# OPERATIONAL AMPLIFIERS Design and Applications

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#### OPERATIONAL AMPLIFIERS

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Contributor Preface Historical I

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Fig. 10.11 Triangle generator waveforms.

would be  $R_2 = 10 \text{ k}\Omega$ ,  $R_1 = 5 \text{ k}\Omega$ ,  $R_3 = 100 \text{ k}\Omega$ ,  $C_1 = 0.1 \mu\text{F}$ . These values will provide a triangle wave of approximately +7.2 to -7.2 V swing at a frequency of 500 Hz.

#### 10.3 Sine-wave Generators<sup>3-5</sup>

One of the most important waveforms that an engineer may be called upon to generate is the sinusoidal waveform. In this section we shall treat a variety of techniques which may be used to perform this task. Specifically we shall investigate the use of Wien-bridge oscillators, quadrature oscillators, and phase-shift oscillators. For the Wien-bridge and the quadrature oscillator case several different circuits are considered. These may be applied to different special requirements as indicated in the discussion of the circuits.

10.3.1 Wien-bridge oscillator—general description A Wien bridge may be combined with an operational amplifier to form an excellent sinewave generator. Some sort of automatic gain control is generally used to stabilize the magnitude of the output sinusoid. A general schematic of a Wien-bridge oscillator is shown in Fig. 10.12. To see how this circuit operates let us assume that the output  $e_o$  is a sinusoid; then the feedback ratio of the bridge is given by

$$\frac{Z_2}{Z_1+Z_2} = \frac{R_2}{R_1+R_2(1+C_2/C_1)+j(\omega R_1 R_2 C_2-1/\omega C_1)}$$

where  $Z_1 = R_1 + 1/j\omega C_1$  and  $Z_2 = R_2/(1 + j\omega R_2 C_2)$ . The operational amplifier will maintain 0 V between its input terminals; thus,

$$\widehat{\mathbf{B}}_{\mathsf{o}} = \frac{\mathbf{Z}_2}{\mathbf{Z}_1 + \mathbf{Z}_2} \,\widehat{\mathbf{E}}_{\mathsf{o}}$$



Fig. 10.12 Wien-bridge oscillator.

where  $\hat{E}_{o}$  is a phasor representing the voltage  $e_{o}(t)$ . The condition for oscillation is

$$\omega_{o} R_{1} R_{2} C_{2} - \frac{1}{\omega_{o} C_{1}} = 0$$
$$\omega_{o} = \frac{1}{\sqrt{R_{1} R_{2} C_{1} C_{2}}}$$

If we make  $R_1 = R_2$  and  $C_1 = C_2$ , then

$$\omega_{0} = \frac{1}{\mathrm{R}_{1}\mathrm{C}_{1}}$$
 and  $\beta = \frac{1}{3}$ 

If  $\beta = \frac{1}{3}$  and the condition of  $R_1 = R_2$  and  $C_1 = C_2$  is met, then the output will be a sinusoid of frequency  $1/2\pi RC$ .

It should be noted that, so long as  $\beta$  is  $\frac{1}{3}$ , the circuit will oscillate at any amplitude. Also, if  $\beta$  is less than  $\frac{1}{3}$ , the oscillation will diverge and if  $\beta$  is more than  $\frac{1}{3}$  the oscillation will converge. Thus it is common practice to provide some sort of automatic amplitude control. This is usually

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done by varying the negative feedback gain ( $\beta$ ) to stabilize the oscillator. Incandescent lamps, thermistors, FETs, diode bridges, or general-purpose multipliers can all be used for such gain control purposes.

10.3.2 Precise Wien-bridge oscillator<sup>4,5</sup> As a typical implementation of the general Wien-bridge oscillator diagram shown in Fig. 10.12, consider the circuit shown in Fig. 10.13. The actual Wien bridge is formed by  $R_1$ ,  $C_1$ ,  $R_2$ , and  $C_2$ . The oscillatory output of amplifier  $A_1$  is amplified by  $A_2$ , and the output level is sensed by the absolute-value circuit of  $A_3$  and  $A_4$ . The amplifier  $A_4$  acts as an error integrator and will stabilize only when the absolute value of the input equals the reference amplitude. A diode bridge is used for varying the negative feedback of  $A_1$ . An FET can be used for gain control rather than the diode bridge if desired.

The integrator gain is set by capacitor  $C_3$ . The choice of  $C_3$  is a tradeoff between response time and distortion. Small values of  $C_3$  will allow the circuit to reach its stable value very rapidly. Also, response to any disturbance is rapid. On the other hand, making  $C_3$  large will minimize distortion. The frequency of oscillation, as discussed previously, will be

$$\mathbf{f}_{o} = \frac{1}{2\pi \mathbf{R}_{1} \mathbf{C}_{1}}$$

where  $R_1 = R_2$  and  $C_1 = C_2$ . Frequencies in the range of 10 Hz to 10 kHz are practical for this circuit. Distortion of less than 0.1 percent and excellent frequency stability are readily achieved. The circuit will operate at frequencies above 10 kHz, but the type of operational amplifier must be carefully chosen and stray capacitances should be considered.

Although, in the circuit shown in Fig. 10.13, five operational amplifiers are used, similar circuits are available in miniature encapsulated packages. In such packages, integrated-circuit operational amplifiers are usually used to minimize the size.

10.3.3 Low-cost Wien-bridge oscillator The Wien-bridge oscillator circuit presented in the preceding paragraphs has the disadvantage of requiring five operational amplifiers. In Fig. 10.14 a circuit diagram for a Wien-bridge oscillator which requires only one operational amplifier is given. The primary virtue of this circuit is that very few components are required. Distortion will be greater than with the previously discussed Wien bridge. But, depending upon care of adjustment, distortion will be in the range of 1 to 5 percent. This circuit has high output impedance, and any loading at  $e_0$  will shift the operating point of the diodes, which will in turn change the amplitude. Thus this circuit must be used with either a fixed load at  $e_0$  or a buffer must be added. As with

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