

Figure 23-6—PCS sublayer to PMA sublayer frame structure

implementation are left to the implementor. The choice of binary encoding for each ternary symbol is left to the implementor.

The following frame elements appear in figure 23-6 (ternary symbols are transmitted leftmost first):

- SOSA The succession of six ternary symbols: [1 -1 1 -1 1 -1], which is the result of encoding the constant sosa.
- SOSB The succession of six ternary symbols: [1 -1 1 -1 -1 1], which is the result of encoding the constant sosb.
- P3 The succession of two ternary symbols: [1 -1].
- P4 The succession of four ternary symbols: [1 -1 1 -1].
- DATA A 6T code group that is the result of encoding a data octet in a packet that is not part of the clause 4 MAC preamble or SFD.
- EOP1-5 A 6T code group that is the result of encoding one of the end-of-packet patterns eop1-5.

23.2.4 PCS state diagrams

The notation used in the state diagrams follows the conventions of 21.5. Transitions shown without source states are evaluated continuously and take immediate precedence over all other conditions.

23.2.4.1 PCS state diagram constants

Register `tsr` may take on any of the nine constant values listed below (`sosa` through `eop5`, `bad_code`, and `zero_code`). These values are used to describe the functional operation of the coding process.

NOTE—Implementors are under no obligation to implement these constants in any particular way. For example, some implementors may choose to implement these codes as special flag bits attached to MII TXD nibble registers. Other implementors may choose to implement insertion of these codes on the downstream side of the coder function, using precoded 6T sequences.

All 6T code words are sent leftmost ternary symbol first.

sosa	A constant that encodes to: [1 -1 1 -1 1 -1].
sosb	A constant that encodes to: [1 -1 1 -1 -1 1].
eop1	A constant that encodes to: [1 1 1 1 1 1].
eop2	A constant that encodes to: [1 1 1 1 -1 -1].
eop3	A constant that encodes to: [1 1 -1 -1 0 0].
eop4	A constant that encodes to: [-1 -1 -1 -1 -1 -1].
eop5	A constant that encodes to: [-1 -1 0 0 0 0].
bad_code	A constant that encodes to: [-1 -1 -1 1 1 1].
zero_code	A constant that encodes to: [0 0 0 0 0 0].

23.2.4.2 PCS state diagram variables

codeword_error

Indicates reception of invalid 6T code group.

Values: ON and OFF

Set by: PCS Receive; error-detecting rules

dc_balance_error

Indicates reception of dc coding violation.

Values: ON and OFF

Set by: PCS Receive; error-detecting rules

eop

Indicates reception of eop1.

A state variable set by the decoding operation. Reset to OFF when in PCS Receive state AWAITING INPUT. When the decoder detects eop1 on any pair, it sets this flag ON. The timing of eop shall be adjusted such that the last nibble of the last decoded data octet in a packet is the last nibble sent across the MII by the PMA Receive state diagram with RX_DV set ON.

Values: ON and OFF

Set by: PCS Receive; error-detecting rules

eop_error

Indicates reception of data with improper end-of-packet coding.

Values: ON and OFF

Set by: PCS Receive; error-detecting rules

ih2, ih4, and ih3 (input holding registers)

A set of holding registers used for the purpose of holding decoded data octets in preparation for sending across the MII one nibble at a time. One register is provided for each of the three receive pairs RX_D2, BI_D4, and BI_D3, respectively.

Value: octet
Set by: PCS Receive

Each time the PCS Receive function decodes a 6T code group, it loads the result (an octet) into one of the ih2-4 registers. These three registers are loaded in round-robin fashion, one register being loaded every two ternary symbol times.

The PCS Receive state diagram reads nibbles as needed from the ih2-4 registers and stuffs them into RXD.

ohr1, ohr3, and ohr4 (output holding registers)

(See figure 23-7.) A set of shift registers used for the purpose of transferring coded 6T ternary symbol groups one ternary symbol at a time into the PMA. One register is provided for each of the three transmit pairs TX_D1, BI_D3, and BI_D4, respectively.

Value: 6T code group. Each of the six cells holds one ternary symbol (i.e., -1, 0, or 1).
Set by: PCS Transmit

Each time the PCS Transmit function encodes a data octet, it loads the result (a 6T code group) into one of the ohr registers. Three registers are loaded in round-robin fashion, one register being loaded every two ternary symbol times. The PCS shall transmit octets on the three transmit pairs in round-robin fashion, in the order TX_D1, BI_D3, and BI_D4, starting with TX_D1.

The PMA_UNITDATA request (DATA) message picks the least significant (rightmost) ternary symbol from each ohr register and sends it to the PMA, as shown below. (Note that 6T code words in annex 23A are listed with lsb on the left, not the right.)

tx_code_vector[TX_D1] = the LSB of ohr1, also called ohr1[0]
tx_code_vector[BI_D3] = the LSB of ohr3, also called ohr3[0]
tx_code_vector[BI_D4] = the LSB of ohr4, also called ohr4[0]

After each PMA_UNITDATA request message, all three ohr registers shift right by one ternary symbol, shifting in zero from the left. The PCS Transmit function loads a new 6T code group into each ohr immediately after the last ternary symbol of the previous group is shifted out.

At the beginning of a preamble, the PCS Transmit function loads the same value (sosa) into all three output holding registers, which causes alternating transitions to immediately appear on all three output pairs. The result on pairs BI_D3 and BI_D4 is depicted by code words P3 and P4 in figure 23-6.

pcs_reset

Causes reset of all PCS functions when ON.

Values: ON and OFF
Set by: PCS Reset

rx_crs

A latched asynchronous variable. Timing for the MII signal CRS is derived from rx_crs.

Values: ON and OFF

Set ON when: carrier_status changes to ON

Set OFF when either of two events occurs:
carrier_status changes to OFF, or
detection of eop1, properly framed, on any of the lines RX_D2, BI_D4, or
BI_D3

Additionally, if, 20 ternary symbol times after rx_crs falls, carrier_status remains set to ON then set rx_crs=ON.

NOTE—A special circuit for the detection of eop1 and subsequent de-assertion of rx_crs, faster than the full 8B6T decoding circuits, is generally required to meet the timing requirements for CRS listed in clause 23.11.

tsr (transmit shift register)

(See figure 23-7.) A shift register defined for the purpose of assembling nibbles from the MII TXD into octets.

Values: The variable tsr always contains both the current nibble of TXD and the previous nibble of TXD. Valid values for tsr therefore include all octets. Register tsr may also take on any of the nine constant values listed in 23.2.4.1.

Nibble order: When encoding the tsr octet, the previous TXD nibble is considered the least significant nibble.

Set by: PCS Transmit

During the first 16 TX_CLK cycles after TX_EN is asserted, tsr shall assume the following values in sequence regardless of TXD: sosa, sosa, sosa, sosa, sosa, sosa, sosa, sosa, sosa, sosa, sosb, sosb, sosb, sosb, sosb. This action substitutes the 100BASE-T4 preamble for the clause 4 MAC preamble. The PCS Transmit state diagram samples the tsr only every other clock, which reduces the number of sosa and sosb constants actually coded to 5 and 3, respectively.

During the first 10 TX_CLK cycles after TX_EN is de-asserted, tsr shall assume the following values in sequence, regardless of TXD: eop1, eop1, eop2, eop2, eop3, eop3, eop4, eop4, eop5, eop5. This action appends the 100BASE-T4 end-of-packet delimiter to each pair. The PCS Transmit state diagram samples the tsr only every other clock, which reduces the number of eop1-5 constants actually coded to 1 each.

Except for the first 16 TX_CLK cycles after TX_EN is asserted, any time TX_ER and TX_EN are asserted, tsr shall assume the value bad_code with such timing as to cause both nibbles of the affected octet to be encoded as bad_code. If TX_ER is asserted at any time during the first 16 TX_CLK cycles after TX_EN is asserted, tsr shall during the 17th and 18th clock cycles assume the value bad_code.

If TX_EN is de-asserted on an odd nibble boundary, the PCS shall extend TX_EN by one TX_CLK cycle, and behave as if TX_ER were asserted during that additional cycle.

Except for the first 10 TX_CLK cycles after TX_EN is de-asserted, any time TX_EN is not asserted, tsr shall assume the value zero_code.

tx_extend

A latched, asynchronous state variable used to extend the TX_EN signal long enough to ensure complete transmission of all nonzero ternary symbols in eop1-5.

Values: ON and OFF

Set ON upon: rising edge of TX_EN

Set OFF upon either of two conditions:
 a) In the event of a collision (COL is asserted at any time during transmission) set tx_extend=OFF when TX_EN de-asserts.
 b) In the event of no collision (COL remains de-asserted throughout transmission) set tx_extend=OFF upon completion of transmission of last ternary symbol in eop4.

NOTES

1—The 6T code group eop5 has four zeroes at the end. The 6T code group eop4 contains the last nonzero ternary symbol to be transmitted.

2—The effect of a collision, if present, is to truncate the frame at the original boundary determined by TX_EN. Noncolliding frames are extended, while colliding frames are not.

23.2.4.3 PCS state diagram timer**tw1_timer**

A continuous free-running timer.

Values: The condition tw1_timer_done goes true when the timer expires.

Restart when: Immediately after expiration (restarting the timer resets condition tw1_timer_done).

Duration: 40 ns nominal.

TX_CLK shall be generated synchronous to tw1_timer (see tolerance required for TX_CLK in 23.5.1.2.10).

On every occurrence of tw1_timer_done, the state diagram advances by one block. The message PMA_UNITDATA request is issued concurrent with tw1_timer_done.

23.2.4.4 PCS state diagram functions**encode()**

The encode operation of 23.2.1.2.

Argument: octet

Returns: 6T code group

decode()

The decode operation of 23.2.1.3.

Argument: 6T code group

Returns: octet

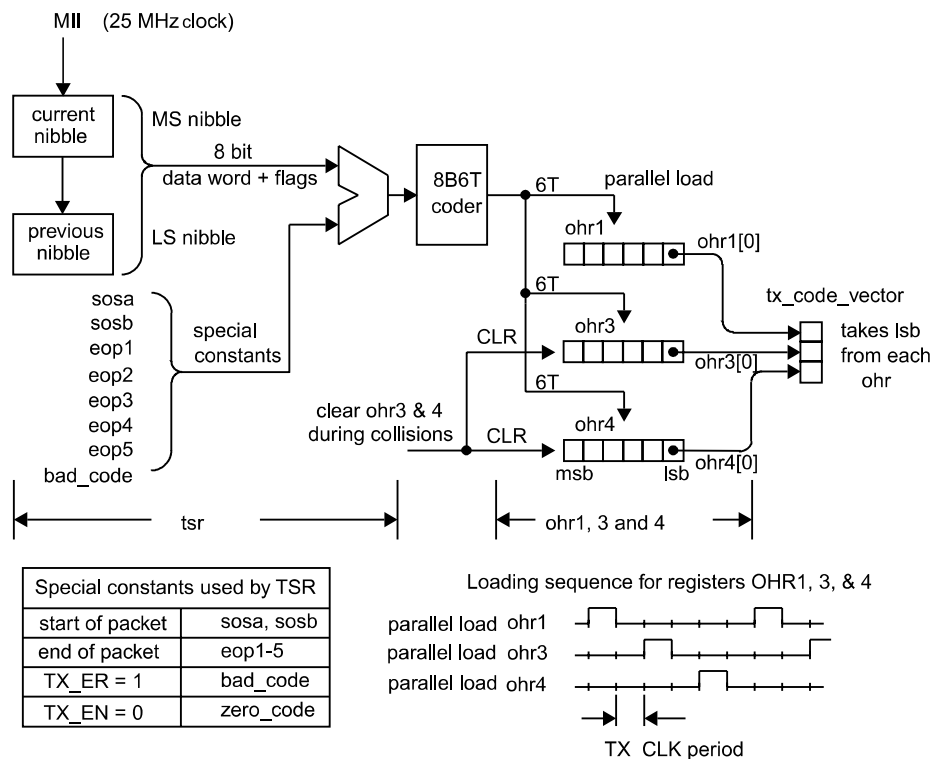


Figure 23-7—PCS Transmit reference diagram

23.2.4.5 PCS state diagrams

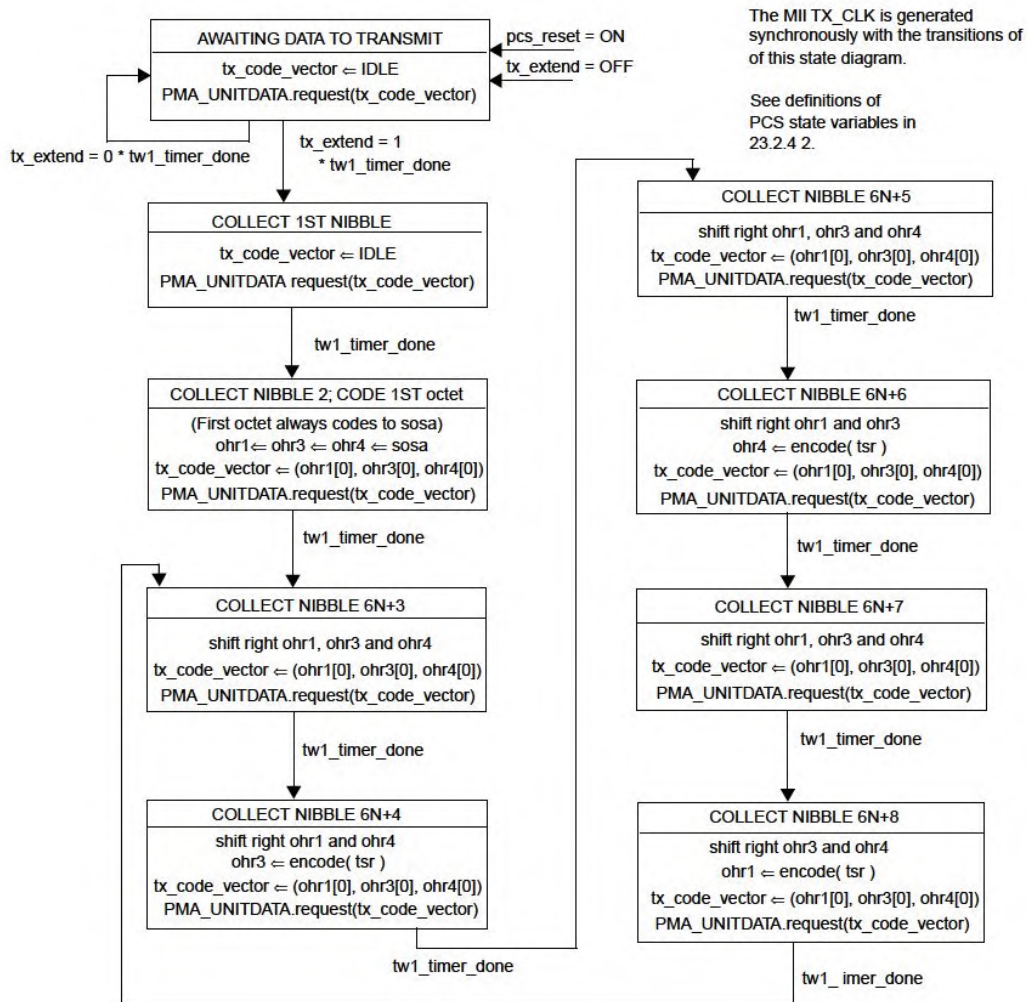


Figure 23-8—PCS Transmit state diagram

This is an Archive IEEE Standard. It has been superseded by a later version of this standard.

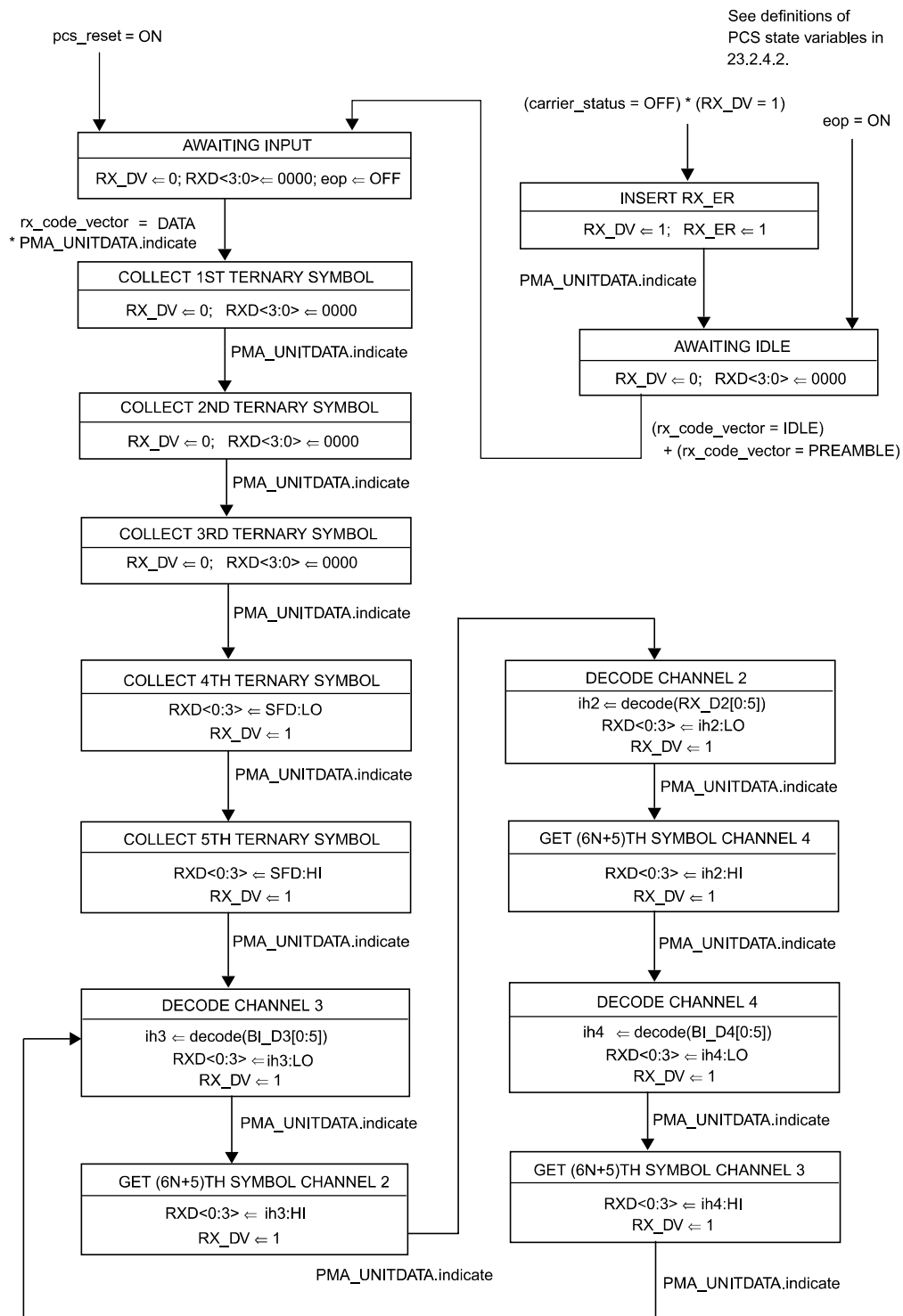


Figure 23-9—PCS Receive state diagram

This is an Archive IEEE Standard. It has been superseded by a later version of this standard.

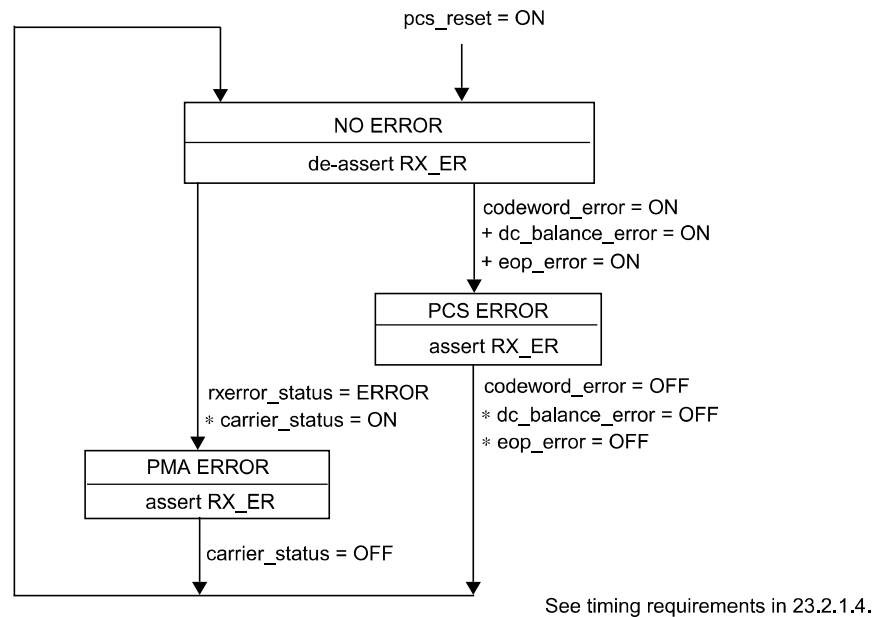


Figure 23-10—PCS Error Sense state diagram

23.2.5 PCS electrical specifications

The interface between PCS and PMA is an abstract message-passing interface, having no specified electrical properties.

Electrical characteristics of the signals passing between the PCS and MII may be found in clause 22.

23.3 PMA service interface

This clause specifies the services provided by the PMA to either the PCS or a Repeater client. These services are described in an abstract manner and do not imply any particular implementation.

The PMA Service Interface supports the exchange of code vectors between the PMA and its client (either the PCS or a Repeater). The PMA also generates status indications for use by the client.

The following primitives are defined:

- PMA_TYPE.indicate
- PMA_UNITDATA.request
- PMA_UNITDATA.indicate
- PMA_CARRIER.indicate
- PMA_LINK.indicate
- PMA_LINK request
- PMA_RXERROR.indicate

23.3.1 PMA_TYPE.indicate

This primitive is generated by the PMA to indicate the nature of the PMA instantiation. The purpose of this primitive is to allow clients to support connections to the various types of 100BASE-T PMA entities in a generalized manner.

23.3.1.1 Semantics of the service primitive

PMA_TYPE.indicate (pma_type)

The pma_type parameter for use with the 100BASE-T4 PMA is T4.

23.3.1.2 When generated

The PMA shall continuously generate this primitive to indicate the value of pma_type.

23.3.1.3 Effect of receipt

The client uses the value of pma_type to define the semantics of the PMA_UNITDATA.request and PMA_UNITDATA.indicate primitives.

23.3.2 PMA_UNITDATA.request

This primitive defines the transfer of data (in the form of tx_code_vector parameters) from the PCS or repeater to the PMA.

23.3.2.1 Semantics of the service primitive

PMA_UNITDATA.request (tx_code_vector)

When transmitting data using 100BASE-T4 signaling, the PMA_UNITDATA.request conveys to the PMA simultaneously the logical output value for each of the three transmit pairs TX_D1, BI_D3, and BI_D4. The value of tx_code_vector during data transmission is therefore a three-element vector, with one element corresponding to each output pair. Each of the three elements of the tx_code_vector may take on one of three logical values: 1, 0, or -1, corresponding to the three ternary possibilities +, 0, and - listed for each ternary symbol in the 8B6T code table (see annex 23A).

Between packets, the 100BASE-T4 PMA layer sends the 100BASE-T4 idle signal, TP_IDLE_100. The PCS informs the PMA layer that it is between packets, thus enabling the PMA idle signal, by setting the tx_code_vector parameter to IDLE.

For pma_type 100BASE-T4, the tx_code_vector parameter can take on either of two forms:

IDLE	A single value indicating to the PMA that there is no data to convey. The PMA generates link integrity pulses during the time that tx_code_vector = IDLE.
DATA	A vector of three ternary symbols, one for each of the three transmit pairs TX_D1, BI_D3, and BI_D4. The ternary symbol for each pair may take on one of three values, 1, 0, or -1.

The ternary symbols comprising tx_code_vector, when they are conveyed using the DATA format, are called, according to the pair on which each will be transmitted, tx_code_vector[BI_D4], tx_code_vector[TX_D1], and tx_code_vector[BI_D3].

23.3.2.2 When generated

The PCS or Repeater client generates PMA_UNITDATA.request synchronous with every MII TX_CLK.

For the purposes of state diagram descriptions, it may be assumed that at the time PMA_UNITDATA request is generated, the MII signals TX_EN, and TX_ER, and TXD instantly become valid and that they retain their values until the next PMA_UNITDATA request.

In the state diagrams, PMA_UNITDATA.request is assumed to occur at the conclusion of each tw1 wait function.

23.3.2.3 Effect of receipt

Upon receipt of this primitive, the PMA transmits the indicated ternary symbols on the MDI.

23.3.3 PMA_UNITDATA.indicate

This primitive defines the transfer of data (in the form of rx_code_vector parameters) from the PMA to the PCS or repeater during the time that link_status=OK.

23.3.3.1 Semantics of the service primitive

PMA_UNITDATA.indicate (rx_code_vector)

When receiving data using 100BASE-T4 signaling, the PMA_UNITDATA.indicate conveys to the PCS simultaneously the logical input value for each of the three receive pairs RX_D2, BI_D4, and BI_D3. The value of rx_code_vector during data reception is therefore a three-element vector, with one element corresponding to each input pair. Each of the three elements of the rx_code_vector may take on one of three logical values: 1, 0, or -1, corresponding to the three ternary possibilities +, 0, and - listed for each ternary symbol in the 8B6T code table (see annex 23A).

Between packets, the rx_code_vector is set by the PMA to the value IDLE.

From the time the PMA asserts carrier_status=ON until the PMA recognizes the SSD pattern (not all of the pattern need be received in order for the PMA to recognize the pattern), the PMA sets rx_code_vector to the value PREAMBLE.

For pma_type 100BASE-T4, the rx_code_vector parameter can take on any of three forms:

IDLE	A single value indicating that the PMA has no data to convey.
PREAMBLE	A single value indicating that the PMA has detected carrier, but has not received a valid SSD.
DATA	A vector of three ternary symbols, one for each of the three receive pairs RX_D2, BI_D3, and BI_D4. The ternary symbol for each pair may take on one of three values, 1, 0, or -1.

The ternary symbols comprising rx_code_vector, when they are conveyed using the DATA format, are called, according to the pair upon which each symbol was received, rx_code_vector[BI_D3], rx_code_vector[RX_D2], and rx_code_vector[BI_D4].

23.3.3.2 When generated

The PMA shall generate PMA_UNITDATA.indicate (DATA) messages synchronous with data received at the MDI.

23.3.3.3 Effect of receipt

The effect of receipt of this primitive is unspecified.

23.3.4 PMA_CARRIER.indicate

This primitive is generated by the PMA to indicate the status of the signal being received from the MDI. The purpose of this primitive is to give the PCS or repeater client the earliest reliable indication of activity on the underlying medium.

23.3.4.1 Semantics of the service primitive

PMA_CARRIER.indicate (carrier_status)

The carrier_status parameter can take on one of two values: OFF or ON, indicating whether the incoming signal should be interpreted as being between packets (OFF) or as a packet in progress (ON).

23.3.4.2 When generated

The PMA shall generate this primitive to indicate the value of carrier_status.

23.3.4.3 Effect of receipt

The effect of receipt of this primitive is unspecified.

23.3.5 PMA_LINK.indicate

This primitive is generated by the PMA to indicate the status of the underlying medium. The purpose of this primitive is to give the PCS or repeater client or Auto-Negotiation algorithm a means of determining the validity of received code elements.

23.3.5.1 Semantics of the service primitive

PMA_LINK.indicate (link_status)

The link_status parameter can take on one of three values: FAIL, READY, or OK:

FAIL	The link integrity function does not detect a valid 100BASE-T4 link.
READY	The link integrity function detects a valid 100BASE-T4 link, but has not been enabled by Auto-Negotiation.
OK	The 100BASE-T4 link integrity function detects a valid 100BASE-T4 link, and has been enabled by Auto-Negotiation.

23.3.5.2 When generated

The PMA shall generate this primitive to indicate the value of link_status.

23.3.5.3 Effect of receipt

The effect of receipt of this primitive is unspecified.

23.3.6 PMA_LINK.request

This primitive is generated by the Auto-Negotiation algorithm. The purpose of this primitive is to allow the Auto-Negotiation algorithm to enable and disable operation of the PHY.

23.3.6.1 Semantics of the service primitive

PMA_LINK request (link_control)

The link_control parameter can take on one of three values: SCAN_FOR_CARRIER, DISABLE, or ENABLE.

SCAN_FOR_CARRIER	Used by the Auto-Negotiation algorithm prior to receiving any fast link pulses. During this mode the PHY reports link_status=READY if it recognizes 100BASE-T4 carrier from the far end, but no other actions are enabled.
DISABLE	Used by the Auto-Negotiation algorithm to disable PHY processing in the event fast link pulses are detected. This gives the Auto-Negotiation algorithm a chance to determine how to configure the link.
ENABLE	Used by Auto-Negotiation to turn control over to the PHY for data processing functions. This is the default mode if Auto-Negotiation is not present.

23.3.6.2 Default value of parameter link_control

Upon power-on, reset, or release from power-down, the link_control parameter shall revert to ENABLE. If the optional Auto-Negotiation algorithm is not implemented, no PMA_LINK.request message will arrive and the PHY will operate indefinitely with link_control=ENABLE.

23.3.6.3 When generated

The Auto-Negotiation algorithm generates this primitive to indicate to the PHY how to behave.

Upon power-on, reset, or release from power down, the Auto-Negotiation algorithm, if present, issues the message PMA_LINK request (SCAN_FOR_CARRIER).

23.3.6.4 Effect of receipt

Whenever link_control=SCAN_FOR_CARRIER, the PHY shall enable the Link Integrity state diagram, but block passage into the state LINK_PASS, while holding rcv=DISABLE, and xmit=DISABLE. While link_control=SCAN_FOR_CARRIER, the PHY shall report link_status=READY if it recognizes 100BASE-T4 link integrity pulses coming from the far end, otherwise it reports link_status=FAIL.

Whenever link_control=DISABLE, the PHY shall report link_status=FAIL and hold the Link Integrity state diagram in the RESET state, while holding rcv=disable and xmit=DISABLE.

While link_control=ENABLE, the PHY shall allow the Link Integrity function to determine if the link is available and, if so, set rcv=ENABLE and xmit=ENABLE.

23.3.7 PMA_RXERROR.indicate

The primitive is generated in the PMA by the PMA Align function to indicate the status of the signal being received from the MDI. The purpose of this primitive is to give the PCS or repeater client an indication of a PMA detectable receive error.

23.3.7.1 Semantics of the service primitive

PMA_RXERROR.indicate (rxerror_status)

The rxerror_status parameter can take on one of two values: ERROR or NO_ERROR, indicating whether the incoming signal contains a detectable error (ERROR) or not (NO_ERROR).

23.3.7.2 When generated

The PMA shall generate this primitive to indicate whether or not each incoming packet contains a PMA detectable error (23.2.1.4).

23.3.7.3 Effect of receipt

The effect of receipt of this primitive is unspecified.

23.4 PMA functional specifications

The PMA couples messages from a PMA service interface (23.3) to the 100BASE-T4 baseband medium (23.6).

The interface between PCS and the baseband medium is the Medium Dependent Interface (MDI), specified in 23.7.

23.4.1 PMA functions

The PMA sublayer comprises one PMA Reset function and six simultaneous and asynchronous operating functions. The PMA operating functions are PMA Transmit, PMA Receive, PMA Carrier Sense, Link Integrity, PMA Align, and Clock Recovery. All operating functions are started immediately after the successful completion of the PMA Reset function. When the PMA is used in conjunction with a PCS, the RESET function may be shared between layers.

The PMA reference diagram, figure 23-11, shows how the operating functions relate to the messages of the PMA Service interface and the signals of the MDI. Connections from the management interface, comprising the signals MDC and MDIO, to other layers are pervasive, and are not shown in figure 23-11. The Management Interface and its functions are specified in clause 22.

23.4.1.1 PMA Reset function

The PMA Reset function shall be executed any time either of two conditions occur. These two conditions are power-on and the receipt of a reset request from the management entity. The PMA Reset function initializes all PMA functions. The PMA Reset function sets pma_reset <= ON for the duration of its reset function. All state diagrams take the open-ended pma_reset branch upon execution of the PMA Reset function. The reference diagrams do not explicitly show the PMA Reset function.

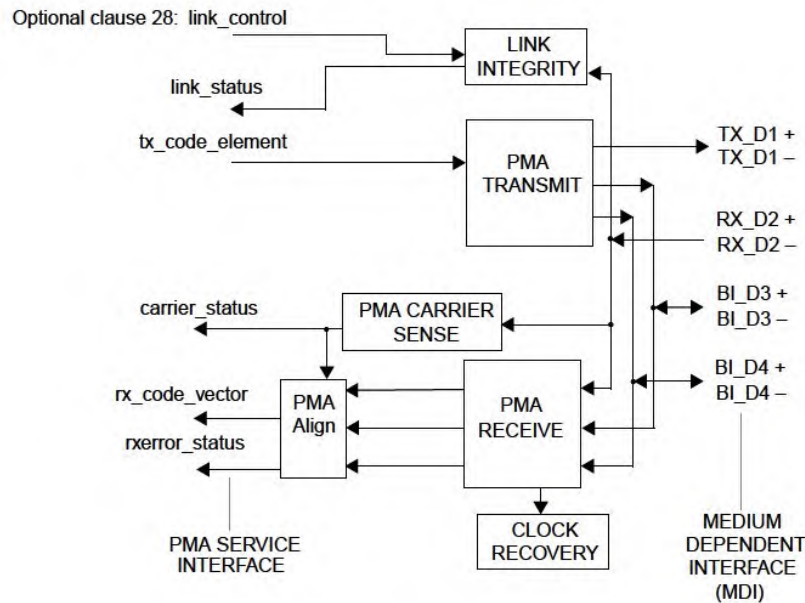


Figure 23-11—PMA reference diagram

23.4.1.2 PMA Transmit function

Except as provided for in the next paragraph, whenever $(tx_code_vector=DATA) \times (pma_carrier=OFF)$, the PMA shall transmit onto the MDI ternary symbols on pairs TX_D1, BI_D3, and BI_D4 equal to $tx_code_vector[TX_D1]$, $tx_code_vector[BI_D3]$, and $tx_code_vector[BI_D4]$, respectively.

Whenever $(tx_code_vector=DATA) \times (pma_carrier=ON)$, the PMA shall transmit onto the MDI ternary symbols on pairs TX_D1, BI_D3, and BI_D4 equal to $tx_code_vector[TX_D1]$, CS0, and CS0, respectively, and continue doing so until $tx_code_vector=IDLE$.

NOTE—This shuts off the transmitters on channels BI_D3 and BI_D4, and keeps them off, in the event of a collision. Shutting off the transmitters prevents overload and saturation of the transmitters, and also reduces the amount of near-end crosstalk present while monitoring for the end of carrier.

Whenever $tx_code_vector=IDLE$, an idle signal shall be transmitted on pair TX_D1 and silence on pairs BI_D3 and BI_D4. The idle signal consists of periods of silence (times where the differential output voltage remains at $0\text{ mV} \pm 50\text{ mV}$) broken by the transmission of link integrity test pulses.

The 100BASE-T4 idle signal is similar to the 10BASE-T idle signal, but with 100BASE-T4 ternary signal levels and a faster repetition rate. The 100BASE-T4 idle signal is called TP_IDL_100. The TP_IDL_100 signal shall be a repeating sequence formed from one $1.2\text{ ms} \pm 0.6\text{ ms}$ period of silence (the time where the differential voltage remains at $0\text{ mV} \pm 50\text{ mV}$) and one link test pulse. Each link test pulse shall be a succession of two ternary symbols having logical values of -1 and 1 transmitted on pair TX_D1 using CS-1 and CS1 as defined in 23.4.3.1. Following a packet, the TP_IDL_100 shall start with a period of silence.

Transmission of TP_IDL_100 may be terminated at any time with respect to the link test pulse. It shall be terminated such that ternary symbols of the subsequent packet are not corrupted, and are not delayed any more than is specified in 23.11.

For any link test pulse occurring within 20 ternary symbol times of the beginning of a preamble, the zero crossing jitter (as defined in 23.5.1.2.5) of the link test pulse when measured along with the zero crossings of the preamble shall be less than 4 ns p-p.

NOTE—The above condition allows clock recovery implementations that optionally begin fast-lock sequences on part of a link integrity pulse to properly acquire lock on a subsequent preamble sequence.

Regardless of other considerations, when the transmitter is disabled (xmit=DISABLE), the PMA Transmit function shall transmit the TP_IDL_100 signal.

23.4.1.3 PMA Receive function

PMA Receive contains the circuits necessary to convert physically encoded ternary symbols from the physical MDI receive pairs (RX_D2, BI_D3 and BI_D4) into a logical format suitable for the PMA Align function. Each receive pair has its own dedicated PMA Receive circuitry.

The PHY shall receive the signals on the receive pairs (RX_D2, BI_D3, and BI_D4) and translate them into one of the PMA_UNITDATA.indicate parameters IDLE, PREAMBLE, or DATA with a ternary symbol error rate of less than one part in 10^8 .

If both pma_carrier=ON and tx_code_vector=DATA, the value of rx_code_vector is unspecified until pma_carrier=OFF.

23.4.1.4 PMA Carrier Sense function

The PMA Carrier Sense function shall set pma_carrier=ON upon reception of the following pattern on pair RX_D2 at the receiving MDI, as measured using a 100BASE-T4 transmit test filter (23.5.1.2.3):

Any signal greater than 467 mV, followed by any signal less than -225 mV, followed by any signal greater than 467 mV, all three events occurring within 2 ternary symbol times.

The operation of carrier sense is undefined for signal amplitudes greater than 4.5 V.

See 23.5.1.3.2 for a list of signals defined *not* to set pma_carrier=ON.

After asserting pma_carrier=ON, PMA Carrier Sense shall set pma_carrier=OFF upon receiving either of these conditions:

- a) Seven consecutive ternary symbols of value CS0 on pair RX_D2.
- b) (tx_code_vector=DATA) has not been true at any time since pma_carrier was asserted, *and* the 6T code group eop1 has been received, properly framed, on any of the lines RX_D2, BI_D4, or BI_D3, *and* enough time has passed to assure passage of all ternary symbols of eop4 across the PMA service interface.

NOTE—Designers may wish to take advantage of the fact that the minimum received packet fragment will include at least 24 ternary symbols of data on pair RX_D2. Therefore, once carrier is activated, it is not necessary to begin searching for seven consecutive zeroes until after the 24th ternary symbol has been received. During the time that the first 24 ternary symbols are being received, the near-end crosstalk from pairs BI_D3 and BI_D4, which are switched off during collisions, decays substantially.

While rcv=ENABLE, the PMA CARRIER function shall set carrier_status = pma_carrier.

While `rcv≠ENABLE`, the PMA CARRIER function shall set `carrier_status = OFF`.

This function operates independently of the Link Integrity function.

23.4.1.5 Link Integrity function

Link Integrity provides the ability to protect the network from the consequences of failure of the simplex link attached to `RX_D2`. While such a failure is present, transfer of data by the Transmit and Receive functions is disabled.

Link Integrity observes the incoming wire pair, `RX_D2`, to determine whether the device connected to the far end is of type 100BASE-T4. Based on its observations, Link Integrity sets two important internal variables:

- a) `pma_type` variable is set to 100BASE-T4.
- b) `link_status` variable is a parameter sent across the PMA Service interface.

The Link Integrity function shall comply with the state diagram of figure 23-12.

Four conditions gate the progression of states toward `LINK_PASS`: (1) reception of at least 31 link integrity test pulses; (2) reception of at least 96 more link integrity test pulses, or reception of carrier; (3) cessation of carrier, if it was present; (4) detection of equals `link_control ENABLE`.

While the PMA is not in the `LINK_PASS` state, the Link Integrity function sets `rcv=DISABLE` and `xmit=DISABLE`, thus disabling the bit transfer of the Transmit and Receive functions.

If a visible indicator is provided on the PHY to indicate the link status, it is recommended that the color be green and that the indicator be labeled appropriately. It is further recommended that the indicator be on when the PHY is in the `LINK_PASS` state and off otherwise.

23.4.1.6 PMA Align function

The PMA Align function accepts received ternary symbols from the PMA Receive function, along with `pma_carrier`. PMA Align is responsible for realigning the received ternary symbols to eliminate the effects of unequal pair propagation time, commonly called pair skew. PMA Align also looks for the SSD pattern to determine the proper alignment of 6T code groups, and then forwards `PMA_UNITDATA.indicate (DATA)` messages to the PCS. The SSD pattern includes referencing patterns on each of the three receive lines that may be used to establish the proper relationship of received ternary symbols (see figure 23-6).

NOTE—The skew between lines is not expected to change measurably from packet to packet.

At the beginning of each received frame, the PMA Carrier Sense function asserts `pma_carrier=ON`. During the preamble, the Clock Recovery function begins synchronizing its receive clock. Until clock is synchronized, data coming from the low-level PMA Receive function is meaningless. The PMA Align function is responsible for waiting for the receiver clock to stabilize and then properly recognizing the 100BASE-T4 coded SSD pattern. The PMA Align function shall send `PMA_UNITDATA.indicate (PREAMBLE)` messages to the PCS from the time `pma_carrier=ON` is asserted until the PMA is ready to transfer the first `PMA_UNITDATA.indicate (DATA)` message. Once the PMA Align function locates a SSD pattern, it begins forwarding `PMA_UNITDATA.indicate (DATA)` messages to the PCS, starting with the first ternary symbol of the first data word on pair `BI_D3`, as defined in figure 23-6. This first `PMA_UNITDATA.indicate (DATA)` message shall transfer the following ternary symbols, as specified in the frame structure diagram, figure 23-6:

- `rx_code_vector[BI_D3]`first ternary symbol of first data code group
- `rx_code_vector[RX_D2]`second ternary symbol prior to start of second data code group

rx_code_vector[BI_D4]fourth ternary symbol prior to start of third data code group

PMA Align shall continue sending PMA_UNITDATA.indicate (DATA) messages until pma_carrier=OFF. While pma_carrier=OFF, PMA Align shall emit PMA_UNITDATA.indicate (IDLE) messages.

If no valid SSD pattern is recognized within 22 ternary symbol times of the assertion of pma_carrier=ON, the PMA Align function shall set rxerror_status=ERROR. The PMA Align function is permitted to begin sending PMA_UNITDATA.indicate (DATA) messages upon receipt of a partially recognized SSD pattern, but it is required to set rxerror_status=ERROR if the complete SSD does not match perfectly the expected ternary symbol sequence. Rxerror_status shall be reset to NO_ERROR when pma_carrier=OFF.

The PMA Align function is permitted to use the first received packet of at least minimum size after RESET or the transition to LINK_PASS to learn the nominal skew between pairs, adjust its equalizer, or perform any other initiation functions. During this first packet, the PMA Align function shall emit PMA_UNITDATA.indicate (PREAMBLE) messages, but may optionally choose to never begin sending PMA_UNITDATA.indicate (DATA) messages.

The PMA Align function shall tolerate a maximum skew between any two pairs of 60 ns in either direction without error.

To protect the network against the consequences of mistaken packet framing, the PMA Align function shall detect the following error and report it by setting rxerror_status=ERROR (optionally, those error patterns already detected by codeword_error, dc_balance_error, or eop_error do not also have to be detected by rxerror_status): *In a series of good packets, any one packet that has been corrupted with three or fewer ternary symbols in error causing its sosb 6T code groups on one or more pairs to appear in the wrong location.*

Several approaches are available for meeting this requirement, including, but not limited to, a) comparing the relative positions of sosb 6T code groups on successive packets; b) measuring the time between the first preamble pulse and reception of sosb on each pair; c) counting the number of zero crossings from the beginning of the preamble until sosb; and d) monitoring for exception strings like "11" and "-1-1-1" in conjunction with one or more of the above techniques.

Regardless of other considerations, when the receive function is disabled (rcv=DISABLE), the PMA Align function shall emit PMA_UNITDATA.indicate (IDLE) messages and no others.

23.4.1.7 Clock Recovery function

The Clock Recovery function couples to all three receive pairs. It provides a synchronous clock for sampling each pair. While it may not drive the MII directly, the Clock Recovery function is the underlying root source of RX_CLK.

The Clock Recovery function shall provide a clock suitable for synchronously decoding ternary symbols on each line within the bit error tolerance provided in 23.4.1.3. During each preamble, in order to properly recognize the frame delimiting pattern formed by code word sosb on each pair, the received clock signal must be stable and ready for use in time to decode the following ternary symbols: the 16th ternary symbol of pair RX_D2, the 18th ternary symbol of pair BI_D4, and the 14th ternary symbol of pair BI_D3.

23.4.2 PMA interface messages

The messages between the PMA and PCS are defined above in 23.3, PMA Service Interface. Communication between a repeater unit and PMA also uses the PMA Service Interface. Communication through the MDI is summarized in tables 23-2 and 23-3.

Table 23-2—MDI signals transmitted by the PHY

Signal	Allowed pair	Meaning
CS1	TX_D1, BI_D3 BI_D4	A waveform that conveys the ternary symbol 1. Nominal voltage level +3.5 V.
CS0	TX_D1, BI_D3 BI_D4	A waveform that conveys the ternary symbol 0. Nominal voltage level 0 V.
CS-1	TX_D1, BI_D3 BI_D4	A waveform that conveys the ternary symbol -1. Nominal voltage level -3.5 V.
TP_IDL_100	TX_D1	Idle signal. Indicates transmitter is currently operating at 100 Mb/s.

Table 23-3—Signals received at the MDI

Signal	Allowed pair	Meaning
CS1	RX_D2, BI_D3 BI_D4	A waveform that conveys the ternary symbol 1. Nominal transmitted voltage level +3.5 V.
CS0	RX_D2, BI_D3 BI_D4	A waveform that conveys the ternary symbol 0. Nominal transmitted voltage level 0 V.
CS-1	RX_D2, BI_D3 BI_D4	A waveform that conveys the ternary symbol -1. Nominal transmitted voltage level -3.5 V.
TP_IDL_100	RX_D2	Idle signal. Indicates transmitter is currently operating at 100 Mb/s.

TP_IDL_100 is defined in 23.4.1.2. The waveforms used to convey CS1, CS0, and CS-1 are defined in 23.5.1.2.

TP_IDL_100 is defined in 23.4.1.2. The encodings for CS1, CS0, and CS-1 are defined in 23.5.1.2.

Re-timing of CS1, CS0, and CS-1 signals within the PMA is required.

23.4.3 PMA state diagrams

The notation used in the state diagrams follows the conventions of 21.5. Transitions shown without source states are evaluated continuously and take immediate precedence over all other conditions.

23.4.3.1 PMA constants

CS0

A waveform that conveys the ternary symbol 0.

Value: CS0 has a nominal voltage of 0 V. See 23.5.1.2.

CS1

A waveform that conveys the ternary symbol 1.

Value: CS1 has a nominal peak voltage of +3.5 V. See 23.5.1.2.

CS-1

A waveform that conveys the ternary symbol -1.

Value: CS-1 has a nominal peak voltage of -3.5 V. See 23.5.1.2.

link_100_max

A constant.

Value: Greater than 5.0 ms and less than 7.0 ms.

Used by link_max_timer to detect the absence of 100BASE-T4 link test pulses on pair RX_D2.

link_100_min

A constant.

Value: Greater than 0.15 ms and less than 0.45 ms.

Used by cnt_link to detect link test pulses on pair RX_D2 that are too close together to be valid 100BASE-T4 link test pulses.

23.4.3.2 State diagram variables**pma_reset**

Causes reset of all PCS functions.

Values: ON and OFF

Set by: PMA Reset

pma_carrier

A version of carrier_status used internally by the PMA sublayer. The variable pma_carrier always functions regardless of the link status. The value of pma_carrier is passed on through the PMA service interface as carrier_status when rcv=ENABLE. At other times, the passage of pma_carrier information to the PMA service interface is blocked.

Values: ON, OFF

Set by: PMA CARRIER

rcv

Controls the flow of data from the PMA to PCS through the PMA_UNITDATA.indicate message.

Values: ENABLE (receive is enabled)
DISABLE (the PMA always sends PMA_UNITDATA.indicate (IDLE), and carrier_status is set to OFF)

xmit

Controls the flow of data from PCS to PMA through the PMA_UNITDATA.request message.

Values: ENABLE (transmit is enabled)
DISABLE (the PMA interprets all PMA_UNITDATA.request messages as PMA_UNITDATA.request (IDLE). The PMA transmits no data, but continues sending TP_IDL_100).

23.4.3.3 State diagram timers**link_max_timer**

A re-triggerable timer.

Values: The condition link_max_timer_done goes true when the timer expires.

Restart when: Timer is restarted for its full duration by every occurrence of either a link test pulse on pair RX_D2 or the assertion of pma_carrier=ON (restarting the timer resets the condition link_max_timer_done).

Duration: link_100_max

Used by Link Integrity to detect the absence of 100BASE-T4 link test pulses on pair RX_D2.

23.4.3.4 State diagram counters

cnt_link

Counts number of 100BASE-T4 link test pulses (see 23.5.1.3.1) received on pair RX_D2.

Values: nonnegative integers

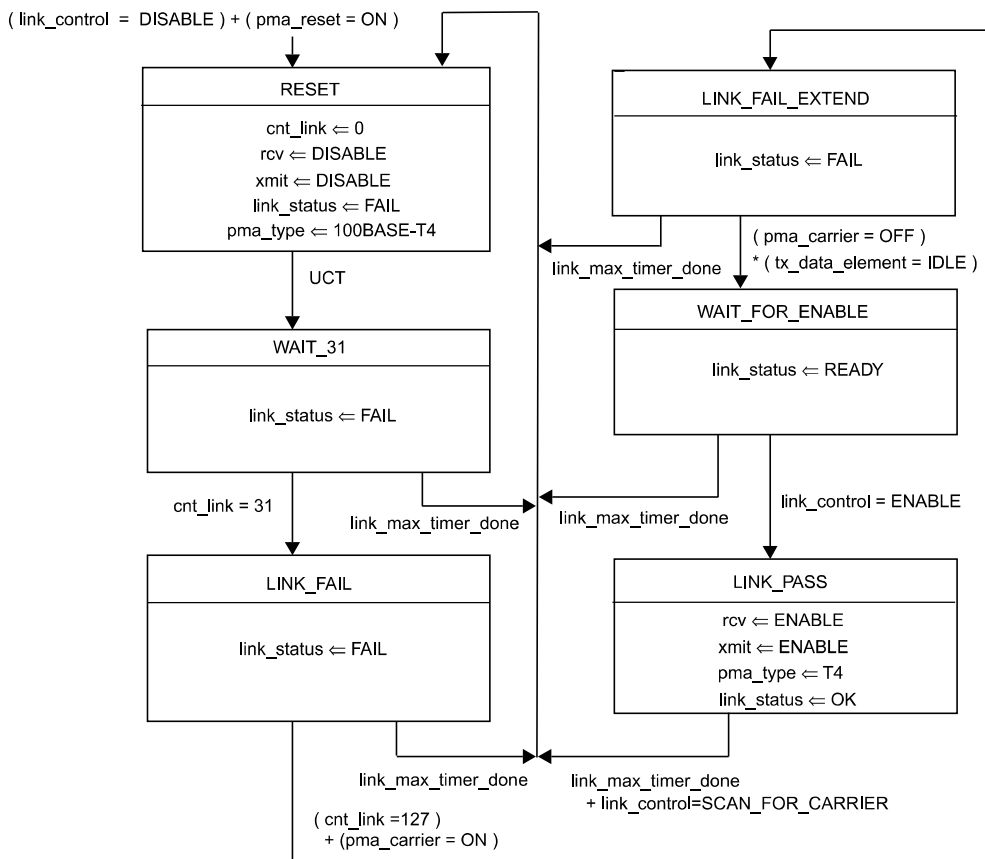
Reset to zero: On either of two conditions:

- a) While in any state other than LINK_PASS, reset counter to zero if successive link test pulses are received within link_100_min.
- b) While in any state, reset to zero if link_max_timer expires.

While in the LINK_PASS state, ignore pulses received within link_100_min (i.e., do not count them).

23.4.3.5 Link Integrity state diagram

The Link Integrity state diagram is shown in figure 23-12.



NOTE—The variables link_control and link_status are designated as link_control_[T4] and link_status_[T4], respectively, by the Auto-Negotiation Arbitration state diagram (figure 28-16).

Figure 23-12—Link integrity state diagram

23.5 PMA electrical specifications

This clause defines the electrical characteristics of the PHY at the MDI.

The ground reference point for all common-mode tests is the MII ground circuit. Implementations without an MII use the chassis ground. The values of all components in test circuits shall be accurate to within $\pm 1\%$ unless otherwise stated.

23.5.1 PMA-to-MDI interface characteristics

23.5.1.1 Isolation requirement

The PHY shall provide electrical isolation between the DTE, or repeater circuits including frame ground, and all MDI leads. This electrical separation shall withstand at least one of the following electrical strength tests:

- a) 1500 V rms at 50 Hz to 60 Hz for 60 s, applied as specified in subclause 5.3.2 of IEC 950: 1991.
- b) 2250 Vdc for 60 s, applied as specified in subclause 5.3.2 of IEC 950: 1991.
- c) A sequence of ten 2400 V impulses of alternating polarity, applied at intervals of not less than 1 s. The shape of the impulses shall be 1.2/50 μ s (1.2 μ s virtual front time, 50 μ s virtual time or half value), as defined in IEC 60.

There shall be no insulation breakdown, as defined in subclause 5.3.2 of IEC 950: 1991, during the test. The resistance after the test shall be at least 2 M Ω , measured at 500 Vdc.

23.5.1.2 Transmitter specifications

The PMA shall provide the Transmit function specified in 23.4.1.2 in accordance with the electrical specifications of this clause.

Where a load is not specified, the transmitter shall meet requirements of this clause when each transmit output is connected to a differentially connected 100 Ω resistive load.

23.5.1.2.1 Peak differential output voltage

While repetitively transmitting the ternary sequence [0 0 1 0 0 0 0 0 -1 0 0 0] (leftmost ternary symbol first), and while observing the differential transmitted output at the MDI, for any pair, with no intervening cable, the absolute value of both positive and negative peaks shall fall within the range of 3.15 V to 3.85 V (3.5 V \pm 10%).

23.5.1.2.2 Differential output templates

While repetitively transmitting the ternary sequence [0 0 1 0 0 0 0 0 -1 0 0 0], and while observing the transmitted output at the MDI, the observed waveform shall fall within the normalized transmit template listed in table 23-4. Portions of this table are represented graphically in figure 23-13. The entire normalized transmit template shall be scaled by a single factor between 3.15 and 3.85. It is a functional requirement that linear interpolation be used between points. The template time axis may be shifted horizontally to attain the most favorable match. In addition to this simple test pattern, all other pulses, including link integrity pulses and also including the first pulse of each packet preamble, should meet this same normalized transmit template, with appropriate shifting and linear superposition of the CS1 and CS-1 template limits. Transmitters are allowed to insert additional delay in the transmit path in order to meet the first pulse requirement, subject to the overall timing limitations listed in 23.11, Timing summary.

While transmitting the TP_IDL_100 signal, and while observing the transmitted output at the MDI, the observed waveform shall fall within the normalized link pulse template listed in table 23-4. Portions of this table are represented graphically in figure 23-14. The entire template shall be scaled by the same factor used for the normalized transmit template test. It is a functional requirement that linear interpolation be used between template points. The template time axis may be shifted horizontally to attain the most favorable match.

After transmitting seven or more consecutive CS0 waveforms during the TP_IDL_100 signal, each pair, as observed using the 100BASE-T4 Transmit Test Filter (23.5.1.2.3) connected to the MDI, shall attain a state within 50 mV of zero.

When the TX_D1, BI_D3, or BI_D4 pair is driven with a repeating pattern (1 -1 1 -1 ...) any harmonic measured at the MDI output shall be at least 27 dB below the fundamental at 12.5 MHz.

NOTES

1—The specification on maximum spectral components is not intended to ensure compliance with regulations concerning RF emissions. The implementor should consider any applicable local, national, or international regulations. Additional filtering of spectral components may therefore be necessary.

2—The repetitive pattern [0 0 1 0 0 0 0 0 -1 0 0 0] (leftmost ternary symbol first) may be synthesized using the 8B6T coding rules from a string of repeating data octets with value 73 hex. The repetitive pattern [1 -1 1 -1 1 -1] (leftmost ternary symbol first) may be synthesized using the 8B6T coding rules from a string of repeating data octets with value 92 hex.

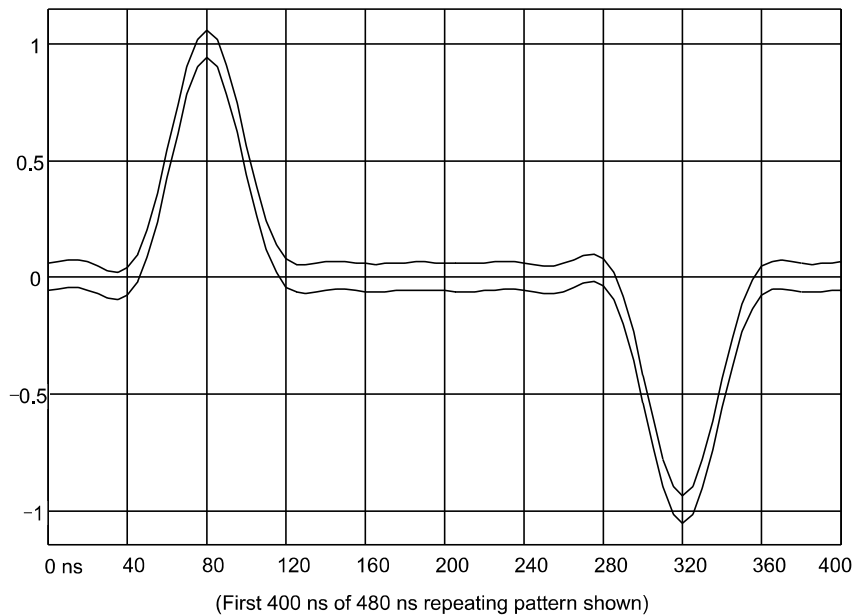


Figure 23-13—Normalized transmit template as measured at MD

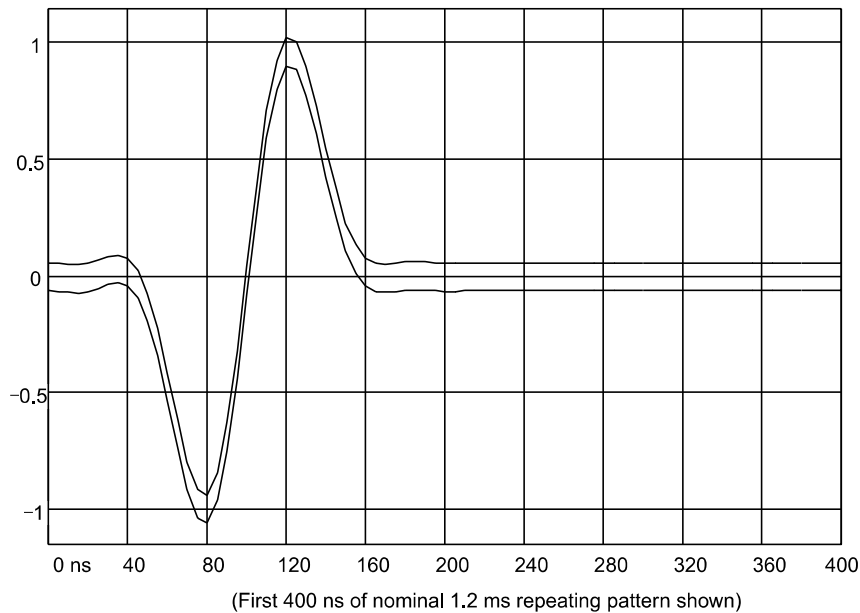


Figure 23-14—Normalized link pulse template as measured at MDI

The ideal template values may be automatically generated from the following equations:

$$\begin{array}{l} \text{Laplace transform of} \\ \text{Ideal transmit response} \end{array} \quad \text{IdealResponse}(s) = \frac{\text{Ideal}(s)}{\text{LPF}(s)}$$

Where $\text{Ideal}(s)$ is a 100% raised cosine system response

Where $\text{LPF}(s)$ is a 3-pole Butterworth low pass filter response with -3 dB point at 25 MHz

Convert $\text{IdealResponse}(s)$ from frequency domain to time domain

Use at least 8 samples per ternary symbol for the conversion

Superimpose alternating positive and negative copies of the ideal time response, separated by 6 ternary symbol times, to form the ideal transmit voltage waveform.

The template limits are formed by offsetting the ideal transmit voltage waveform by plus and minus 6% of its peak.

23.5.1.2.3 Differential output ISI (intersymbol interference)

While observing a pseudo-random 8B6T coded data sequence (with every 6T code group represented at least once) preceded by at least 128 octets and followed by at least 128 octets of data, and while observing the transmitted output through a 100BASE-T4 Transmit Test Filter (one implementation of which is depicted in figure 23-16), the ISI shall be less than 9%. The ISI for this test is defined by first finding the largest of the three peak-to-peak ISI error voltages marked in figure 23-15 as TOP ISI, MIDDLE ISI, and BOTTOM ISI.

The largest of these peak-to-peak ISI error voltages is then divided by the overall peak-to-peak signal voltage. (The technique of limiting the ratio of worst ISI to overall peak-to-peak voltage at 9% accomplishes the same end as limiting the ratio of worst ISI to nominal peak-to-peak at 10%.)

Table 23-4—Normalized voltage templates as measured at the MDI

Time, ns	Normalized transmit template, pos. limit	Normalized transmit template, neg. limit	Normalized link template, pos. limit	Normalized link template, neg. limit
0	0.060	-0.061	0.061	-0.060
5	0.067	-0.054	0.056	-0.065
10	0.072	-0.049	0.052	-0.069
15	0.072	-0.049	0.052	-0.069
20	0.063	-0.058	0.058	-0.063
25	0.047	-0.074	0.071	-0.050
30	0.030	-0.091	0.086	-0.035
35	0.023	-0.098	0.094	-0.027
40	0.041	-0.080	0.080	-0.041
45	0.099	-0.022	0.027	-0.094
50	0.206	0.085	-0.076	-0.197
55	0.358	0.237	-0.231	-0.352
60	0.544	0.423	-0.428	-0.549
65	0.736	0.615	-0.640	-0.761
70	0.905	0.784	-0.829	-0.950
75	1.020	0.899	-0.954	-1.075
80	1.060	0.940	-0.977	-1.098
85	1.020	0.899	-0.876	-0.997
90	0.907	0.786	-0.653	-0.774
95	0.744	0.623	-0.332	-0.453
100	0.560	0.439	0.044	-0.077
105	0.384	0.263	0.419	0.298
110	0.239	0.118	0.738	0.617
115	0.137	0.016	0.959	0.838
120	0.077	-0.044	1.060	0.940
125	0.053	-0.068	1.044	0.923
130	0.050	-0.071	0.932	0.811
135	0.057	-0.064	0.759	0.638
140	0.064	-0.057	0.565	0.444
145	0.067	-0.054	0.383	0.262
150	0.065	-0.056	0.238	0.117
155	0.061	-0.060	0.138	0.017
160	0.057	-0.064	0.081	-0.040
165	0.055	-0.066	0.057	-0.064
170	0.056	-0.065	0.054	-0.067
175	0.059	-0.062	0.058	-0.063

This is an Archive IEEE Standard. It has been superseded by a later version of this standard.

Table 23-4—Normalized voltage templates as measured at the MDI (Continued)

Time, ns	Normalized transmit template, pos. limit	Normalized transmit template, neg. limit	Normalized link template, pos. limit	Normalized link template, neg. limit
180	0.062	-0.059	0.063	-0.058
185	0.064	-0.057	0.064	-0.057
190	0.064	-0.057	0.063	-0.058
195	0.062	-0.059	0.060	-0.061
200	0.060	-0.061	0.058	-0.063
205	0.057	-0.064	0.058	-0.063
210	0.056	-0.065	0.059	-0.062
215	0.058	-0.063	0.060	-0.061
220	0.061	-0.060	0.062	-0.059
225	0.064	-0.057	0.062	-0.059
230	0.066	-0.055	0.062	-0.059
235	0.065	-0.056	0.061	-0.060
240	0.061	-0.060	0.060	-0.061
245	0.054	-0.067	0.060	-0.061
250	0.049	-0.072	0.060	-0.061
255	0.049	-0.072	0.060	-0.061
260	0.058	-0.063	0.061	-0.060
265	0.074	-0.047	0.061	-0.060
270	0.091	-0.030	0.061	-0.060
275	0.099	-0.022	0.061	-0.060
280	0.080	-0.041	0.060	-0.061
285	0.022	-0.099	0.060	-0.061
290	-0.085	-0.206	0.060	-0.061
295	-0.238	-0.359	0.060	-0.061
300	-0.423	-0.544	0.061	-0.060
305	-0.615	-0.736	0.061	-0.060
310	-0.783	-0.904	0.061	-0.060
315	-0.899	-1.020	0.061	-0.060
320	-0.940	-1.061	0.060	-0.061
325	-0.899	-1.020	0.060	-0.061
330	-0.786	-0.907	0.060	-0.061
335	-0.623	-0.744	0.060	-0.061
340	-0.439	-0.560	0.061	-0.060
345	-0.263	-0.384	0.061	-0.060
350	-0.118	-0.239	0.061	-0.060
355	-0.016	-0.137	0.061	-0.060

This is an Archive IEEE Standard. It has been superseded by a later version of this standard.

Table 23-4—Normalized voltage templates as measured at the MDI (Continued)

Time, ns	Normalized transmit template, pos. limit	Normalized transmit template, neg. limit	Normalized link template, pos. limit	Normalized link template, neg. limit
360	0.044	-0.077	0.060	-0.061
365	0.068	-0.053	0.060	-0.061
370	0.070	-0.051	0.060	-0.061
375	0.064	-0.057	0.060	-0.061
380	0.057	-0.064	0.061	-0.060
385	0.054	-0.067	0.061	-0.060
390	0.056	-0.065	0.061	-0.060
395	0.060	-0.061	0.061	-0.060
400	0.064	-0.057	0.060	-0.061
405	0.065	-0.056	0.060	-0.061
410	0.064	-0.057	0.060	-0.061
415	0.061	-0.060	0.060	-0.061
420	0.059	-0.062	0.061	-0.060
425	0.058	-0.063	0.061	-0.060
430	0.059	-0.062	0.061	-0.060
435	0.060	-0.061	0.061	-0.060
440	0.061	-0.060	0.060	-0.061
445	0.062	-0.059	0.060	-0.061
450	0.062	-0.059	0.060	-0.061
455	0.061	-0.060	0.060	-0.061
460	0.060	-0.061	0.061	-0.060
465	0.059	-0.062	0.061	-0.060
470	0.060	-0.061	0.061	-0.060
475	0.060	-0.061	0.061	-0.060
480	0.061	-0.060	0.060	-0.061

It is a mandatory requirement that the peak-to-peak ISI, and the overall peak-to-peak signal voltage, be measured at a point in time halfway between the nominal zero crossings of the observed eye pattern.

It is a mandatory requirement that the 100BASE-T4 Transmit Test Filter perform the function of a third-order Butterworth filter with its -3 dB point at 25.0 MHz.

One acceptable implementation of a 100BASE-T4 Transmit Test Filter appears in figure 23-16. That implementation uses the 100BASE-T4 Transmit Test Filter as a line termination. The output of the filter is terminated in 100 Ω. It is a mandatory requirement that such implementations of the 100BASE-T4 Transmit Test Filter be designed such that the reflection loss of the filter, when driven by a 100 Ω source, exceeds 17 dB across the frequency range 2 to 12.5 MHz.

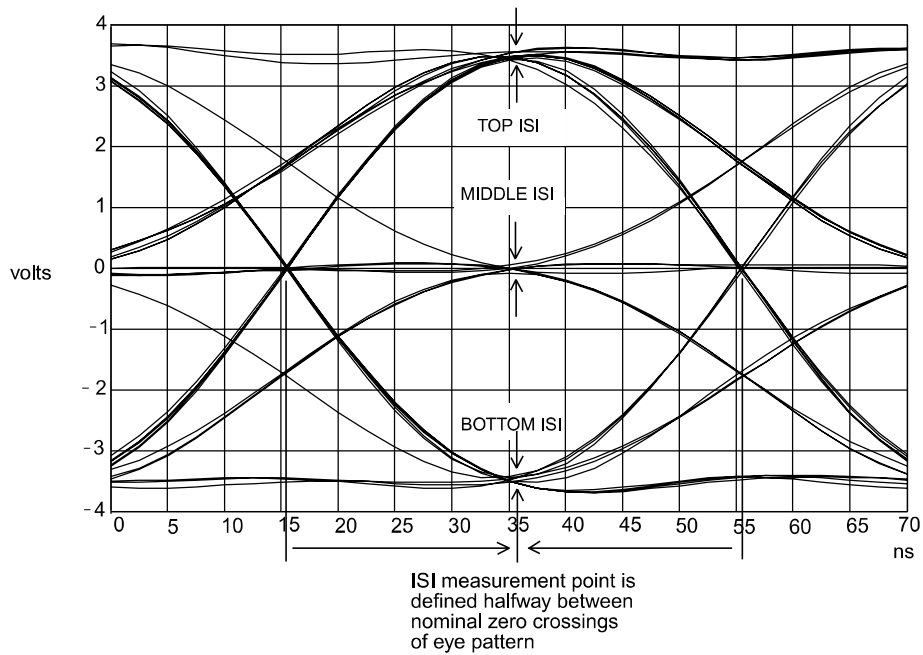


Figure 23-15—Definition of sampling points for ISI measurement

Equivalent circuits that implement the same overall transfer function are also acceptable. For example, the 100BASE-T4 Transmit Test Filter may be tapped onto a line in parallel with an existing termination. It is a mandatory requirement that such implementations of the 100BASE-T4 Transmit Test Filter be designed with an input impedance sufficiently high that the reflection loss of the parallel combination of filter and 100 Ω termination, when driven by 100 Ω, exceeds 17 dB across the frequency range 2 to 12.5 MHz.

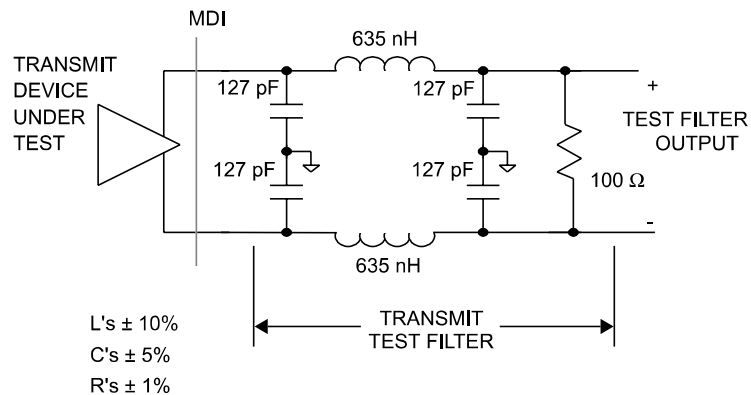


Figure 23-16—Acceptable implementation of transmit test filter

This is an Archive IEEE Standard. It has been superseded by a later version of this standard.

23.5.1.2.4 Transmitter differential output impedance

The differential output impedance as measured at the MDI for each transmit pair shall be such that any reflection due to differential signals incident upon the MDI from a balanced cable having an impedance of $100\ \Omega$ is at least 17 dB below the incident signal, over the frequency range of 2.0 MHz to 12.5 MHz. This return loss shall be maintained at all times when the PHY is fully powered.

With every transmitter connected as in figure 23-17, and while transmitting a repeating sequence of packets as specified in table 23-3, the amount of droop on any transmit pair as defined in figure 23-18 during the transmission of eop1 and eop4 shall not exceed 6.0%.

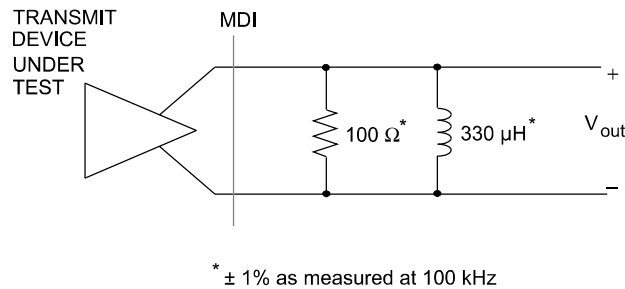


Figure 23-17—Output impedance test setup

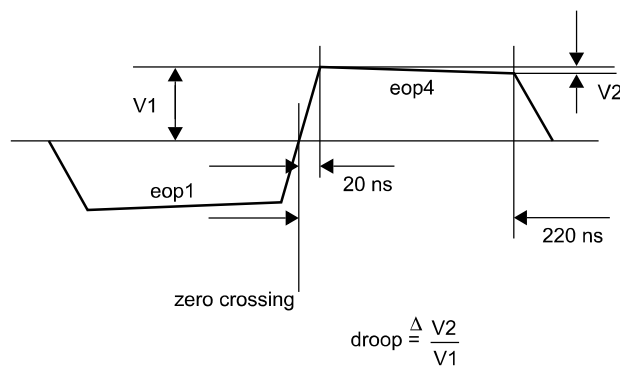


Figure 23-18—Measurement of output droop

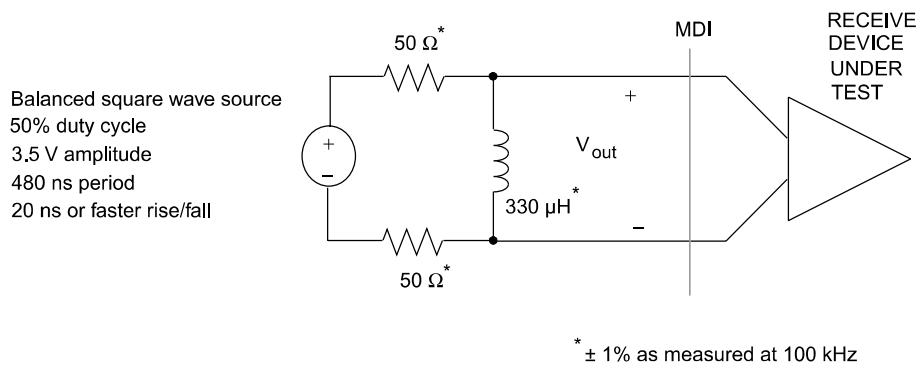


Figure 23-19—Input impedance test setup

Table 23-5—Sequence of packets for droop test

Packet sequence (Transmit this sequence of packets in a repetitive loop)	Packet length (Number of data octets)	Data, hex (All octets in each packet are the same)
first packet	64	AA
second packet	65	AA
third packet	66	AA

23.5.1.2.5 Output timing jitter

While repetitively transmitting a random sequence of valid 8B6T code words, and while observing the output of a 100BASE-T4 Transmit Test Filter connected at the MDI to any of the transmit pairs as specified in 23.5.1.2.3, the measured jitter shall be no more than 4 ns p-p. For the duration of the test, each of the other transmit pairs shall be connected to either a 100BASE-T4 Transmit Test Filter or a 100 Ω resistive load.

NOTES

1—Jitter is the difference between the actual zero crossing point in time and the ideal time. For various ternary transitions, the zero crossing time is defined differently. For transitions between +1 and -1 or vice versa, the zero crossing point is defined as that point in time when the voltage waveform crosses zero. For transitions between zero and the other values, or from some other value to zero, the zero crossing time is defined as that point in time when the voltage waveform crosses the boundary between logical voltage levels, halfway between zero volts and the logical +1 or logical -1 ideal level.

2—The ideal zero crossing times are contained in a set of points $\{t_n\}$ where $t_n = t_0 + n/f$, where n is an integer, and f is in the range 25.000 MHz \pm 0.01%. A collection of zero crossing times satisfies the jitter requirement if there exists a pair (t_0, f) such that each zero crossing time is separated from some member of $\{t_n\}$ by no more than 4 ns.

23.5.1.2.6 Transmitter impedance balance

The common-mode to differential-mode impedance balance of each transmit output shall exceed

$$29 - 17 \log\left(\frac{f}{10}\right) \text{ dB}$$

where f is the frequency (in MHz) over the frequency range 2.0 MHz to 12.5 MHz. The balance is defined as

$$20 \log\left(\frac{E_{\text{cm}}}{E_{\text{dif}}}\right)$$

where E_{cm} is an externally applied sine-wave voltage as shown in figure 23-20.

NOTE—The balance of the test equipment (such as the matching of the test resistors) must be insignificant relative to the balance requirements.

23.5.1.2.7 Common-mode output voltage

The implementor should consider any applicable local, national, or international regulations. Driving unshielded twisted pairs with high-frequency, common-mode voltages may result in interference to other equipment. FCC conducted and radiated emissions tests may require that, while transmitting data, the magnitude of the total common-mode output voltage, $E_{\text{cm(out)}}$, on any transmit circuit, be less than a few millivolts when measured as shown in figure 23-21.

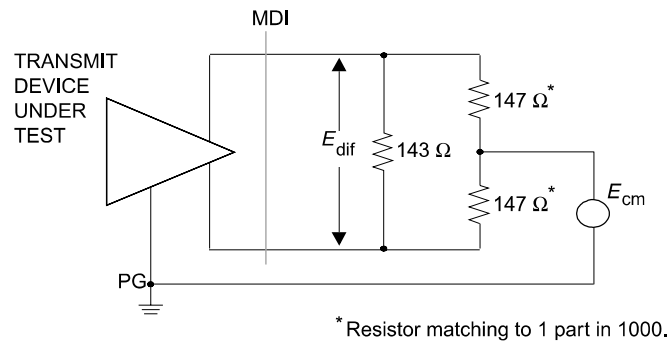


Figure 23-20—Transmitter impedance balance and common-mode rejection test circuit

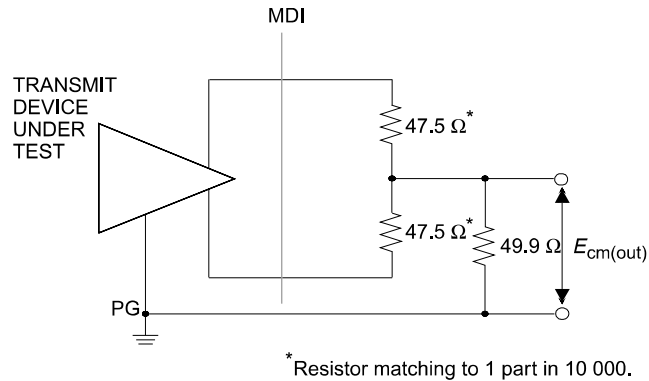


Figure 23-21—Common-mode output voltage test circuit

23.5.1.2.8 Transmitter common-mode rejection

The application of E_{cm} as shown in figure 23-20 shall not change the differential voltage at any transmit output, E_{dif} , by more than 100 mV for all data sequences while the transmitter is sending data. Additionally, the edge jitter added by the application of E_{cm} shall be no more than 1.0 ns. E_{cm} shall be a 15 V peak 10.1 MHz sine wave.

23.5.1.2.9 Transmitter fault tolerance

Transmitters, when either idle or nonidle, shall withstand without damage the application of short circuits across any transmit output for an indefinite period of time and shall resume normal operation after such faults are removed. The magnitude of the current through such a short circuit shall not exceed 420 mA.

Transmitters, when either idle or nonidle, shall withstand without damage a 1000 V common-mode impulse applied at E_{cm} of either polarity (as indicated in figure 23-22). The shape of the impulse shall be 0.3/50 μ s (300 ns virtual front time, 50 μ s virtual time of half value), as defined in IEC 60.

23.5.1.2.10 Transmit clock frequency

The ternary symbol transmission rate on each pair shall be 25.000 MHz \pm 0.01%.