 probability of clipping. The PAR of a transmission signal transmitted and received in a DMT communication system is an important consideration in the design of the DMT communication system because the PAR of a signal affects the communication system's total power consumption and component linearity requirements of the system.

If the phase of the modulated carriers is not random, then the PAR can increase greatly. Examples of cases where the phases of the modulated carrier signals are not random are when bit scramblers are not used, multiple carrier signals are used to modulate the same

input data bits, and the constellation maps, which are mappings of input data bits to the phase of a carrier signal, used for modulation are not random enough (i.e., a zero value for a data bit corresponds to a 90 degree phase characteristic of the DMT carrier signal and a one value for a data bit corresponds to a -90 degree phase characteristic of the DMT carrier signal). An increased PAR can result in a system with high power consumption and/or with high probability of clipping the transmission signal. Thus, there remains a need for a system and method that can effectively scramble the phase of the modulated carrier signals in order to provide a low PAR for the transmission signal.

Summary of the Invention

The present invention features a system and method that scrambles the phase characteristics of the modulated carrier signals in a transmission signal. In one aspect, a value is associated with each carrier signal. A phase shift is computed for each carrier signal based on the value associated with that carrier signal. The value is determined independently of any input bit value carried by that carrier signal. The phase shift computed for each carrier signal is combined with the phase characteristic of that carrier signal to substantially scramble the phase characteristics of the carrier signals.

In one embodiment, the input bit stream is modulated onto the carrier signals having the substantially scrambled phase characteristic to produce a transmission signal with a reduced peak-to-average power ratio (PAR). The value is derived from a predetermined parameter, such as a random number generator, a carrier number, a DMT symbol count, a superframe count, and a hyperframe count. In another embodiment, a predetermined transmission signal is transmitted when the amplitude of the transmission signal exceeds a certain level.

In another aspect, the invention features a method wherein a value is associated with each carrier signal. The value is determined independently of any input bit value carried by that carrier signal. A phase shift for each carrier signal is computed based on the value associated with that carrier signal. The transmission signal is demodulated using the phase shift computed for each carrier signal.

In another aspect, the invention features a system comprising a phase scrambler that computes a phase shift for each carrier signal based on a value associated with that carrier signal. The phase scrambler also combines the phase shift computed for each carrier signal with the phase characteristic of that carrier signal to substantially scramble the phase characteristic of the carrier signals. In one embodiment, a modulator, in communication with the phase scrambler, modulates bits of an input signal onto the carrier signals having the substantially scrambled phase characteristics to produce a transmission signal with a reduced PAR.

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Description of the Drawings

The invention is pointed out with particularity in the appended claims. The advantages of the invention described above, as well as further advantages of the invention, may be better understood by reference to the following description taken in conjunction with the accompanying drawings, in which:

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FIG. 1 is a block diagram of an embodiment of a digital subscriber line communications system including a DMT (discrete multitone modulation) transceiver, in communication with a remote transceiver, having a phase scrambler for substantially scrambling the phase characteristics of carrier signals; and

FIG. 2 is a flow diagram of an embodiment of a process for scrambling the phase characteristics of the carrier signals in a transmission signal.

Detailed Description

FIG. 1 shows a digital subscriber line (DSL) communication system 2 including a discrete multitone (DMT) transceiver 10 in communication with a remote transceiver 14 over a communication channel 18 using a transmission signal 38 having a plurality of carrier signals. The DMT transceiver 10 includes a DMT transmitter 22 and a DMT receiver 26. The remote transceiver 14 includes a transmitter 30 and a receiver 34. Although described with respect to discrete multitone modulation, the principles of the invention apply also to other types of multicarrier modulation, such as, but not limited to, orthogonally multiplexed quadrature amplitude modulation (OQAM), discrete wavelet multitone (DWMT) modulation, and orthogonal frequency division multiplexing (OFDM).

The communication channel 18 provides a downstream transmission path from the DMT transmitter 22 to the remote receiver 34, and an upstream transmission path from the remote transmitter 30 to the DMT receiver 26. In one embodiment, the communication channel 18 is a pair of twisted wires of a telephone subscriber line. In other embodiments, the communication channel 18 can be a fiber optic wire, a quad cable, consisting of two pairs of twisted wires, or a quad cable that is one of a star quad cable, a Dieselhorst-Martin quad cable, and the like. In a wireless communication system wherein the transceivers 10, 14 are

wireless modems, the communication channel 18 is the air through which the transmission signal 38 travels between the transceivers 10, 14.

By way of example, the DMT transmitter 22 shown in FIG. 1 includes a quadrature amplitude modulation (QAM) encoder 42, a modulator 46, a bit allocation table (BAT) 44, and a phase scrambler 66. The DMT transmitter 22 can also include a bit scrambler 74, as described further below. The remote transmitter 30 of the remote transceiver 14 comprises equivalent components as the DMT transmitter 22. Although this embodiment specifies a detailed description of the DMT transmitter 22, the inventive concepts apply also to the receivers 34, 2436 which have similar components to that of the DMT transmitter 22, but perform inverse functions in a reverse order.

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The QAM encoder 42 has a single input for receiving an input serial data bit stream 54 and multiple parallel outputs to transmit QAM symbols 58 generated by the QAM encoder 42 from the bit stream 54. In general, the QAM encoder 42 maps the input serial bit-stream 54 in the time domain into parallel QAM symbols 58 in the frequency domain. In particular, the QAM encoder 42 maps the input serial data bit stream 54 into N parallel quadrature amplitude modulation (QAM) constellation points 58, or QAM symbols 58, where N represents the number of carrier signals generated by the modulator 46. The BAT 44 is in communication with the QAM encoder 42 to specify the number of bits carried by each carrier signal. The QAM symbols 58 represent the amplitude and the phase characteristic of each carrier signal.

The modulator 46 provides functionality associated with the DMT modulation and transforms the QAM symbols 58 into DMT symbols 70 each comprised of a plurality of time-domain samples. The modulator 46 modulates each carrier signal with a different QAM symbol 58. As a result of this modulation, carrier signals have phase and amplitude characteristics based on the QAM symbol 58 and therefore based on the input-bit stream 54. In particular, the modulator 46 uses an inverse fast Fourier transform (IFFT) to change the QAM symbols 58 into a transmission signal 38 comprised of a sequence of DMT symbols 70. The modulator 46 changes the QAM symbols 58 into DMT symbols 70 through modulation of the carrier signals. In another embodiment, the modulator 46 uses the inverse discrete Fourier transform (IDFT) to change the QAM symbols 58 into DMT symbols 70. In

one embodiment, a pilot tone is included in the transmission signal 38 to provide a reference signal for coherent demodulation of the carrier signals in the remote receiver 34 during reception of the transmission signal 38.

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The modulator 46 also includes a phase scrambler 66 that combines a phase shift computed for each QAM-modulated carrier signal with the phase characteristic of that carrier signal. Combining phase shifts with phase characteristics, in accordance with the principles of the invention, substantially scrambles the phase characteristics of the carrier signals in the transmission signal 38. By scrambling the phase characteristics of the carrier signals, the resulting transmission signal 38 has a substantially minimized peak-to-average (PAR) power ratio. The phase scrambler 66 can be part of or external to the modulator 46. Other embodiments of the phase scrambler 66 include, but are not limited to, a software program that is stored in local memory and is executed on the modulator 46, a digital signal processor (DSP) capable of performing mathematical functions and algorithms, and the like. The remote receiver 34 similarly includes a phase scrambler descrambler 66' for use when demodulating carrier signals that have had their phase characteristics adjusted by the phase scrambler 66 of the DMT transceiver 10.

To compute a phase shift for each carrier signal, the phase scrambler 66 associates one or more values with that carrier signal. The phase scrambler 66 determines each value for a carrier signal independently of the QAM symbols 58, and, therefore, independently of the bit value(s) modulated onto the carrier signal. The actual value(s) that the phase scrambler 66 associates with each carrier signal can be derived from one or more predefined parameters, such as a pseudo-random number generator (pseudo-RNG), a DMT carrier number, a DMT symbol count, a DMT superframe count, a DMT hyperframe count, and the like, as described in more detail below. Irrespective of the technique used to produce each value, the same technique is used by the DMT transmitter 22 and the remote receiver 34 so that the value associated with a given carrier signal is known at both ends of the communication channel 18.

The phase scrambler 66 then solves a predetermined equation to compute a phase shift for the carrier signal, using the value(s) associated with that carrier signal as input that effects the output of the equation. Any equation suitable for computing phase shifts can be

used to compute the phase shifts. When the equation is independent of the bit values of the input serial bit stream 54, the computed phase shifts are also independent of such bit values.

In one embodiment (shown in phantom), the DMT transmitter 22 includes a bit scrambler 74, which receives the input serial bit stream 54 and outputs data bits 76 that are substantially scrambled. The substantially scrambled bits 76 are then passed to the QAM encoder 42. When the bit scrambler 74 is included in the DMT transmitter 22, the operation of the phase scrambler 66 further assures that the transmission signal 38 has a Gaussian probability distribution and, therefore, a substantially minimized PAR.

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FIG. 2 shows embodiments of a process used by the DMT transmitter 22 for adjusting the phase characteristic of each carrier signal and combining these carrier signals to produce the transmission signal 38. The DMT transmitter 22 generates (step 100) a value that is associated with a carrier signal. Because the value is being used to alter the phase characteristics of the carrier signal, both the DMT transmitter 22 and the remote receiver 34 must recognize the value as being associated with the carrier signal. Either the DMT transmitter 22 and the remote receiver 34 independently derive the associated value, or one informs the other of the associated value. For example, in one embodiment the DMT transmitter 22 can derive the value from a pseudo-RNG and then transmit the generated value to the remote receiver 34. In another embodiment, the remote receiver 34 similarly derives the value from the same pseudo-RNG and the same seed as used by the transmitter (i.e., the transmitter pseudo-RNG produces the same series of random numbers as the receiver pseudo-RNG).

As another example, the DMT transmitter 22 and the remote receiver 34 can each maintain a symbol counter for counting DMT symbols. The DMT transmitter 22 increments its symbol counter upon transmitting a DMT symbol; the remote receiver 34 upon receipt. Thus, when the DMT transmitter 22 and the remote receiver 34 both use the symbol count as a value for computing phase shifts, both the DMT transmitter 22 and remote receiver 34 "know" that the value is associated with a particular DMT symbol and with each carrier signal of that DMT symbol.

Values can also be derived from other types of predefined parameters. For example, if the predefined parameter is the DMT carrier number, then the value associated with a particular carrier signal is the carrier number of that signal within the DMT symbol. The number of a carrier signal represents the location of the frequency of the carrier signal relative to the frequency of other carrier signals within a DMT symbol. For example, in one embodiment the DSL communication system 2 provides 256 carrier signals, each separated by a frequency of 4.3125 kHz and spanning the frequency bandwidth from 0 kHz to 1104 kHz. The DMT transmitter 22 numbers the carrier signals from 0 to 255. Therefore, "DMT carrier number 50" represents the 51st DMT carrier signal which is located at the frequency of 215.625 kHz (i.e., 51 x 4.3125 kHz).

Again, the DMT transmitter 22 and the remote receiver 34 can know the value that is associated with the carrier signal because both the DMT transmitter 22 and the remote receiver 34 use the same predefined parameter (here, the DMT carrier number) to make the value-carrier signal association. In other embodiments (as exemplified above with the transmitter pseudo-RNG), the DMT transmitter 22 can transmit the value to the remote receiver 34 (or vice versa) over the communication channel 18.

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In other embodiments, other predefined parameters can be used in conjunction with the symbol count. One example of such a predefined parameter is the superframe count that increments by one every 69 DMT symbols. One exemplary implementation that achieves the superframe counter is to perform a modulo 68 operation on the symbol count. As another example, the DMT transmitter 22 can maintain a hyperframe counter for counting hyperframes. An exemplary implementation of the hyperframe count is to perform a modulo 255 operation on the superframe count. Thus, the hyperframe count increments by one each time the superframe count reaches 255.

Accordingly, it is seen that some predefined parameters produce values that vary from carrier signal to carrier signal. For example, when the predefined parameter is the DMT carrier number, values vary based on the frequency of the carrier signal. As another example, the pseudo-RNG generates a new random value for each carrier signal.

Other predefined parameters produce values that vary from DMT symbol 70 to DMT symbol 70. For example, when the predefined parameter is the symbol count, the superframe

count, or hyperframe count, values vary based on the numerical position of the DMT symbol 70 within a sequence of symbols, superframes, or hyperframes. Predefined parameters such as the pseudo-RNG, symbol count, superframe count, and superframe can also be understood to be parameters that vary values over time. Any one or combination of the predefined parameters can provide values for input to the equation that computes a phase shift for a given carrier signal.

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In one embodiment, the phase scrambling is used to avoid clipping of the transmission signal 38 on a DMT symbol 70 by DMT symbol 70 basis. In this embodiment, the DMT transmitter 22 uses a value based on a predefined parameter that varies over time, such as the symbol count, to compute the phase shift. It is to be understood that other types of predefined parameters that vary the values associated with carrier signals can be used to practice the principles of the invention. As described above, the transceivers 10, 14 may communicate (step 110) the values to synchronize their use in modulating and demodulating the carrier signals.

The DMT transmitter 22 then computes (step 115) the phase shift that is used to adjust the phase characteristic of each carrier signal. The amount of the phase shift combined with the phase characteristic of each QAM-modulated carrier signal depends upon the equation used and the one or more values associated with that carrier signal.

The DMT transmitter 22 then combines (step 120) the phase shift computed for each carrier signal with the phase characteristic of that carrier signal. By scrambling the phase characteristics of the carrier signals, the phase scrambler 66 reduces (with respect to unscrambled phase characteristics) the combined PAR of the plurality of carrier signals and, consequently, the transmission signal 38. The following three phase shifting examples, PS #1-PS #3, illustrate methods used by the phase scrambler 66 to combine a computed phase shift to the phase characteristic of each carrier signal.

Phase Shifting Example #1

Phase shifting example #1 (PS #1) corresponds to adjusting the phase characteristic of

the QAM-modulated carrier signal associated with a carrier number N by $\frac{N}{3}$, modulo (mod) 2π . In this example, a carrier signal having a carrier number N equal to 50 has a phase

shift added to the phase characteristic of that carrier signal equal to $\frac{\pi}{3}$ (mod 2π) = $\frac{2}{3}\pi$.

The carrier signal with a carrier number N equal to 51 has a phase shift added to the phase

characteristic of that carrier signal equal to $51 \frac{\pi}{3}$ (mod 2π)= π . The carrier signal with the carrier number N equal to 0 has no phase shift added to the phase characteristic of that carrier signal.

Phase Shifting Example #2

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Phase shifting example #2 (PS #2) corresponds to adjusting the phase characteristic of the QAM-modulated carrier signal associated with a carrier number N by $\frac{(N+M)\times\frac{\pi}{4}}{4}$, mod 2π , where M is the symbol count. In this example, a carrier signal having a carrier number N equal to 50 on DMT symbol count M equal to 8 has a phase shift added to the phase

characteristic of that carrier signal equal to $(50+8) \times \frac{\pi}{4} \pmod{2\pi} = \frac{\pi}{2}$. The carrier signal with the same carrier number N equal to 50 on the next DMT symbol count M equal to 9 has a

phase shift added to the phase characteristic of that carrier signal equal to $\frac{3\pi}{4} \pmod{2\pi} = \frac{4}{4}.$

20 Phase Shifting Example #3

Phase shifting example #3 (PS #3) corresponds to adjusting the phase characteristic of the QAM-modulated carrier signal associated with a carrier number N by $\frac{\pi}{6}$, mod 2π ,

where $X \gg is$ an array of N pseudo-random numbers. In this example, a carrier signal having a carrier number N equal to 5 and $X \gg is$ equal to [3, 8, 1, 4, 9, 5, . . .] has a phase shift added

to the phase characteristic of the carrier signal that is equal to $\frac{\pi}{6} \pmod{2\pi} = \frac{3\pi}{2}$ (Note that 9 is the 5% value in XM.) The carrier signal with a carrier number N equal to 6 has a phase shift added to the phase characteristic of the carrier signal equal to

$$(5) \times \frac{\pi}{6} \pmod{2\pi} = \frac{5\pi}{6}$$

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It is to be understood that additional and/or different phase shifting techniques can be used by the phase scrambler 66, and that PS #1, #2, and #3 are merely illustrative examples of the principles of the invention. The DMT transmitter 22 then combines (step 130) the carrier signals to form the transmission signal 38. If the transmission signal is not clipped, as described below, the DMT transmitter 22 consequently transmits (step 160) the transmission signal 38 to the remote receiver 34.

15 Clipping of Transmission Signals

A transmission signal 38 that has high peak values of voltage (i.e., a high PAR) can induce non-linear distortion in the DMT transmitter 22 and the communication channel 18. One form of this non-linear distortion of the transmission signal 38 that may occur is the limitation of the amplitude of the transmission signal 38 (i.e., clipping). For example, a particular DMT symbol 70 clips in the time domain when one or more time domain samples in that DMT symbol 70 are larger than the maximum allowed digital value for the DMT symbols 70. In multicarrier communication systems when clipping occurs, the transmission signal 38 does not accurately represent the input serial data bit signal 54.

In one embodiment, the DSL communication system 2 avoids the clipping of the transmission signal 38 on a DMT symbol 70 by DMT symbol 70 basis. The DMT transmitter 22 detects (step 140) the clipping of the transmission signal 38. If a particular DMT symbol 70 clips in the time domain to produce a clipped transmission signal 38, the DMT transmitter

22 substitutes (step 150) a predefined transmission signal 78 for the clipped transmission signal 38.

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The predefined transmission signal 78 has the same duration as a DMT symbol 70 (e.g., 250 ms) in order to maintain symbol timing between the DMT transmitter 22 and the remote receiver 34. The predefined transmission signal 78 is not based on (i.e., independent of) the modulated input data bit stream 54; it is a bit value pattern that is recognized by the remote receiver 34 as a substituted signal. In one embodiment, the predefined transmission signal 78 is a known pseudo-random sequence pattern that is easily detected by the remote receiver 34. In another embodiment, the predefined transmission signal 78 is an "all zeros" signal, which is a zero voltage signal produced at the DMT transmitter 22 output (i.e., zero volts modulated on all the carrier signals). In addition to easy detection by the remote receiver 34, the zero voltage signal reduces the power consumption of the DMT transmitter 22 when delivered by the DMT transmitter 22. Further, a pilot tone is included in the predefined transmission signal 78 to provide a reference signal for coherent demodulation of the carrier signals in the remote receiver 34 during reception of the predefined transmission signal 78.

After the remote receiver 34 receives the transmission signal 38, the remote receiver 34 determines if the transmission signal 38 is equivalent to the predefined transmission signal 78. In one embodiment, when the remote receiver 34 identifies the predefined transmission signal 78, the remote receiver 34 ignores (i.e., discards) the predefined transmission signal 78.

Following the transmission of the predefined transmission signal 78, the phase scrambler 66 shifts (step 120) the phase characteristic of the QAM-modulated carrier signals (based on one of the predefined parameters that varies over time). For example, consider that a set of QAM symbols 58 produces a DMT symbol 70 comprising a plurality of time domain samples, and that one of the time domain samples is larger than the maximum allowed digital value for the DMT symbol 70. Therefore, because the transmission signal 38 would be clipped when sent to the remote receiver 34, the DMT transmitter 22 sends the predefined transmission signal 78 instead.

After transmission of the predefined transmission signal 78, the DMT transmitter 22 again attempts to send the same bit values that produced the clipped transmission signal 38 in a subsequent DMT symbol 70'. Because the generation of phase shifts in this embodiment is based on values that vary over time, the phase shifts computed for the subsequent DMT symbol 70' are different than those that were previously computed for the DMT symbol 70 with the clipped time domain sample. These different phase shifts are combined to the phase characteristics of the modulated carrier signals to produce carrier signals of the subsequent DMT symbol 70' with different phase characteristics than the carrier signals of the DMT symbol 70 with the clipped time domain sample.

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DMT communication systems 2 infrequently produce transmission signals 38 that clip (e.g., approximately one clip every 10⁷ time domain samples 70). However, if the subsequent DMT symbol 70' includes a time domain sample that clips, then the predefined transmission signal 78 is again transmitted (step 150) to the remote receiver 34 instead of the clipped transmission signal 38. The clipping time domain sample may be on the same or on a different carrier signal than the previously clipped DMT symbol 70. The DMT transmitter 22 repeats the transmission of the predefined transmission signal 78 until the DMT transmitter 22 produces a subsequent DMT symbol 70' that is not clipped. When the DMT transmitter 22 produces a DMT symbol 70' that is not clipped, the DTM transmitter 22 transmits (step 160) the transmission signal 38 to the remote receiver 34. The probability of a DMT symbol 70 producing a transmission signal 38 that clips in the time domain depends on the PAR of the transmission signal 38.

For example, the following phase shifting example, PST #4, illustrates the method used by the phase scrambler 66 to combine a different phase shift to the phase characteristic of each carrier signal to avoid the clipping of the transmission signal 38.

Phase Shifting Example #4

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Phase shifting example #4 (PS #4) corresponds to adjusting the phase characteristic of

the carrier signal associated with a carrier number N by $\frac{\pi}{3}$, mod 2π , where M is the DMT symbol count. In this example, if the DMT symbol 70 clips when the DMT symbol count M equals 5, the predefined transmission signal 78 is transmitted instead of the current clipped transmission signal 38. On the following DMT symbol period, the DMT count M equals 6, thereby causing a different set of time domain samples to be generated for the subsequent DMT symbol 70', although the QAM symbols 58 used to produce both DMT symbols 70, 70' are the same.

If this different set of time domain samples (and consequently the transmission signal 38) is not clipped, the DMT transmitter 22 sends the transmission signal 38. If one of the time domain samples in the different set of time domain samples 70 (and consequently the transmission signal 38) is clipped, then the DMT transmitter 22 sends the predefined transmission signal 78 again. The process continues until a DMT symbol 70 is produced without a time domain sample 70 that is clipped. In one embodiment, the transmitter 22 stops attempting to produce a non-clipped DMT symbol 70' for the particular set of QAM symbols 58 after generating a predetermined number of clipped DMT symbols 70'. At that moment, the transmitter 22 can transmit the most recently produced clipped DMT symbol 70' or the predetermined transmission signal 78.

The PAR of the DSL communication system 2 is reduced because the predefined transmission signal 78 is sent instead of the transmission signal 38 when the DMT symbol 70 clips. For example, a DMT communication system 2 that normally has a clipping probability of 10-7 for the time domain transmission signal 38 can therefore operate with a 10-5 probability of clipping and a lower PAR equal to 12.8 dB (as compared to 14.5 dB). When operating at a 10-5 probability of clipping, assuming a DMT symbol 70 has 512 time-domain samples 70, the DMT transmitter 22 experiences one clipped DMT symbol 70 out of every

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512, or 195 DMT symbols 70. This results in the predefined (non-data carrying) transmission signal 78 being transmitted, on average, once every 195 DMT symbols.

Although increasing the probability of clipping to 10 results in approximately a 0.5% (1/195) decrease in throughput, the PAR of the transmission signal 38 is reduced by 1.7 dB, which reduces transmitter complexity in the form of power consumption and component linearity.

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While the invention has been shown and described with reference to specific preferred embodiments, it should be understood by those skilled in the art that various changes in form and detail may be made therein without departing from the spirit and scope of the invention as defined by the following claims. For example, although the specification uses DSL to describe the invention, it is to be understood that various form of DSL can be used, e.g., ADSL, VDSL, SDSL, HDSL, HDSL2, or SHDSL. It is also to be understood that the principles of the invention apply to various types of applications transported over DSL systems (e.g., telecommuting, video conferencing, high speed Internet access, video-on demand).

Abstract

A system and method that scrambles the phase characteristic of a carrier signal are described. The scrambling of the phase characteristic of each carrier signal includes associating a value with each carrier signal and computing a phase shift for each carrier signal based on the value associated with that carrier signal. The value is determined independently of any input bit value carried by that carrier signal. The phase shift computed for each carrier signal is combined with the phase characteristic of that carrier signal so as to substantially scramble the phase characteristic of the carrier signals. Bits of an input signal are modulated onto the carrier signals having the substantially scrambled phase characteristic to produce a transmission signal with a reduced PAR.

A System and Method for Scrambling the Phase of the Carriers in a Multicarrier Communications System

Related Application

This application is a Continuation of U.S. Patent Application No. 12/783,725, filed May 20, 2010, which is a Continuation of U.S. Patent Application No. 12/255,713, filed October 22, 2008, now U.S. Patent No. 7,769,104, which is a Continuation of U.S. Patent Application No. 11/863,581, filed September 28, 2007, now U.S. Patent No. 7,471,721, which is a Continuation of U.S. Application No. 11/211,535, filed August 26, 2005, now U.S. Patent No. 7,292,627, which is a Continuation of U.S. Patent Application No. 09/710,310, filed November 9, 2000, now U.S. Patent No. 6,961,369, which claims the benefit of the filing date of copending U.S. Provisional Application, Serial No. 60/164,134, filed November 9, 1999, entitled "A Method For Randomizing The Phase Of The Carriers In A Multicarrier Communications System To Reduce The Peak To Average Power Ratio Of The Transmitted Signal," each of which are incorporated herein by reference in their entirety.

Field of the Invention

This invention relates to communications systems using multicarrier modulation. More particularly, the invention relates to multicarrier communications systems that lower the peak-to-average power ratio (PAR) of transmitted signals.

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Background of the Invention

In a conventional multicarrier communications system, transmitters communicate over a communication channel using multicarrier modulation or Discrete Multitone Modulation (DMT). Carrier signals (carriers) or sub-channels spaced within a usable frequency band of the communication channel are modulated at a symbol (i.e., block) transmission rate of the system. An input signal, which includes input data bits, is sent to a DMT transmitter, such as a DMT modem. The DMT transmitter typically modulates the phase characteristic, or phase, and amplitude of the carrier signals using an Inverse Fast Fourier Transform (IFFT) to generate a time domain signal, or transmission signal, that represents the input signal. The DMT transmitter transmits the transmission signal, which is a

linear combination of the multiple carriers, to a DMT receiver over the communication channel.

The phase and amplitude of the carrier signals of DMT transmission signal can be considered random because the phase and amplitude result from the modulation of an arbitrary sequence of input data bits comprising the transmitted information. Therefore, under the condition that the modulated data bit stream is random, the DMT transmission signal can be approximated as having a Gaussian probability distribution. A bit scrambler is often used in the DMT transmitter to scramble the input data bits before the bits are modulated to assure that the transmitted data bits are random and, consequently, that the modulation of those bits produces a DMT transmission signal with a Gaussian probability distribution.

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With an appropriate allocation of transmit power levels to the carriers or subchannels, such a system provides a desirable performance. Further, generating a transmission signal with a Gaussian probability distribution is important in order to transmit a transmission signal with a low peak-to-average ratio (PAR), or peak-to-average power ratio. The PAR of a transmission signal is the ratio of the instantaneous peak value (i.e., maximum magnitude) of a signal parameter (e.g., voltage, current, phase, frequency, power) to the time-averaged value of the signal parameter. In DMT systems, the PAR of the transmitted signal is determined by the probability of the random transmission signal reaching a certain peak voltage during the time interval required for a certain number of symbols. An example of the PAR of a transmission signal transmitted from a DMT transmitter is 14.5 dB, which is equivalent to having a 1E-7 probability of clipping. The PAR of a transmission signal transmitted and received in a DMT communication system is an important consideration in the design of the DMT communication system because the PAR of a signal affects the communication system's total power consumption and component linearity requirements of the system.

If the phase of the modulated carriers is not random, then the PAR can increase greatly. Examples of cases where the phases of the modulated carrier signals are not random are when bit scramblers are not used, multiple carrier signals are used to modulate the same input data bits, and the constellation maps, which are mappings of input data bits to the phase

of a carrier signal, used for modulation are not random enough (i.e., a zero value for a data bit corresponds to a 90 degree phase characteristic of the DMT carrier signal and a one value for a data bit corresponds to a -90 degree phase characteristic of the DMT carrier signal). An increased PAR can result in a system with high power consumption and/or with high probability of clipping the transmission signal. Thus, there remains a need for a system and method that can effectively scramble the phase of the modulated carrier signals in order to provide a low PAR for the transmission signal.

Summary of the Invention

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The present invention features a system and method that scrambles the phase characteristics of the modulated carrier signals in a transmission signal. In one aspect, a value is associated with each carrier signal. A phase shift is computed for each carrier signal based on the value associated with that carrier signal. The value is determined independently of any input bit value carried by that carrier signal. The phase shift computed for each carrier signal is combined with the phase characteristic of that carrier signal to substantially scramble the phase characteristics of the carrier signals.

In one embodiment, the input bit stream is modulated onto the carrier signals having the substantially scrambled phase characteristic to produce a transmission signal with a reduced peak-to-average power ratio (PAR). The value is derived from a predetermined parameter, such as a random number generator, a carrier number, a DMT symbol count, a superframe count, and a hyperframe count. In another embodiment, a predetermined transmission signal is transmitted when the amplitude of the transmission signal exceeds a certain level.

In another aspect, the invention features a method wherein a value is associated with each carrier signal. The value is determined independently of any input bit value carried by that carrier signal. A phase shift for each carrier signal is computed based on the value associated with that carrier signal. The transmission signal is demodulated using the phase shift computed for each carrier signal.

In another aspect, the invention features a system comprising a phase scrambler that computes a phase shift for each carrier signal based on a value associated with that carrier signal. The phase scrambler also combines the phase shift computed for each carrier signal

with the phase characteristic of that carrier signal to substantially scramble the phase characteristic of the carrier signals. In one embodiment, a modulator, in communication with the phase scrambler, modulates bits of an input signal onto the carrier signals having the substantially scrambled phase characteristics to produce a transmission signal with a reduced PAR.

Description of the Drawings

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The invention is pointed out with particularity in the appended claims. The advantages of the invention described above, as well as further advantages of the invention, may be better understood by reference to the following description taken in conjunction with the accompanying drawings, in which:

FIG. 1 is a block diagram of an embodiment of a digital subscriber line communications system including a DMT (discrete multitone modulation) transceiver, in communication with a remote transceiver, having a phase scrambler for substantially scrambling the phase characteristics of carrier signals; and

FIG. 2 is a flow diagram of an embodiment of a process for scrambling the phase characteristics of the carrier signals in a transmission signal.

Detailed Description

FIG. 1 shows a digital subscriber line (DSL) communication system 2 including a discrete multitone (DMT) transceiver 10 in communication with a remote transceiver 14 over a communication channel 18 using a transmission signal 38 having a plurality of carrier signals. The DMT transceiver 10 includes a DMT transmitter 22 and a DMT receiver 26. The remote transceiver 14 includes a transmitter 30 and a receiver 34. Although described with respect to discrete multitone modulation, the principles of the invention apply also to other types of multicarrier modulation, such as, but not limited to, orthogonally multiplexed quadrature amplitude modulation (OQAM), discrete wavelet multitone (DWMT) modulation, and orthogonal frequency division multiplexing (OFDM).

The communication channel 18 provides a downstream transmission path from the DMT transmitter 22 to the remote receiver 34, and an upstream transmission path from the remote transmitter 30 to the DMT receiver 26. In one embodiment, the communication channel 18 is a pair of twisted wires of a telephone subscriber line. In other embodiments, the

communication channel 18 can be a fiber optic wire, a quad cable, consisting of two pairs of twisted wires, or a quad cable that is one of a star quad cable, a Dieselhorst-Martin quad cable, and the like. In a wireless communication system wherein the transceivers 10, 14 are wireless modems, the communication channel 18 is the air through which the transmission signal 38 travels between the transceivers 10, 14.

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By way of example, the DMT transmitter 22 shown in FIG. 1 includes a quadrature amplitude modulation (QAM) encoder 42, a modulator 46, a bit allocation table (BAT) 44, and a phase scrambler 66. The DMT transmitter 22 can also include a bit scrambler 74, as described further below. The remote transmitter 30 of the remote transceiver 14 comprises equivalent components as the DMT transmitter 22. Although this embodiment specifies a detailed description of the DMT transmitter 22, the inventive concepts apply also to the receivers 34, 24 which have similar components to that of the DMT transmitter 22, but perform inverse functions in a reverse order.

The QAM encoder 42 has a single input for receiving an input serial data bit stream 54 and multiple parallel outputs to transmit QAM symbols 58 generated by the QAM encoder 42 from the bit stream 54. In general, the QAM encoder 42 maps the input serial bit-stream 54 in the time domain into parallel QAM symbols 58 in the frequency domain. In particular, the QAM encoder 42 maps the input serial data bit stream 54 into N parallel quadrature amplitude modulation (QAM) constellation points 58, or QAM symbols 58, where N represents the number of carrier signals generated by the modulator 46. The BAT 44 is in communication with the QAM encoder 42 to specify the number of bits carried by each carrier signal. The QAM symbols 58 represent the amplitude and the phase characteristic of each carrier signal.

The modulator 46 provides functionality associated with the DMT modulation and transforms the QAM symbols 58 into DMT symbols 70 each comprised of a plurality of time-domain samples. The modulator 46 modulates each carrier signal with a different QAM symbol 58. As a result of this modulation, carrier signals have phase and amplitude characteristics based on the QAM symbol 58 and therefore based on the input-bit stream 54. In particular, the modulator 46 uses an inverse fast Fourier transform (IFFT) to change the QAM symbols 58 into a transmission signal 38 comprised of a sequence of DMT symbols

70. The modulator 46 changes the QAM symbols 58 into DMT symbols 70 through modulation of the carrier signals. In another embodiment, the modulator 46 uses the inverse discrete Fourier transform (IDFT) to change the QAM symbols 58 into DMT symbols 70. In one embodiment, a pilot tone is included in the transmission signal 38 to provide a reference signal for coherent demodulation of the carrier signals in the remote receiver 34 during reception of the transmission signal 38.

The modulator 46 also includes a phase scrambler 66 that combines a phase shift computed for each QAM-modulated carrier signal with the phase characteristic of that carrier signal. Combining phase shifts with phase characteristics, in accordance with the principles of the invention, substantially scrambles the phase characteristics of the carrier signals in the transmission signal 38. By scrambling the phase characteristics of the carrier signals, the resulting transmission signal 38 has a substantially minimized peak-to-average (PAR) power ratio. The phase scrambler 66 can be part of or external to the modulator 46. Other embodiments of the phase scrambler 66 include, but are not limited to, a software program that is stored in local memory and is executed on the modulator 46, a digital signal processor (DSP) capable of performing mathematical functions and algorithms, and the like. The remote receiver 34 similarly includes a phase descrambler 66' for use when demodulating carrier signals that have had their phase characteristics adjusted by the phase scrambler 66 of the DMT transceiver 10.

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To compute a phase shift for each carrier signal, the phase scrambler 66 associates one or more values with that carrier signal. The phase scrambler 66 determines each value for a carrier signal independently of the QAM symbols 58, and, therefore, independently of the bit value(s) modulated onto the carrier signal. The actual value(s) that the phase scrambler 66 associates with each carrier signal can be derived from one or more predefined parameters, such as a pseudo-random number generator (pseudo-RNG), a DMT carrier number, a DMT symbol count, a DMT superframe count, a DMT hyperframe count, and the like, as described in more detail below. Irrespective of the technique used to produce each value, the same technique is used by the DMT transmitter 22 and the remote receiver 34 so that the value associated with a given carrier signal is known at both ends of the communication channel 18.

The phase scrambler 66 then solves a predetermined equation to compute a phase shift for the carrier signal, using the value(s) associated with that carrier signal as input that effects the output of the equation. Any equation suitable for computing phase shifts can be used to compute the phase shifts. When the equation is independent of the bit values of the input serial bit stream 54, the computed phase shifts are also independent of such bit values.

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In one embodiment (shown in phantom), the DMT transmitter 22 includes a bit scrambler 74, which receives the input serial bit stream 54 and outputs data bits 76 that are substantially scrambled. The substantially scrambled bits 76 are then passed to the QAM encoder 42. When the bit scrambler 74 is included in the DMT transmitter 22, the operation of the phase scrambler 66 further assures that the transmission signal 38 has a Gaussian probability distribution and, therefore, a substantially minimized PAR.

FIG. 2 shows embodiments of a process used by the DMT transmitter 22 for adjusting the phase characteristic of each carrier signal and combining these carrier signals to produce the transmission signal 38. The DMT transmitter 22 generates (step 100) a value that is associated with a carrier signal. Because the value is being used to alter the phase characteristics of the carrier signal, both the DMT transmitter 22 and the remote receiver 34 must recognize the value as being associated with the carrier signal. Either the DMT transmitter 22 and the remote receiver 34 independently derive the associated value, or one informs the other of the associated value. For example, in one embodiment the DMT transmitter 22 can derive the value from a pseudo-RNG and then transmit the generated value to the remote receiver 34. In another embodiment, the remote receiver 34 similarly derives the value from the same pseudo-RNG and the same seed as used by the transmitter (i.e., the transmitter pseudo-RNG produces the same series of random numbers as the receiver pseudo-RNG).

As another example, the DMT transmitter 22 and the remote receiver 34 can each maintain a symbol counter for counting DMT symbols. The DMT transmitter 22 increments its symbol counter upon transmitting a DMT symbol; the remote receiver 34 upon receipt. Thus, when the DMT transmitter 22 and the remote receiver 34 both use the symbol count as a value for computing phase shifts, both the DMT transmitter 22 and remote receiver 34

"know" that the value is associated with a particular DMT symbol and with each carrier signal of that DMT symbol.

Values can also be derived from other types of predefined parameters. For example, if the predefined parameter is the DMT carrier number, then the value associated with a particular carrier signal is the carrier number of that signal within the DMT symbol. The number of a carrier signal represents the location of the frequency of the carrier signal relative to the frequency of other carrier signals within a DMT symbol. For example, in one embodiment the DSL communication system 2 provides 256 carrier signals, each separated by a frequency of 4.3125 kHz and spanning the frequency bandwidth from 0 kHz to 1104 kHz. The DMT transmitter 22 numbers the carrier signals from 0 to 255. Therefore, "DMT carrier number 50" represents the 51st DMT carrier signal which is located at the frequency of 215.625 kHz (i.e., 51 x 4.3125 kHz).

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Again, the DMT transmitter 22 and the remote receiver 34 can know the value that is associated with the carrier signal because both the DMT transmitter 22 and the remote receiver 34 use the same predefined parameter (here, the DMT carrier number) to make the value-carrier signal association. In other embodiments (as exemplified above with the transmitter pseudo-RNG), the DMT transmitter 22 can transmit the value to the remote receiver 34 (or vice versa) over the communication channel 18.

In other embodiments, other predefined parameters can be used in conjunction with the symbol count. One example of such a predefined parameter is the superframe count that increments by one every 69 DMT symbols. One exemplary implementation that achieves the superframe counter is to perform a modulo 68 operation on the symbol count. As another example, the DMT transmitter 22 can maintain a hyperframe counter for counting hyperframes. An exemplary implementation of the hyperframe count is to perform a modulo 255 operation on the superframe count. Thus, the hyperframe count increments by one each time the superframe count reaches 255.

Accordingly, it is seen that some predefined parameters produce values that vary from carrier signal to carrier signal. For example, when the predefined parameter is the DMT carrier number, values vary based on the frequency of the carrier signal. As another example, the pseudo-RNG generates a new random value for each carrier signal.

Other predefined parameters produce values that vary from DMT symbol 70 to DMT symbol 70. For example, when the predefined parameter is the symbol count, the superframe count, or hyperframe count, values vary based on the numerical position of the DMT symbol 70 within a sequence of symbols, superframes, or hyperframes. Predefined parameters such as the pseudo-RNG, symbol count, superframe count, and superframe can also be understood to be parameters that vary values over time. Any one or combination of the predefined parameters can provide values for input to the equation that computes a phase shift for a given carrier signal.

In one embodiment, the phase scrambling is used to avoid clipping of the transmission signal 38 on a DMT symbol 70 by DMT symbol 70 basis. In this embodiment, the DMT transmitter 22 uses a value based on a predefined parameter that varies over time, such as the symbol count, to compute the phase shift. It is to be understood that other types of predefined parameters that vary the values associated with carrier signals can be used to practice the principles of the invention. As described above, the transceivers 10, 14 may communicate (step 110) the values to synchronize their use in modulating and demodulating the carrier signals.

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The DMT transmitter 22 then computes (step 115) the phase shift that is used to adjust the phase characteristic of each carrier signal. The amount of the phase shift combined with the phase characteristic of each QAM-modulated carrier signal depends upon the equation used and the one or more values associated with that carrier signal.

The DMT transmitter 22 then combines (step 120) the phase shift computed for each carrier signal with the phase characteristic of that carrier signal. By scrambling the phase characteristics of the carrier signals, the phase scrambler 66 reduces (with respect to unscrambled phase characteristics) the combined PAR of the plurality of carrier signals and, consequently, the transmission signal 38. The following three phase shifting examples, PS #1-PS #3, illustrate methods used by the phase scrambler 66 to combine a computed phase shift to the phase characteristic of each carrier signal.

Phase Shifting Example #1

Phase shifting example #1 (PS #1) corresponds to adjusting the phase characteristic of the QAM-modulated carrier signal associated with a carrier number N by $N \times \frac{\pi}{3}$, modulo (mod) 2π . In this example, a carrier signal having a carrier number N equal to 50 has a phase shift added to the phase characteristic of that carrier signal equal to $50 \times \frac{\pi}{3}$ (mod 2π) = $\frac{2}{3}\pi$. The carrier signal with a carrier number N equal to 51 has a phase shift added to the phase characteristic of that carrier signal equal to $51 \times \frac{\pi}{3}$ (mod 2π) = π . The carrier signal with the carrier number N equal to 0 has no phase shift added to the phase characteristic of that carrier signal.

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Phase Shifting Example #2

Phase shifting example #2 (PS #2) corresponds to adjusting the phase characteristic of the QAM-modulated carrier signal associated with a carrier number N by $(N+M)\times\frac{\pi}{4}$, mod 2π , where M is the symbol count. In this example, a carrier signal having a carrier number N equal to 50 on DMT symbol count M equal to 8 has a phase shift added to the phase characteristic of that carrier signal equal to $(50+8)\times\frac{\pi}{4}$ (mod 2π) = $\frac{\pi}{2}$. The carrier signal with the same carrier number N equal to 50 on the next DMT symbol count M equal to 9 has a phase shift added to the phase characteristic of that carrier signal equal to $(50+9)\times\frac{\pi}{4}$ (mod 2π) = $\frac{3\pi}{4}$.

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Phase Shifting Example #3

Phase shifting example #3 (PS #3) corresponds to adjusting the phase characteristic of the QAM-modulated carrier signal associated with a carrier number N by $(X_N) \times \frac{\pi}{6}$, mod 2π , where X_N is an array of N pseudo-random numbers. In this example, a carrier signal having a carrier number N equal to 5 and X_N equal to [3, 8, 1, 4, 9, 5, . . .] has a phase shift added to the phase characteristic of the carrier signal that is equal to $(9) \times \frac{\pi}{6} \pmod{2\pi} = \frac{3\pi}{2}$ (Note that 9 is the 5th value in X_N .) The carrier signal with a carrier number N equal to 6 has a phase shift added to the phase characteristic of the carrier signal equal to $(5) \times \frac{\pi}{6} \pmod{2\pi} = \frac{5\pi}{6}$.

It is to be understood that additional and/or different phase shifting techniques can be used by the phase scrambler 66, and that PS #1, #2, and #3 are merely illustrative examples of the principles of the invention. The DMT transmitter 22 then combines (step 130) the carrier signals to form the transmission signal 38. If the transmission signal is not clipped, as described below, the DMT transmitter 22 consequently transmits (step 160) the transmission signal 38 to the remote receiver 34.

Clipping of Transmission Signals

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A transmission signal 38 that has high peak values of voltage (i.e., a high PAR) can induce non-linear distortion in the DMT transmitter 22 and the communication channel 18. One form of this non-linear distortion of the transmission signal 38 that may occur is the limitation of the amplitude of the transmission signal 38 (i.e., clipping). For example, a particular DMT symbol 70 clips in the time domain when one or more time domain samples in that DMT symbol 70 are larger than the maximum allowed digital value for the DMT symbols 70. In multicarrier communication systems when clipping occurs, the transmission signal 38 does not accurately represent the input serial data bit signal 54.

In one embodiment, the DSL communication system 2 avoids the clipping of the transmission signal 38 on a DMT symbol 70 by DMT symbol 70 basis. The DMT transmitter 22 detects (step 140) the clipping of the transmission signal 38. If a particular DMT symbol 70 clips in the time domain to produce a clipped transmission signal 38, the DMT transmitter 22 substitutes (step 150) a predefined transmission signal 78 for the clipped transmission signal 38.

The predefined transmission signal 78 has the same duration as a DMT symbol 70 (e.g., 250 ms) in order to maintain symbol timing between the DMT transmitter 22 and the remote receiver 34. The predefined transmission signal 78 is not based on (i.e., independent of) the modulated input data bit stream 54; it is a bit value pattern that is recognized by the remote receiver 34 as a substituted signal. In one embodiment, the predefined transmission signal 78 is a known pseudo-random sequence pattern that is easily detected by the remote receiver 34. In another embodiment, the predefined transmission signal 78 is an "all zeros" signal, which is a zero voltage signal produced at the DMT transmitter 22 output (i.e., zero volts modulated on all the carrier signals). In addition to easy detection by the remote

receiver 34, the zero voltage signal reduces the power consumption of the DMT transmitter 22 when delivered by the DMT transmitter 22. Further, a pilot tone is included in the predefined transmission signal 78 to provide a reference signal for coherent demodulation of the carrier signals in the remote receiver 34 during reception of the predefined transmission signal 78.

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After the remote receiver 34 receives the transmission signal 38, the remote receiver 34 determines if the transmission signal 38 is equivalent to the predefined transmission signal 78. In one embodiment, when the remote receiver 34 identifies the predefined transmission signal 78, the remote receiver 34 ignores (i.e., discards) the predefined transmission signal 78.

Following the transmission of the predefined transmission signal 78, the phase scrambler 66 shifts (step 120) the phase characteristic of the QAM-modulated carrier signals (based on one of the predefined parameters that varies over time). For example, consider that a set of QAM symbols 58 produces a DMT symbol 70 comprising a plurality of time domain samples, and that one of the time domain samples is larger than the maximum allowed digital value for the DMT symbol 70. Therefore, because the transmission signal 38 would be clipped when sent to the remote receiver 34, the DMT transmitter 22 sends the predefined transmission signal 78 instead.

After transmission of the predefined transmission signal 78, the DMT transmitter 22 again attempts to send the same bit values that produced the clipped transmission signal 38 in a subsequent DMT symbol 70'. Because the generation of phase shifts in this embodiment is based on values that vary over time, the phase shifts computed for the subsequent DMT symbol 70' are different than those that were previously computed for the DMT symbol 70 with the clipped time domain sample. These different phase shifts are combined to the phase characteristics of the modulated carrier signals to produce carrier signals of the subsequent DMT symbol 70' with different phase characteristics than the carrier signals of the DMT symbol 70 with the clipped time domain sample.

DMT communication systems 2 infrequently produce transmission signals 38 that clip (e.g., approximately one clip every 10⁷ time domain samples 70). However, if the subsequent DMT symbol 70' includes a time domain sample that clips, then the predefined

transmission signal 78 is again transmitted (step 150) to the remote receiver 34 instead of the clipped transmission signal 38. The clipping time domain sample may be on the same or on a different carrier signal than the previously clipped DMT symbol 70. The DMT transmitter 22 repeats the transmission of the predefined transmission signal 78 until the DMT transmitter 22 produces a subsequent DMT symbol 70' that is not clipped. When the DMT transmitter 22 produces a DMT symbol 70' that is not clipped, the DTM transmitter 22 transmits (step 160) the transmission signal 38 to the remote receiver 34. The probability of a DMT symbol 70 producing a transmission signal 38 that clips in the time domain depends on the PAR of the transmission signal 38.

For example, the following phase shifting example, PST #4, illustrates the method used by the phase scrambler 66 to combine a different phase shift to the phase characteristic of each carrier signal to avoid the clipping of the transmission signal 38.

Phase Shifting Example #4

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Phase shifting example #4 (PS #4) corresponds to adjusting the phase characteristic of the carrier signal associated with a carrier number N by $\frac{\pi}{3} \times (M+N)$, mod 2π , where M is the DMT symbol count. In this example, if the DMT symbol 70 clips when the DMT symbol count M equals 5, the predefined transmission signal 78 is transmitted instead of the current clipped transmission signal 38. On the following DMT symbol period, the DMT count M equals 6, thereby causing a different set of time domain samples to be generated for the subsequent DMT symbol 70', although the QAM symbols 58 used to produce both DMT symbols 70, 70' are the same.

If this different set of time domain samples (and consequently the transmission signal 38) is not clipped, the DMT transmitter 22 sends the transmission signal 38. If one of the time domain samples in the different set of time domain samples 70 (and consequently the transmission signal 38) is clipped, then the DMT transmitter 22 sends the predefined transmission signal 78 again. The process continues until a DMT symbol 70 is produced without a time domain sample 70 that is clipped. In one embodiment, the transmitter 22 stops attempting to produce a non-clipped DMT symbol 70' for the particular set of QAM symbols 58 after generating a predetermined number of clipped DMT symbols 70'. At that moment,

the transmitter 22 can transmit the most recently produced clipped DMT symbol 70' or the predetermined transmission signal 78.

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The PAR of the DSL communication system 2 is reduced because the predefined transmission signal 78 is sent instead of the transmission signal 38 when the DMT symbol 70 clips. For example, a DMT communication system 2 that normally has a clipping probability of 10-7 for the time domain transmission signal 38 can therefore operate with a 10-5 probability of clipping and a lower PAR equal to 12.8 dB (as compared to 14.5 dB). When operating at a 10-5 probability of clipping, assuming a DMT symbol 70 has 512 time-domain samples 70, the DMT transmitter 22 experiences one clipped DMT symbol 70 out of every $\frac{10^5}{512}$, or 195 DMT symbols 70. This results in the predefined (non-data carrying) transmission signal 78 being transmitted, on average, once every 195 DMT symbols. Although increasing the probability of clipping to 10^{-5} results in approximately a 0.5% (1/195) decrease in throughput, the PAR of the transmission signal 38 is reduced by 1.7 dB, which reduces transmitter complexity in the form of power consumption and component linearity.

While the invention has been shown and described with reference to specific preferred embodiments, it should be understood by those skilled in the art that various changes in form and detail may be made therein without departing from the spirit and scope of the invention as defined by the following claims. For example, although the specification uses DSL to describe the invention, it is to be understood that various form of DSL can be used, e.g., ADSL, VDSL, SDSL, HDSL, HDSL2, or SHDSL. It is also to be understood that the principles of the invention apply to various types of applications transported over DSL systems (e.g., telecommuting, video conferencing, high speed Internet access, video-on demand).

Abstract

A system and method that scrambles the phase characteristic of a carrier signal are described. The scrambling of the phase characteristic of each carrier signal includes associating a value with each carrier signal and computing a phase shift for each carrier signal based on the value associated with that carrier signal. The value is determined independently of any input bit value carried by that carrier signal. The phase shift computed for each carrier signal is combined with the phase characteristic of that carrier signal so as to substantially scramble the phase characteristic of the carrier signals. Bits of an input signal are modulated onto the carrier signals having the substantially scrambled phase characteristic to produce a transmission signal with a reduced PAR.

FIG. 1

FIG. 1

Electronic Patent /	Ap p	lication Fee	Transmi	ttal		
Application Number:						
Filing Date:						
Title of Invention:	SYSTEM AND METHOD FOR SCRAMBLING THE PHASE OF THE CARRIERS IN MULTICARRIER COMMUNICATIONS SYSTEM					
First Named Inventor/Applicant Name:	Ma	rcos C. Tzannes				
Filer:	Jas	on Vick/Joanne Vos	;			
Attorney Docket Number:	55	50-47-CON-5				
Filed as Large Entity						
Utility under 35 USC 111(a) Filing Fees						
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Utility application filing		1011	1	380	380	
Utility Search Fee		1111	1	620	620	
Utility Examination Fee		1311	1	250	250	
Pages:						
Claims:						
Miscellaneous-Filing:						
Petition:						
Patent-Appeals-and-Interference:						

Description	Fee Code	Quantity	Amount	Sub-Total in USD(\$)
Post-Allowance-and-Post-Issuance:				
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Miscellaneous:				
	(\$)	1250		

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First Named Inventor/Applicant Name:	Marcos C. Tzannes						
Customer Number:	62574						
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Attorney Docket Number:	5550-47-CON-5						
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Application Data Sheet 37 CFR 1.70					1 76	Attorney Docket Number			5550-47-CON-5						
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Application Data Sheet 37 CFR 1.76		Application Number				
Title of Invention	SYSTEM AND METHOD FOR COMMUNICATIONS SYSTEM		OF THE CARRIERS IN A MULTICARRIER			
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Request Not to Publish. I hereby request that the attached application not be published under 35 U.S. C. 122(b) and certify that the invention disclosed in the attached application has not and will not be the subject of an application filed in another country, or under a multilateral international agreement, that requires publication at eighteen months after filing.

Representative Information:

Representative information should be provided for all practitioners having a power of attorney in the application. Providing this information in the Application Data Sheet does not constitute a power of attorney in the application (see 37 CFR 1.32). Enter either Customer Number or complete the Representative Name section below. If both sections are completed the Customer Number will be used for the Representative Information during processing.							
Please Select One:	Customer Number	US Patent Practitioner	Limited Recognition (37 CFR 11.9)				
Customer Number	62574						

Domestic Benefit/National Stage Information:

This section allows for the applicant to either claim benefit under 35 U.S.C. 119(e), 120, 121, or 365(c) or indicate National Stage entry from a PCT application. Providing this information in the application data sheet constitutes the specific reference required by 35 U.S.C. 119(e) or 120, and 37 CFR 1.78(a)(2) or CFR 1.78(a)(4), and need not otherwise be made part of the specification.

33 0.3.C. 113(e) (or 120, and	31 Cl IX 1.10(a)	(2) Of Cl (C 1.70(a)(4),	33 0.3.0. 113(e) of 120, and 37 of 10 1.70(a)(2) of 0110 1.70(a)(4), and freed not otherwise be made part of the specification.								
Prior Applicati	on Status	Pending		Remove								
Application Number Cor		Cont	inuity Type	Prior Application Number		Filing Date (YYYY-MM-DD						
		Continuation of	of	12/783725 2010-05-								
Prior Applicati	on Status	Patented		Remove								
Application Number			Prior Application Number	Filing Date (YYYY-MM-DD)	Pate	ent Number	Issue Date (YYYY-MM-DD)					
12/783725	Continua	tion of	12/255713	2008-10-22	776	9104	2010-08-03					
Prior Applicati	on Status	Patented		Remove								
Application Number	Continuity Type		Prior Application Number	Filing Date (YYYY-MM-DD)	Patent Number		Issue Date (YYYY-MM-DD)					
12/255713	Continua	tion of	11/863581	2007-09-28 74		1721	2008-12-30					
Prior Applicati	on Status	Patented		Remove								
Application Number	Con	tinuity Type	Prior Application Number	Filing Date (YYYY-MM-DD)	Patent Number		Issue Date (YYYY-MM-DD)					
11/863581	Continua	tion of	11/211535	2005-08-26	7292627		2007-11-06					
Prior Applicati	Prior Application Status Patented			Remove			nove					
Application Number	'' Continuity Lyne		Prior Application Number	Filing Date (YYYY-MM-DD)	Patent Number		Issue Date (YYYY-MM-DD)					
11/211535	Continua	tion of	09/710310	2000-11-09	696	1369	2005-11-01					
				•								

Application Data Sheet 37 CFR 1.76		Attorney Docket Number	5550-47-CON-5
		Application Number	
Title of Invention	SYSTEM AND METHOD FOR SCRAMBLING THE PHASE COMMUNICATIONS SYSTEM		OF THE CARRIERS IN A MULTICARRIER

Prior Application Status	Expired		Remove
Application Number	Continuity Type	Prior Application Number	Filing Date (YYYY-MM-DD)
09/710310	non provisional of	61/164134	1999-11-09
Additional Domestic Benefi by selecting the Add button	enerated within this form	Add	

Foreign Priority Information:

This section allows for the applicant to claim benefit of foreign priority and to identify any prior foreign application for which priority is not claimed. Providing this information in the application data sheet constitutes the claim for priority as required by 35 U.S.C. 119(b) and 37 CFR 1.55(a).

		Rei	move			
Application Number	Country i	Parent Filing Date (YYYY-MM-DD)	Priority Claimed			
			◯ Yes ⊙ No			
Additional Foreign Priority Data may be generated within this form by selecting the Add button.						

Assignee Information:

ŭ .	n the application data sheet doingnment recorded in the Office	•	th any requirement of part 3 of Title 37
Assignee 1			Remove
If the Assignee is an Org	ganization check here.	X	
Organization Name	AWARE, INC.		
Mailing Address Inforr	nation:		
Address 1	40 Middlesex Tumpike		
Address 2			
City	Bedford	State/Province	MA
Country i US		Postal Code	01730-1432
Phone Number		Fax Number	
Email Address		·	
Additional Assignee Da button.	ta may be generated within	n this form by selecting the Ad	d Add

Signature:

A signature of the applicant or representative is required in accordance with 37 CFR 1.33 and 10.18. Please see 37 CFR 1.4(d) for the form of the signature.							
Signature	/Jason H. Vick/			Date (YYYY-MM-DD)	2011-11-23		
First Name	Jason H.	Last Name	Vick	Registration Number	45285		

Application Data Sheet 37 CFR 1.76		Attorney Docket Number	5550-47-CON-5
		Application Number	
Title of Invention SYSTEM AND METHOD FOR SO COMMUNICATIONS SYSTEM			OF THE CARRIERS IN A MULTICARRIER

This collection of information is required by 37 CFR 1.76. The information is required to obtain or retain a benefit by the public which is to file (and by the USPTO to process) an application. Confidentiality is governed by 35 U.S.C. 122 and 37 CFR 1.14. This collection is estimated to take 23 minutes to complete, including gathering, preparing, and submitting the completed application data sheet form to the USPTO. Time will vary depending upon the individual case. Any comments on the amount of time you require to complete this form and/or suggestions for reducing this burden, should be sent to the Chief Information Officer, U.S. Patent and Trademark Office, U.S. Department of Commerce, P.O. Box 1450, Alexandria, VA 22313-1450. DO NOT SEND FEES OR COMPLETED FORMS TO THIS ADDRESS. **SEND TO: Commissioner for Patents, P.O. Box 1450, Alexandria, VA 22313-1450.**

Privacy Act Statement

The Privacy Act of 1974 (P.L. 93-579) requires that you be given certain information in connection with your submission of the attached form related to a patent application or patent. Accordingly, pursuant to the requirements of the Act, please be advised that: (1) the general authority for the collection of this information is 35 U.S.C. 2(b)(2); (2) furnishing of the information solicited is voluntary; and (3) the principal purpose for which the information is used by the U.S. Patent and Trademark Office is to process and/or examine your submission related to a patent application or patent. If you do not furnish the requested information, the U.S. Patent and Trademark Office may not be able to process and/or examine your submission, which may result in termination of proceedings or abandonment of the application or expiration of the patent.

The information provided by you in this form will be subject to the following routine uses:

- 1. The information on this form will be treated confidentially to the extent allowed under the Freedom of Information Act (5 U.S.C. 552) and the Privacy Act (5 U.S.C. 552a). Records from this system of records may be disclosed to the Department of Justice to determine whether the Freedom of Information Act requires disclosure of these records.
- A record from this system of records may be disclosed, as a routine use, in the course of presenting evidence to a court, magistrate, or administrative tribunal, including disclosures to opposing counsel in the course of settlement negotiations.
- A record in this system of records may be disclosed, as a routine use, to a Member of Congress submitting a request involving an
 individual, to whom the record pertains, when the individual has requested assistance from the Member with respect to the subject matter of
 the record
- 4. A record in this system of records may be disclosed, as a routine use, to a contractor of the Agency having need for the information in order to perform a contract. Recipients of information shall be required to comply with the requirements of the Privacy Act of 1974, as amended, pursuant to 5 U.S.C. 552a(m).
- 5. A record related to an International Application filed under the Patent Cooperation Treaty in this system of records may be disclosed, as a routine use, to the International Bureau of the World Intellectual Property Organization, pursuant to the Patent Cooperation Treaty.
- 6. A record in this system of records may be disclosed, as a routine use, to another federal agency for purposes of National Security review (35 U.S.C. 181) and for review pursuant to the Atomic Energy Act (42 U.S.C. 218(c)).
- 7. A record from this system of records may be disclosed, as a routine use, to the Administrator, General Services, or his/her designee, during an inspection of records conducted by GSA as part of that agency's responsibility to recommend improvements in records management practices and programs, under authority of 44 U.S.C. 2904 and 2906. Such disclosure shall be made in accordance with the GSA regulations governing inspection of records for this purpose, and any other relevant (i.e., GSA or Commerce) directive. Such disclosure shall not be used to make determinations about individuals.
- 8. A record from this system of records may be disclosed, as a routine use, to the public after either publication of the application pursuant to 35 U.S.C. 122(b) or issuance of a patent pursuant to 35 U.S.C. 151. Further, a record may be disclosed, subject to the limitations of 37 CFR 1.14, as a routine use, to the public if the record was filed in an application which became abandoned or in which the proceedings were terminated and which application is referenced by either a published application, an application open to public inspections or an issued patent.
- A record from this system of records may be disclosed, as a routine use, to a Federal, State, or local law enforcement agency, if the USPTO becomes aware of a violation or potential violation of law or regulation.

·							
DECLARATION A	ND POWER	Attorne	y Docket No.	AWR	-017 (457/19)		
OF ATTORNEY FO	R UTILITY	First Na	amed Inventor	Tzanr	ies		
OR DESIG	GN		C	OMPLE	TE IF KNOW	7V	
PATENT APPLI	CATION	Applica	ation Serial Number	Not Y	et Assigned		
□ Declaration □ Dec □ Declaration □ Declarati	laration	Filing I	Date	Herev	vith		
Submitted with Subm	itted after Initial	Group	Art Unit	Not Y	et Assigned		
Initial Filing Filing	(surcharge	Examin	ier Name	Not Y	et Assigned		
37 CE	R 1.16(e) required)			1			
							
As a below named inventor, [
My residence, post office addre	•		•				
I believe I am the original, first names are listed below) of the s							f plural
A System And Me	thod For Scrambling	The Phase	Of The Carriers In A	Multic	arrier Commun	ications System	.]
	(Title of the Invention)						
the specification of which	the specification of which						
is attached hereto							
was filed on (MM/DD/YYYY)			as United States App	lication	Scrial Number	or PCT Internal	tional
Application Number	and	was amer	ided on (MM/DD/YYY	Y) [(if applicat	ole).
I hereby state that I have review by any amendment specifically		contents	of the above identifie	d specif	ication, includi	ng the claims, a	s amended
I acknowledge the duty to discl	ose to the Patent Offic	e all infor	mation known by me	to be m	aterial to paten	tability as defin	cd in 37 CFR
I hereby claim foreign priority l	oenefits under 35 U.S.	C. 119(a)	-(d) or 365(b) of any	foreign :	application(s) f	or patent or inve	entor's
certificate, or 365(a) of any PC listed below and have also iden	l'international applica tified below, by check	ition which ing the bo	th designated at least of ox, any foreign applica	one cour ation for	ntry other than i r patent or inve	ine United State ntor's certificate	:s of America, e, or of any
PCT international application h	aving a filing date bef	ore that o	f the application on w	hich pri	ority is claimed	1.	
Prior Foreign Application Number(s)	Country		Foreign Filing Da (MM/DD/YYYY		Priority Not Claimed	Certified Cop YES	py Attached? NO
	plication numbers are						
I hereby claim the benefit unde		any Unite	d States provisional a	pplicati	on(s) listed belo	ow.	
Application Serial Nun 60/164,134	iber(s)	<u>Filing Da</u>	ate (MM/DD/YYYY) 11/09/1999)	serial nu	nal provisional a imbers are listed ental priority de hereto.	d on a

DECLAR	DECLARATION – Utility or Design Patent Application					
I hereby claim the benefit under 35 U United States of America, listed below States or PCT International application which is material to patentability as d PCT international filing date of this a	v and, insofar as the subject on in the manner provided by efined in 37 CFR 1.56 which pplication.	matter of	each of the claims of this applicat paragraph of 35 U.S.C. 112, I ack	ion is not nowledge	disclosed in the prior United the duty to disclose information	
U.S. Parent Application o	r PCT Parent	I	Parent Filing Date		Parent Patent Number	
Serial Number			(MM/DD/YYYY)		(if applicable)	
☐ Additional U.S. or PCT internation		listed on	o qualifyental priority data chee	attached	hereto	
As a named inventor, I hereby appoint the following registered practitioners to prosecute this application and to transact all business in the Patent and Trademark Office connected therewith: OR Registered practitioner(s) name/registration number listed below						
	Registration				Registration	
Name	Number		Name		Number	
Steven M. Bauer	31,481		Thomas C. Meyers		6,989	
John V. Bianco	36,748		Joseph B. Milstein		2,897	
Isabelle A.S. Blundell	43,321		David G. Miranda		2,898	
Maureen A. Bresnahan	44,559		Ronda P. Moore		4,244	
Michael H. Brodowski	41,640		Indranil Mukerji		°-46,944	
Jennifer A. Camacho	43,526		Edmund R. Pitcher		.7,829	
Joseph A. Capraro, Jr.	36,471		Michael A. Rodriguez	1	1,274	
John J. Cotter	38,116		Jamie H. Rose		5,054	
John V. Forcier	42,545		R. Stephen Rosenholm		5,283	
Steven J. Frank	33,497		Christopher W. Stamos		5,370	
Brian M. Gaff	44,691		Joseph P. Sullivan		15,349	
Michael J. Giannetta	42,574		Robert J. Tosti		15,393	
Duncan A. Greenhalgh	38,678		Thomas A. Turano		15,722	
William G. Guerin	41,047		Michael J. Twomey		38,349	
Jonathan A. Harris	44,744		Christine C. Vito		39,061	
Ira V. Heffan	41,059		Patrick R.H. Waller		11,418	
Danielle L. Herritt	43,670		Daniel A. Wilson		15,508	
Douglas J. Kline	35,574		Yin P. Zhang	1 '	14,372	
John D. Lanza	40,060			1		
Kurt W. Lockwood	40,704					
Additional registered practiti	ioners named on suppleme	ental Re	gistered Practitioner Information	on sheet	attached hereto.	
Direct all correspondence to:	Patent Administra Testa, Hurwitz & High Street Towe 125 High Street Boston, MA 021 Tel. No.: (617) 24 Fax No.: (617) 24	Thibeau er 10 18-7000	ilt, LLP			

I hereby declare that all statements made herein of my own knowledge are true and that all statements made on information and belief are believed to be true; and further that these statements were made with the knowledge that willful false statements and the like so made are punishable by fine or imprisonment, or both, under 18 U.S.C. 1001 and that such willful false statements may jeopardize the validity of the application or any patent issued thereon.

Name of Sole or First	icst Inventor:				A petition has been filed for this unsigned inventor				
Given Na	anc (firs	t and middle [if any])		Family Name or Surname				
Marcos C.					Tzannes				
Inventor's Signature	1	MB					Date	No.	19,2000
Residence	City	Orinda Z	State	CA	Country	USA	Citiz	enship	USA
Post Office Address	121 L	*Espiral			 -				1
P.O. Address (line 2)	City	Orinda	Stato	CA	ZIP	94563	Country	1	USA
Additional inventor	rs are be	no teman gni	theso	pplement	al Additional I	inventor(s) she	ct(s) attache	d hereto.	-
Name of Additional Joint Inventor, if any:				A petition	n has heen filed	I for this uns	lgned in	ventor	
Given N	ame (fir	st and middle	[if any])			Famil	ly Name or S	umanc	
			ŀ						
Inventor's Signature							Date		
Residence	City		State		Country	T	Citiz	caship	T
Post Office Address		'				4			
P.O. Address (line 2)	City	<u> </u>	State		ZIP		Country		
Name of Additional Jo	olnt Inv	entor, if any:			A petition has been filed for this unsigned inventor				
Givca Na	me (firs	t and middle [īf any])		Family Name or Surgame				
				!					
Inventor's Signature							Date		
Residence	City		State	•	Country	T :	Citiz	cnship	T
Post Office Address									.1
P.O. Address (line 2)	City	T.:-	State		ZIP	7	Country		

PTO/SB/80 (01-06)
Approved for use through 12/31/2008, OMB 0651-0035
U.S. Palent and Trademark Office; U.S. DEPARTMENT OF COMMERCE

Under the Paperwork Reduction Act of 1995, no persons are required to respond to a collection of information unless it displays a valid OMB control number.

POWER OF ATTORNEY TO PROSECUTE APPLICATIONS BEFORE THE USPTO I hereby revoke all previous powers of attorney given in the application identified in the attached statement under 37 CFR 3.73(b) I hereby appoint: 62574 Practitioners associated with the Customer Number: Practitioner(s) named below (if more than ten patent practitioners are to be named, then a customer number must be used): Registration Registration Name Name Number Number as attorney(s) or agent(s) to represent the undersigned before the United States Patent and Trademark Office (USPTO) in connection with any and all patent applications assigned only to the undersigned according to the USPTO assignment records or assignment documents attached to this form in accordance with 37 CFR 3.73(b). Please change the correspondence address for the application identified in the attached statement under 37 CFR 3.73(b) to: 62574 The address associated with Customer Number. OR Firm or Individual Name Address Zip State City Country Email Telephone Assignee Name and Address: AWARE, INC. 40 Middlesex Turnpike Bedford, MA 07130-1423 A copy of this form, together with a statement under 37 CFR 3.73(b) (Form PTO/SB/96 or equivalent) is required to be filed in each application in which this form is used. The statement under 37 CFR 3.73(b) may be completed by one of the practitioners appointed in this form if the appointed practitioner is authorized to act on behalf of the assignee, and must identify the application in which this Power of Attorney is to be filed. SIGNATURE of Assignee of Record The individual whose signature and title is supplied below is authorized to act on behalf of the assignee Signature Telephone

This collection of information is required by 37 CFR 1.31, 1.32 and 1.33. The information is required to obtain or retain a benefit by the public which is to file (and by the USPTO to process) an application. Confidentiality is governed by 35 U.S.C. 122 and 37 CFR 1.11 and 1.14. This collection is estimated to take 3 minutes to complete, including gathering, preparing, and submitting the completed application form to the USPTO. Time will vary depending upon the individual case. Any comments on the amount of time you require to complete this form and/or suggestions for reducing hits burden, should be sent to the Chief Information Officer, U.S. Patent and Trademark Office, U.S. Department of Commerce, P.O. Box 1450, Alexandria, VA 22313-1450. DO NOT SEND FEES OR COMPLETED FORMS TO THIS ADDRESS. SEND TO: Commissioner for Patents, P.O. Box 1450, Alexandria, VA 22313-1450.

if you need assistance in completing the form, call 1-800-PTO-9199 and select option 2.

Electronic Ack	Electronic Acknowledgement Receipt				
EFS ID:	11474475				
Application Number:	13303417				
International Application Number:					
Confirmation Number:	3035				
Title of Invention:	SYSTEM AND METHOD FOR SCRAMBLING THE PHASE OF THE CARRIERS IN A MULTICARRIER COMMUNICATIONS SYSTEM				
First Named Inventor/Applicant Name:	Marcos C. Tzannes				
Customer Number:	62574				
Filer:	Jason Vick/Joanne Vos				
Filer Authorized By:	Jason Vick				
Attorney Docket Number:	5550-47-CON-5				
Receipt Date:	23-NOV-2011				
Filing Date:					
Time Stamp:	12:48:07				
Application Type:	Utility under 35 USC 111(a)				

Payment information:

Submitted with Payment	no
File Listing:	

Document Number	Document Description	File Name	File Size(Bytes)/ Message Digest	Multi Part /.zip	Pages (if appl.)
1		Statement_Under_373b_w_PO	269415	ves	2
'		A.pdf	22a5c8bec9716d6a28224120e11cfad7321 4eaf0	'	2

	Multipart Description/PDF files in .zip description		
	Document Description	Start	End
	Assignee showing of ownership per 37 CFR 3.73(b).	1	1
	Power of Attorney	2	2
Warnings			ı

Warnings:

Information:

Total Files Size (in bytes):	269415
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This Acknowledgement Receipt evidences receipt on the noted date by the USPTO of the indicated documents, characterized by the applicant, and including page counts, where applicable. It serves as evidence of receipt similar to a Post Card, as described in MPEP 503.

New Applications Under 35 U.S.C. 111

If a new application is being filed and the application includes the necessary components for a filing date (see 37 CFR 1.53(b)-(d) and MPEP 506), a Filing Receipt (37 CFR 1.54) will be issued in due course and the date shown on this Acknowledgement Receipt will establish the filing date of the application.

National Stage of an International Application under 35 U.S.C. 371

If a timely submission to enter the national stage of an international application is compliant with the conditions of 35 U.S.C. 371 and other applicable requirements a Form PCT/DO/EO/903 indicating acceptance of the application as a national stage submission under 35 U.S.C. 371 will be issued in addition to the Filing Receipt, in due course.

New International Application Filed with the USPTO as a Receiving Office

If a new international application is being filed and the international application includes the necessary components for an international filing date (see PCT Article 11 and MPEP 1810), a Notification of the International Application Number and of the International Filing Date (Form PCT/RO/105) will be issued in due course, subject to prescriptions concerning national security, and the date shown on this Acknowledgement Receipt will establish the international filing date of the application.

PTO/SB/96 (07-09)
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U.S. Patent and Trademark Office; U.S. DEPARTMENT OF COMMERCE
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STATEMENT UNDER 37 CF	R 3.73(b)			
Applicant/Patent Owner: AWARE, INC.				
Application No./Patent No.: 13/303,417 Filed/I	ssue Date: November 23, 2011			
Titled: SYSTEM AND METHOD FOR SCRAMBLING THE PHASE O COMMUNICATIONS SYSTEM	F THE CARRIERS IN A MULTICARRIER			
AWARE, INC. , a Corporation				
	e.g., corporation, partnership, university, government agency, etc.			
states that it is:				
1. X the assignee of the entire right, title, and interest in;				
an assignee of less than the entire right, title, and interest in (The extent (by percentage) of its ownership interest is	_ %); or			
3. the assignee of an undivided interest in the entirety of (a complete	assignment from one of the joint inventors was made)			
the patent application/patent identified above, by virtue of either:	,			
A. X An assignment from the inventor(s) of the patent application/patent identified above. The assignment was recorded in the United States Patent and Trademark Office at Reel 010877 Frame 0307 or for which a				
copy therefore is attached. OR				
B. A chain of title from the inventor(s), of the patent application/patent	identified above, to the current assignee as follows:			
1. From:	То:			
The document was recorded in the United States Patent and Trademark Office at				
Reel, Frame	or for which a copy thereof is attached.			
2. From:	То:			
The document was recorded in the United States Patent				
Reel, Frame	, or for which a copy thereof is attached.			
3. From:	То:			
The document was recorded in the United States Patent				
Reel, Frame	, or for which a copy thereof is attached.			
Additional documents in the chain of title are listed on a suppleme	ntal sheet(s).			
As required by 37 CFR 3.73(b)(1)(i), the documentary evidence of the	chain of title from the original owner to the assignee was			
or concurrently is being, submitted for recordation pursuant to 37 CFR 3	3.11.			
[NOTE: A separate copy (i.e., a true copy of the original assignment do accordance with 37 CFR Part 3, to record the assignment in the records	s of the USPTO. <u>See</u> MPEP 302.08]			
The undersigned (whose title is supplied below) is authorized to act on behalf	of the assignee.			
	November 23, 2011			
Signature	Date			
Jason H. Vick	Attorney for Assignee			
Printed or Typed Name	Title			

This collection of information is required by 37 CFR 3.73(b). The information is required to obtain or retain a benefit by the public which is to file (and by the USPTO to process) an application. Confidentiality is governed by 35 U.S.C. 122 and 37 CFR 1.11 and 1.14. This collection is estimated to take 12 minutes to complete, including gathering, preparing, and submitting the completed application form to the USPTO. Time will vary depending upon the individual case. Any comments on the amount of time you require to complete this form and/or suggestions for reducing this burden, should be sent to the Chief Information Officer, U.S. Patent and Trademark Office, U.S. Department of Commerce, P.O. Box 1450, Alexandria, VA 22313-1450. DO NOT SEND FEES OR COMPLETED FORMS TO THIS ADDRESS. SEND TO: Commissioner for Patents, P.O. Box 1450, Alexandria, VA 22313-1450.

If you need assistance in completing the form, call 1-800-PTO-9199 and select option 2.