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MULTIRATE DIGITAL SIGNAL PROCESSING

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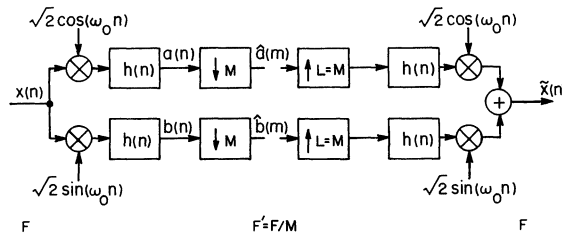


Figure 2.25 Block diagram of a quadrature modulator and sampling of bandpass signals using integer decimation and interpolation.

$$1 > \frac{L}{M} \geq \frac{\omega_{\Delta}}{2\pi} \quad (2.110)$$

As seen in Fig. 2.26, the outputs of the cosine and sine modulators are first increased in sampling rate by a factor L . They are then lowpass filtered by the filters $h(k)$, which must approximate the ideal characteristic (referenced to the sampling rate $F' = LF$)

$$\tilde{H}(e^{j\omega'}) = \begin{cases} 1, & 0 \leq |\omega'| \leq \frac{\omega_{\Delta}}{2L} \\ 0, & \text{otherwise} \end{cases} \quad (2.111)$$

and then decimated by M . The resulting quadrature signals are at the sampling rate $F'' = (L/M)F$. In the quadrature demodulator the reverse process takes place, as seen in Fig. 2.26.

2.4.4 Single-Sideband Modulation

The quadrature modulation approach coupled with methods of interpolation and decimation as discussed in the preceding section permits the sampling of bandpass signals, with arbitrary band edges, at their minimum required sampling rate. One consequence of this method, however, is that the resulting sampled signal $\hat{a}(m) + j\hat{b}(m)$ is a complex signal, and in some applications this may not be desired. By a slight modification of the quadrature approach, however, it can be shown that a real signal output can be obtained which corresponds to a “single-sideband” modulated format at the minimal sampling rate [2.5, 2.6].

Figure 2.27 illustrates this approach. The complex signal $a(n) + jb(n)$ is modulated by $e^{-j(\omega_{\Delta}/2)n}$ and its conjugate is modulated by $e^{j(\omega_{\Delta}/2)n}$ to produce the respective sideband signals X_{γ}^{-} and X_{γ}^{+} , as shown in Fig. 2.27(b). Summing these

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